



# Final Project Report

CSCE 325-511: Electronics | Fall 2021 | Dr. Silva

# 1. Introduction

The purpose of this report is to show the developmental process and final results of the lab. The requirement of the project was to build a multi-stage amplifier with the gain 30dB assuming that the load impedance is  $20\Omega$ , with maximum peak-to-peak input voltage of 20mV. The Bipolar Junction diode that will be used in this project is 2N3904. The input impedance of the configurations should be more than  $20k\Omega$ , and result in amplification of the input.

## 2. Design & Calculations

In the advanced report we recognized the given values and analysis of the circuit.

Furthermore, the three stages of amplification are used, since the gain is in dB, it needs to be changed in voltage, the formula  $\text{gain} = 20 \log(\text{Av})$  is used to find out the voltage gain in  $\text{Av}$ . Since  $\text{gain} = 30\text{dB}$  we have the following value:

$$30 = 20 \log(\text{Av})$$

$$1.5 = \log(\text{Av})$$

$$\text{Av} = 10^{1.5}$$

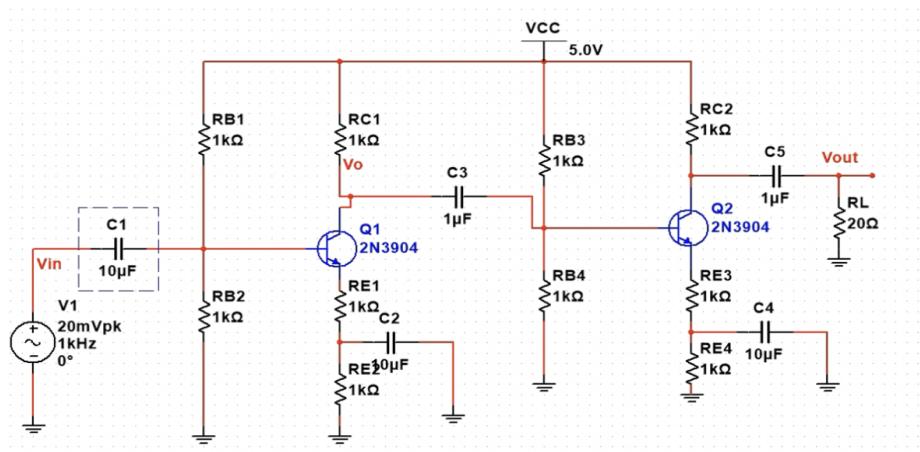
$$\text{Av} = 31.63$$

$$\text{Av} \approx 32$$

From the above calculation of the voltage gain  $\text{Av}$  should be around 32. So to get that, our team has decided to use first stage amplification voltage gain ( $\text{Av1}$ ) = 16, second stage amplification voltage gain ( $\text{Av2}$ ) = 2, hence getting components values of:

|                            |                         |
|----------------------------|-------------------------|
| Voltage gain $ \text{Av} $ | 32                      |
| Input Impedance $R_i$      | $\geq 20\text{k}\Omega$ |
| Load Impedance $R_L$       | $= 20\text{ ohm}$       |
| $\beta$                    | $= 100$                 |
| $V_{o,\text{pk}}$          | 600mA $\mu\text{K}$     |

Our original design looked like:



Following similar formulas from the advanced report, which were

$$A_{V2} = (R_{C2} || R_L) / (R_{E3} + r_{e2})$$

$$A_{V1} = (R_{C1} || Z_{in}) / (R_{E1} + r_{e1})$$

$Z_{in}$  is the input impedance of the second stage and  $A_v = A_{V1} \times A_{V2}$ .

For the final report in the multistage circuit for audio amplification we implemented 3 BJT's instead of 2, so our first attempt for calculations are attached below:

$$V_{CQ3} - V_{opk} \geq 0 \quad V_{CQ3} \geq 0.6V$$

$$V_{CQ} - V_{opk} \leq V_{cc} - V_{RE} - V_{ce\ min} \Rightarrow V_{CQ} \leq 3.6V - V_{RE}$$

$$V_{CQ3} - 0.7V - V_{opk} \geq V_{CQ} = 1.3V$$

$$V_{CQ3} - 0.7V + 0.6V \leq 5 - 0.2V \Rightarrow 4.7V$$

$$1.3 \leq V_{CQ} \leq 3.6V - V_{RE}$$

$$V_{RE} = 1V$$

$$V_{CQ3} = 2.6V \quad (\text{Linear Range})$$

$$(i_H + i_L)_m = \frac{V_{CQ} - 0.7V - V_{opk}}{R_H} - \frac{V_{opk}}{R_L} > 0$$

