

## 2.1: Plane Wave Solutions

Assume wave propagation in  $x$ -axis, thus:  $\frac{\partial}{\partial y} = \frac{\partial}{\partial z} = 0$

$$\therefore \nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\therefore \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & 0 & 0 \\ E_x & E_y & E_z \end{vmatrix} = -\mu \frac{\partial}{\partial t} (H_x \hat{x} + H_y \hat{y} + H_z \hat{z})$$

$$\therefore 0 \hat{x} = -\mu \frac{\partial}{\partial t} H_x \hat{x}$$

$$\therefore \frac{\partial H_x}{\partial t} = 0 \rightarrow \begin{cases} H_x = \text{const. } x \\ H_x = 0 \end{cases} \checkmark$$

no mag. field comp. along the direction of Prop.

• lossless medium

$$\therefore \nabla \times \vec{H} = \epsilon \frac{\partial \vec{E}}{\partial t} \quad \text{"Source Free + lossless"}$$

$$\therefore \frac{\partial E_x}{\partial t} = 0 \rightarrow \begin{cases} E_x = \text{const. } x \\ E_x = 0 \end{cases} \checkmark$$

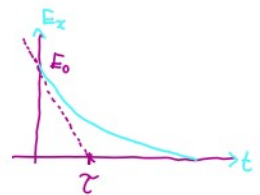
• lossy medium:

$$\nabla \times \vec{H} = \left[ \sigma + j\omega\epsilon \right] \vec{E} = \sigma \langle E_x, E_y, E_z \rangle + j\omega\epsilon \langle E_x, E_y, E_z \rangle$$

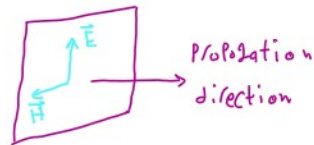
$$\therefore \sigma E_x + j\omega\epsilon E_x = 0$$

$$\therefore \frac{\partial E_x}{\partial t} = -\frac{\sigma}{\epsilon} E_x$$

$$\therefore E_x = E_0 e^{-\gamma t}, \quad \gamma = \frac{\sigma}{\epsilon} \rightarrow \infty \quad \therefore E_x \rightarrow 0$$



→ there's no component of  $\vec{E}, \vec{H}$  along the direction of Propagation,  $\vec{E}, \vec{H}$  are found in the plane normal to Prop. direction.  $\vec{P} = \vec{E} \times \vec{H}$



→ there for,  $\vec{E}$  wave equation become:

$$\nabla_x^2 \vec{E} + \omega^2 \mu \epsilon \vec{E} = 0, \quad \vec{E} = E_y \hat{y} + E_z \hat{z}$$

$$\therefore \frac{\partial^2 E_y}{\partial x^2} - \gamma^2 E_y = 0, \quad \gamma = -\omega^2 \mu \epsilon$$

$$\therefore E_y(x) = \underbrace{E_y^+ e^{-\gamma x}}_{\text{Incident Forward}} + \underbrace{E_y^- e^{\gamma x}}_{\text{Reflected backward}}$$

Similar for  $z$  Component:

$$E_z(x) = \underbrace{E_z^+ e^{-\gamma x}}_{\text{Incident Forward}} + \underbrace{E_z^- e^{\gamma x}}_{\text{Reflected backward}}$$

→  $\vec{H}$  wave eq:

$$\nabla_x^2 \vec{H} - \gamma^2 \vec{H} = 0, \quad \vec{H} = H_y \hat{y} + H_z \hat{z}$$

→  $y$  comp:

$$\therefore \frac{\partial^2 H_y}{\partial x^2} - \gamma^2 H_y = 0$$

$$\therefore H_y(x) = \underbrace{H_y^+ e^{-\gamma x}}_{\text{Incident Forward}} + \underbrace{H_y^- e^{\gamma x}}_{\text{Reflected backward}}$$

→  $z$  comp:

$$\therefore \frac{\partial^2 H_z}{\partial x^2} - \gamma^2 H_z = 0$$

$$\therefore H_z(x) = \underbrace{H_z^+ e^{-\gamma x}}_{\text{Incident Forward}} + \underbrace{H_z^- e^{\gamma x}}_{\text{Reflected backward}}$$

→ We send the electric signal in the Tx, so we know  $E_{oy}, E_{oz}$  constants. Mag field is generated due to the electric field.

