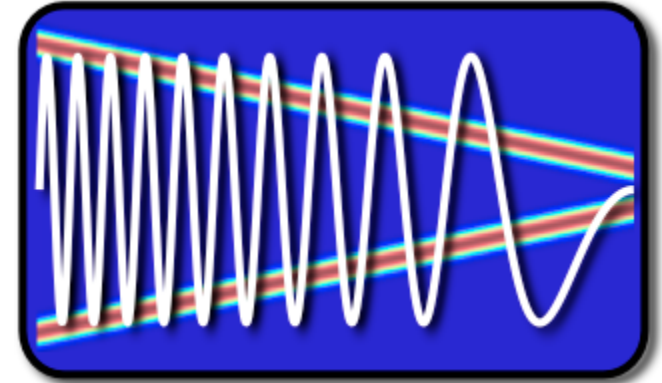


EE123



Digital Signal Processing

Lecture 16 Resampling

Topics

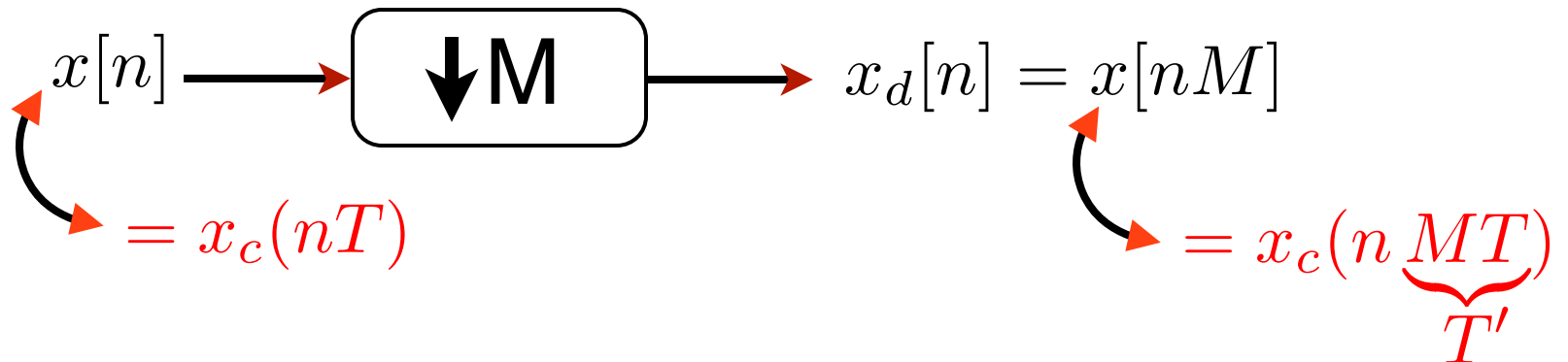
- <http://rtl-sdr.com>
- Did you sign up for the ham exam?
- Last time
 - D.T processing of C.T signals
 - C.T processing of D.T signals (ha?????)
- Today
 - Downsampling
 - Changing Sampling Rate via DSP
 - Upsampling
 - Rational resampling

DownSampling

- Much like C/D conversion
- Expect similar effects:
 - Aliasing
 - mitigate by antialiasing filter
- Finely sampled signal \Rightarrow almost continuous
 - Downsample in that case is like sampling!

Changing Sampling-rate via D.T Processing

Downsampling:



The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T} k}_{\Omega_s} \right) \right)$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

Changing Sampling-rate via D.T Processing

The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T}}_{\Omega_s} k \right) \right)$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

we would like to bypass X_c and go from $X(e^{j\omega}) \Rightarrow X_d(e^{j\omega})$

substitute counter to

$$k = rM + i$$

$$i=0,1,\dots,M-1$$

$$r=-\infty,\dots,\infty$$

two counters

e.g., r : hours, i : minutes

Changing Sampling-rate via D.T Processing

$$\begin{aligned} X_d(e^{j\omega}) &= \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right) \\ &= \frac{1}{M} \sum_{i=0}^{M-1} \underbrace{\frac{1}{T} \sum_{r=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} i - \frac{2\pi}{T} r \right) \right)} \end{aligned}$$

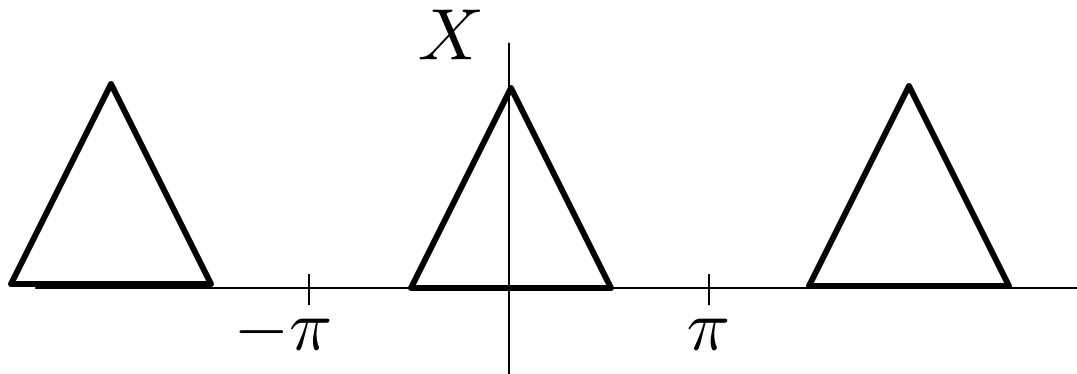
$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}} - \underbrace{\frac{2\pi}{T} k} \right) \right) \quad X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})$$

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})$$

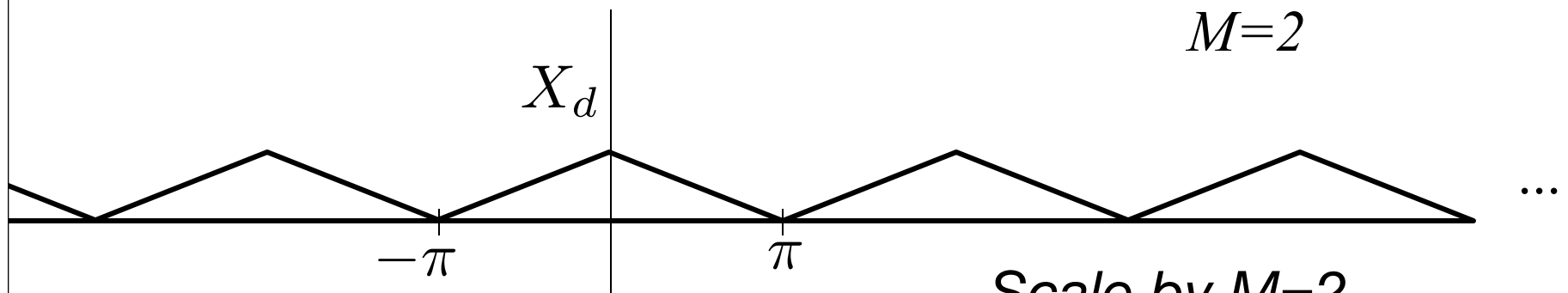
stretch by M replicate

Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j \left(\frac{\omega}{M} - \frac{2\pi}{M} i \right)} \right)$$



