

Due Sunday, 17 May 2020, by 11:59pm to CCLE.

100 points total.

This homework covers questions relate to Fourier series and Fourier transform.

1. (18 points) **Fourier Series**

- (a) (7 points) When the periodic signal  $f(t)$  is real, you have seen in class some properties of symmetry for the Fourier series coefficients of  $f(t)$  (see the Lecture 11 slide titled: Fourier Series Properties: Fourier Symmetry (cont.)). How do these properties of symmetry change when  $f(t)$  is imaginary (with no real component)?

**Solution:** Since  $f(t)$  is pure imaginary, it can equivalently written as  $f(t) = jg(t)$ , where  $g(t)$  is real.

$$\begin{aligned} c_k &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} f(t) \left[ \cos\left(\frac{2\pi k}{T_0}t\right) - j \sin\left(\frac{2\pi k}{T_0}t\right) \right] dt \\ &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} jg(t) \left[ \cos\left(\frac{2\pi k}{T_0}t\right) - j \sin\left(\frac{2\pi k}{T_0}t\right) \right] dt \\ &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} jg(t) \cos\left(\frac{2\pi k}{T_0}t\right) + g(t) \sin\left(\frac{2\pi k}{T_0}t\right) dt \\ &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} g(t) \sin\left(\frac{2\pi k}{T_0}t\right) dt + j \frac{1}{T_0} \int_{t_0}^{t_0+T_0} g(t) \cos\left(\frac{2\pi k}{T_0}t\right) dt \end{aligned}$$

Now, because  $g(t)$  is real:

$$\begin{aligned} \text{Re}(c_k) &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} g(t) \sin\left(\frac{2\pi k}{T_0}t\right) dt \\ \text{Im}(c_k) &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} g(t) \cos\left(\frac{2\pi k}{T_0}t\right) dt \end{aligned}$$

Therefore,

$$\begin{aligned} \text{Re}(c_k) &= -\text{Re}(c_{-k}) \\ \text{Im}(c_k) &= \text{Im}(c_{-k}) \\ c_k^* &= -c_{-k} \\ |c_k| &= |c_{-k}| \\ \angle c_k &= -\angle c_{-k}^* \pm \pi \end{aligned}$$

- (b) (7 points) A *real* and *odd* signal  $x(t)$  has the following properties:

- it is a periodic signal with period 1 s;

- it has one positive frequency component (positive frequency component meaning  $c_k$  with  $k > 0$ );
- it has a power of 9 (hint: consider Parseval's relation. The power of the signal in the time domain is the same as the sum of the powers of its frequency components).

What is  $x(t)$ ?

**Solution:**

The signal has a fundamental period of 1 s, and its frequency is:  $\omega_0 = 2\pi$ . Therefore, it can be written as:

$$x(t) = \sum_{k=-\infty}^{+\infty} c_k e^{jk\omega_0 t} = \sum_{k=-\infty}^{+\infty} c_k e^{j2\pi k t}$$

Since  $x(t)$  is real and odd, the coefficients  $c_k$ 's are imaginary and odd. Since  $x(t)$  has one positive frequency component, it can be reduced to the following:

$$x(t) = c_{-1} e^{-j2\pi t} + c_1 e^{j2\pi t} = c_1 (e^{j2\pi t} - e^{-j2\pi t}) = \dots$$

We know  $c_1$  to be purely imaginary, so let  $c_1 = jb$ , where  $b$  is a real coefficient.

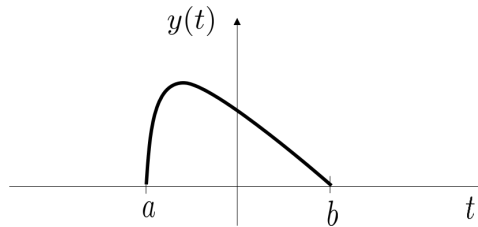
$$\dots = -\frac{2b}{2j} (e^{j2\pi t} - e^{-j2\pi t}) = -2b \sin(2\pi t)$$

Using Parseval's relation, we have:

$$2b^2 = 9 \implies b = \pm 3\sqrt{2}/2$$

Therefore,  $x(t) = 3\sqrt{2} \sin(2\pi t)$  or  $x(t) = -3\sqrt{2} \sin(2\pi t)$