$$R_H \leq \frac{2.6 - 0.7 - 0.6}{V_{opk}} \cdot 20$$

$$R_H \leq \frac{1.3}{0.6} \times 20 = 43.33$$

$$R_H = 42\Omega$$

$$I_{CQ3} = \frac{V_{CQ3} - 0.7}{R_H} = \frac{1.9V}{42\Omega} = 45.23mA$$

$$R_{i3} = 100 \left( \frac{25mV}{45.23mA} + 42 || 20 \right) = 1.4k\Omega$$

$$R_L \ll R_{i2} \Rightarrow R_Q \leq 140\Omega$$

$$I_{C2} = \frac{5V}{140\Omega} = 35mA.$$

$$R_{H2} = \frac{V_{RE2}}{I_E} = 142\Omega$$

$$\beta_1 = \frac{R_{C2}}{\left(\frac{V_T}{I_{CQ2}}\right) \frac{25mV}{35mA} + 142 \parallel R_{L2}}$$

$$\beta = \frac{140\Omega}{0.71 + \frac{142 \parallel R_{L2}}{R_{L2} + 142}}$$

$$R_{L2} = 19.04\Omega$$

$$R_{i2} = R_{B3} \parallel R_{B4} \parallel (1+\beta) (r_{e2} + R_{H2} \parallel R_{L2})$$

$$R_{i2} = R_{B3} \parallel R_{B4} \parallel 100 \left( \frac{R_C}{AV_e} \right)$$

$$R_{i2} = R_{B3} \parallel R_{B4} \parallel 100 \left( \frac{140}{\beta} \right)$$

$$R_{i2} = R_{B3} \parallel R_{B4} \parallel 1.75k\Omega$$

$$\text{Suppose } R_{B3} = 2k\Omega$$

Now

$$\frac{R_{B4}}{R_{B3} + R_{B4}} \times 5V = V_{CC} - V_{RE2} - 0.7$$

$$\frac{R_{B4}}{1k\Omega + R_{B4}} \times 5V = 3.3V$$

$$\frac{RB_4}{1k\Omega + RB_4} = 0.66$$

$$RB_4 = 660\Omega + 0.66RB_4$$

$$RB_4 = 1.9k\Omega$$

$$R_{i2} = RB_3 \parallel RB_4 \parallel 3.5k\Omega$$

$$R_{i2} = \left( \frac{1k\Omega \cdot 1.9k\Omega}{1k\Omega + 1.9k\Omega} \right) \parallel 3.5k\Omega$$

$$R_{i2} = (0.655k\Omega) \parallel 3.5k\Omega$$

$$R_{i2} = 551\Omega$$

$$RC_1 = 300\Omega$$

$$IC_1 = \frac{5V}{300\Omega} = 16mA$$

$$RH_1 = \frac{VRE}{IE} = \frac{1}{16.9A} = 59.99\Omega = 60$$

$$4 = \frac{RC_2}{\frac{25mV}{16.67mA} + 60\Omega \parallel RL_1}$$

$$4 = \frac{\frac{300\Omega}{1.5 + \frac{60 \times RL_1}{RL_1 + 60}}}{RL_1 + 60}$$

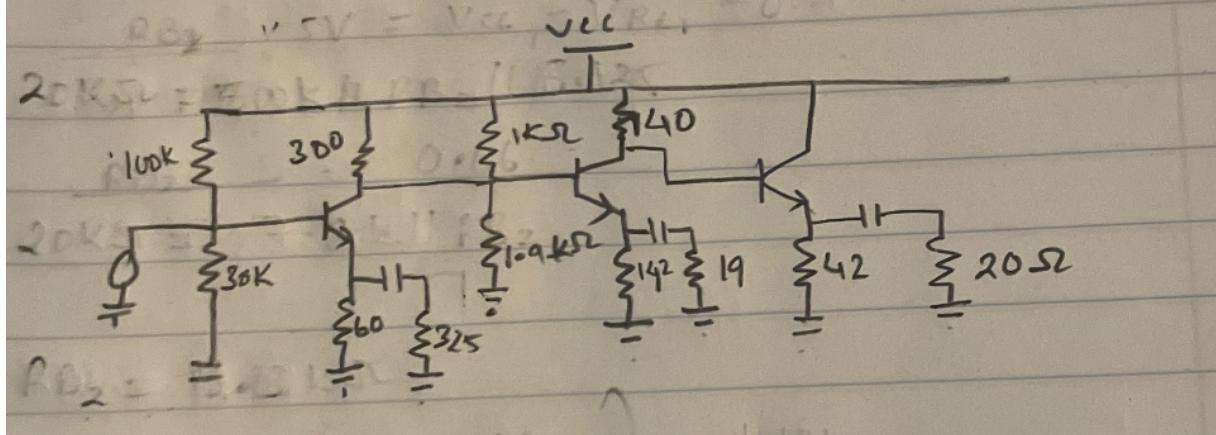
$$RL_1 = 326.66 \approx 325$$

$$R_i = R_{B1} \parallel R_{B2} \parallel (1 + \beta) (r_{e2} + R_H \parallel R_L)$$

