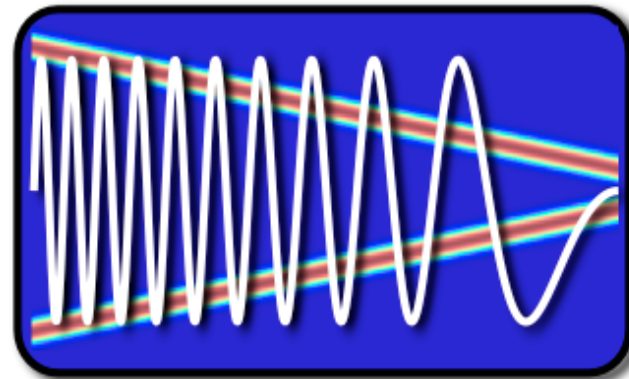


EE123



# Digital Signal Processing

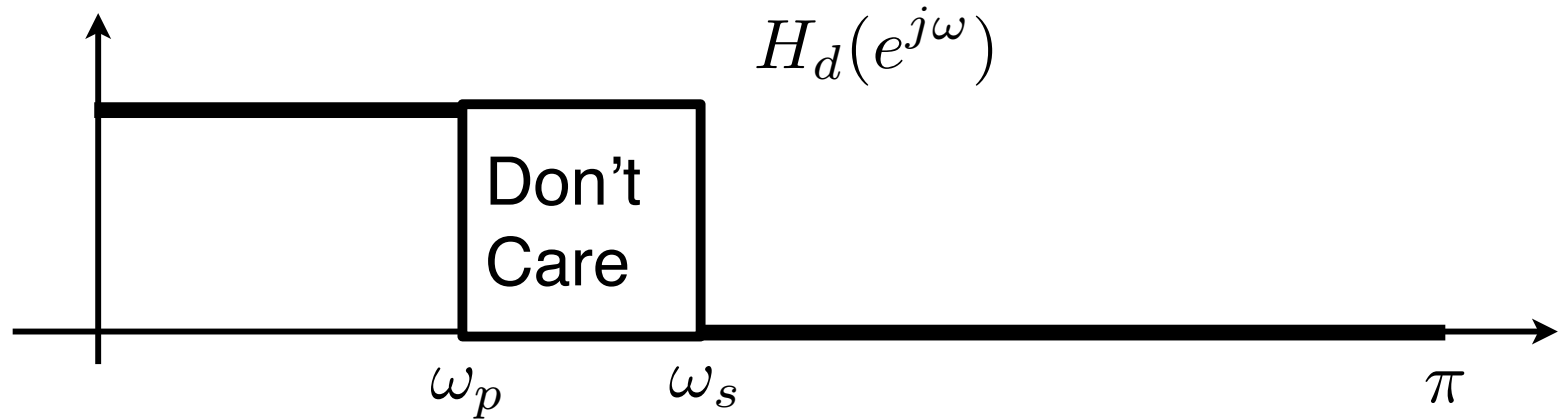
## Lecture 21 Optimal Filter Design

# Optimal Filter Design

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- Window method
  - Design Filters heuristically using windowed sinc functions
- Optimal design
  - Design a filter  $h[n]$  with  $H(e^{j\omega})$
  - Approximate  $H_d(e^{j\omega})$  with some optimality criteria - or satisfies specs.

## Optimality



- Least Squares:

$$\text{minimize} \int_{\omega \in \text{care}} |H(e^{j\omega}) - H_d(e^{j\omega})|^2 d\omega$$

Variation: weighted least-squares

$$\text{minimize} \int_{-\pi}^{\pi} W(\omega) |H(e^{j\omega}) - H_d(e^{j\omega})|^2 d\omega$$

## Optimality

---

- Chebychev Design (min-max)

