

Digital Filter Design

(Subject Code - EET 3134)

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Digital Filters



Digital Filtering System

- Types of Filters
 - 1. Time-domain
 - 2. Frequency-domain
- Simplest kind of filters are non-recursive filters defined as

$$y(n) = \sum_{k=0}^{M-1} h(k)x(n - k) \quad (1)$$

where, $h(k)$ are the constants of the filter, the $x(n - k)$ are the input data, and the $y(n)$ are the outputs.

Moving Averaging Filter

$$y(n) = \sum_{k=0}^{M-1} x(k)h(n-k) \quad (2)$$

- Implement a de-noising filter for a noisy sinusoidal signal using moving averaging Filter for different size of rectangular gates.



Introduction to Digital Filters

- In general, digital filters are used for
 1. Separation
 2. Restoration
- Every linear filter has
 1. Impulse response
 2. Unit step response
 3. Frequency response
- All these representation are important because they describe that how the filter will react under different circumstances.



Representation of Signals

- Information can be represented in
 - Time-domain
 - Frequency-domain
- Types of filters
 - Finite impulse response (FIR) filter

$$h(n) \neq 0 \quad N_1 \leq n \leq N_2 \quad (3)$$

$$= 0 \quad \text{everywhere else} \quad (4)$$

- Infinite impulse response (IIR) filter

$$h(n) \neq 0 \quad \infty \leq n \leq \infty \quad (5)$$

Advantage of FIR over IIR filters

- We can easily design the FIR filter to meet the required magnitude response in such a way that it achieves a constant group delay, i.e., they can have exactly linear phase.
- Always stable, even when quantized
- Design methods are generally linear
- They can be realized efficiently in hardware.
- The filter startup transients have finite duration.

Disadvantages:

- Often require higher filter order than IIR filters to achieve a good level of performance.
- The delay of these filters is often much greater than IIR filters having equal performance.

z Transform

The z transform of a sequence $x(n)$ is simply defined as

$$\mathcal{Z}[x(n)] = x(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (6)$$

and the inverse z transform defined as

$$\mathcal{Z}[x(z)] = x(n) = \frac{1}{2\pi j} \int_C x(z)z^{n-1} dz \quad (7)$$

This is the double-sided or bilateral z transform of sequence $x(n)$ defined for $-\infty < n < \infty$.



Group Delay

- Group delay is defined as

$$\tau = -(d\theta/d\omega) \quad (8)$$

where ω is the phase response of the filter.

- The phase response of a filter with a constant group delay is therefore a linear function of frequency.
- It transmits all frequencies with the same amount of delay, which means that there will not be any phase distortion and the input signal will be delayed by a constant when it is transmitted to the output.
- A filter with a constant group delay is highly desirable in the transmission of digital signals.

Notation



$$H(z^{-1}) = \sum_{n=0}^N h(n)z^{-n} \quad (9)$$

The order of the FIR filter or degree of the polynomial = N

The length of the filter or the number of coefficient = $N + 1$

- Example: $H(z^{-1}) = 0.3z^{-4} + 0.1z^{-5} + 0.5z^{-6}$, order and length?
- Notation often used in MATLAB

$$\begin{aligned} H(z^{-1}) &= h(1) + h(2)z^{-1} + h(3)z^{-2} + \cdots + h(N + 1)z^{-N} \\ &= \sum_{n=0}^N h(n + 1)z^{-n} \end{aligned}$$

Linear time-invariant causal system

Difference Equation

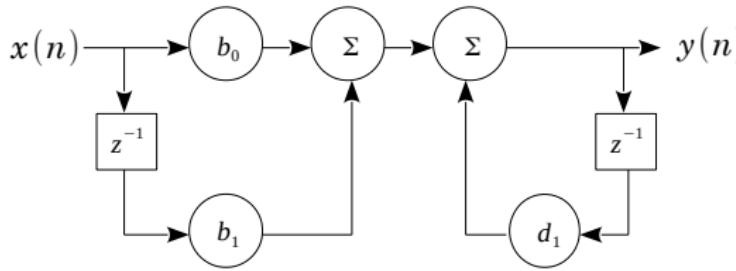
$$y(n) = - \sum_{k=1}^N a(k)y(n-k) + \sum_{k=0}^M b(k)x(n-k) \quad (10)$$