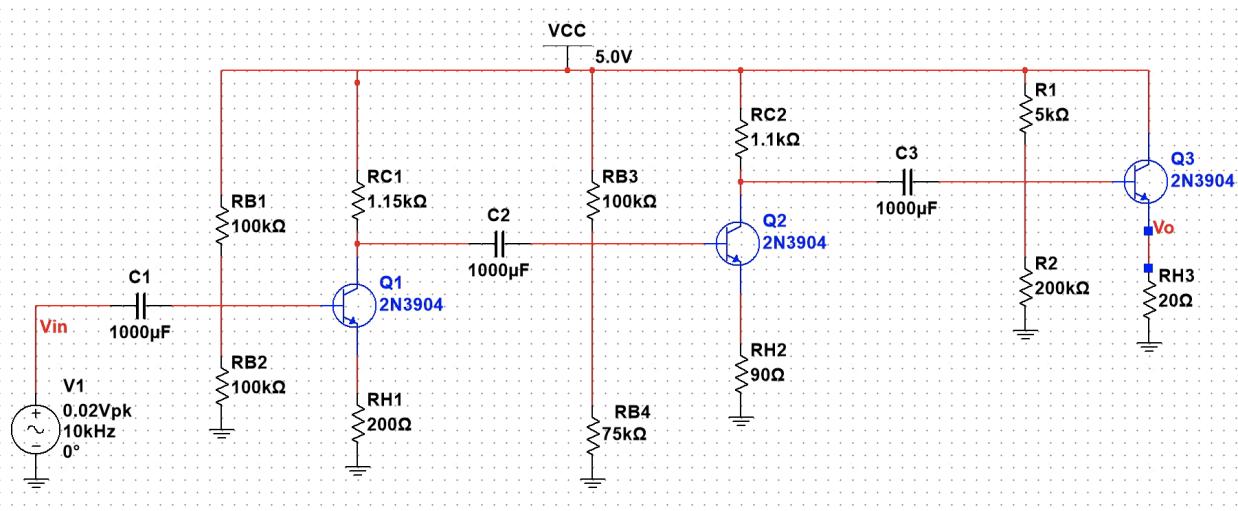
$$R_i = R_{B1} \parallel R_{B2} \parallel 100 \left( \frac{325}{4} \right)$$

$$20\text{ k}\Omega = R_{B1} \parallel R_{B2} \parallel 8 + 125$$

Suppose  $R_{B1} = 100\text{k}$ . and  $R_{B2} = 30\text{ k}\Omega$ .

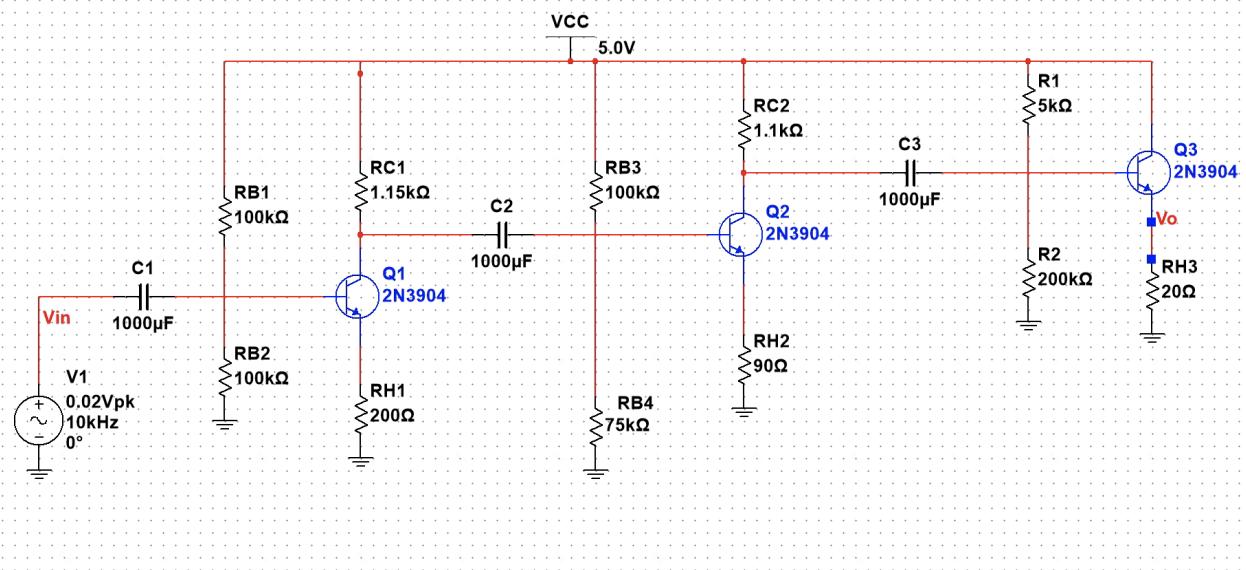


However, with few more alterations of components to get reasonable simulation results our final design was the following:

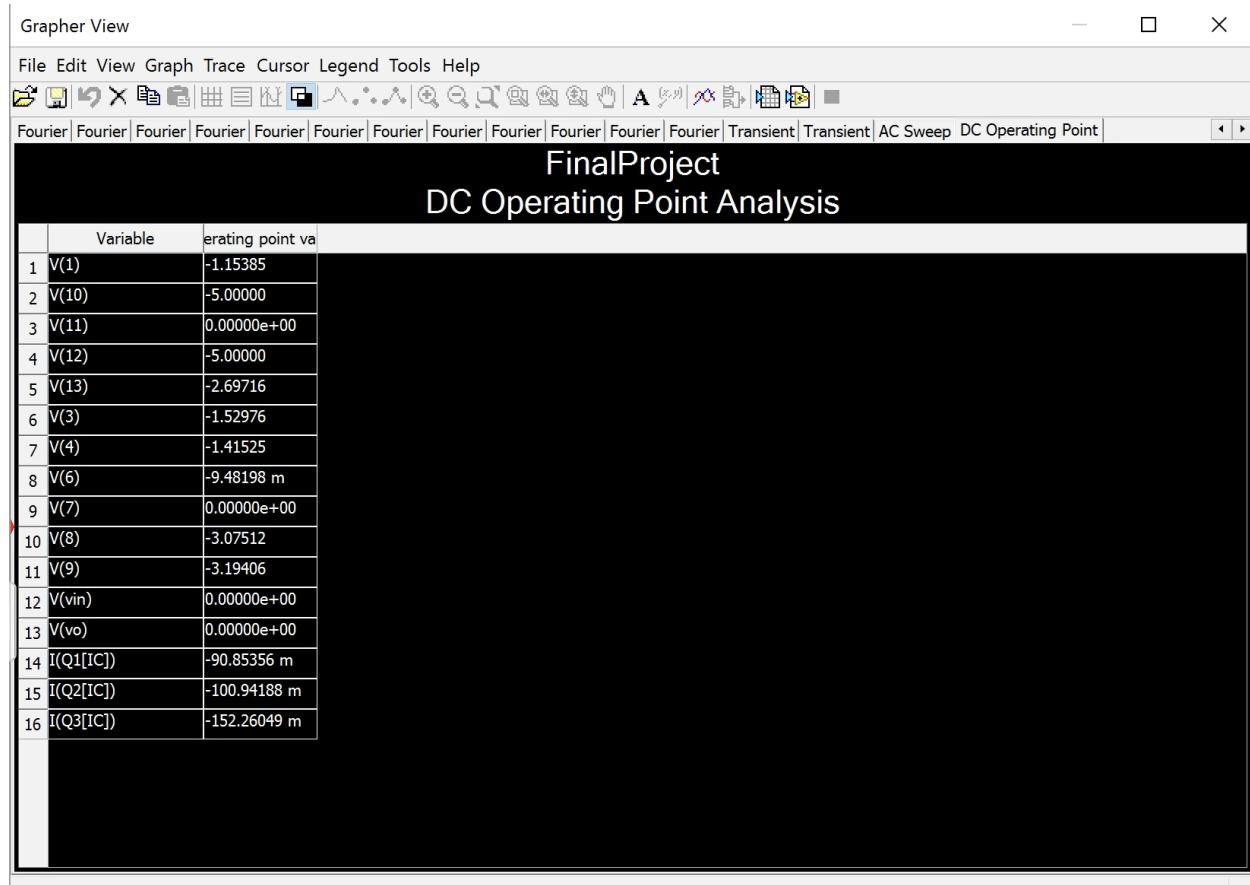


## Simulations

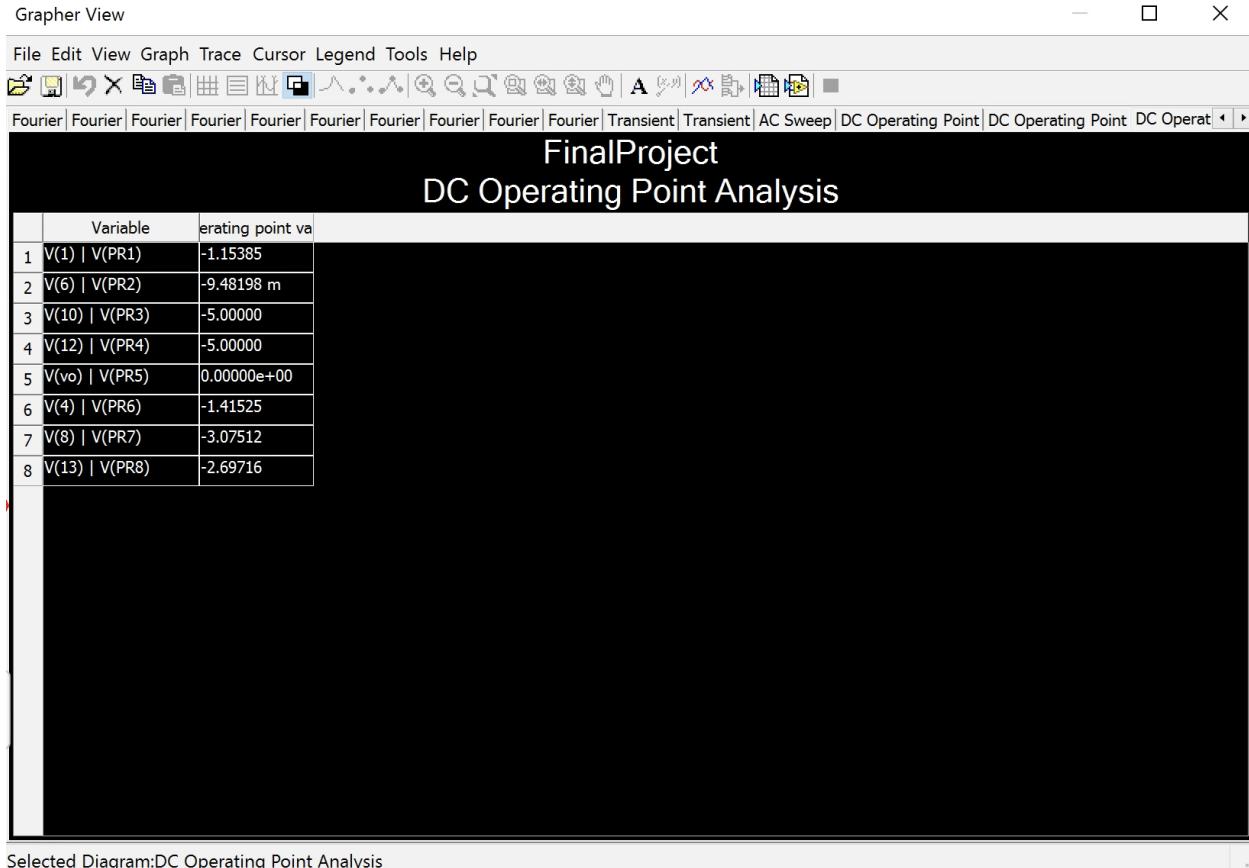
### 3.1 Schematic



### 3.2 DC Operating Point

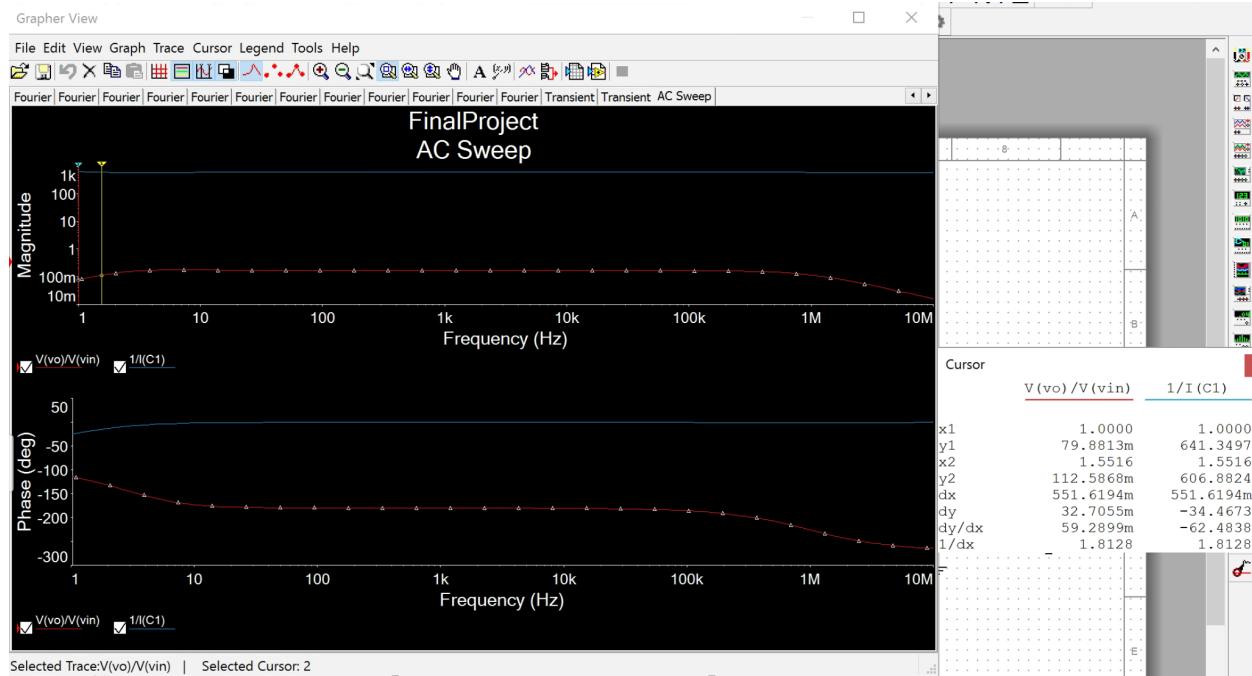


