

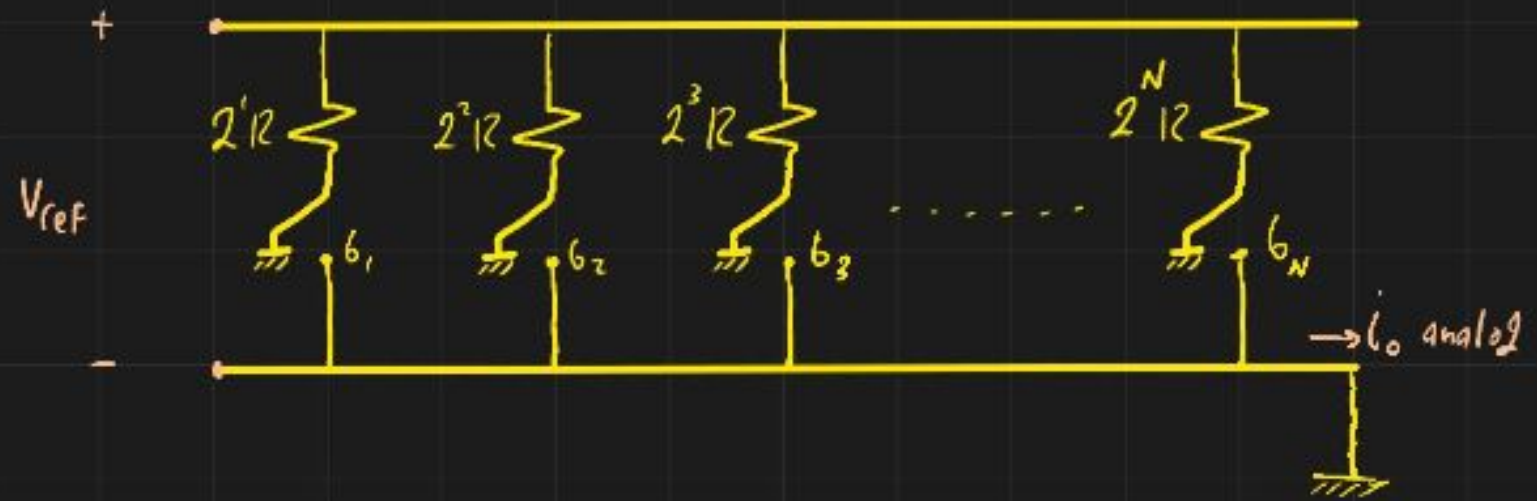
## \* Digital to Analog Converter (DAC):

→ total current  $I_o$ :

→ resultant current in each branch if it's switch is grounded.

$$I_o = \frac{V_{ref}}{2R} b_1 + \frac{V_{ref}}{4R} b_2 + \dots + \frac{V_{ref}}{2^N R} b_N$$

$$\therefore I_o = \frac{V_{ref}}{R} \left[ \frac{b_1}{2^1} + \frac{b_2}{2^2} + \dots + \frac{b_N}{2^N} \right] = \frac{V_{ref}}{R} \sum_{n=1}^N \frac{b_n}{2^n}$$



c) for  $V_{ref} = 10V$ ,  $R = 10K\Omega$  and  $N = 8$  find the maximum value of  $I_o$  obtained.  $I_{o_{max}}$

→ all bits are on  $\sim b = 1$

$$\therefore I_{o_{max}} = \frac{V_{ref}}{R} \sum_{n=1}^8 \frac{1}{2^n} = \frac{10}{10} (0.996) \rightarrow I_{o_{max}} = 0.996 \text{ mA}$$