Transfer function

$$H(z^{-1}) = \frac{b(0) + b(1)z^{-1} + b(2)z^{-2} + \cdots + b(M)z^{-M}}{1 + a(1)z^{-1} + a(2)z^{-2} + \cdots + a(N)z^{-N}} \quad (11)$$

Example

$$y(n) = b_0x(n) + b_1x(n-1) + d_1y(n-1) \quad (12)$$



Recursive and non-recursive filters

$\text{FIR} \rightarrow \text{non-recursive}$ } not necessarily,
 $\text{IIR} \rightarrow \text{recusive}$ } it depends on representation of output of the system (13)

$$y(n) = x(n) + x(n-1) \rightarrow \text{non-recursive}$$

$$\Rightarrow y(n-1) = x(n-1) + x(n-2)$$

$$\text{So, } y(n) - y(n-1) = x(n) - x(n-2)$$

$$y(n) = y(n-1) + x(n) - x(n-2) \rightarrow \text{recursive}$$

Ideal filters

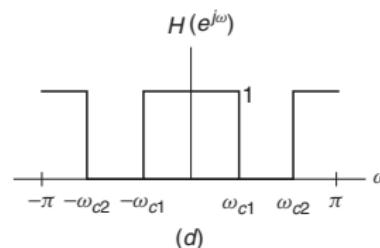
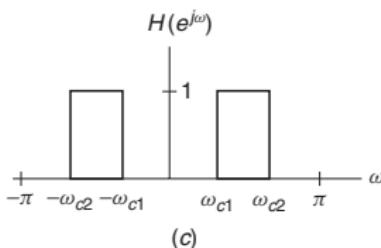
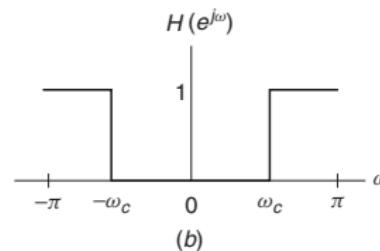
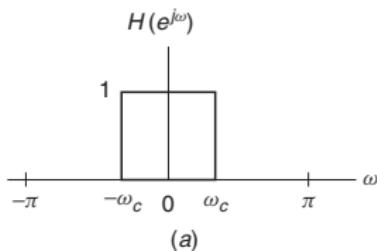


Figure: Basic filters (a) Lowpass filter, (b) Highpass filter, (c) Bandpass filter, and (d) Bandstop filter



Impulse responses of ideal filters

$$h(n) = \begin{cases} \frac{\omega_c}{\pi}; & n = 0 \\ \frac{\sin(\omega_c n)}{\pi n}; & |n| > 0 \end{cases}$$

(a) Lowpass filter

$$h(n) = \begin{cases} \frac{\omega_{c2} - \omega_{c1}}{\pi}; & n = 0 \\ \frac{1}{\pi n} [\sin(\omega_{c2} n) - \sin(\omega_{c1} n)]; & |n| > 0 \end{cases}$$

(b) Bandpass filter

$$h(n) = \begin{cases} 1 - \frac{\omega_c}{\pi}; & n = 0 \\ -\frac{\sin(\omega_c n)}{\pi n}; & |n| > 0 \end{cases}$$

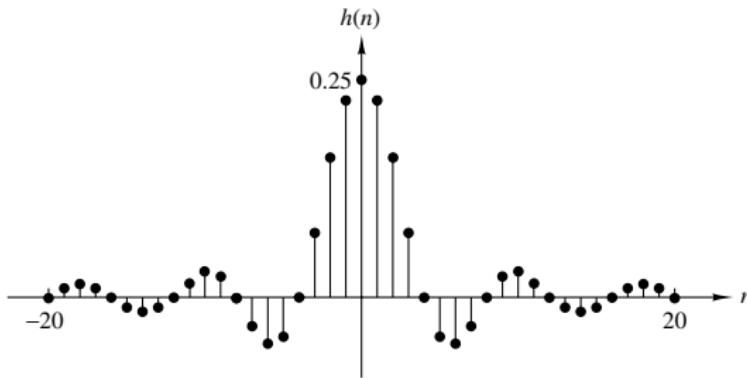
(c) Highpass filter

$$h(n) = \begin{cases} 1 - \frac{(\omega_{c2} - \omega_{c1})}{\pi}; & n = 0 \\ \frac{1}{\pi n} [\sin(\omega_{c1} n) - \sin(\omega_{c2} n)]; & |n| > 0 \end{cases}$$