Selected Diagram:DC Operating Point Analysis

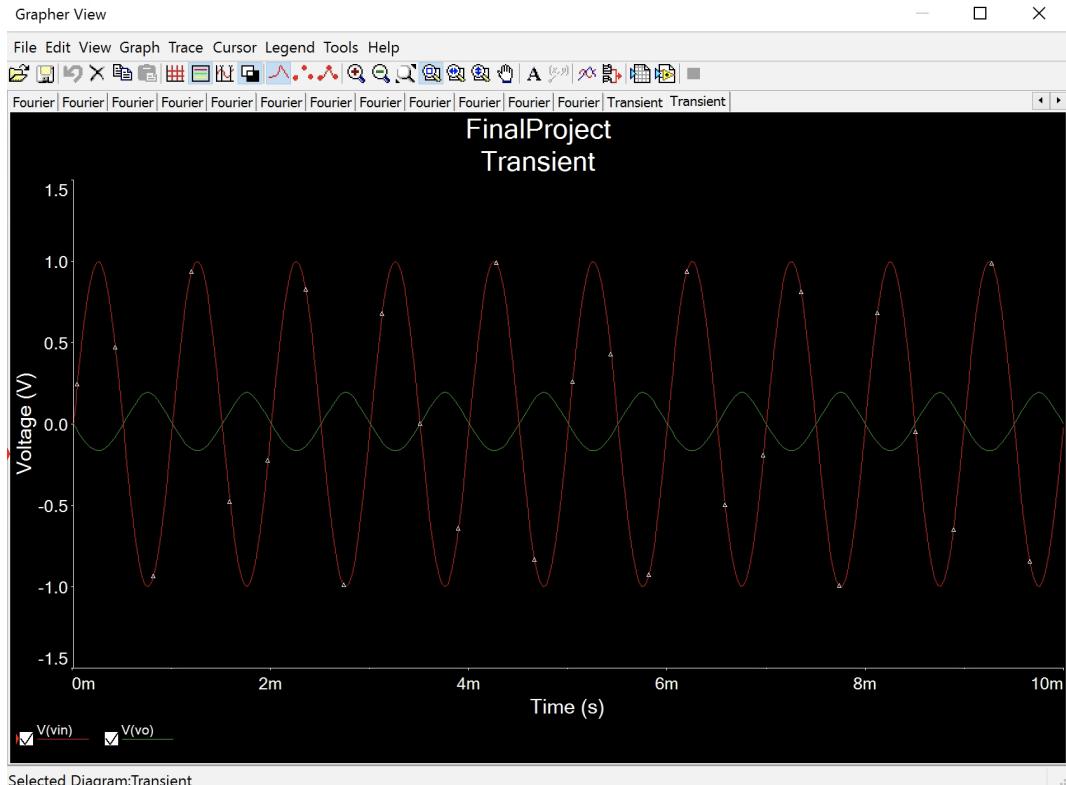


Selected Diagram:DC Operating Point Analysis

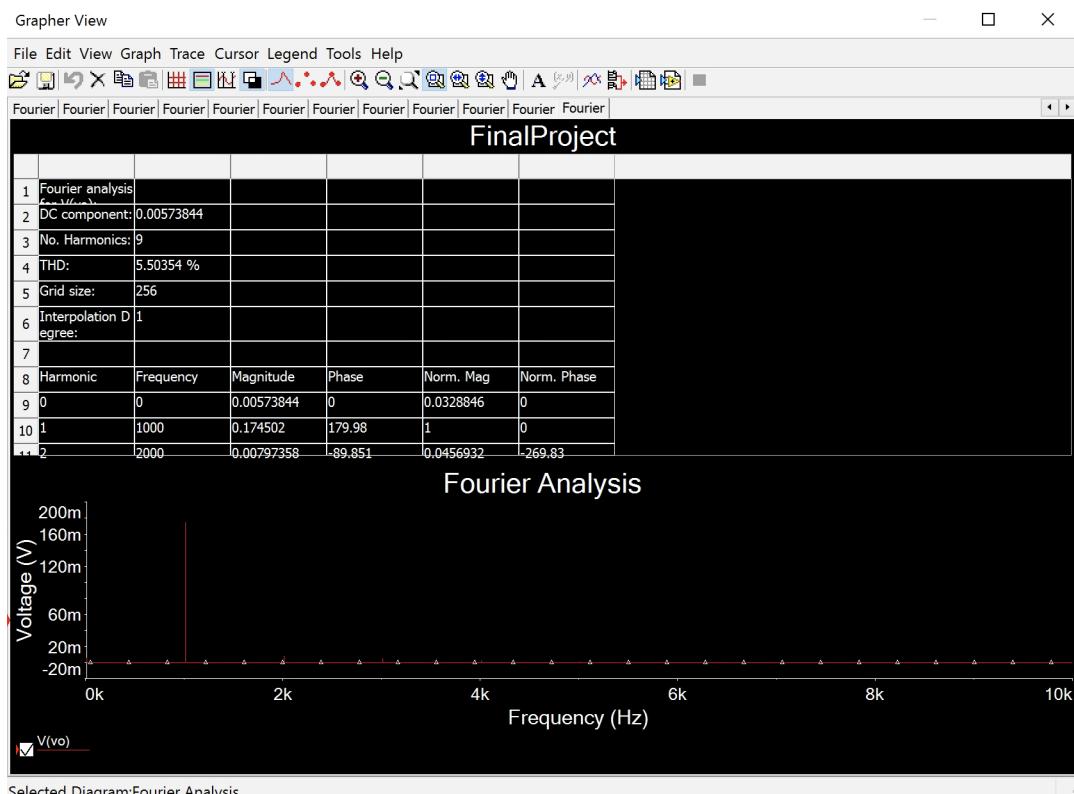
### 3.3 AC Sweep



## 3.4 