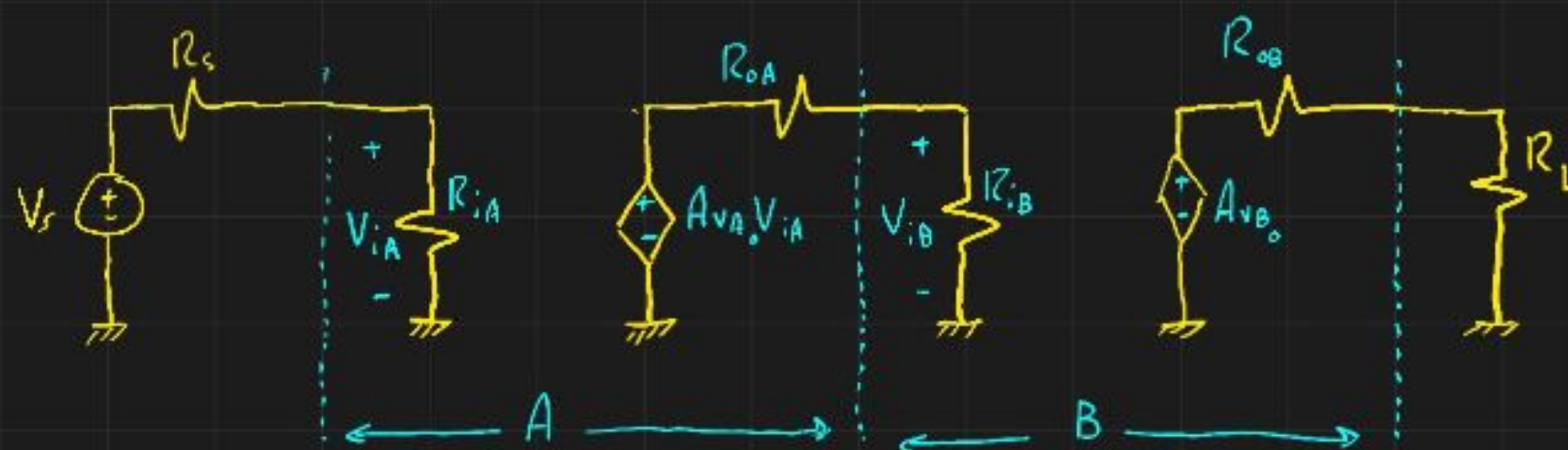
→ what is the change in  $I_o$  resulting from the LSB change from 0 to 1.

$$\rightarrow \text{LSB is } b_N = b_8 \rightarrow I_o = \frac{V_{ref}}{R} \left[ \frac{b_8}{2^8} \right] = 3.9 \mu A$$

$\therefore$  the op will change by  $3.9 \mu A$

## \* A, B Amplifier:

→ it's an amplifier of 2 cascaded amplifiers to increase the gain.





### 1.53) A, B amplifiers

$$V_s = 10 \text{ mV}$$

$$R_s = 100 \text{ k}\Omega \quad R_L = 100 \Omega$$

هل التوصيلة الأفضل SABL ولا ؟

Source → SABL → Load

	A	B
$A_{v_o}$	100	10
$R_i$	100 k $\Omega$	10 k $\Omega$
$R_o$	10 k $\Omega$	1 k $\Omega$

We'll get overall gain in both cases

and the bigger is the better.



$$A_v = A_{vA} \cdot A_{vB} = \frac{V_o}{V_s} = \frac{V_{iA}}{V_s} \cdot \frac{V_{iB}}{V_{iA}} \cdot \frac{V_o}{V_{iB}}$$

$$\rightarrow \frac{V_{iA}}{V_s} = \frac{R_{iA}}{R_{iA} + R_s} = \frac{100}{100 + 100} = 0,5$$

$$\rightarrow \frac{V_{iB}}{V_{iA}} = \frac{R_{iB}}{R_{iB} + R_{oA}} = \frac{(10)(100)}{10 + 10} = 50$$

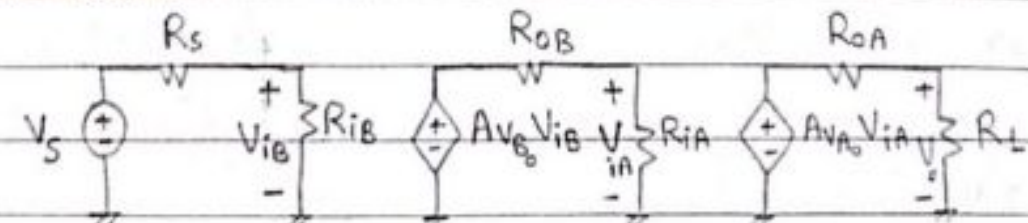
$$\rightarrow \frac{V_o}{V_{iB}} = \frac{R_L}{R_L + R_{oB}} (A_{vB_o}) = \frac{(100)(10)}{100 + 1000} = 0,909$$

$$\therefore A_v = (0,5)(50)(0,909) \rightarrow \therefore A_v = 22,73$$

$$\therefore A_v = 20 \log(22,7) \rightarrow \therefore A_v = 27,1 \text{ dB}$$

SABL

② for SBAL



$$\frac{V_{iB}}{V_s} = \frac{R_{iB}}{R_s + R_{iB}} = \frac{10}{100 + 10} = \frac{1}{11}$$

$$\frac{V_{iA}}{V_{iB}} = \frac{A_{vB_o} R_{iA}}{R_{oB} + R_{iA}} = \frac{10 * 100}{1 + 100} = \frac{1000}{101}$$

$$\frac{V_o}{V_{iA}} = \frac{A_{vA_o} R_L}{R_{oA} + R_L} = \frac{100 * 100}{100 + 10.000} = \frac{100}{101}$$

$$\frac{V_o}{V_s} = \frac{1}{11} * \frac{1000}{101} * \frac{100}{101} = 0,891$$

$$A_v = 20 \log(0,891) = -1 \text{ dB}$$

SBAL

$$\therefore A_{v_{SABL}} > A_{v_{SBAL}}$$

$\therefore$  SABL is better

note that we can know without calculating anything, we know that the higher the  $R_{i_{in}}$  the better and also the lower the  $R_{o_{out}}$  the better