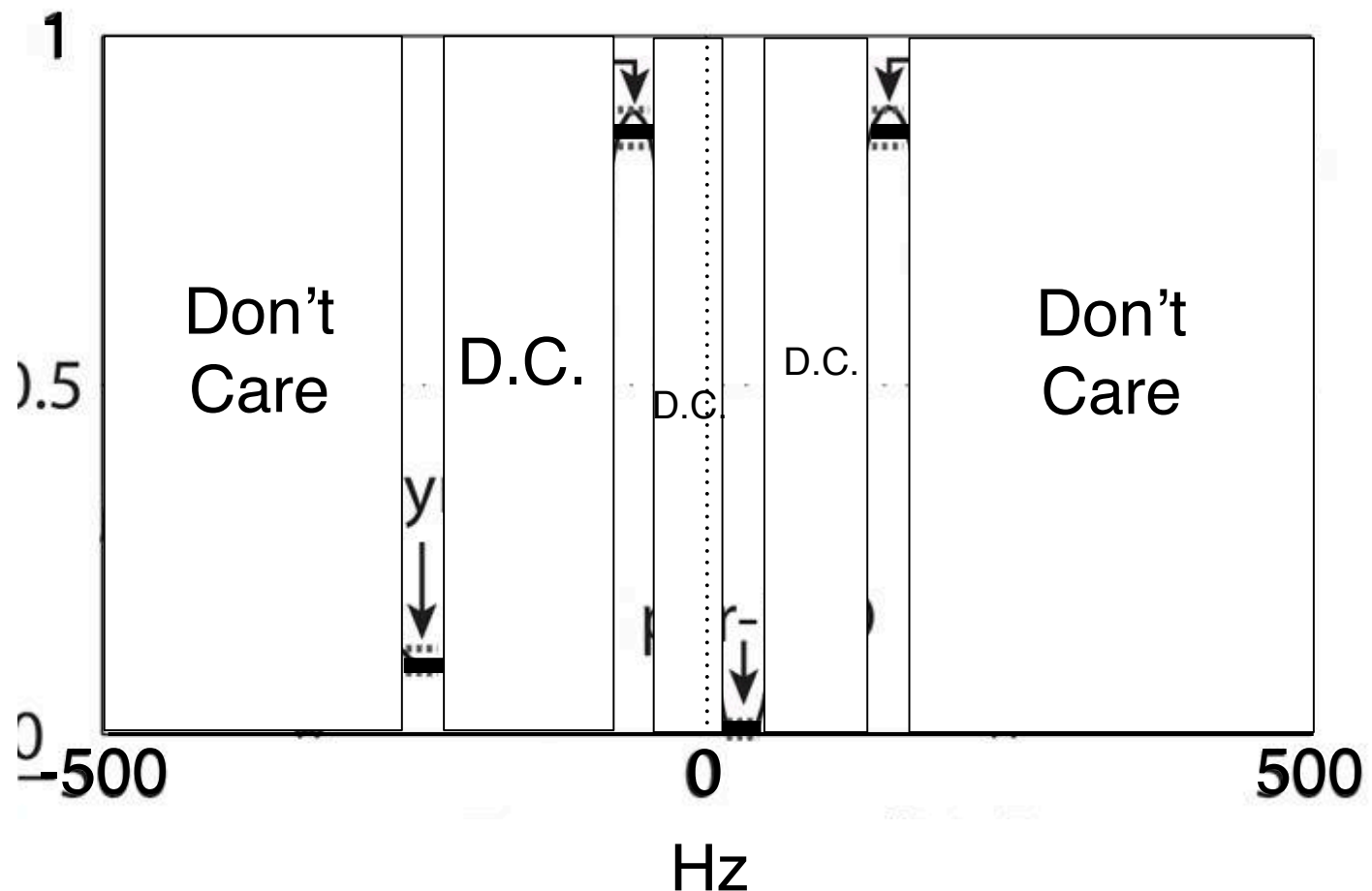
$$\text{minimize}_{\omega \in \text{care}} \quad \max |H(e^{j\omega}) - H_d(e^{j\omega})|$$