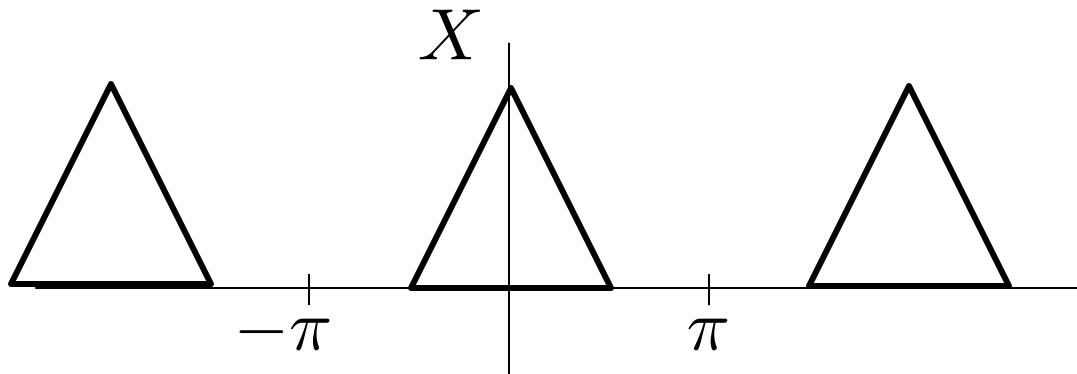
$M=2$



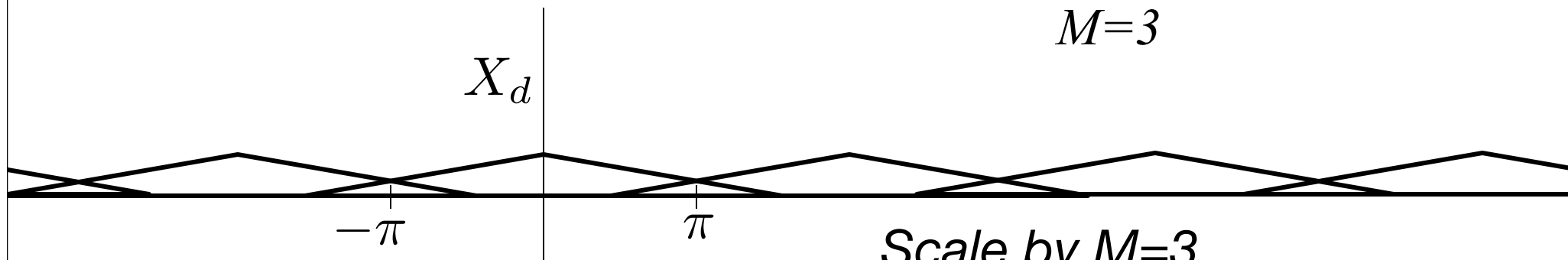
Scale by $M=2$
Shift by $(i=1)^*2\pi/(M=2)$

Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j \left(\frac{\omega}{M} - \frac{2\pi}{M} i \right)} \right)$$