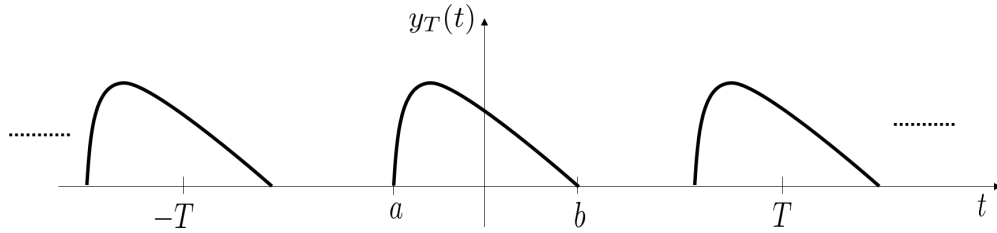
- (c) (4 points) Consider the signal  $y(t)$  shown below and let  $Y(j\omega)$  denote its Fourier transform.



Let  $y_T(t)$  denote its periodic extension:

How can the Fourier series coefficients of  $y_T(t)$  can be obtained from the Fourier transform  $Y(j\omega)$  of  $y(t)$ ? (Note that the figures given in this problem are for illustrative purposes, the question is for any arbitrary  $y(t)$ ).

**Solution:**



The Fourier transform of  $y(t)$  is given by:

$$Y(j\omega) = \int_{-\infty}^{\infty} y(t)e^{-j\omega t} dt = \int_a^b y(t)e^{-j\omega t} dt \quad (1)$$

The coefficients of the Fourier series for  $y_T(t)$  are given by:

$$Y_k = \frac{1}{T} \int_a^b y(t)e^{-j(2\pi k/T)t} dt \quad (2)$$

for any integer  $k$ . Therefore, by comparing (2) to (1), we conclude:

$$Y_k = \frac{1}{T} Y(j\omega) \Big|_{\omega = \frac{2\pi k}{T}}$$

## 2. (32 points) Symmetry properties of Fourier transform

(a) (16 points) Determine which of the signals, whose Fourier transforms are depicted in Fig. 1, satisfy each of the following:

i.  $x(t)$  is even

**Solution:** If  $x(t)$  is even, then its Fourier transform should be even. Since  $X(j\omega)$  in (a), (d) and (e) are even, signals (a), (d) and (e) are all even in the time domain.

ii.  $x(t)$  is odd

**Solution:** If  $x(t)$  is odd, then its Fourier transform should be odd. Since  $X(j\omega)$  in (f) is odd, signal in (f) is odd in the time domain.

iii.  $x(t)$  is real

**Solution:** If  $x(t)$  is real, then  $X(j\omega)$  is Hermitian, i.e.,  $X(-j\omega) = X^*(j\omega)$ . This means the real part of  $X(j\omega)$  is even and the imaginary part of  $X(j\omega)$  is odd. It also means that the magnitude of  $X(j\omega)$  is even and the phase of  $X(j\omega)$  is odd. Since  $X(j\omega)$  in (c) and (e) are both Hermitian, signals (c) and (e) are real in the time domain.

iv.  $x(t)$  is complex (neither real, nor pure imaginary)

**Solution:** For  $x(t)$  to be complex (not real neither pure imaginary),  $X(j\omega)$  should not be Hermitian or anti-Hermitian. We know from the previous part that  $X(j\omega)$  in (c) and (e) are Hermitian. Signals in (d) and (f) are anti-Hermitian. Therefore, signals in (a) and (b) are both complex in the time domain.

v.  $x(t)$  is real and even

**Solution:** If  $x(t)$  is real and even, then  $X(j\omega)$  is real and even. Therefore, it is (e).

vi.  $x(t)$  is imaginary and odd

**Solution:** If  $x(t)$  is imaginary and odd, then  $X(j\omega)$  is real and odd. Therefore, it is (f).

vii.  $x(t)$  is imaginary and even

**Solution:** If  $x(t)$  is imaginary and even, then  $X(j\omega)$  is imaginary and even. Therefore, it is (d).

viii. There exists a non-zero  $\omega_0$  such that  $e^{j\omega_0 t}x(t)$  is real and even

**Solution:** If  $e^{j\omega_0 t}x(t)$  is real and even, then  $X(j(\omega - \omega_0))$  is real and even. In (b), it is symmetric if we shift  $X(j\omega)$  to the left by 2 (i.e.,  $\omega_0 = -2$ ), giving a real and even Fourier signal. Elsewhere it will not be even if we shift the Fourier signal by a non-zero increment.