- Parks-McClellan algorithm - equi-ripple
- Also known as Remez exchange algorithms (signal.remez)
- Can also use convex optimization

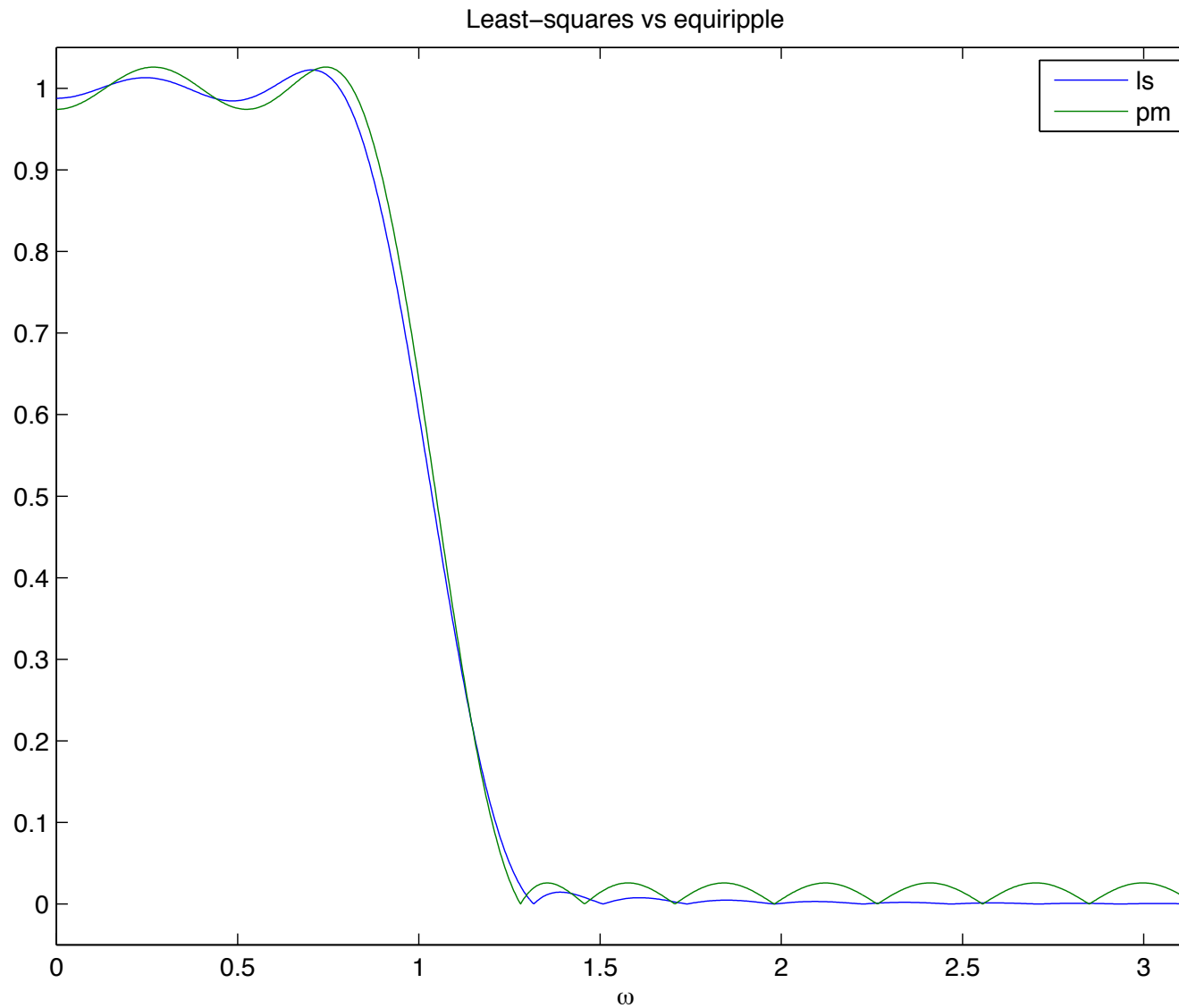
# Example of Complex Filter

Larson et. al, "Multiband Excitation Pulses for Hyperpolarized  $^{13}\text{C}$  Dynamic Chemical Shift Imaging" JMR 2008;194(1):121-127