→ We send the electric signal in the Tx, so we know  $E_{oy}, E_{oz}$  constants. Mag field is generated due to the electric field, and we don't know the  $H_{oy}, H_{oz}$  const, so we need to find a relation between  $H, E$  const. to know  $H_{oy}, H_{oz}$ .

### \* Relation between E, H Constants

We will get the relation from Maxwell eq.

$$\nabla \times \vec{E} = -\mu_0 \epsilon_0 \frac{\partial \vec{H}}{\partial t}$$

$$\begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & 0 & 0 \\ E_x & E_y & E_z \end{vmatrix} = -\mu_0 \epsilon_0 \frac{\partial}{\partial t} \langle 0, H_y, H_z \rangle$$

y-dir:  $-(\frac{\partial E_z}{\partial x}) = -\mu_0 \epsilon_0 \frac{\partial H_y}{\partial t}$

z-dir:  $\frac{\partial E_y}{\partial x} = -\mu_0 \epsilon_0 \frac{\partial H_z}{\partial t}$

$$\therefore \frac{\partial E_y}{\partial x} = -\mu_0 \epsilon_0 \frac{\partial H_z}{\partial t} \quad \therefore E_y = E_{oy} e^{-j\omega t} + E_{oz} e^{j\omega t} \quad \therefore H_z = H_{oz} e^{-j\omega t} + H_{oy} e^{j\omega t}$$

→ For incident wave

$$\therefore +8 E_{oy} e^{-j\omega t} = +\mu_0 \epsilon_0 \frac{\partial}{\partial t} (H_{oz} e^{-j\omega t})$$

$$\therefore H_{oz} = \frac{E_{oy}}{\mu_0 \epsilon_0 \omega} = \frac{E_{oy}}{\eta}$$

$$\therefore H_{oz} = \frac{E_{oy}}{\eta}$$

$\eta$ : wave Impedance,  $\eta = \frac{\mu_0 \epsilon_0}{\epsilon}$

→ For reflected wave

$$\therefore H_{oz} = \frac{E_{oz}}{\eta}$$

$$\therefore E_y(x) = E_{oy} e^{-j\omega t} + E_{oz} e^{j\omega t}$$

$$\therefore H_z(x) = \frac{E_{oy}}{\eta} e^{-j\omega t} + \frac{E_{oz}}{\eta} e^{j\omega t}$$

$$\therefore E_z(x) = E_{oz} e^{-j\omega t} + E_{oy} e^{j\omega t}$$

$$\therefore H_z(x) = \frac{E_{oy}}{\eta} e^{-j\omega t} - \frac{E_{oz}}{\eta} e^{j\omega t}$$

→ it's important to notice the signs of each forward & backward component in the above two eqs. The direction of propagation is  $\hat{z}$  in this case  $E_z, H_z$ , meaning the propagation direction is in  $\hat{z}$ , now if we get the cross product of  $E_z$  (or  $H_z$ ) of the forward part, it will be  $(E_{oz} e^{-j\omega t}) \hat{z} \times (\frac{E_{oy}}{\eta} e^{-j\omega t}) \hat{y}$  which is  $\hat{z} \times \hat{y}$ , which is  $\hat{x}$  "the propagation direction", for the backward component,  $\hat{z} \times \hat{y} = -\hat{x}$  which is the direction of the reflected wave.