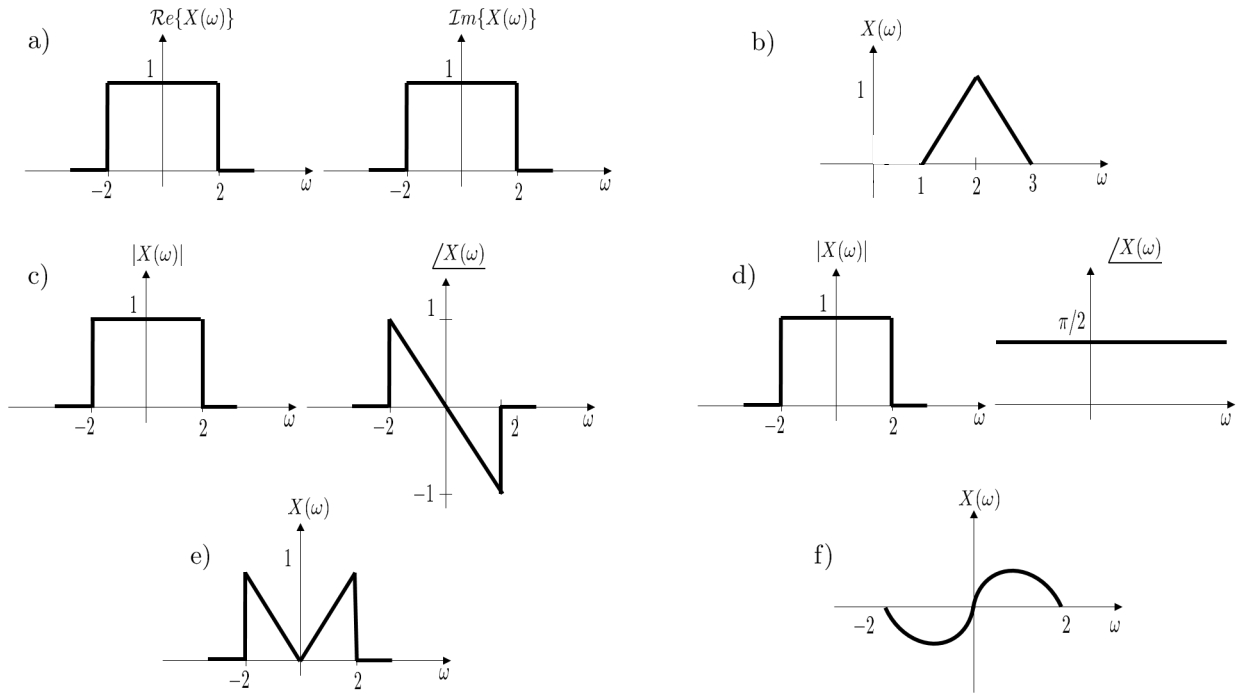


Figure 1: P2.a

(b) (8 points) Using the properties of the Fourier transform, determine whether the assertions are true or false.

i. The convolution of a real and even signal and a real and odd signal, is odd.

**Solution:** Let  $f(t)$  be a real and even signal, and  $g(t)$  be a real and odd signal. Then  $F(j\omega)$  is real and even, and  $G(j\omega)$  is imaginary and odd. The convolution  $h(t) = (f * g)(t)$  has the Fourier transform

$$H(j\omega) = F(j\omega)G(j\omega)$$

If  $F(j\omega)$  is real and even, and  $G(j\omega)$  is imaginary and odd, then  $H(j\omega)$  is imaginary and odd, and  $h(t)$  is real and odd. The assertion is true.

- ii. The convolution of a signal and the same signal reversed is an even signal.

**Solution:** Let  $f(t)$  be a signal, and  $f_R(t) = f(-t)$ . Let  $h(t) = (f * f_R)(t)$ . Then

$$H(j\omega) = F(j\omega)F_R(j\omega) = F(j\omega)F(-j\omega)$$

which is even (replacing  $\omega$  by  $-\omega$  results in the same expression). As the Fourier spectrum is even, by symmetry the output of the convolutions is even. The assertion is true.