Need to design 11 taps filter with following frequency response:



# Least-Squares v.s. Min-Max



## Design Through Optimization

---

- Idea: Sample/discretize the frequency response

$$H(e^{j\omega}) \Rightarrow H(e^{j\omega_k})$$

- Sample points are fixed  $\omega_k = k \frac{\pi}{P}$

$$-\pi \leq \omega_1 < \dots < \omega_p \leq \pi$$

- $M+1$  is the filter order
- $P \gg M + 1$  ( rule of thumb  $P=15M$ )
- Yields a (good) approximation of the original problem

## Example: Least Squares

---

- Target: Design  $M+1 = 2N+1$  filter
- First design non-causal  $\tilde{H}(e^{j\omega})$  and hence  $\tilde{h}[n]$
- Then, shift to make causal

$$h[n] = \tilde{h}[n - M/2]$$

$$H(e^{j\omega}) = e^{-j\frac{M}{2}} \tilde{H}(e^{j\omega})$$



## Example: Least Squares

---

- Matrix formulation:

$$\tilde{h} = \left[ \tilde{h}[-N], \tilde{h}[-N+1], \dots, \tilde{h}[N] \right]^T$$

$$b = \left[ H_d(e^{j\omega_1}), \dots, H_d(e^{j\omega_P}) \right]^T$$

$$A = \begin{bmatrix} e^{-j\omega_1(-N)} & \dots & e^{-j\omega_1(+N)} \\ \vdots & & \\ e^{-j\omega_P(-N)} & \dots & e^{-j\omega_P(+N)} \end{bmatrix}$$

$$\operatorname{argmin}_{\tilde{h}} ||A\tilde{h} - b||^2$$

## Least Squares

---

$$\operatorname{argmin}_{\tilde{h}} \quad ||A\tilde{h} - b||^2$$

Solution:

$$\tilde{h} = (A^* A)^{-1} A^* b$$

- Result will generally be non-symmetric and complex valued.
- However, if  $\tilde{H}(e^{j\omega})$  is real,  $\tilde{h}[n]$  should have symmetry!