(d) Bandstop filter

Ideal lowpass filter realization

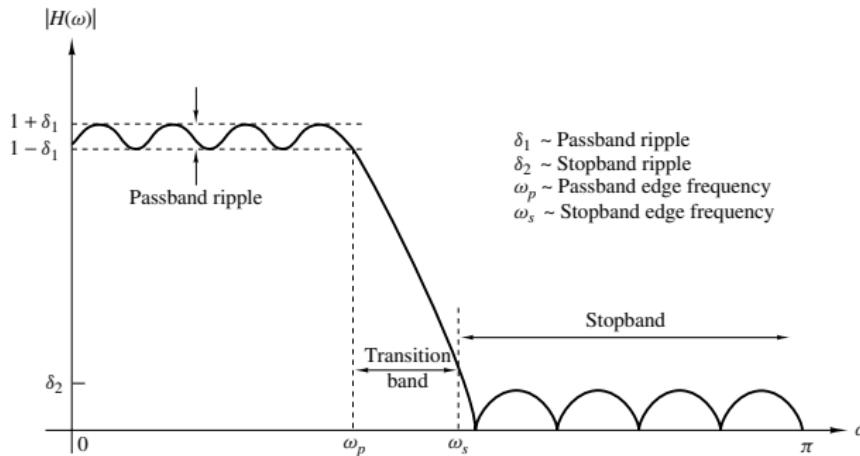
$$H(\omega) = \begin{cases} 1, & |\omega| \leq \omega_c \\ 0, & \omega_c < \omega \leq \pi \end{cases} \quad h(n) = \begin{cases} \frac{\omega_c}{\pi}, & n = 0 \\ \frac{\omega_c}{\pi} \frac{\sin \omega_c n}{\omega_c n}, & n \neq 0 \end{cases}$$



- Ideal filters are not realizable due to its non-causal property of their impulse response.



What we can realize?



$$\Delta\omega = \omega_s - \omega_p$$

$$\omega_c = \omega_p + 0.5\Delta\omega$$

$$A_p = 20 \log_{10}(1 + \delta_p)$$

$$A_s = -20 \log_{10}(\delta_s)$$

Figure: Magnitude characteristics of physically realizable filters



Linear Phase FIR Filters

Linear Phase FIR Filters

Consider the special types of FIR filters in which the coefficients $h(n)$ of the transfer function

$$H(z^{-1}) = \sum_{n=0}^N h(n)z^{-n} \quad (14)$$

Type I. The coefficients are symmetric, and N is even.

Type II. The coefficients are symmetric, and N is odd.

Type III. The coefficients are antisymmetric, and N is even.

Type IV. The coefficients are antisymmetric, and N is odd.



Type I Linear phase FIR Filter

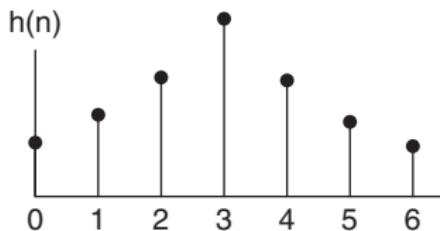
The coefficients are symmetric [i.e., $h(n) = h(N - n)$], and the order N is even.

Type I Linear phase FIR Filter

The coefficients are symmetric [i.e., $h(n) = h(N - n)$], and the order N is even.