- (c) (8 points) Show the following statements:

- i. If  $x(t) = x^*(-t)$ , then  $X(j\omega)$  is real.

**Solution:** When we reverse this,

$$\begin{aligned} X^*(j\omega) &= \left[ \int_{-\infty}^{+\infty} x(t)e^{-j\omega t} dt \right]^* \\ &= \int_{-\infty}^{+\infty} [x(t)e^{-j\omega t}]^* dt \\ &= \int_{-\infty}^{+\infty} x^*(t)e^{j\omega t} dt \\ &= \int_{-\infty}^{+\infty} x^*(-\tau)e^{-j\omega\tau} d\tau, \text{ here we did the variable change } \tau = -t \\ &= \int_{-\infty}^{+\infty} x(\tau)e^{-j\omega\tau} d\tau, \text{ here we used the fact that } x(\tau) = x^*(-\tau) \\ &= X(j\omega), \end{aligned}$$

Since  $X^*(j\omega) = X(j\omega)$ , we conclude that  $X(j\omega)$  is real.

- ii. If  $x(t)$  is a real signal with  $X(j\omega)$  its Fourier transform, then the Fourier transforms  $X_e(j\omega)$  and  $X_o(j\omega)$  of the even and odd components of  $x(t)$  satisfy the following:

$$X_e(j\omega) = \text{Re}\{X(j\omega)\}$$

and

$$X_o(j\omega) = j\text{Im}\{X(j\omega)\}$$

**Solution:**

Since  $x(t) = x_e(t) + x_o(t)$ , the Fourier transform of  $x(t)$  is given by:

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt = \underbrace{\int_{-\infty}^{\infty} x_e(t)e^{-j\omega t} dt}_{X_e(j\omega)} + \underbrace{\int_{-\infty}^{\infty} x_o(t)e^{-j\omega t} dt}_{X_o(j\omega)}$$

Now using Euler,

$$X_e(j\omega) = \int_{-\infty}^{+\infty} x_e(t)(\cos(\omega t) - j \sin(\omega t))dt = \int_{-\infty}^{+\infty} x_e(t) \cos(\omega t)dt$$

$$X_o(j\omega) = \int_{-\infty}^{+\infty} x_o(t)(\cos(\omega t) - j \sin(\omega t))dt = -j \int_{-\infty}^{+\infty} x_o(t) \sin(\omega t)dt$$

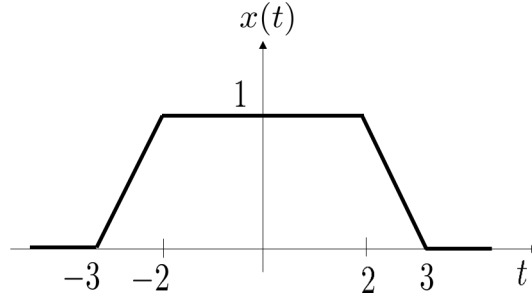
Since  $x(t)$  is real,  $x_e(t)$  and  $x_o(t)$  are both real. Therefore,  $X_e(j\omega)$  is real and  $X_o(j\omega)$  is pure imaginary. Therefore,

$$\mathcal{Re}\{X(j\omega)\} = \int_{-\infty}^{+\infty} x_e(t) \cos(\omega t)dt = X_e(j\omega)$$

$$\mathcal{Im}\{X(j\omega)\} = - \int_{-\infty}^{+\infty} x_o(t) \sin(\omega t)dt = -jX_o(j\omega)$$

### 3. (15 points) **Fourier transform properties**

Let  $X(j\omega)$  denote the Fourier transform of the signal  $x(t)$  sketched below:



Evaluate the following quantities without explicitly finding  $X(j\omega)$ :

(a)  $\int_0^{\infty} X(j\omega) d\omega$

Hint: Consider the properties of  $x(t)$ .

#### **Solution:**

Intuitively, when we see the integral on  $X(j\omega)$ , we will recall the inverse Fourier transform:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$$

Now that we do not have the exponential term in the integral, it comes to mind that when we let  $t = 0$ , the exponential term becomes a constant value 1:

$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) d\omega$$

Also, when we observe the signal  $x(t)$ , we can find that  $x(t)$  is even and real. According to what we have learned in lectures,  $X(j\omega)$  should also be even and real.