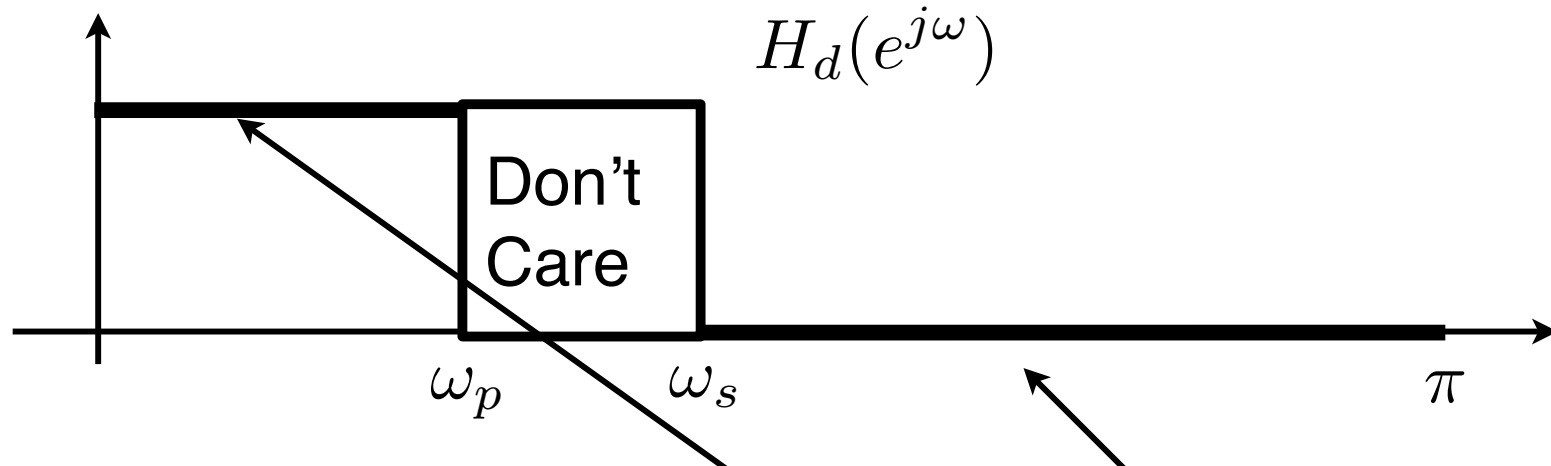
## Design of Linear-Phase L.P Filter

---

- Suppose:
  - $\tilde{H}(e^{j\omega})$  is real-symmetric
  - M is even (M+1 taps)
- Then:
  - $\tilde{h}[n]$  is real-symmetric around midpoint
- So:

$$\begin{aligned}\tilde{H}(e^{j\omega}) &= \tilde{h}[0] + \tilde{h}[1]e^{-j\omega} + \tilde{h}[-1]e^{+j\omega} \\ &\quad + \tilde{h}[2]e^{-j2\omega} + \tilde{h}[-2]e^{+j2\omega} \dots \\ &= \tilde{h}[0] + 2\cos(\omega)\tilde{h}[1] + 2\cos(2\omega)\tilde{h}[2] + \dots\end{aligned}$$

# Least-Squares Linear-Phase Filter



Given M,  $\omega_P$ ,  $\omega_s$  find the best LS filter:

$$A = \begin{bmatrix} \text{ } \\ \text{ } \end{bmatrix}$$

$$b = \begin{bmatrix} \text{ } \end{bmatrix}, \begin{bmatrix} \text{ } \end{bmatrix}^T$$

# Least-Squares Linear-Phase Filter

Given  $M$ ,  $\omega_P$ ,  $\omega_S$  find the best LS filter:

$$A = \begin{bmatrix} \phantom{0} & \phantom{0} \\ \phantom{0} & \phantom{0} \end{bmatrix}$$

$$\tilde{h}_+ = [\tilde{h}[0], \dots, \tilde{h}[\frac{M}{2}]]^T = (A^* A)^{-1} A^* b$$

$$\tilde{h} = \begin{cases} \tilde{h}_+[n] & n \geq 0 \\ \tilde{h}_+[-n] & n < 0 \end{cases}$$

$$h[n] = \tilde{h}[n - M/2]$$

## Extension:

---

- LS has no preference for pass band or stop band
- Use weighting of LS to change ratio

want to solve the discrete version of:

$$\text{minimize } \int_{-\pi}^{\pi} W(\omega) |H(e^{j\omega}) - H_d(e^{j\omega})|^2 d\omega$$

where  $W(\omega)$  is  $\delta_p$  in the pass band and  $\delta_s$  in stop band

Similarly:  $W(\omega)$  is 1 in the pass band and  $\delta_p/\delta_s$  in stop band

## Weighted Least-Squares

---

$$\operatorname{argmin}_{\tilde{h}_+} (A\tilde{h}_+ - b)^* W^2 (A\tilde{h}_+ - b)$$

Solution:

$$\tilde{h}_+ = (A^* W^2 A)^{-1} W^2 A^* b$$

$$W = \begin{bmatrix} 1 & & & & & 0 \\ & 1 & & & & \\ & & \dots & & & \\ & & & \frac{\delta_p}{\delta_s} & & \\ & & & & \dots & \\ 0 & & & & & \frac{\delta_p}{\delta_s} \end{bmatrix}$$