Examples: $N = 6$

$$H(z^{-1}) = h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + h(4)z^{-4} + h(5)z^{-5} + h(6)z^{-6}$$



$$h(0) = h(6)$$

$$h(1) = h(5)$$

$$h(2) = h(4)$$



Type I Linear phase FIR Filter

$$\begin{aligned}H(z^{-1}) &= \{h(0)[1 + z^{-6}] + h(1)[z^{-1} + z^{-5}] + h(2)[z^{-2} + z^{-4}] + h(3)z^{-3}\} \\&= z^{-3}\{h(0)[z^3 + z^{-3}] + h(1)[z^2 + z^{-2}] + h(2)[z + z^{-1}] + h(3)\}\end{aligned}$$

Type I Linear phase FIR Filter

$$\begin{aligned}H(z^{-1}) &= \{h(0)[1 + z^{-6}] + h(1)[z^{-1} + z^{-5}] + h(2)[z^{-2} + z^{-4}] + h(3)z^{-3}\} \\&= z^{-3}\{h(0)[z^3 + z^{-3}] + h(1)[z^2 + z^{-2}] + h(2)[z + z^{-1}] + h(3)\}\end{aligned}$$

put $z = e^{jw}$ to find frequency response (DTFT), we get

Type I Linear phase FIR Filter

$$\begin{aligned}H(z^{-1}) &= \{h(0)[1 + z^{-6}] + h(1)[z^{-1} + z^{-5}] + h(2)[z^{-2} + z^{-4}] + h(3)z^{-3}\} \\&= z^{-3}\{h(0)[z^3 + z^{-3}] + h(1)[z^2 + z^{-2}] + h(2)[z + z^{-1}] + h(3)\}\end{aligned}$$

put $z = e^{j\omega}$ to find frequency response (DTFT), we get

$$\begin{aligned}H(e^{-j\omega}) &= e^{-3j\omega}\{h(0)[e^{3j\omega} + e^{-3j\omega}] + h(1)[e^{2j\omega} + e^{-2j\omega}] \\&\quad + h(2)[e^{j\omega} + e^{-j\omega}] + h(3)\}\end{aligned}$$



Type I Linear phase FIR Filter

$$\begin{aligned}
 H(z^{-1}) &= \{h(0)[1 + z^{-6}] + h(1)[z^{-1} + z^{-5}] + h(2)[z^{-2} + z^{-4}] + h(3)z^{-3}\} \\
 &= z^{-3}\{h(0)[z^3 + z^{-3}] + h(1)[z^2 + z^{-2}] + h(2)[z + z^{-1}] + h(3)\}
 \end{aligned}$$

put $z = e^{j\omega}$ to find frequency response (DTFT), we get

$$\begin{aligned}
 H(e^{-j\omega}) &= e^{-3j\omega}\{h(0)[e^{3j\omega} + e^{-3j\omega}] + h(1)[e^{2j\omega} + e^{-2j\omega}] \\
 &\quad + h(2)[e^{j\omega} + e^{-j\omega}] + h(3)\} \\
 &= e^{-3j\omega}\{2h(0)\cos(3\omega) + 2h(1)\cos(2\omega) + 2h(2)\cos(\omega) + h(3)\} \\
 &= e^{j\theta(\omega)}\{H_R(\omega)\}
 \end{aligned}$$



Type I Linear phase FIR Filter

$$\begin{aligned} H(z^{-1}) &= \{h(0)[1 + z^{-6}] + h(1)[z^{-1} + z^{-5}] + h(2)[z^{-2} + z^{-4}] + h(3)z^{-3}\} \\ &= z^{-3}\{h(0)[z^3 + z^{-3}] + h(1)[z^2 + z^{-2}] + h(2)[z + z^{-1}] + h(3)\} \end{aligned}$$

put $z = e^{j\omega}$ to find frequency response (DTFT), we get

$$\begin{aligned} H(e^{-j\omega}) &= e^{-3j\omega}\{h(0)[e^{3j\omega} + e^{-3j\omega}] + h(1)[e^{2j\omega} + e^{-2j\omega}] \\ &\quad + h(2)[e^{j\omega} + e^{-j\omega}] + h(3)\} \\ &= e^{-3j\omega}\{2h(0)\cos(3\omega) + 2h(1)\cos(2\omega) + 2h(2)\cos(\omega) + h(3)\} \\ &= e^{j\theta(\omega)}\{H_R(\omega)\} \end{aligned}$$

phase angle, $\theta(\omega) = -3\omega \Rightarrow$ linear function of ω

$$\text{group delay} = -\frac{d\theta(\omega)}{d\omega} = 3 \text{ samples} = 3T \text{ seconds} \quad (15)$$

