

Aiman	2 <sup>nd</sup>	year	EE
Tom	2 <sup>nd</sup>	year	CS
Travis	2 <sup>nd</sup>	year	CS
Johnathon	2 <sup>nd</sup>	year	CS
Umar	2 <sup>nd</sup>	EE	

# EE102

## Lecture 4

# EE102 Announcements

---

- Syllabus link is ~~[tinyurl.com/ucla102](https://tinyurl.com/ucla102)~~ *tiny.cc/ucla102*
- CCLE difficulties, please email [help@seas.ucla.edu](mailto:help@seas.ucla.edu)
- ~~My office hour meeting minutes are sent out weekly~~
- **First Homework due this Friday**

**Slide Credits:** This lecture adapted from Prof. Jonathan Kao (UCLA) and recursive credits apply to Prof. Cabric (UCLA), Prof. Yu (Carnegie Mellon), Prof. Pauly (Stanford), and Prof. Fragouli (UCLA)

## Sidebar: Regarding Periodic Signals

The sum or product of a periodic signal is itself periodic if:

$$\left( \begin{array}{l} x_1: \text{period } T_1 \\ x_2: \text{period } T_2 \end{array} \right) \rightarrow \exists T, f(t+T) = f(t) \quad \forall t.$$

cyv  $z = x_1 + x_2 \dots$  when is  $z$  periodic and what's the period?

□ We have established that  $z$  is periodic if  $\exists T$  s.t.

$$T = k_1 T_1 = k_2 T_2.$$

$$\frac{T_1}{T_2} = \frac{k_2}{k_1}$$

cyv: See if you can find  $k_2, k_1$  that make  $z$  aperiodic

CyU: Come up with an  $x_1$ ,  $x_2$  s.t. sum is aperiodic.

$$\left(\frac{T_1}{T_2}\right) = c\pi$$

irrational number

$$\cos(\pi t) + \cos(2t) = z$$

aperiodic

# Signal Models

$$\omega = 2\pi f$$

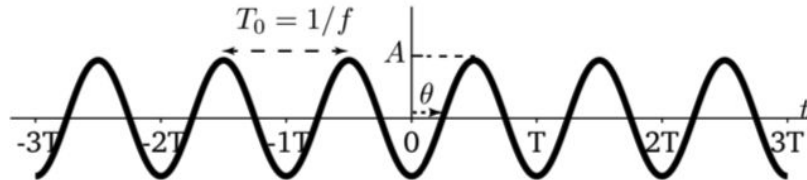
## Real sinusoids (cont.)

We illustrate a sinusoid signal below:

$$x(t) = A \cos(\omega t - \theta)$$

$$f = \frac{\omega}{2\pi}, \quad T_0 = \frac{1}{f} = \frac{2\pi}{\omega}$$

↓  
Fundamental  
frequency



# Signal Models

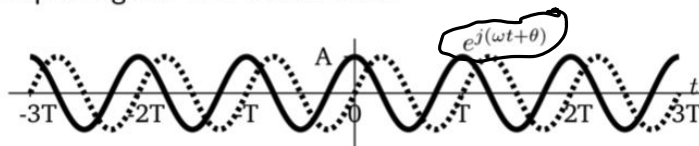
## Complex sinusoids

Ex 1

The complex sinusoid is given by:

$$\parallel Ae^{j(\omega t + \theta)} = A \cos(\omega t + \theta) + jA \sin(\omega t + \theta)$$

We draw complex signals with dotted lines.



The real part of the complex sinusoid (solid line) is:

$$\parallel \Re(Ae^{j(\omega t + \theta)}) = A \cos(\omega t + \theta)$$

The imaginary part of the complex sinusoid (dotted line) is:

$$\parallel \Im(Ae^{j(\omega t + \theta)}) = A \sin(\omega t + \theta)$$

# Signal Models

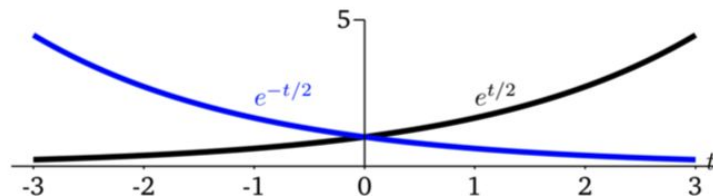
---

## Exponential

An exponential signal is given by

$$x(t) = e^{\sigma t}$$

- If  $\sigma > 0$ , this signal grows with increasing  $t$  (black signal in plot below). This is called exponential growth.
- If  $\sigma < 0$ , this signal decays with increasing  $t$  (blue signal in plot below). This is called exponential decay.



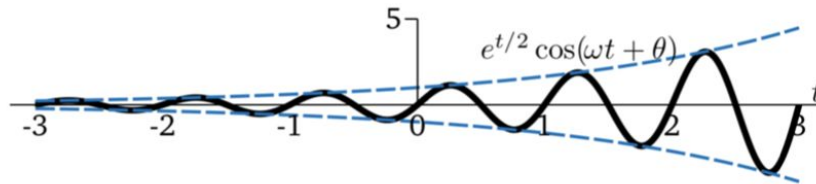
# Signal Models

## Damped or growing sinusoids

A damped or growing sinusoid is denoted

$$x(t) = e^{\sigma t} \cos(\omega t + \theta)$$

The sinusoid will grow exponentially if  $\sigma > 0$  and decay exponentially if  $\sigma < 0$ .