Using this property, we can reformulate the previous equality to:

$$x(0) = \frac{1}{\pi} \int_0^{+\infty} X(j\omega) d\omega$$

Therefore, we find that the integral in the question can be calculated as:

$$\int_0^{\infty} X(j\omega) d\omega = x(0)\pi = \pi$$

(b)  $X(j\omega)|_{\omega=0}$

**Solution:**

Since

$$X(j\omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t} dt$$

we then have:

$$X(j\omega)|_{\omega=0} = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t} dt|_{\omega=0} = \int_{-\infty}^{+\infty} x(t) dt = \frac{1}{2} + 4 + \frac{1}{2} = 5$$

(c)  $\angle X(j\omega)$

**Solution:** The Fourier transform of a real and even function is real and even. Therefore the phase of  $X(j\omega)$  is either 0 or  $\pi$ . It is zero when  $X(j\omega) \geq 0$  and it is  $\pi$  when  $X(j\omega) < 0$ .

(d)  $\int_{-\infty}^{\infty} e^{-j\omega} X(j\omega) d\omega$

**Solution:**

$$\begin{aligned} x(t) &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega)e^{j\omega t} d\omega \\ x(t)|_{t=-1} &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega)e^{j\omega t} d\omega|_{t=-1} \\ &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega)e^{-j\omega} d\omega \end{aligned}$$

Therefore,

$$\int_{-\infty}^{+\infty} X(j\omega)e^{-j\omega} d\omega = 2\pi x(-1) = 2\pi$$

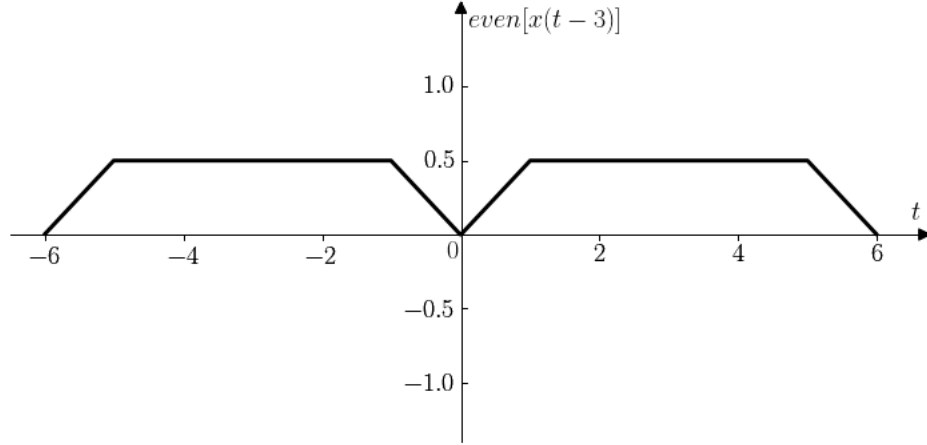
- (e) Plot the inverse Fourier transform of  $\mathcal{R}e\{e^{-3j\omega}X(j\omega)\}$

Hint: Consider the 'even and odd' properties of the Fourier transform

**Solution:** Let  $Y(j\omega) = e^{-3j\omega}X(j\omega)$ , then  $y(t) = x(t-3)$ . Since  $y(t)$  is real,

$$\mathcal{R}e\{e^{-3j\omega}X(j\omega)\} = \mathcal{R}e\{Y(j\omega)\} = Y_e(j\omega)$$

where  $Y_e(j\omega)$  is the Fourier transform of the even component of  $y(t)$ . Therefore, the inverse Fourier transform of  $\mathcal{R}e\{e^{-3j\omega}X(j\omega)\}$  is the even component of  $x(t-3)$ .