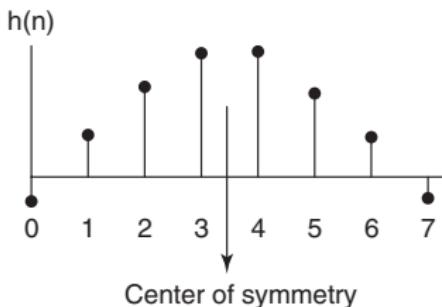
Total group delay is a constant $= \frac{N}{2}$ in the general case for a type I FIR filters.

Type II Linear phase FIR Filter

The coefficients are symmetric [i.e., $h(n) = h(N - n)$], and the order N is odd.

Examples: $N = 7$

$$H(z^{-1}) = h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + h(4)z^{-4} \\ + h(5)z^{-5} + h(6)z^{-6} + h(7)z^{-7}$$



$$h(0) = h(7) \\ h(1) = h(6) \\ h(2) = h(5) \\ h(3) = h(4)$$



Type II Linear phase FIR Filter

$$\begin{aligned}
 H(e^{-j\omega}) &= e^{-3.5j\omega} \{2h(0) \cos(3.5\omega) + 2h(1) \cos(2.5\omega) + 2h(2) \cos(1.5\omega) \\
 &\quad + 2h(3) \cos(0.5\omega)\} \\
 &= e^{j\theta(\omega)} \{H_R(\omega)\}
 \end{aligned}$$

phase angle, $\theta(\omega) = -3.5\omega \Rightarrow$ linear function of ω

$$\text{group delay} = -\frac{d\theta(\omega)}{d\omega} = 3.5 \text{ samples} = 3.5T \text{ seconds} \quad (16)$$

- In general, a linear phase $\theta(\omega) = -[\frac{N}{2}]\omega$
- Constant group delay = $\frac{N}{2}$

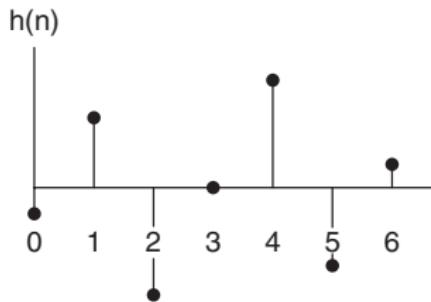


Type III Linear phase FIR Filter

The coefficients are antisymmetric [i.e., $h(n) = -h(N - n)$], and the order N is even. Here, $h(\frac{N}{2}) = 0$ for antisymmetric

Examples: $N = 6$

$$H(z^{-1}) = h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + h(4)z^{-4} + h(5)z^{-5} + h(6)z^{-6}$$



$$h(0) = -h(6)$$

$$h(1) = -h(5)$$

$$h(2) = -h(4)$$

$$h(3) = 0$$

Type III Linear phase FIR Filter

$$\begin{aligned} H(z^{-1}) &= \{h(0)[1 - z^{-6}] + h(1)[z^{-1} - z^{-5}] + h(2)[z^{-2} - z^{-4}]\} \\ &= z^{-3}\{h(0)[z^3 - z^{-3}] + h(1)[z^2 - z^{-2}] + h(2)[z - z^{-1}]\} \end{aligned}$$

put $z = e^{j\omega}$ to find frequency response (DTFT), we get

$$\begin{aligned} H(e^{-j\omega}) &= e^{-3j\omega}\{h(0)[e^{3j\omega} - e^{-3j\omega}] + h(1)[e^{2j\omega} - e^{-2j\omega}] + h(2)[e^{j\omega} - e^{-j\omega}]\} \\ &= e^{-3j\omega}e^{j\frac{\pi}{2}}\{2h(0)\sin(3\omega) + 2h(1)\sin(2\omega) + 2h(2)\sin(\omega)\} \\ &= e^{j(-3\omega+\frac{\pi}{2})}\{2h(0)\sin(3\omega) + 2h(1)\sin(2\omega) + 2h(2)\sin(\omega)\} \\ &= e^{j\theta(\omega)}\{H_R(\omega)\} \end{aligned}$$

phase angle, $\theta(\omega) = -3\omega + \frac{\pi}{2} \Rightarrow$ linear function of ω

$$\text{group delay} = -\frac{d\theta(\omega)}{d\omega} = 3 \text{ samples} = 3T \text{ seconds} \quad (17)$$