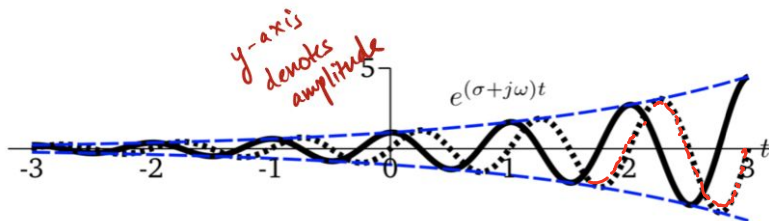


## Complex exponential

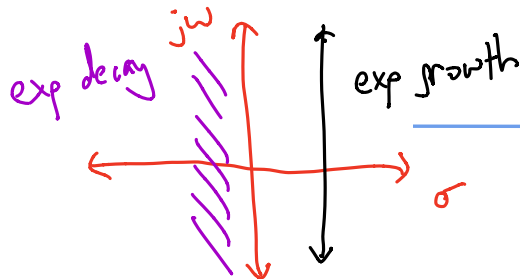
A complex sinusoid is denoted

$$x(t) = e^{(\sigma + j\omega)t}$$

It is a combination of the complex sinusoid and an exponential. All prior signals are special cases of the complex exponential signal.



It is helpful to think of  $\sigma$  and  $j\omega$  in the complex plane.  $\sigma$  is the x-axis and  $j\omega$  is the y-axis. Then complex exponentials in the left complex plane are decreasing signals and those in the right are increasing signals.

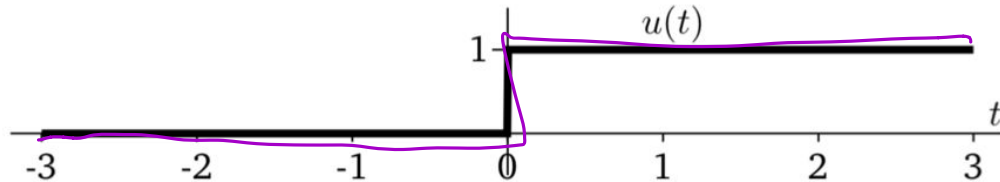


# Heaviside Step Function

The unit step function, denoted  $u(t)$  in this class, is given by

$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

It is also called the Heaviside step function. Drawn below:



cyu: Is this causal

Answer: Yes

LIDAR



# Unit Rectangle

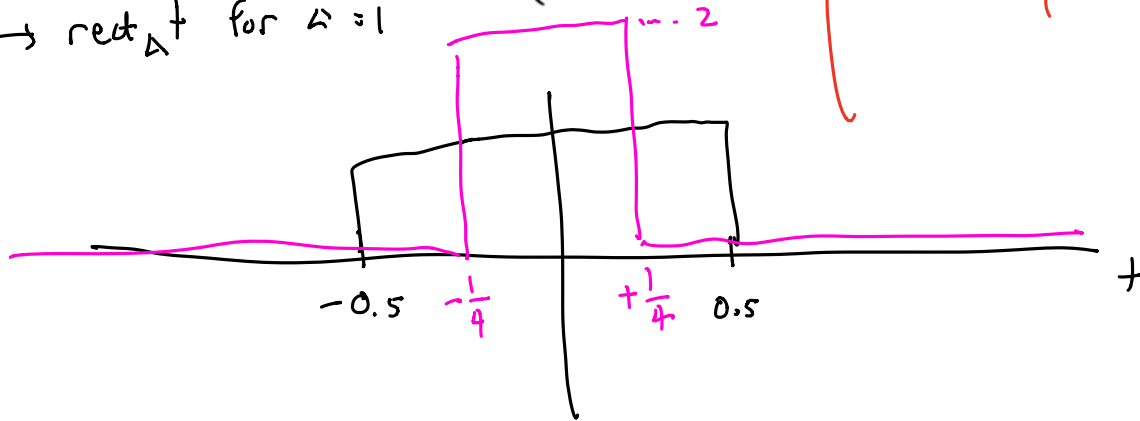
"Boxcar"

$\text{rect}(t)$  width of 1

$$\text{rect}(t) = \begin{cases} 1, & |t| \leq 1/2 \\ 0, & \text{else} \end{cases}$$

$\text{rect}(t) \rightarrow \text{rect}_{\Delta}(t)$  for  $\Delta = 1$

$$\text{rect}_{\Delta}(t) = \begin{cases} \frac{1}{\Delta}, & |t| \leq \frac{\Delta}{2} \\ 0, & \text{o.w.} \end{cases}$$