$M=3$



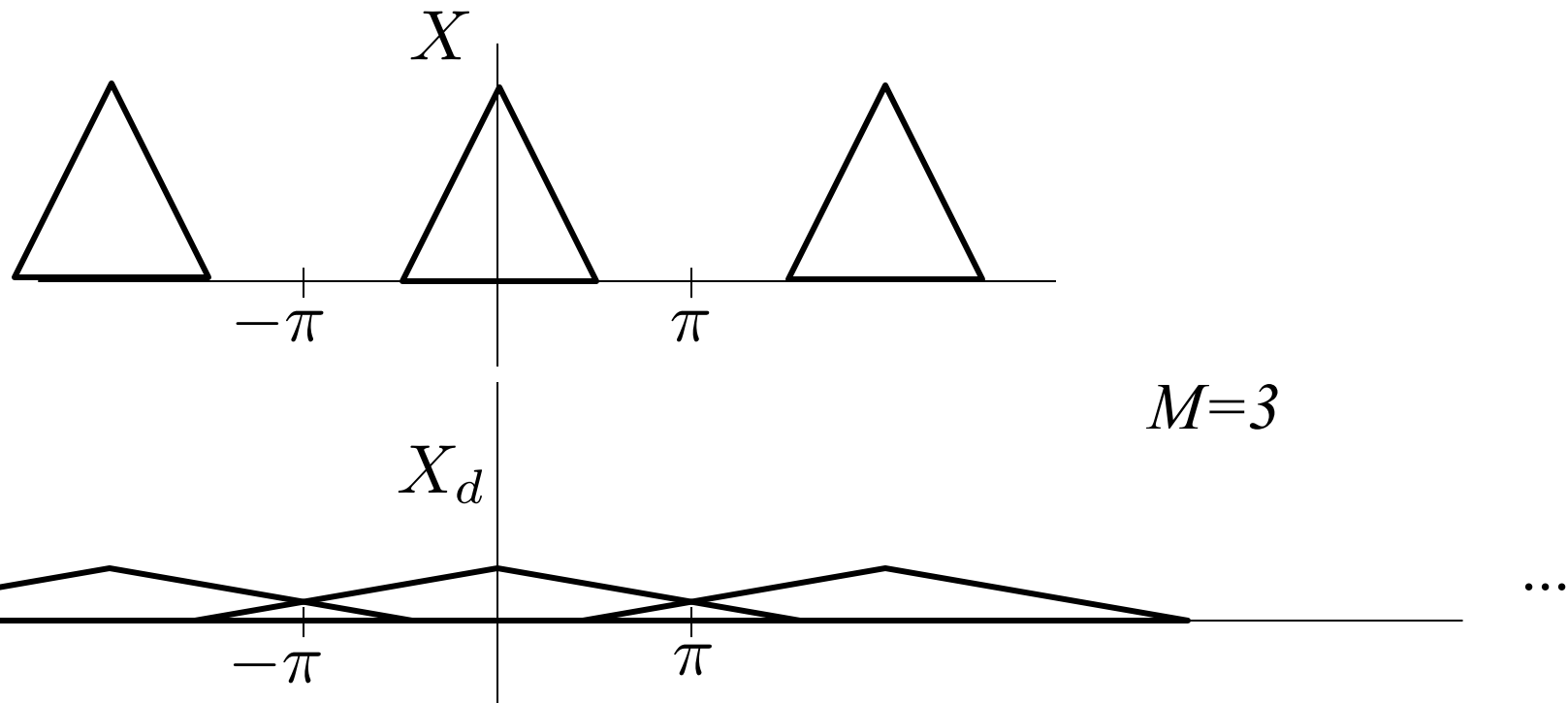
Scale by $M=3$

*Shift by $(i=1)*2\pi/(M=3)$*

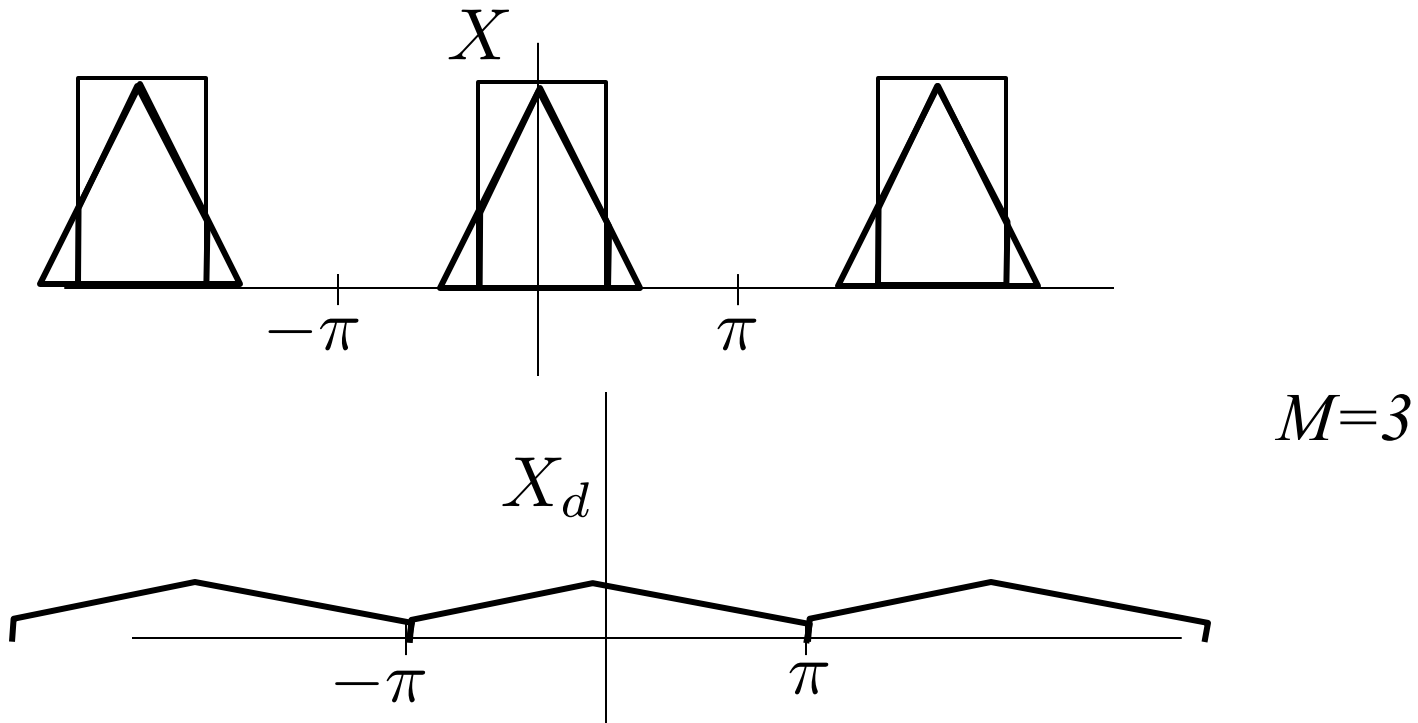
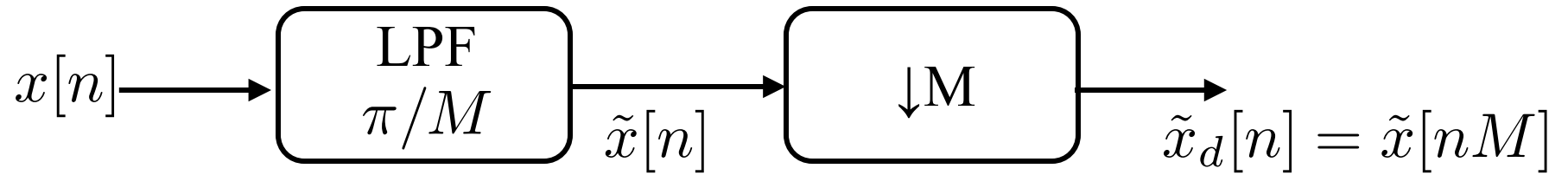
*Shift by $(i=2)*2\pi/(M=3)$*

Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j \left(\frac{\omega}{M} - \frac{2\pi}{M} i \right)} \right)$$



Anti-Aliasing



UpSampling

- Much like D/C converter
- Upsample by A LOT \Rightarrow almost continuous
- Intuition:
 - Recall our D/C model: $x[n] \Rightarrow x_s(t) \Rightarrow x_c(t)$
 - Approximate “ $x_s(t)$ ” by placing zeros between samples
 - Convolve with a sinc to obtain “ $x_c(t)$ ”

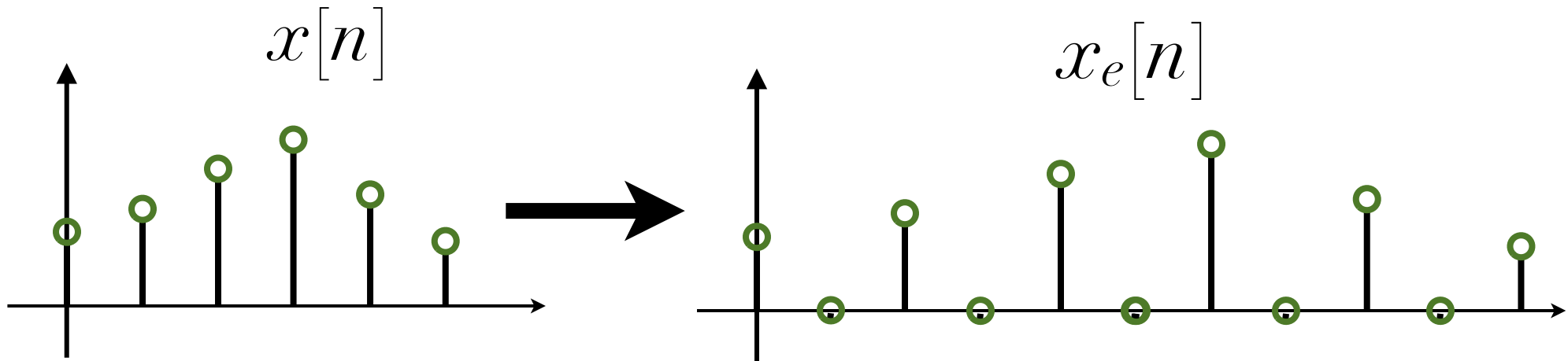
Up-sampling

$$x[n] = \overset{\text{lower x}}{X_c(nT)}$$

$$x_i[n] = X_c(nT') \quad \text{where } T' = \frac{T}{L} \quad L \text{ integer}$$

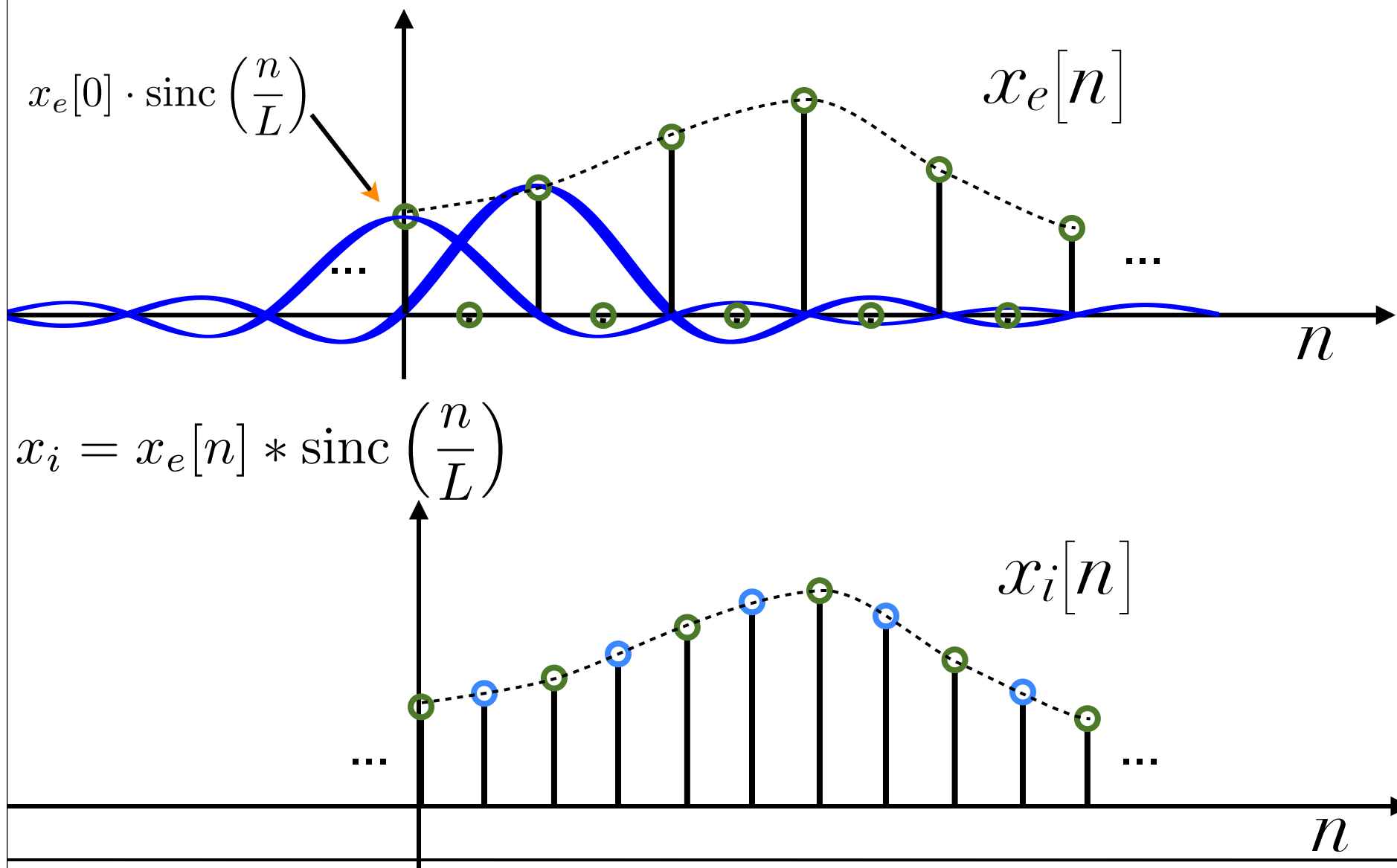
Obtain $x_i[n]$ from $x[n]$ in two steps:

(1) Generate:
$$x_e = \begin{cases} x[n/L] & n = 0, \pm L, \pm 2L, \dots \\ 0 & \text{otherwise} \end{cases}$$



Up-Sampling

(2) Obtain $x_i[n]$ from $x_e[n]$ by bandlimited interpolation:



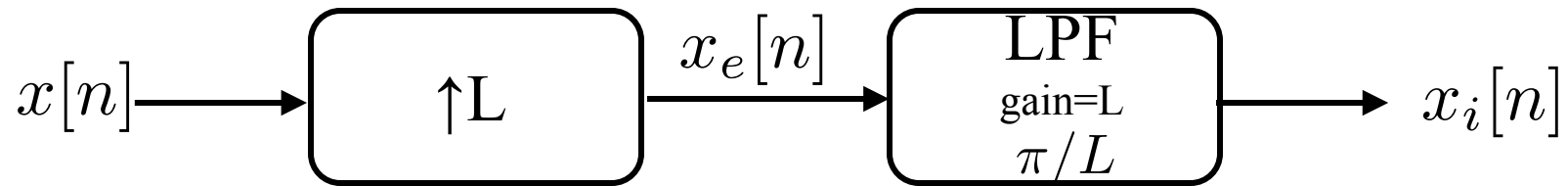
Up-Sampling

$$x_i[n] = x_e[n] * \text{sinc}(n/L)$$

$$x_e[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - kL]$$

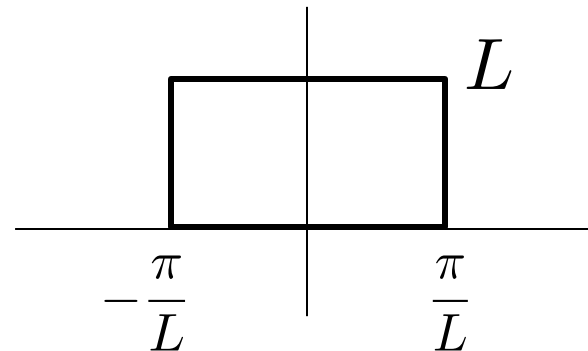
$$x_i[n] = \sum_{k=-\infty}^{\infty} x[k] \text{sinc}\left(\frac{n - kL}{L}\right)$$