Type III Linear phase FIR Filter

$$\begin{aligned} H(z^{-1}) &= \{h(0)[1 - z^{-6}] + h(1)[z^{-1} - z^{-5}] + h(2)[z^{-2} - z^{-4}]\} \\ &= z^{-3}\{h(0)[z^3 - z^{-3}] + h(1)[z^2 - z^{-2}] + h(2)[z - z^{-1}]\} \end{aligned}$$

put $z = e^{j\omega}$ to find frequency response (DTFT), we get

$$\begin{aligned} H(e^{-j\omega}) &= e^{-3j\omega}\{h(0)[e^{3j\omega} - e^{-3j\omega}] + h(1)[e^{2j\omega} - e^{-2j\omega}] + h(2)[e^{j\omega} - e^{-j\omega}]\} \\ &= e^{-3j\omega}e^{j\frac{\pi}{2}}\{2h(0)\sin(3\omega) + 2h(1)\sin(2\omega) + 2h(2)\sin(\omega)\} \\ &= e^{j(-3\omega+\frac{\pi}{2})}\{2h(0)\sin(3\omega) + 2h(1)\sin(2\omega) + 2h(2)\sin(\omega)\} \\ &= e^{j\theta(\omega)}\{H_R(\omega)\} \end{aligned}$$

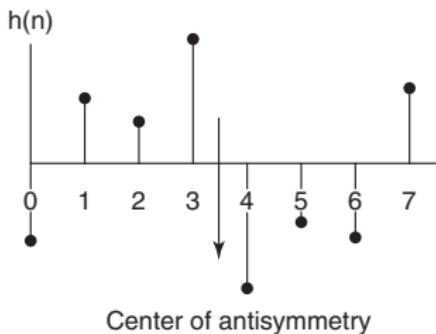
- In general, a linear phase $\theta(\omega) = -\left(\frac{N\omega-\pi}{2}\right)$
- Constant group delay $\tau = \frac{N}{2}$

Type IV Linear phase FIR Filter

The coefficients are symmetric [i.e., $h(n) = -h(N - n)$], and the order N is odd.

Examples: $N = 7$

$$H(z^{-1}) = h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + h(4)z^{-4} \\ + h(5)z^{-5} + h(6)z^{-6} + h(7)z^{-7}$$



$$h(0) = -h(7) \\ h(1) = -h(6) \\ h(2) = -h(5) \\ h(3) = -h(4)$$



Type IV Linear phase FIR Filter

$$\begin{aligned}
 H(e^{-j\omega}) &= e^{-3.5j\omega} \{2jh(0) \sin(3.5\omega) + 2jh(1) \sin(2.5\omega) + 2jh(2) \sin(1.5\omega) \\
 &\quad + 2jh(3) \sin(0.5\omega)\} \\
 &= e^{-3.5j\omega} e^{j\frac{\pi}{2}} \{2h(0) \sin(3.5\omega) + 2h(1) \sin(2.5\omega) + 2h(2) \sin(1.5\omega) \\
 &\quad + 2h(3) \sin(0.5\omega)\} \\
 &= e^{j\theta(\omega)} \{H_R(\omega)\}
 \end{aligned}$$

phase angle, $\theta(\omega) = -3.5\omega + \frac{\pi}{2} \Rightarrow$ linear function of ω

$$\text{group delay} = -\frac{d\theta(\omega)}{d\omega} = 3.5 \text{ samples} = 3.5T \text{ seconds} \quad (18)$$