Transient



## 3.5 Fourier

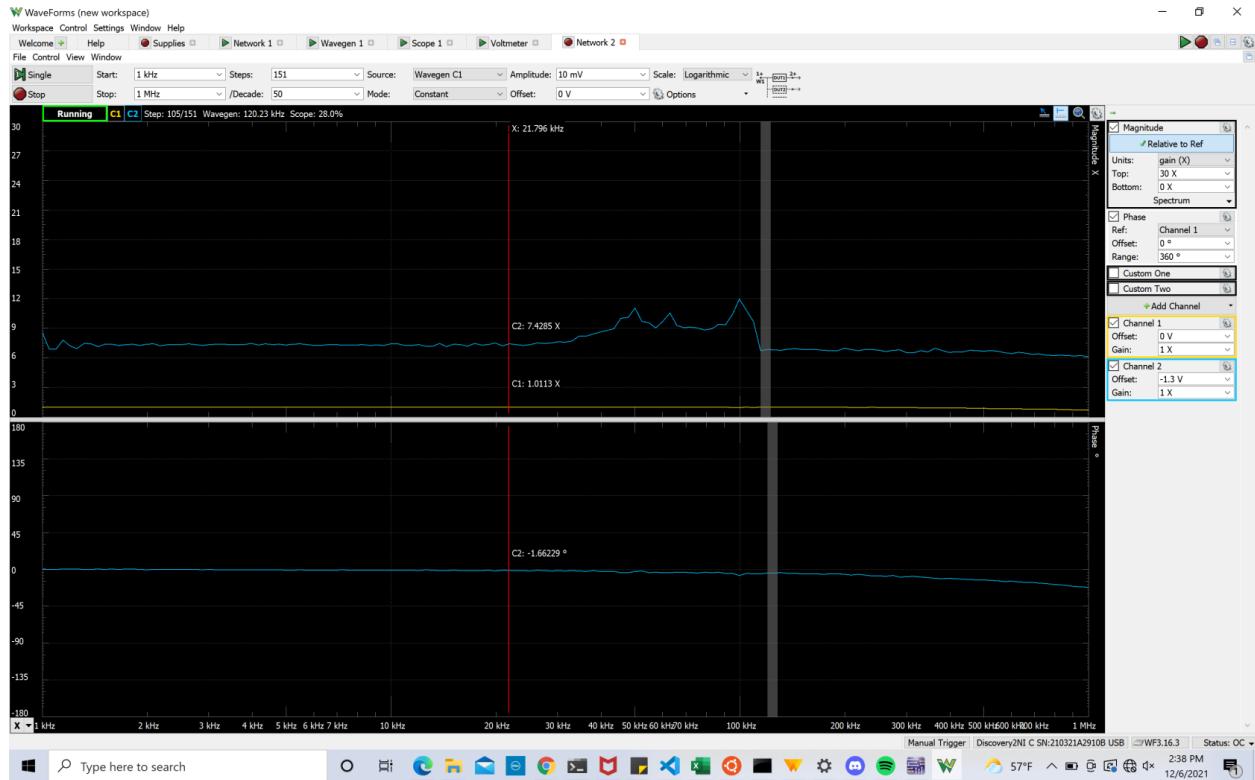


### 3. Measurements / Demo

#### 4.1 DC Values

|           | DC Values |
|-----------|-----------|
| $V_{RB2}$ | 1.01 V    |
| $V_{RE1}$ | 1.175V    |
| $V_{RE2}$ | 3.13 V    |
| $V_{RE3}$ | 2.531 V   |
| $V_{RC}$  | 4.34 V    |
| $V_o$     | .213 V    |
| $I_{C1}$  | 97.34 mA  |
| $I_{C2}$  | 114.23 mA |
| $I_{C3}$  | 167.16 mA |