Frequency Domain Interpretation

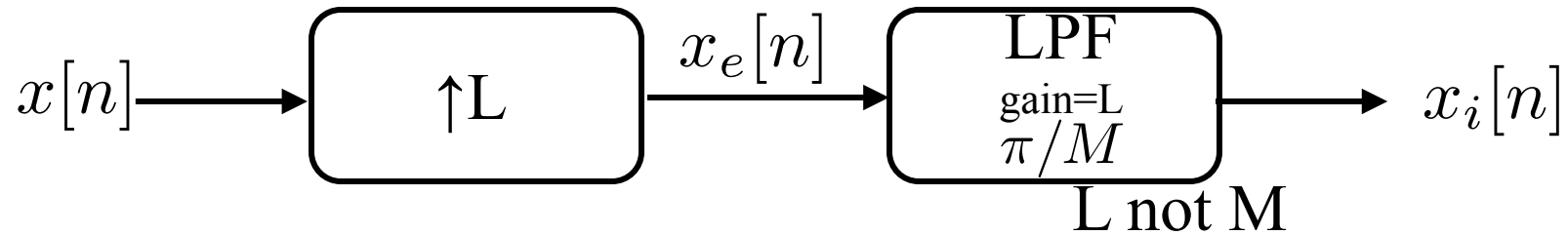


$$\text{sinc}(n/L)$$

DTFT \Rightarrow



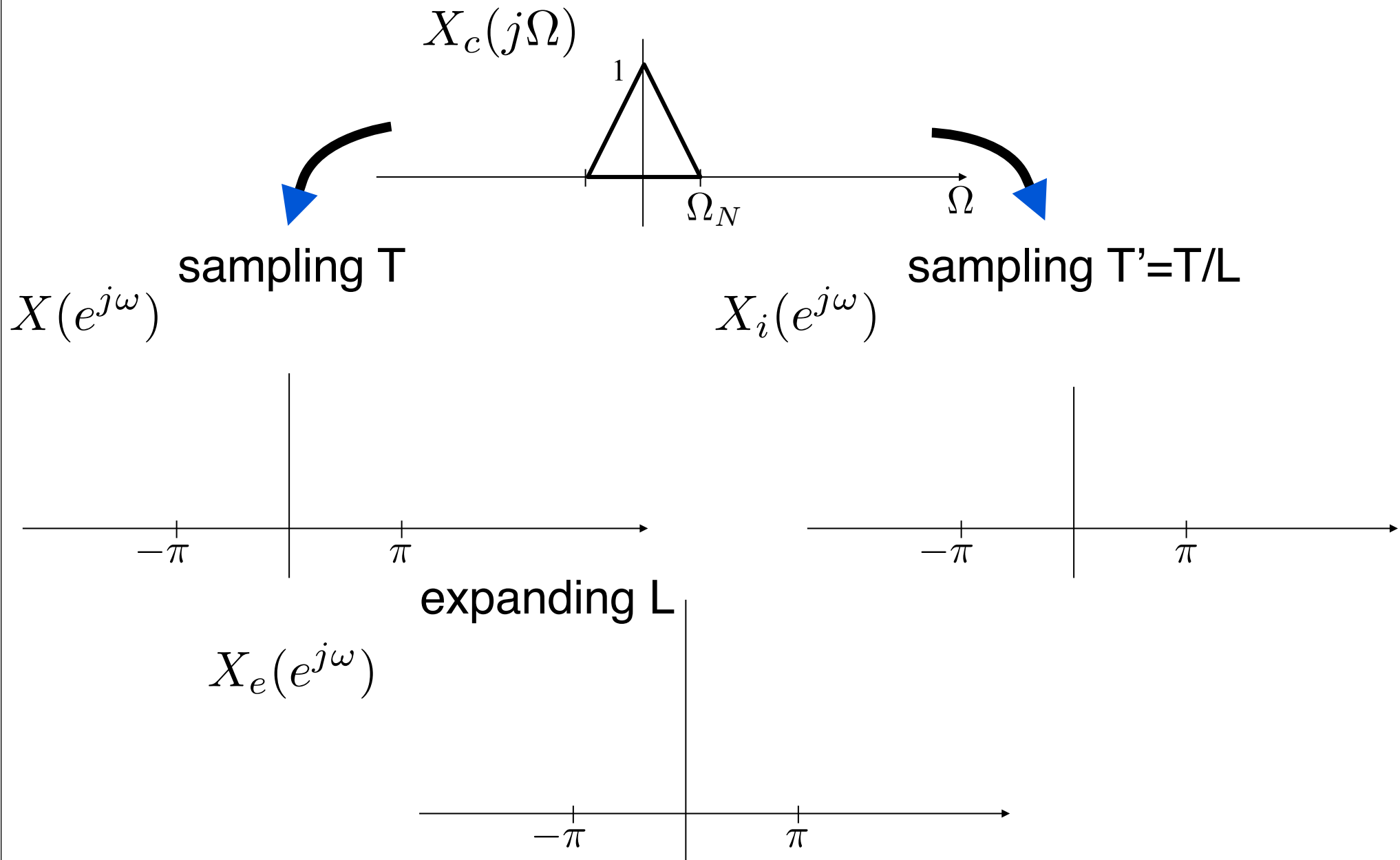
Frequency Domain Interpretation



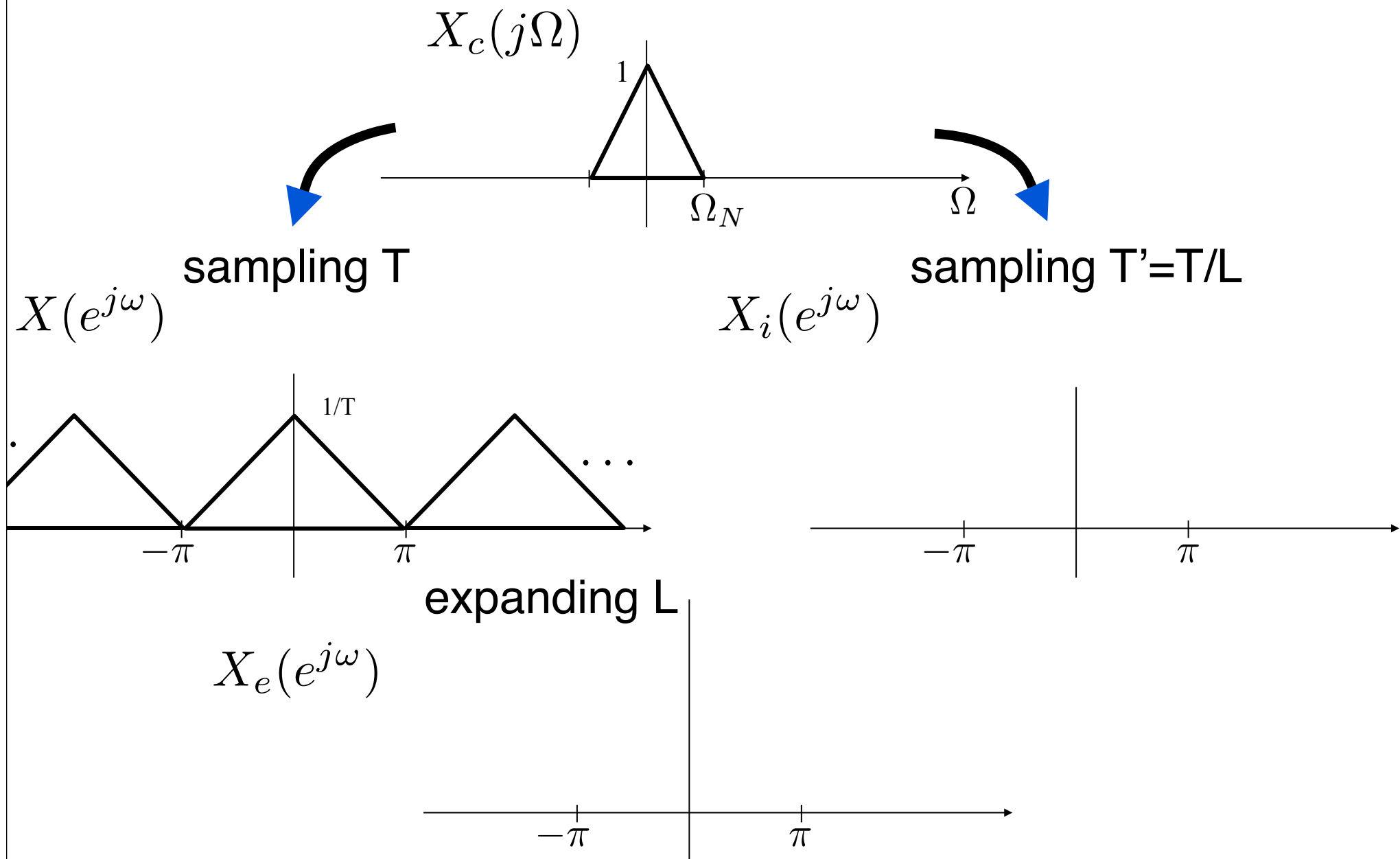
$$\begin{aligned} X_e(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\substack{\neq 0 \text{ only for } n=mL \\ (\text{integer } m)}} e^{-j\omega n} \\ &= \sum_{m=-\infty}^{\infty} \underbrace{x_e[mL]}_{=x[m]} e^{-j\omega mL} = X(e^{j\omega L}) \end{aligned}$$

Compress DTFT by a factor of L !

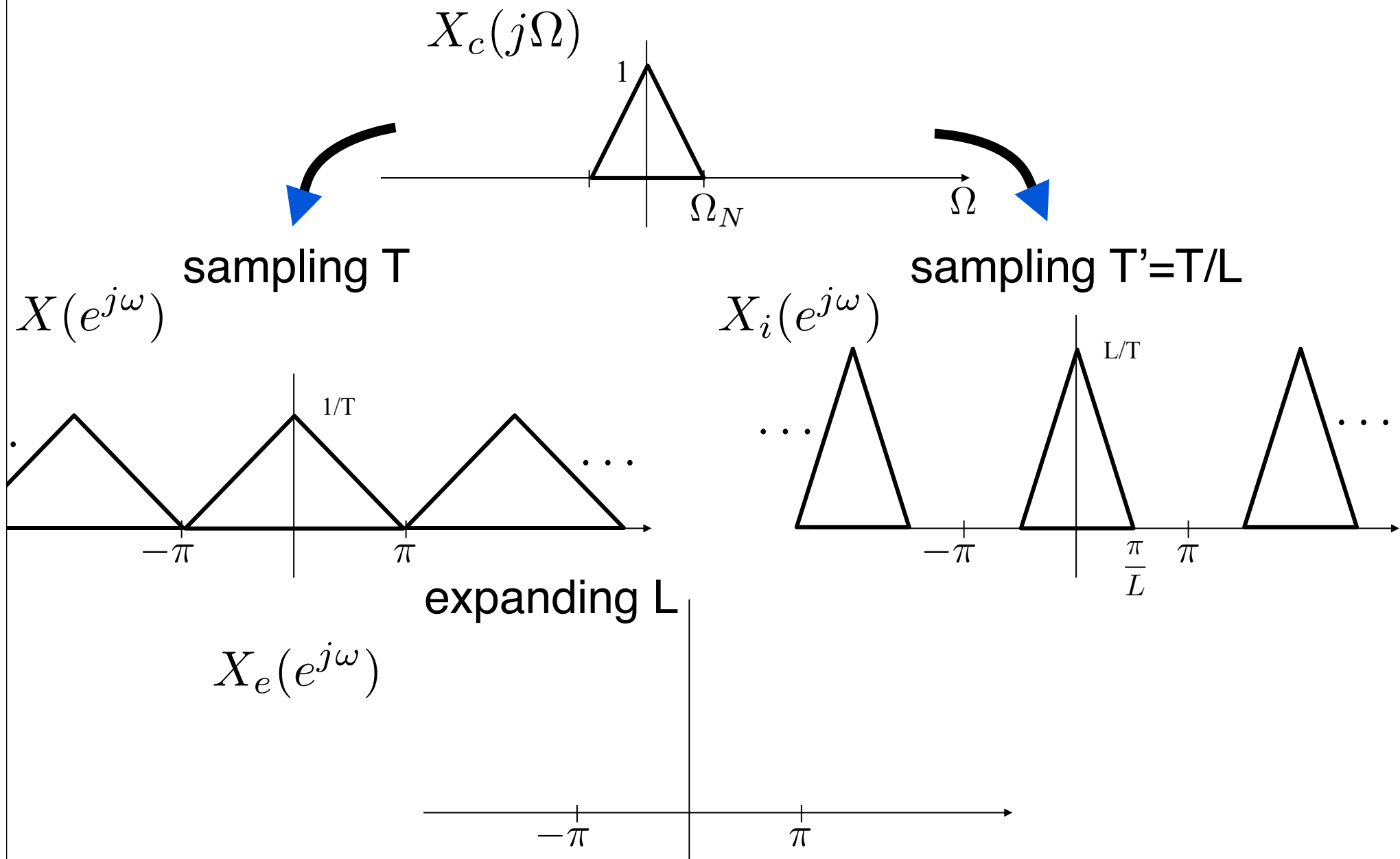
Example:



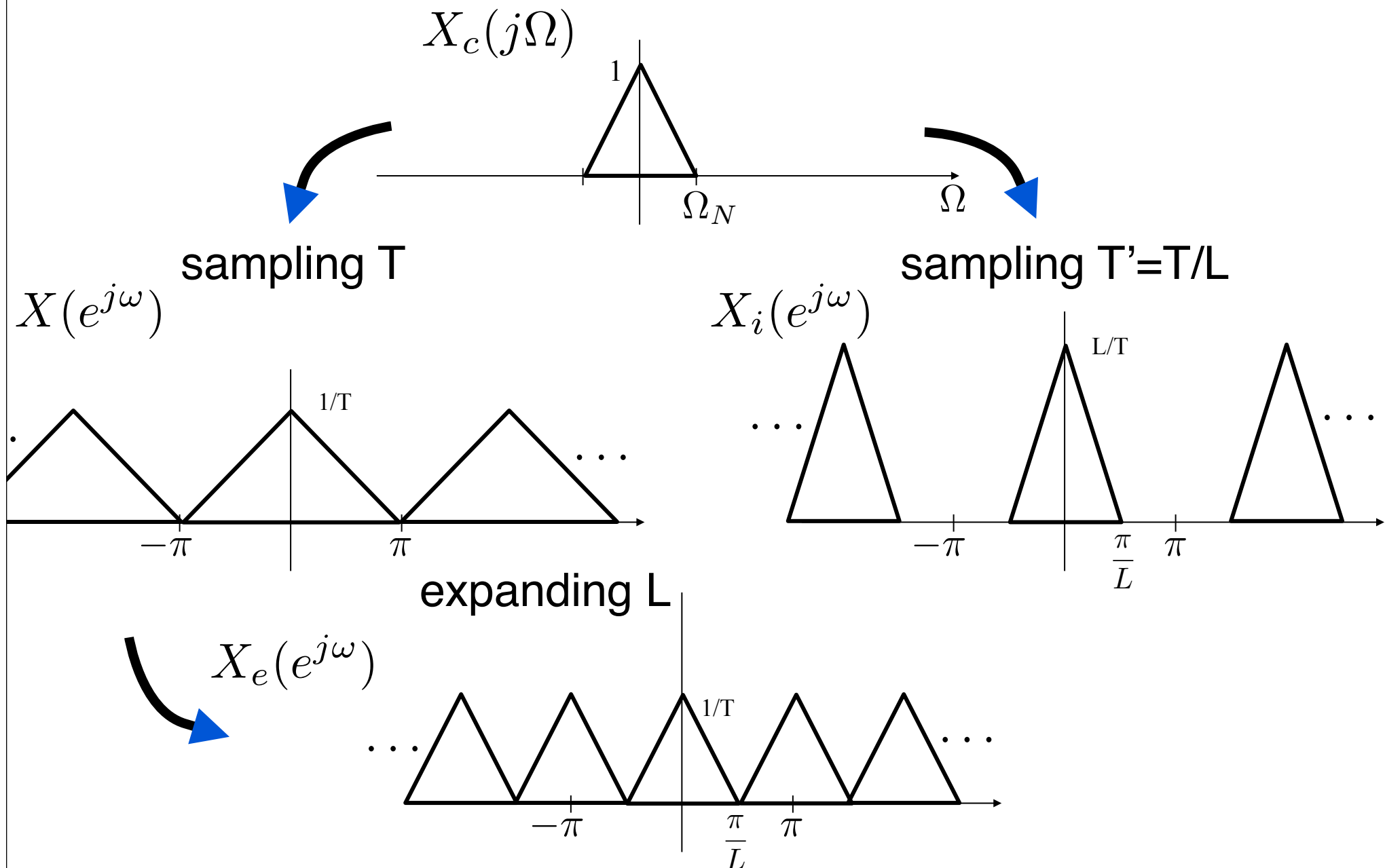
Example:



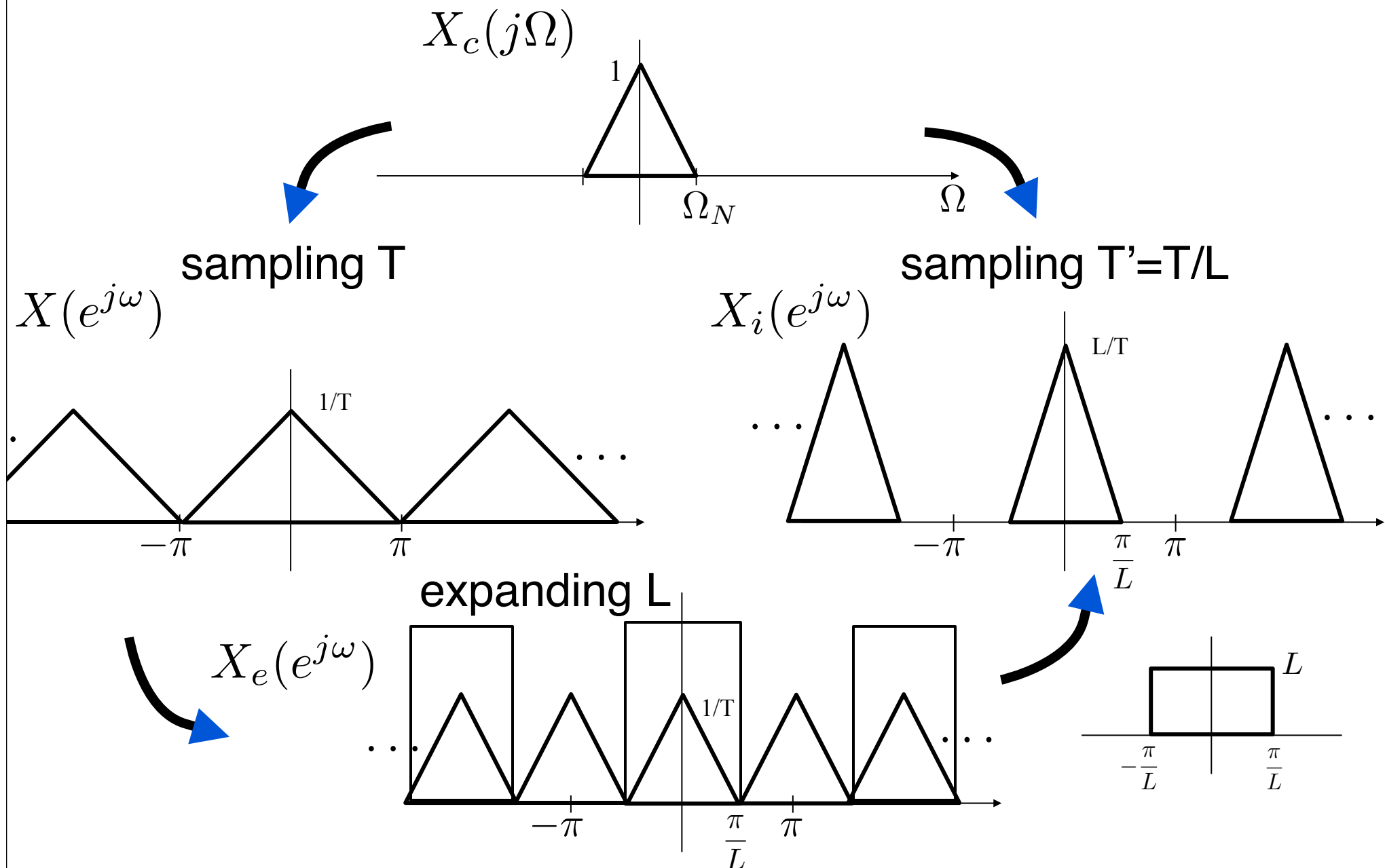
Example:



Example:

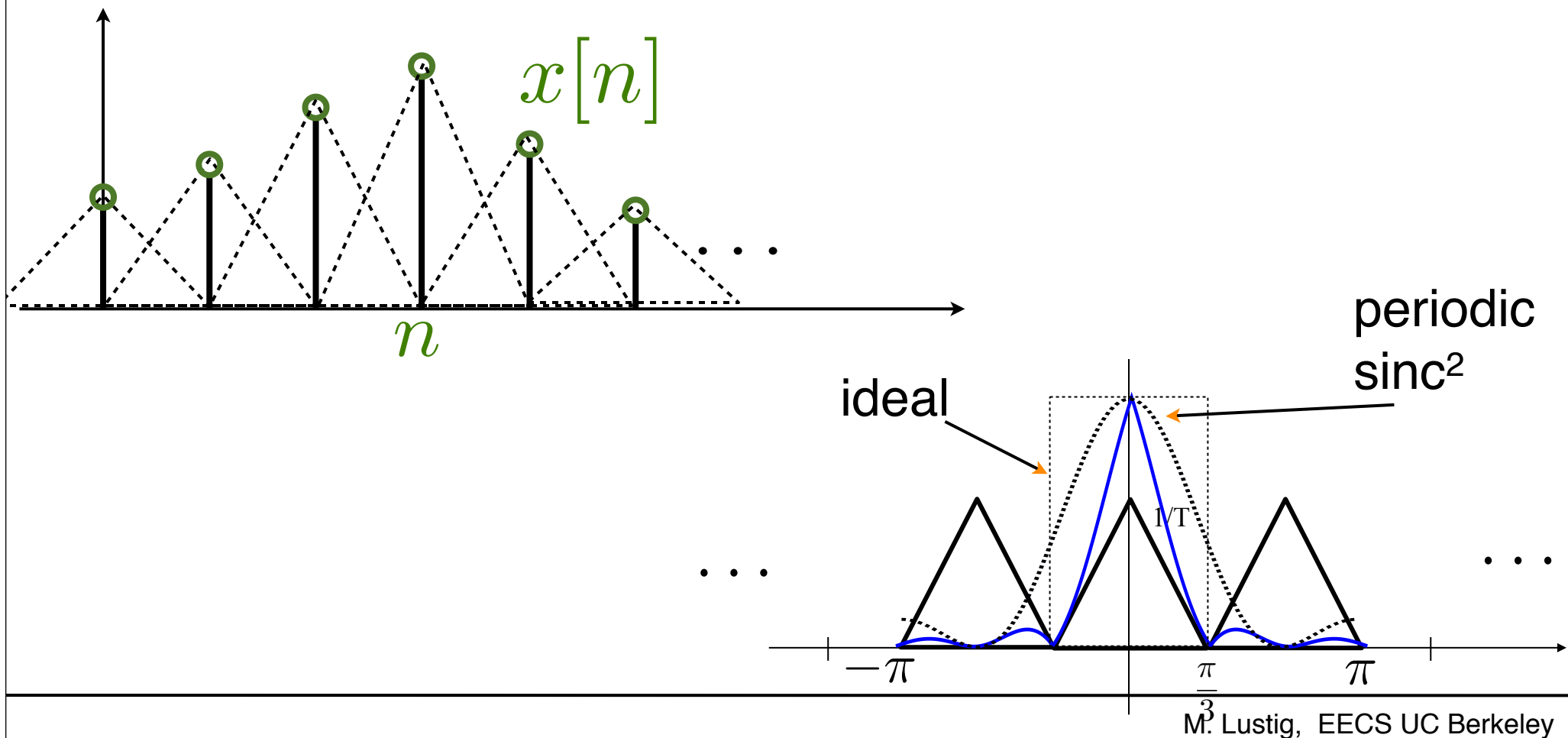


Example:



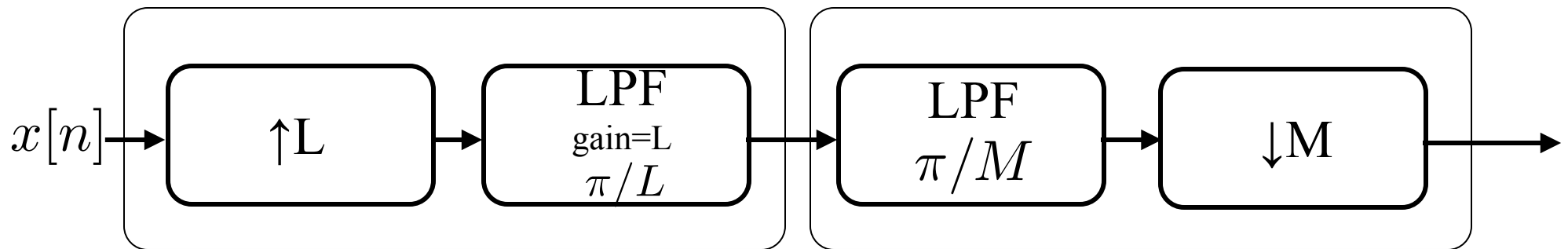
Practical Upsampling

- Can interpolate with simple, practical filters. What's happening?
- Example: $L=3$, linear interpolation - convolve with triangle

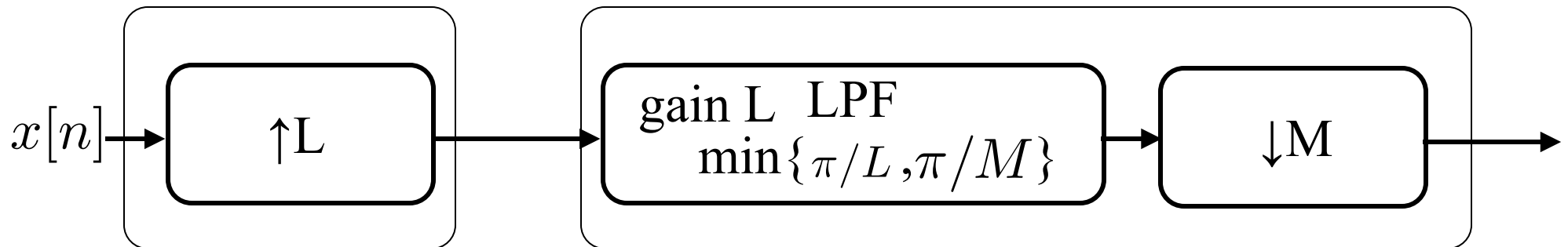


Resampling by non-integer

- $T' = TM/L$ (upsample L , downsample M)