- In general, a linear phase $\theta(\omega) = -\left(\frac{N\omega - \pi}{2}\right)$
- Constant group delay $\tau = \frac{N}{2}$

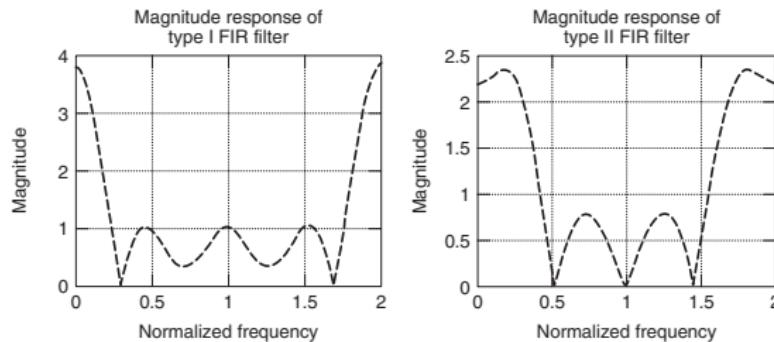


Frequency Response and Group Delay of FIR Filters

	Frequency Response	Group Delay
Type I (Symmetric and Even)	$H(e^{j\omega}) = e^{-j[(N/2)\omega]} \left\{ h\left(\frac{N}{2}\right) + 2 \sum_{n=1}^{N/2} h\left[\frac{N}{2} - n\right] \cos(n\omega) \right\}$	$N/2$
Type II (Symmetric and Odd)	$H(e^{-j\omega}) = e^{-j[(N/2)\omega]} \left\{ 2 \sum_{n=1}^{(N+1)/2} h\left[\frac{N+1}{2} - n\right] \cos\left(\left(n - \frac{1}{2}\right)\omega\right) \right\}$	$N/2$
Type III (Antisymmetric and Even)	$H(e^{-j\omega}) = e^{-j[(N\omega - \pi)/2]} \left\{ 2 \sum_{n=1}^{N/2} h\left[\frac{N}{2} - n\right] \sin(n\omega) \right\}$	$N/2$
Type IV (Antisymmetric and Odd)	$H(e^{-j\omega}) = e^{-j[(N\omega - \pi)/2]} \left\{ 2 \sum_{n=1}^{(N+1)/2} h\left[\frac{N+1}{2} - n\right] \sin\left(\left(n - \frac{1}{2}\right)\omega\right) \right\}$	$N/2$

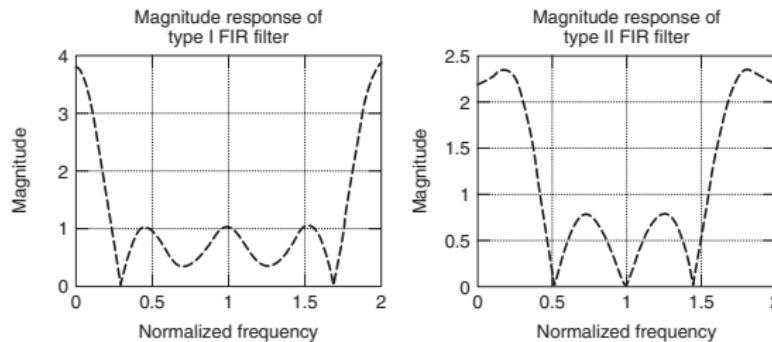


Magnitude Response of Linear Phase FIR Filters



- Type I filters have a non-zero magnitude at $\omega = 0$ and also a non-zero value at the normalized frequency $\omega/\pi = 1$.
- Type II filters have non-zero magnitude at $\omega = 0$ but a zero value at the Nyquist frequency.