A has a higher  $R_{i_{in}}$ , and B has the lower  $R_{o_{out}} \rightarrow \therefore$  A at i/p and B at o/p  $\rightarrow \therefore$  SABL



## Amplifier Model:

- it's preferred to make  $R_i \gg$  to neglect the drop out voltage on  $R_s$
- ideal  $\rightarrow V_i = V_s$  "  $R_i = \infty$  "
- $R_o \ll$



## MOSFET Circuit at DC:

EX (5.3)

$$I_D = 0.2 \text{ mA}$$

$$V_D = 0.2 \text{ V}$$

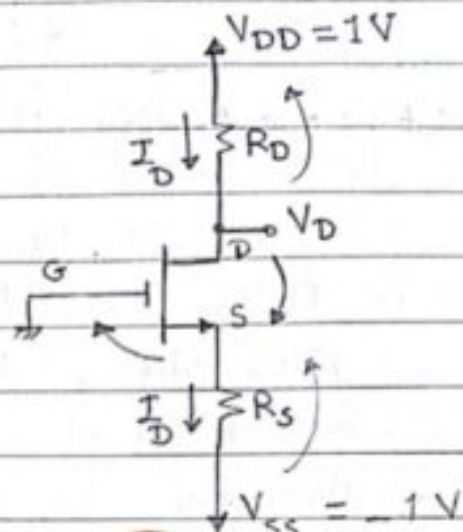
$$V_t = 0.5 \text{ V}$$

$$\mu_n C_{ox} = 400 \text{ } \mu\text{A/V}^2$$

$$L = 0.5 \text{ } \mu\text{m}, W = 15 \text{ } \mu\text{m}$$

$$\lambda = 0$$

find:  $R_D, R_S$



$\rightarrow$  Assume saturation "  $V_{DS} > V_{ov}$  "

$$\therefore V_{DS} - V_{GS} = V_D - V_S - V_t + V_S$$

$$= 0.2 - 0$$

$$\therefore V_{DS} - V_{GS} = 0.2 \rightarrow \textcircled{1}$$

$$\therefore V_{GS} = V_{ov} + V_t$$

$$\therefore V_{GS} - 0.2 = V_{ov} + 0.5$$

$$\therefore V_{GS} - V_{ov} = 0.7 \quad \therefore V_{DS} > V_{ov} \quad \therefore \text{saturation}$$

$$R_D = \frac{V_{DD} - V_D}{I_D} = \frac{1 - 0.2}{0.2} = 4 \text{ k}\Omega$$

$$\therefore I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} V_{ov}^2$$

$$\therefore 0.2 \times 10^{-3} = \frac{1}{2} (400 \times 10^{-6}) \left( \frac{15}{0.5} \right) V_{ov}^2$$

$$\therefore V_{ov}^2 = \frac{1}{30} \rightarrow \therefore V_{ov} = 0.18 \text{ V}$$