## Min-Max optimal Filters

---

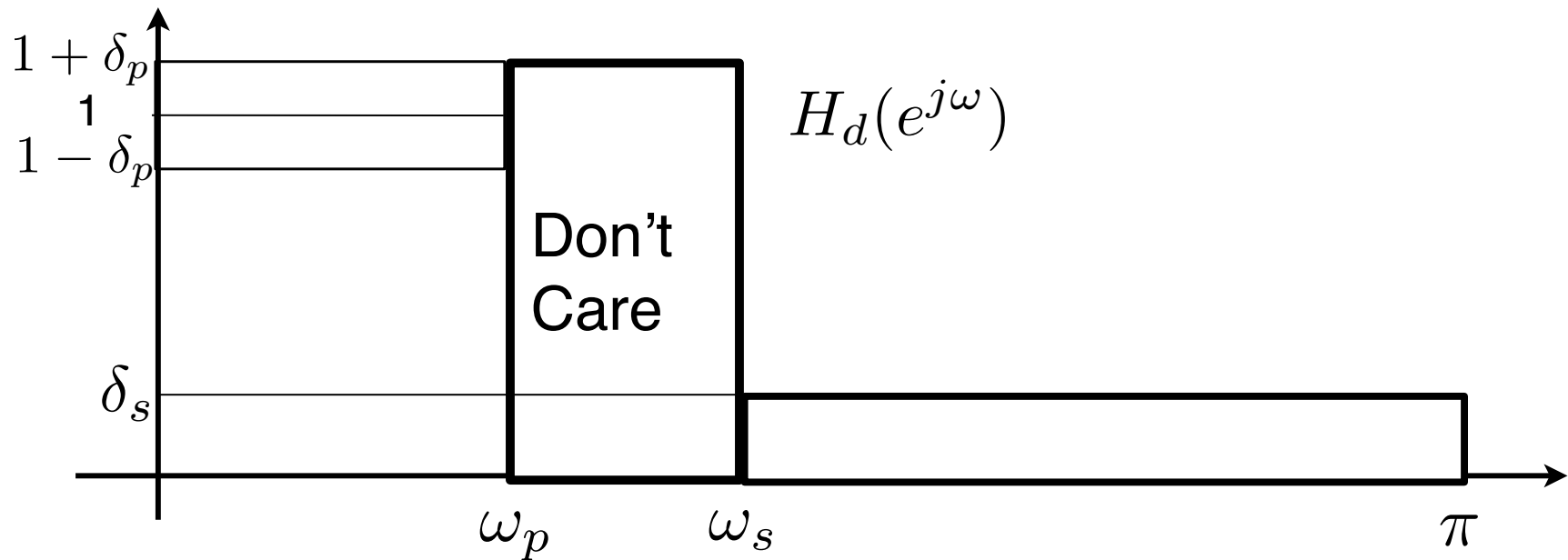
- Chebychev Design (min-max)

$$\text{minimize}_{\omega \in \text{care}} \quad \max |H(e^{j\omega}) - H_d(e^{j\omega})|$$

- Parks-McClellan algorithm - equi-ripple
- Also known as Remez exchange algorithms (signal.remez)
- Also with convex optimization



# Specifications



- Filter specifications are given in terms of boundaries

# Min-Max Filter Design

---

- Minimize:

- max pass-band ripple

$$1 - \delta_p \leq |H(e^{j\omega})| \leq 1 + \delta_p, \quad 0 \leq \omega \leq \omega_p$$