\* Prove that  $\vec{E}$  &  $\vec{H}$  are orthogonal

$$\therefore \vec{E} = \langle 0, E_{oy} e^{-j\omega t}, E_{oz} e^{j\omega t} \rangle$$

$$\therefore \vec{H} = \langle 0, \frac{E_{oy}}{\eta} e^{-j\omega t}, \frac{E_{oz}}{\eta} e^{j\omega t} \rangle$$

$$\vec{E} \cdot \vec{H} = E_{oy} e^{-j\omega t} \cdot \frac{E_{oy}}{\eta} e^{-j\omega t} + E_{oz} e^{j\omega t} \cdot \frac{E_{oz}}{\eta} e^{j\omega t}$$

$$\vec{E} \cdot \vec{H} = 0$$

$$\therefore \boxed{\vec{E} \perp \vec{H}}$$

$$\therefore \vec{E} = \langle 0, E_{oy} e^{-j\omega t}, E_{oz} e^{j\omega t} \rangle$$

$$\therefore \vec{H} = \langle 0, \frac{E_{oy}}{\eta} e^{-j\omega t}, \frac{E_{oz}}{\eta} e^{j\omega t} \rangle$$

$$\therefore \vec{E} \cdot \vec{H} = 0$$

$$\therefore \boxed{\vec{E} \perp \vec{H}}$$

## Complex Propagation Constant

For a lossy medium, The Maxwell equation:

$$\nabla \times \vec{H} = \sigma \vec{E} + j\omega \epsilon \vec{E}$$

$$= j\omega (\epsilon + \frac{\sigma}{j\omega}) \vec{E}$$

$$= j\omega \epsilon_{eff} \vec{E}$$

$\epsilon_{eff} = \epsilon - j\frac{\sigma}{\omega}$ , therefore, for a dielectric medium  $\sigma = j\omega \epsilon_{eff}$  but for a lossy medium  $\sigma = \sqrt{j\omega \epsilon_{eff}}$  Complex

$\gamma$  is now a complex number

$$\gamma = \sqrt{-j\omega \mu (\epsilon - j\frac{\sigma}{\omega})} = \alpha + j\beta$$

attenuation constant      phase shift

$\gamma = \sqrt{j\omega \mu \epsilon_{eff}}$   $\gamma$  is complex too due to  $\epsilon_{eff}$

The final form of the wave eqs

$$E_y(x, z) = E_{y0} e^{-\alpha x} e^{j(\omega t - \beta x)} + E_{y0} e^{-\alpha x} e^{j(\omega t + \beta x)}$$

$$E_z(x, z) = E_{z0} e^{-\alpha x} e^{j(\omega t - \beta x)} + E_{z0} e^{-\alpha x} e^{j(\omega t + \beta x)}$$

$$H_y(x, z) = \frac{E_{y0}}{\eta} e^{-\alpha x} e^{j(\omega t - \beta x)} + \frac{E_{y0}}{\eta} e^{-\alpha x} e^{j(\omega t + \beta x)}$$

$$H_z(x, z) = \frac{E_{z0}}{\eta} e^{-\alpha x} e^{j(\omega t - \beta x)} + \frac{E_{z0}}{\eta} e^{-\alpha x} e^{j(\omega t + \beta x)}$$

Note that  $\alpha$  has an undesirable effect on the wave, as the wave propagates the  $x$  dir,  $e^{-\alpha x}$  meaning the magnitude is "attenuation", this is halves in the lossy medium. before in the lossless dielectric medium  $\gamma$  were was pure imaginary, therefore  $\alpha = 0$  thus there were no attenuation, only the change was in the phase.

### Velocities:

Phase Velocity  $V_{ph} = \frac{\omega}{\beta}$  is the velocity of the wave front of the plane

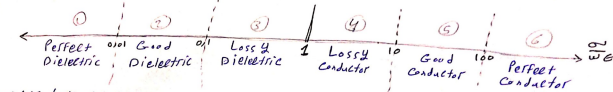
Group Velocity  $V_g = \frac{\omega}{\alpha}$  ----- energy

The optimum value of  $\alpha$  is 0 or constant that is known to account for at the  $T_x$  &  $R_x$ , for  $\beta$  is linear with freq.