$$\therefore V_{ov} = V_{GS} - V_t$$

$$\therefore V_{GS} = 0.18 + 0.5$$

$$\therefore V_{GS} = 0.68 \text{ V}$$

$$\therefore V_{GS} - V_S = 0.68 \rightarrow \therefore V_S = -0.68 \text{ V}$$

$$R_S = \frac{V_S - V_{SS}}{I_D} = \frac{-0.68 + 1}{0.2} = 1.6 \text{ k}\Omega$$

$$R_S = \frac{V_S - V_{SS}}{I_D} = \frac{-0.68 + (-1)}{0.2 \text{ mA}}$$

$$R_S = 1.6 \text{ k}\Omega$$

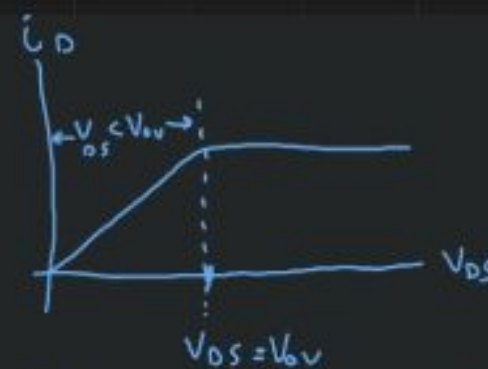
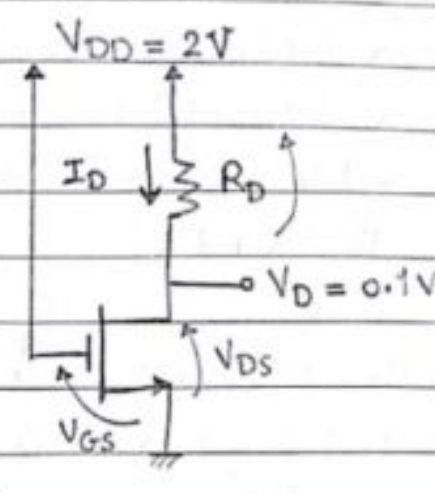
EX (5.5)

$$V_D = 0.1 \text{ V}$$

$$V_t = 0.5 \text{ V} \quad K_n \left( \frac{W}{L} \right) = 2 \text{ mA/V}^2$$

find: effective resistance between

drain and source.  $R_{DS} = \frac{V_{DS}}{I_D}$



$$r_{os} = \frac{1}{\text{slope}} = \frac{1}{K_n V_{ov}} = \frac{1}{(2 \times 10^{-3}) (1.5)} = 333.3 \text{ } \Omega$$

1) determine operation mode

$$V_G = 2 \text{ V}, V_S = 0 \quad \therefore V_{GS} = 2 \text{ V}$$

$$V_D = 0.1 \text{ V}, V_S = 0 \quad \therefore V_{DS} = 0.1 \text{ V}$$

$$V_{ov} = V_{GS} - V_t = 2 - 0.5 \rightarrow \therefore V_{ov} = 1.5 \text{ V} \quad \therefore V_{DS} < V_{ov} \rightarrow \therefore \text{triode mode}$$

$$I_D = K_n \frac{W}{L} \left[ V_{ov} V_{DS} - \frac{1}{2} V_{DS}^2 \right]$$

$$\therefore I_D = 2 \times 10^{-3} \left[ (1.5) (0.1) - \frac{1}{2} (0.1)^2 \right]$$

$$\therefore I_D = 0.29 \text{ mA}$$

$$R_{eff} = \frac{V_{DS}}{I_D} = \frac{0.1}{0.29} = r_{os}$$

$$\therefore R_{eff} = 345 \text{ } \Omega = r_{os}$$

$$R_D = \frac{2 - 0.1}{0.29}$$

$$\therefore R_D = 6.5 \text{ k}\Omega$$



### Prob (5.29)

for a Particular MOSFET operating in the Saturation Region at a Constant  $V_{GS}$ ,  $I_D$  is found to be 200  $\mu A$  for  $V_{DS} = 1 V$  and 205  $\mu A$  for  $V_{DS} = 1.5 V$  find the values of  $r_o$ ,  $V_A$  and  $\lambda$



$I_D$	200	205
$V_{DS}$	1	1.5

$\mu A$

- in saturation,  $I_D$  is constant with changing  $V_{DS}$ , but here  $I_D$  increased.  
- So there's  $\lambda$  in this example.

$$-r_o = \frac{1}{\text{slope}} = \frac{1}{K_n V_{ov}} = \frac{\Delta V_{DS}}{\Delta I_D} = \frac{1.5 - 1}{205 - 200} \rightarrow \therefore r_o = 100 \text{ k}\Omega$$

•  $V_{DS} = V_A$  when  $I_D = 0$

$$\therefore I_D = \frac{1}{2} K_n (V_{GS} - V_t)^2 (1 + \lambda V_{DS}) = 0$$

$$\therefore 1 + \lambda V_{DS} = 0 \rightarrow \therefore V_{DS} = \frac{-1}{\lambda} = -V_A$$

$$\therefore r_o = \frac{V_{DS}}{I_D}, \therefore r_o = \frac{1}{\lambda I_D} = \frac{V_A}{I_D}$$

$$\therefore 100 \text{ k} = \frac{V_A}{200 \mu A}, \therefore V_A = 20 \text{ V}, \therefore \lambda = \frac{1}{20}$$

or • From line eq we get the intersection with x-axis " $V_{DS}$ " and this point will be  $V_A$

$$\therefore \frac{I_D - 200}{V_{DS} - 1} = \frac{205 - 200}{1.5 - 1}$$

$$\therefore V_{DS} - 1 = (I_D - 200) \frac{1.5 - 1}{205 - 200}$$

$$\therefore V_{DS} = -V_A = 19 \text{ V}$$

$$\therefore \lambda = \frac{1}{19}$$