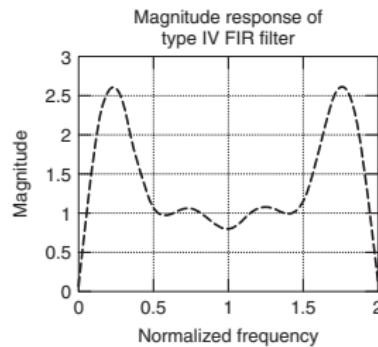
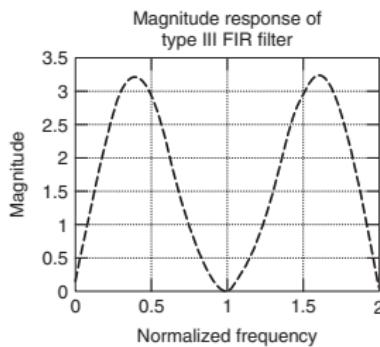
Magnitude Response of Linear Phase FIR Filters



- Type I filters have a non-zero magnitude at $\omega = 0$ and also a non-zero value at the normalized frequency $\omega/\pi = 1$.
- Type II filters have non-zero magnitude at $\omega = 0$ but a zero value at the Nyquist frequency.
- So it is obvious that these filters are not suitable for designing bandpass and highpass filters, where both of them are suitable for low-pass filters.



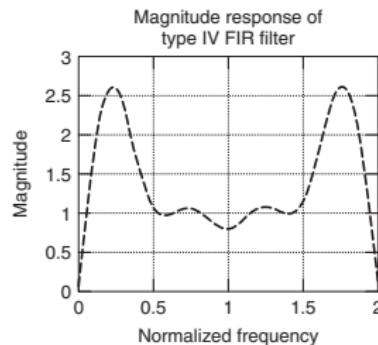
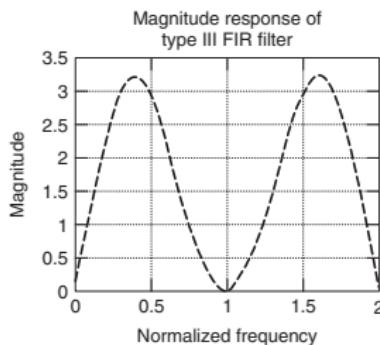
Magnitude Response of Linear Phase FIR Filters



- Type III filters have zero magnitude at $\omega = 0$ and also at $\omega/\pi = 1$ (Nyquist frequency).
- Type IV filters have zero magnitude at $\omega = 0$ and a non-zero magnitude at $\omega/\pi = 1$.



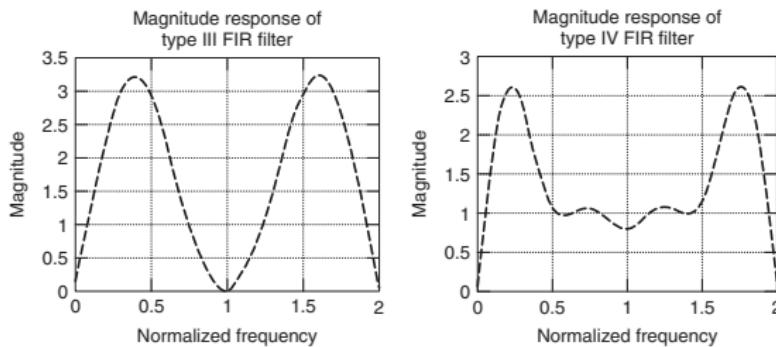
Magnitude Response of Linear Phase FIR Filters



- Type III filters have zero magnitude at $\omega = 0$ and also at $\omega/\pi = 1$ (Nyquist frequency). **So they are suitable for designing bandpass filters but not lowpass and bandstop filters.**
- Type IV filters have zero magnitude at $\omega = 0$ and a non-zero magnitude at $\omega/\pi = 1$.



Magnitude Response of Linear Phase FIR Filters



- Type III filters have zero magnitude at $\omega = 0$ and also at $\omega/\pi = 1$ (Nyquist frequency). **So they are suitable for designing bandpass filters but not lowpass and bandstop filters.**
- Type IV filters have zero magnitude at $\omega = 0$ and a non-zero magnitude at $\omega/\pi = 1$. **They are not suitable for designing lowpass and bandstop filters but are candidates for bandpass and highpass filters.**



Properties of Linear Phase FIR Filters

- An FIR filter with symmetric or antisymmetric coefficients has a linear phase and therefore a constant group delay.
- The reverse statement, an FIR filter with a constant group delay must have symmetric or antisymmetric coefficients.



Thank You
Queries?