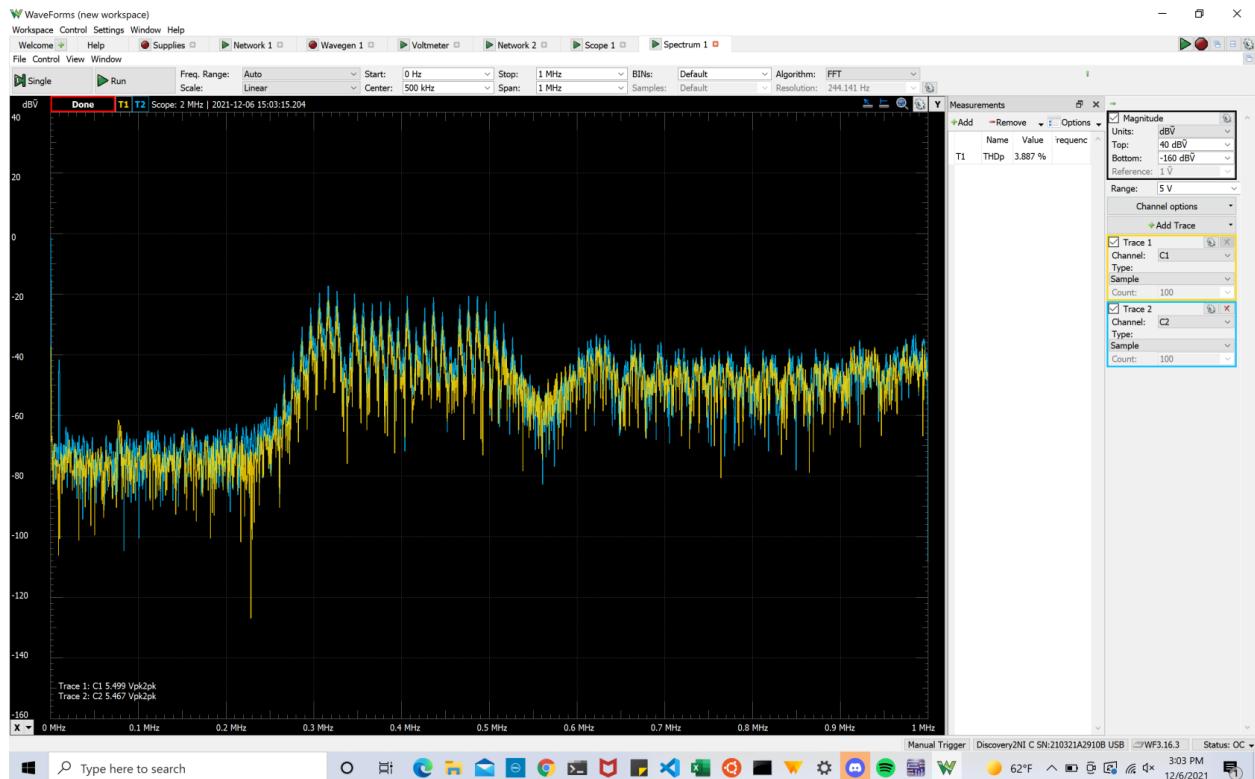
#### 4.2 Network Analyzer



## 4.3 Scope

The scope is shown in the demo video where channel 2 is amplified as it receives an input. We can see the changes in amplification as the volume is increased or decreased in the input.

## 4.4 THD



## 4. Data Tables

### 5.1 Calculations / Simulations / Measurements

|                      | Calculations | Simulations | Measurements |
|----------------------|--------------|-------------|--------------|
| <b>Gain</b>          | 30 dB        | 19.98 dB    | 7.4285       |
| <b>R<sub>i</sub></b> | -            | 606.88 m    | 510.12 m     |
| <b>Transient</b>     | -            | 4.053k      | Demo video   |
| <b>THD</b>           | -            | 5.50354%    | 3.887%       |

### 5.2 DC values

| DC Values        | Simulations | Measurements |
|------------------|-------------|--------------|
| V <sub>RB2</sub> | 1.15 V      | 1.01 V       |

|           |           |           |
|-----------|-----------|-----------|
| $V_{RE1}$ | 1.41 V    | 1.175V    |
| $V_{RE2}$ | 3.07 V    | 3.13 V    |
| $V_{RE3}$ | 2.73 V    | 2.531 V   |
| $V_{RC}$  | 5 V       | 4.34 V    |
| $V_o$     | 180 mV    | .213 V    |
| $I_{C1}$  | 90.85 mA  | 97.34 mA  |
| $I_{C2}$  | 100.94 mA | 114.23 mA |
| $I_{C3}$  | 152.26 mA | 167.16 mA |

## 5. Conclusion

For this project most of the values between the simulations and measurements were pretty consistent for the circuits. If there were any minor differences, that's probably because of component differences, since we did not use exact values for resistors and capacitors. Despite these differences, we were able to get clear sound using the speaker, which was connected to a MacBook Pro as the input sound. As seen in the demo video, changing the volume on the input changed the amplification on the output, hence making the speaker louder, and showing more fluctuations on the scope (channel 2).