#### 4. (35 points) **Fourier transform and its inverse**

- (a) (18 points) Find the Fourier transform of each of the signals given below:

Hint: you may use Fourier Transforms derived in class.

i.  $x_1(t) = \begin{cases} 1 + \cos(\pi t), & |t| < 1 \\ 0, & \text{otherwise} \end{cases}$

**Solution:**

We can compute  $X_1(j\omega)$  by applying the definition of Fourier transform:

$$\begin{aligned} X_1(j\omega) &= \int_{-1}^1 [1 + \cos(\pi t)] e^{-j\omega t} dt \\ &= -\frac{1}{j\omega} [e^{-j\omega} - e^{j\omega}] + \frac{1}{j2(\pi - \omega)} [e^{j(\pi - \omega)} - e^{-j(\pi - \omega)}] - \frac{1}{j2(\pi + \omega)} [e^{-j(\pi + \omega)} - e^{j(\pi + \omega)}] \\ &= \frac{2 \sin(\omega)}{\omega} + \frac{\sin(\pi - \omega)}{\pi - \omega} + \frac{\sin(\pi + \omega)}{\pi + \omega} = 2\text{sinc}\left(\frac{\omega}{\pi}\right) + \text{sinc}\left(\frac{\omega - \pi}{\pi}\right) + \text{sinc}\left(\frac{\omega + \pi}{\pi}\right) \end{aligned}$$

Or we can see that:

$$x_2(t) = \text{rect}\left(\frac{t}{2}\right) + \cos(\pi t) \text{rect}\left(\frac{t}{2}\right)$$

so that,

$$X_2(j\omega) = 2\text{sinc}\left(\frac{\omega}{\pi}\right) + \text{sinc}\left(\frac{\omega - \pi}{\pi}\right) + \text{sinc}\left(\frac{\omega + \pi}{\pi}\right)$$



ii.  $x_2(t) = e^{(1+3j)t}u(-t+1)$

**Solution:**

We can write  $x_2(t)$  as follows:

$$x_2(t) = e^{j3t}e^t u(-t+1) = e^{j3t}e^1 e^{t-1} u(-(t-1))$$

We know:

$$\begin{aligned} e^{-t}u(t) &\longleftrightarrow \frac{1}{1+j\omega} \\ e^t u(-t) &\longleftrightarrow \frac{1}{1-j\omega} \\ e^{(t-1)}u(-(t-1)) &\longleftrightarrow \frac{e^{-j\omega}}{1-j\omega} \\ e^{j3t}e^{(t-1)}u(-(t-1)) &\longleftrightarrow \frac{e^{-j(\omega-3)}}{1-j(\omega-3)} \end{aligned}$$

Therefore,

$$X_2(j\omega) = \frac{e^{1-j(\omega-3)}}{1-j(\omega-3)}$$

iii.  $x_3(t) = 2te^{-2t}u(t)$

Hint: You can consider Fourier transform of the derivative and its dual.

**Solution:**

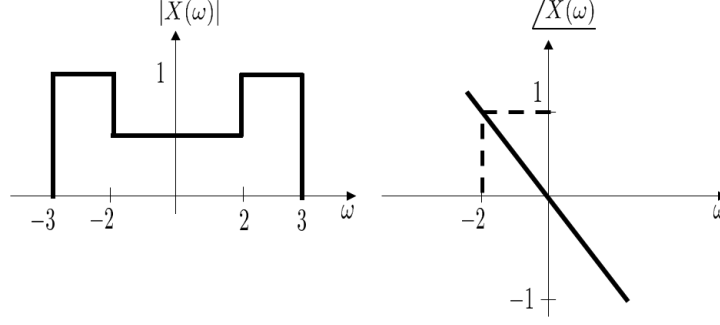
We know the following mapping regarding Fourier transform:

$$\begin{aligned} -jtf(t) &\longleftrightarrow F'(j\omega) \\ e^{-2t}u(t) &\longleftrightarrow \frac{1}{2+j\omega} \end{aligned}$$

Therefore, we can let  $f(t) = e^{-2t}u(t)$ . Then  $x_3(t) = -\frac{2}{j}(-jtf(t))$ . Correspondingly,

$$X_3(j\omega) = -\frac{2}{j}F'(j\omega) = -\frac{2}{j} \left( \frac{d}{d\omega} \frac{1}{2+j\omega} \right) = \frac{2}{(2+j\omega)^2}.$$

(b) (7 points) Find the inverse Fourier transform of the signal shown below:



**Solution:** We have:

$$\begin{aligned}
 x(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)| e^{j\angle X(\omega)} e^{j\omega t} d\omega \\
 &= \frac{1}{2\pi} \left( \int_{-3}^{-2} e^{-j\frac{1}{2}\omega} e^{j\omega t} d\omega + \int_{-2}^2 \frac{1}{2} e^{-j\frac{1}{2}\omega} e^{j\omega t} d\omega + \int_2^3 e^{-j\frac{1}{2}\omega} e^{j\omega t} d\omega \right) \\
 &= \frac{1}{2\pi} \left( \int_{-3}^{-2} e^{j(t-\frac{1}{2})\omega} d\omega + \int_{-2}^2 \frac{1}{2} e^{j(t-\frac{1}{2})\omega} d\omega + \int_2^3 e^{j(t-\frac{1}{2})\omega} d\omega \right) \\
 &= \frac{1}{2\pi} \left( \frac{e^{-j2(t-\frac{1}{2})} - e^{-j3(t-\frac{1}{2})}}{j(t-\frac{1}{2})} + \frac{e^{j2(t-\frac{1}{2})} - e^{-j2(t-\frac{1}{2})}}{j2(t-\frac{1}{2})} + \frac{e^{j3(t-\frac{1}{2})} - e^{j2(t-\frac{1}{2})}}{j(t-\frac{1}{2})} \right) \\
 &= \frac{1}{2\pi} \left( \frac{e^{-2j(t-\frac{1}{2})} - e^{2j(t-\frac{1}{2})}}{j2(t-\frac{1}{2})} + \frac{e^{j3(t-\frac{1}{2})} - e^{-j3(t-\frac{1}{2})}}{j(t-\frac{1}{2})} \right) \\
 &= \frac{1}{2\pi} \left( -\frac{\sin(2(t-\frac{1}{2}))}{(t-\frac{1}{2})} + \frac{2\sin(3(t-\frac{1}{2}))}{(t-\frac{1}{2})} \right)
 \end{aligned}$$

(c) (10 points) Two signals  $f_1(t)$  and  $f_2(t)$  are defined as

$$\begin{aligned}
 f_1(t) &= \text{sinc}(2t) \\
 f_2(t) &= \text{sinc}(t) \cos(3\pi t)
 \end{aligned}$$

Let the convolution of the two signals be

$$f(t) = (f_1 * f_2)(t)$$

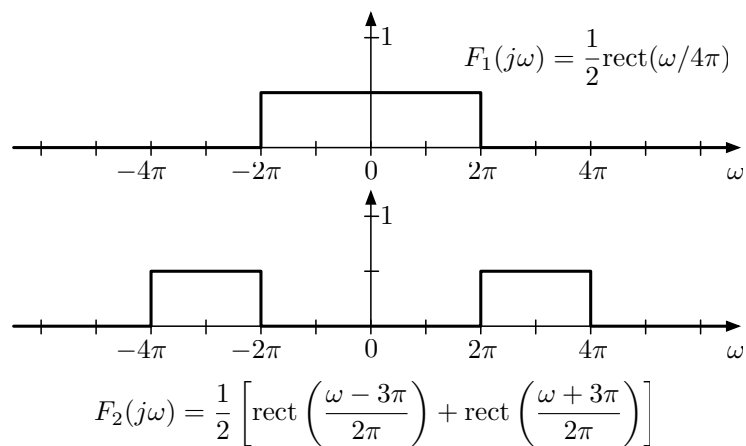
i. Find  $F(j\omega)$ , the Fourier transform of  $f(t)$ .

**Solution:** We know that:

$$\begin{aligned}
 f_1(t) = \text{sinc}(2t) &\longleftrightarrow F_1(j\omega) = \frac{1}{2} \text{rect}\left(\frac{\omega}{4\pi}\right) \\
 f_2(t) = \text{sinc}(t) \cos(3\pi t) &\longleftrightarrow F_2(j\omega) = \frac{1}{2} \left( \text{rect}\left(\frac{\omega - 3\pi}{2\pi}\right) + \text{rect}\left(\frac{\omega + 3\pi}{2\pi}\right) \right)
 \end{aligned}$$

We then have:

$$f(t) = (f_1 * f_2)(t) \longleftrightarrow F(j\omega) = F_1(j\omega) F_2(j\omega)$$



To see what the multiplication of  $F_1(j\omega)$  and  $F_2(j\omega)$  gives us, let us first plot them. We clear see that  $F_1(j\omega)$  and  $F_2(j\omega)$  do not overlap, therefore  $F(j\omega) = 0$ .

ii. Find  $f(t)$ .

**Solution:**

Since  $F(j\omega) = 0$ ,  $f(t)$  is then 0