- min-max stop-band ripple

$$|H(e^{j\omega})| \leq \delta_s, \quad \omega_s \leq \omega \leq \pi$$

# Min-max Ripple Design

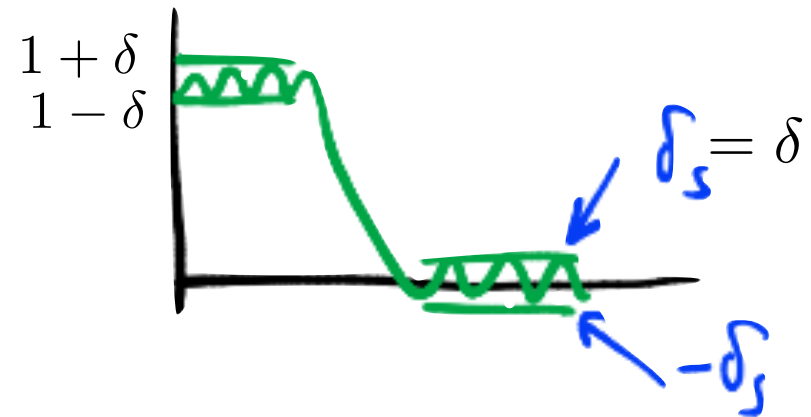
- Recall,  $\tilde{H}(e^{j\omega})$  is symmetric and real

- Given  $\omega_p$   $\omega_s$   $M$ , find  $\delta, \tilde{h}_+$ :

minimize  $\delta$

Subject to :

$$\begin{aligned} 1 - \delta &\leq \tilde{H}(e^{j\omega_k}) \leq 1 + \delta & 0 \leq \omega_k \leq \omega_p \\ -\delta &\leq \tilde{H}(e^{j\omega_k}) \leq \delta & \omega_s \leq \omega_k \leq \pi \\ \delta &> 0 \end{aligned}$$



- Solution is a linear program in  $\delta, \tilde{h}_+$
- A well studied class of problems

## Min-Max Ripple via Linear Programming

$$\begin{aligned} & \text{minimize} && \delta \\ & \text{subject to :} && \\ & && 1 - \delta \preceq A_p \tilde{h}_+ \preceq 1 + \delta \\ & && -\delta \preceq A_s \tilde{h}_+ \preceq \delta \\ & && \delta > 0 \end{aligned}$$

$$A_p = \begin{bmatrix} 1 & 2 \cos(\omega_1) & \cdots & 2 \cos(\frac{M}{2} \omega_1) \\ & \vdots & & \\ 1 & 2 \cos(\omega_p) & \cdots & 2 \cos(\frac{M}{2} \omega_p) \end{bmatrix}$$

$$A_s = \begin{bmatrix} 1 & 2 \cos(\omega_s) & \cdots & 2 \cos(\frac{M}{2} \omega_1) \\ & \vdots & & \\ 1 & 2 \cos(\omega_P) & \cdots & 2 \cos(\frac{M}{2} \omega_P) \end{bmatrix}$$

capital P



# Convex Optimization

---

- Many tools and Solvers
- Tools:
  - CVX (Matlab) <http://cvxr.com/cvx/>
  - CVXOPT, CVXMOD (Python)
- Engines:
  - Sedumi (Free)
  - MOSEK (commercial)
- Take EE127!

# Using CVX (in Matlab)

```
M = 16;
wp = 0.5*pi;
ws = 0.6*pi;
MM = 15*M;
w = linspace(0,pi,MM);
```

```
idxp = find(w <= wp);
idxs = find(w >= ws);
```

```
Ap = [ones(length(idxp),1) 2*cos(kron(w(idxp)',
[1:M/2]))];
As = [ones(length(idxs),1) 2*cos(kron(w(idxs)',
[1:M/2]))];
```

% optimization

```
cvx_begin
```

```
    variable hh(M/2+1,1);
```

```
    variable d(1,1);
```

```
    minimize(d)
```

```
    subject to
```

```
        Ap*hh <= 1+d;
```

```
        Ap*hh >= 1-d;
```