## 2.2. Propagation in Dielectric Medium

Conduction current density  $J_c = \sigma \vec{E}$ , displacement current density  $J_d = \frac{\partial D}{\partial t} = j\omega \epsilon \vec{E}$   
The ratio between the magnitudes  $|\frac{J_c}{J_d}| = \frac{\sigma}{\omega \epsilon}$   
if  $\frac{\sigma}{\omega \epsilon} \ll 1$ , the medium is considered as dielectric, while if  $\frac{\sigma}{\omega \epsilon} \gg 1$  the medium is considered conducting.

The classification of mediums can be summarized in the following graph:



We have 6 types of mediums, we need to know how to get  $\alpha, \beta$  for each.

### 1. Lossy Dielectric medium, $0 < \frac{\sigma}{\omega \epsilon} < 1$

The propagation constant  $\gamma$  is given as  $\gamma = \sqrt{-j\omega \mu \epsilon_{eff}} = \sqrt{-j\omega \mu (\epsilon - j\frac{\sigma}{\omega})}$   
 $\gamma = j\omega \sqrt{\mu \epsilon} \sqrt{1 - j\frac{\sigma}{\omega \epsilon}}$   
 $\sqrt{1 - j\frac{\sigma}{\omega \epsilon}} \approx 1 - \frac{j}{2} \frac{\sigma}{\omega \epsilon} + \frac{1}{8} (\frac{\sigma}{\omega \epsilon})^2 + \frac{j}{16} (\frac{\sigma}{\omega \epsilon})^3$   
 $\gamma = j\omega \sqrt{\mu \epsilon} [\alpha + j\beta]$  "Series define"

$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} [1 - \frac{\sigma^2}{8\omega^2 \epsilon^2}]$ ,  $\beta = \omega \sqrt{\mu \epsilon} [1 + \frac{\sigma^2}{8\omega^2 \epsilon^2}]$

neither  $\alpha$  is constant nor  $\beta$  is linear with  $\omega$

### 2. Good Dielectric medium, $0 < \frac{\sigma}{\omega \epsilon} < 1$

In this case  $\sqrt{1 - j\frac{\sigma}{\omega \epsilon}} \approx 1 - \frac{j}{2} \frac{\sigma}{\omega \epsilon}$  as  $\frac{\sigma}{\omega \epsilon} \ll 1$   
thus,  $\gamma = j\omega \sqrt{\mu \epsilon} \sqrt{1 - j\frac{\sigma}{\omega \epsilon}} \approx j\omega \sqrt{\mu \epsilon} (1 - \frac{j}{2} \frac{\sigma}{\omega \epsilon})$   
 $\gamma = \alpha + j\beta$

$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} \rightarrow$  Constant "solved by approximation"

$\beta = \omega \sqrt{\mu \epsilon} \rightarrow$  Linear with freq.

### 3. Perfect Dielectric, $\frac{\sigma}{\omega \epsilon} \ll 1$

$\sqrt{1 - j\frac{\sigma}{\omega \epsilon}} \approx 1$  as  $\frac{\sigma}{\omega \epsilon} \ll 1$

$\gamma = j\omega \sqrt{\mu \epsilon} \sqrt{1 - j\frac{\sigma}{\omega \epsilon}} \approx j\omega \sqrt{\mu \epsilon}$

$\gamma = \alpha + j\beta$

$\alpha = 0$ ,  $\beta = \omega \sqrt{\mu \epsilon}$  } Optimum values

## Skin Depth (6.6)

$\delta$  is the depth at which the magnitude of  $\vec{E}$  reaches 37% of its initial value.

$$|E(x)| = E_0 e^{-\alpha x}$$

at  $x = \delta$ :

$$E_0 e^{-\alpha \delta} = 0.37 E_0 = E_0 e^{-1}$$

$$\therefore \alpha \delta = 1 \rightarrow \delta = \frac{1}{\alpha}$$

We conclude: we know that the attenuation is  $\alpha$  for a medium, we know that the maximum distance between the  $T_x$  &  $R_x$  is  $\delta = \frac{1}{\alpha}$ , any further the  $R_x$  will consider it as noise.