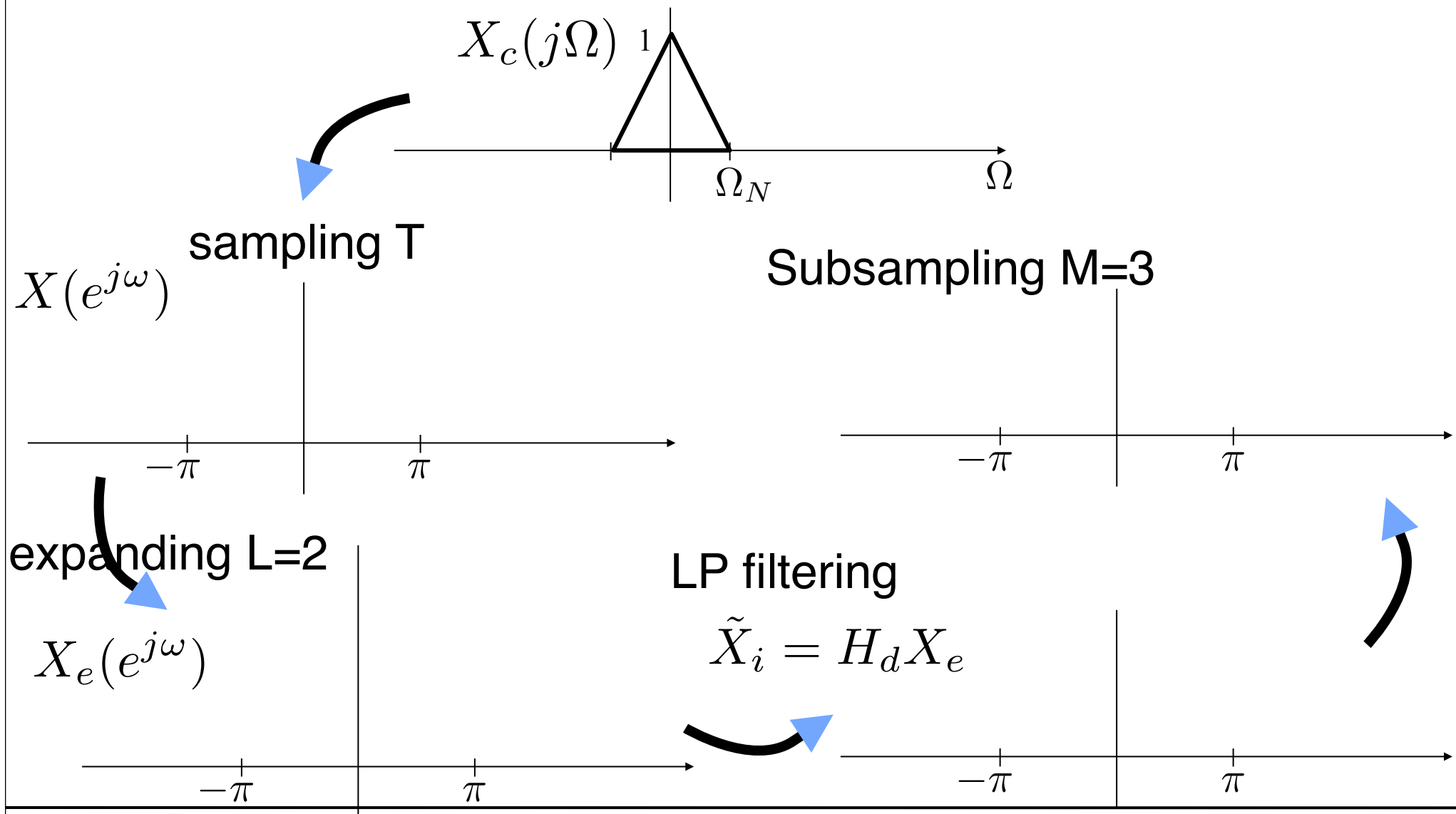
Or,



- What would happen if change order?

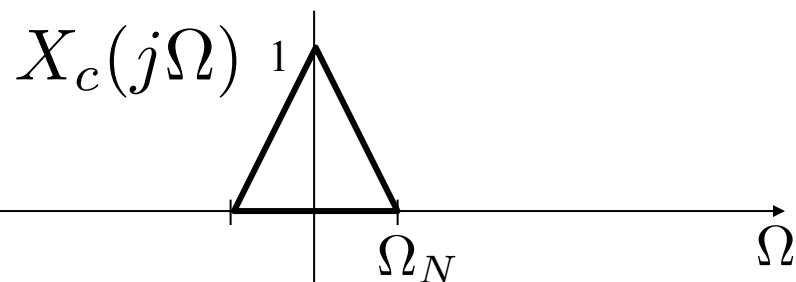
Example:

- $L = 2, M=3, T'=3/2T$ (fig 4.30)

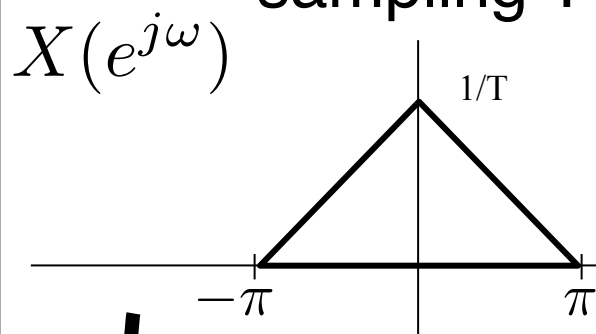


Example:

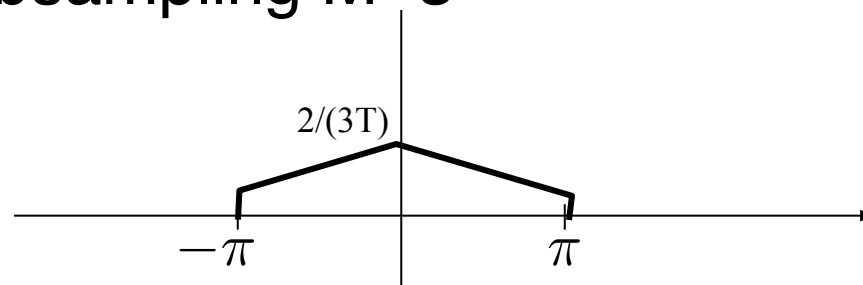
- $L = 2, M=3, T'=3/2T$ (fig 4.30)



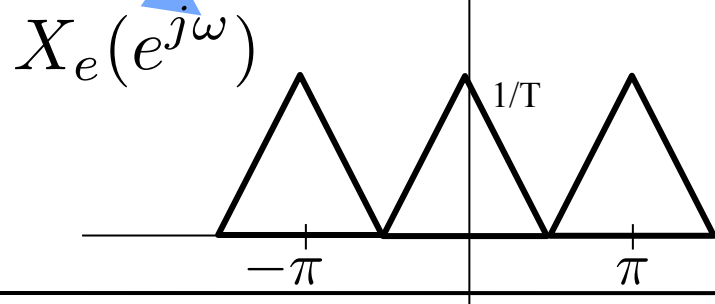
sampling T



Subsampling $M=3$

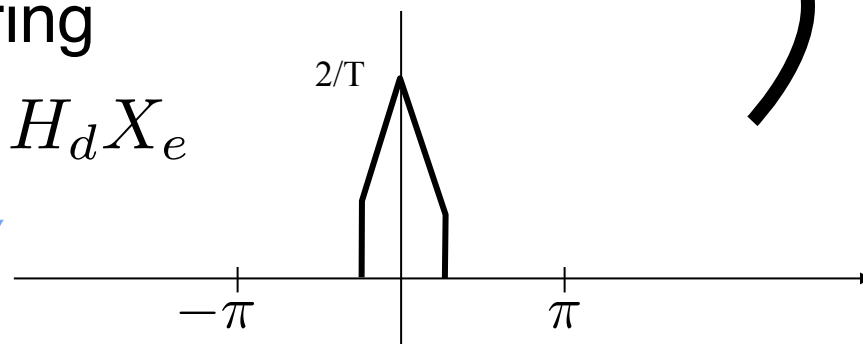


expanding $L=2$



LP filtering

$$\tilde{X}_i = H_d X_e$$



Multi-Rate Signal Processing

- What if we want to resample by 1.01T?
 - Expand by $L=100$
 - Filter $\pi/101$ (\$\$\$\$\$)
 - Downsample by $M=101$
- Fortunately there are ways around it!
 - Called multi-rate
 - Uses compressors, expanders and filtering