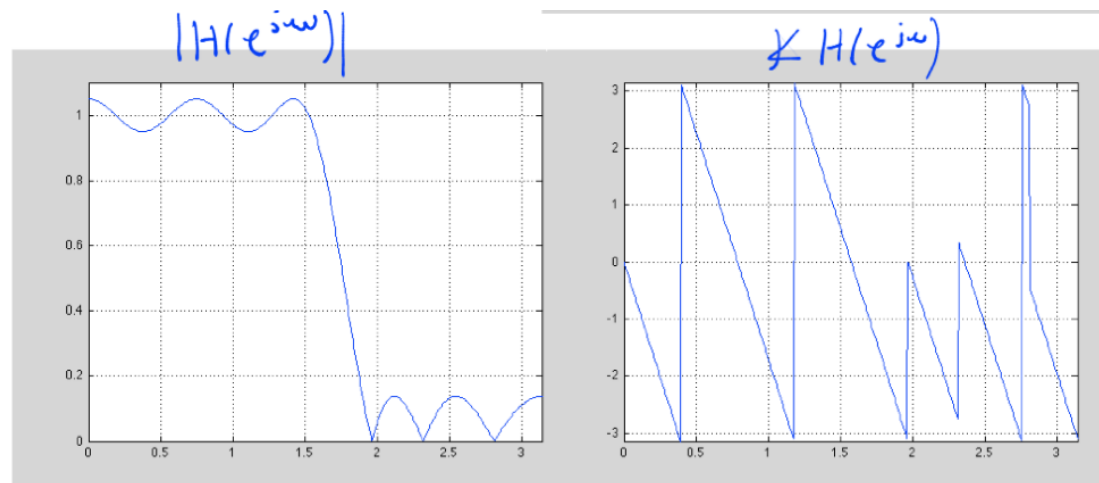
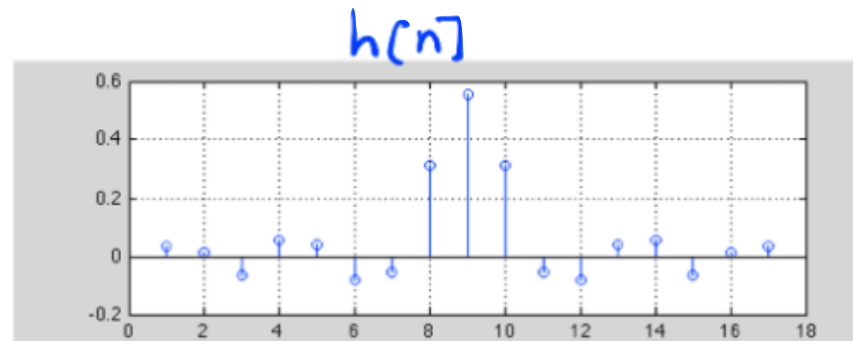
```
        As*hh < d;
```

```
        As*hh > -d;
```

```
        ds > 0;
```

```
cvx_end
```

```
h = [hh(end:-1:1) ; hh(2:end)];
```



## Variations:

---

- Convex Problems:
  - Fix  $\delta_s$  optimize for  $\delta_p$
  - Fix  $\delta_p$  optimize for  $\delta_s$
  - Linear constraints on  $h[n]$
- Quasi-Convex (feasible through bisection)
  - Fix  $\delta_p, \delta_s, M$ , minimize  $\Delta\omega = \omega_s - \omega_p$
  - Fix  $\delta_p, \delta_s, \Delta\omega = \omega_s - \omega_p$ , minimize  $M$

## Bisection Example: Minimize M

---

- given  $\delta_p$ ,  $\delta_s$ ,  $\Delta\omega = \omega_s - \omega_p$  Initialize problem with:
  - Set  $M_{\min}$  to be small and hence infeasible
  - Set  $M_{\max}$  to be large and hence feasible
  - Set  $M = \text{floor}(M_{\max}/2 + M_{\min}/2)$
- Given  $M$ ,  $\delta_p$ ,  $\Delta\omega = \omega_s - \omega_p$  solve for minimum  $\delta_s$ 
  - If  $\delta_s$  violates constraints, set  $M_{\min} = M$
  - if  $\delta_s$  within constraints, set  $M_{\max} = M$
  - Set  $M = \text{floor}(M_{\max}/2 + M_{\min}/2)$
  - Repeat till M is tight