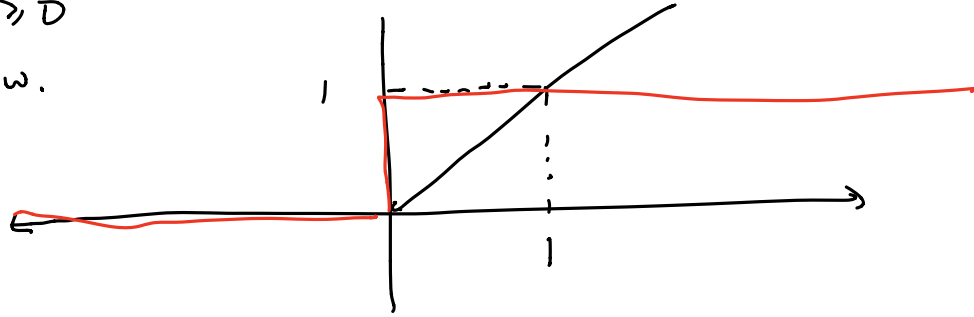


$\text{rect}_{\Delta} = \frac{1}{\Delta}(t)$  ?

# Unit Ramp Function

"ReLU"

$$r(t) = \begin{cases} t, & t \geq 0 \\ 0, & \text{o.w.} \end{cases}$$



cyu: How can I express  $r(t)$   
in terms of the building blocks I've learned?

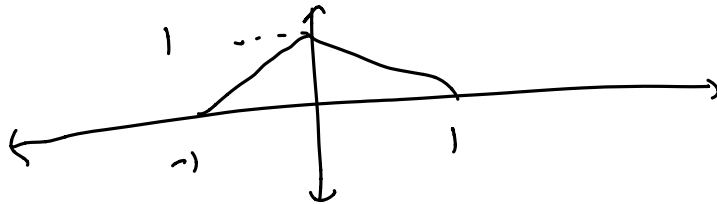
$$r(t) = \int_{-\infty}^t u(\tau) d\tau$$

$$r(t) = t u(t)$$

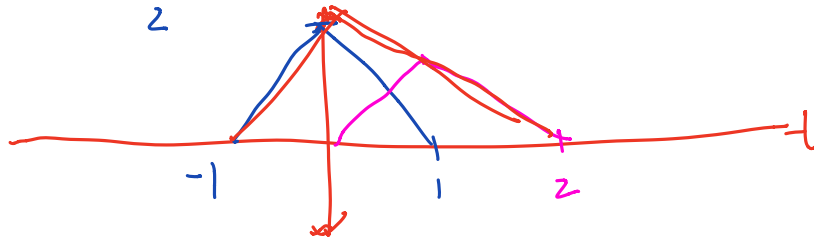
$\tau$ : dummy  
variable.  
integral bounds  
contain time

# Unit Triangle

$$\Delta(t) = \begin{cases} 1 - |t|, & |t| \leq 1 \\ 0, & \text{o.w.} \end{cases}$$



//  $z = 2\Delta(t) + \Delta(t-1)$



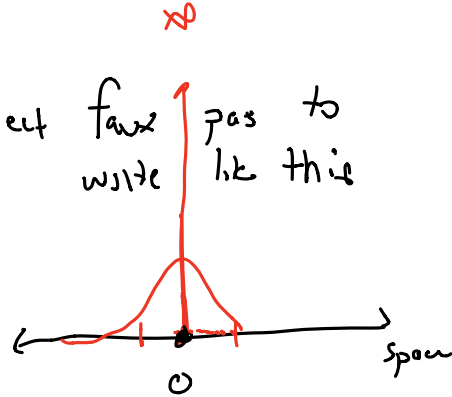
Dirac

# Impulse Function (Important!)

Impulse Function (Dirac Delta  $\delta(t)$ )

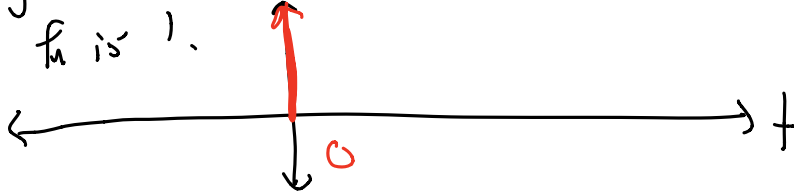
$$\delta(t) = \begin{cases} \infty & t=0 \\ 0 & \text{o.w.} \end{cases}$$

Incorrect way to write like this



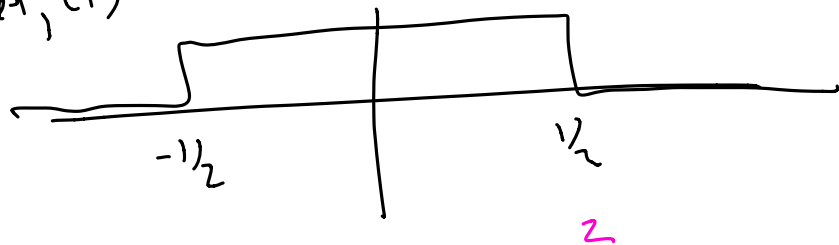
Properties of  $\delta(t)$

- ① It's very large (i.e. it approaches  $\infty$  when  $t=0$ )
- ② It's zero everywhere  $t \neq 0$
- ③ The area of this  $\delta(t)$  is 1.

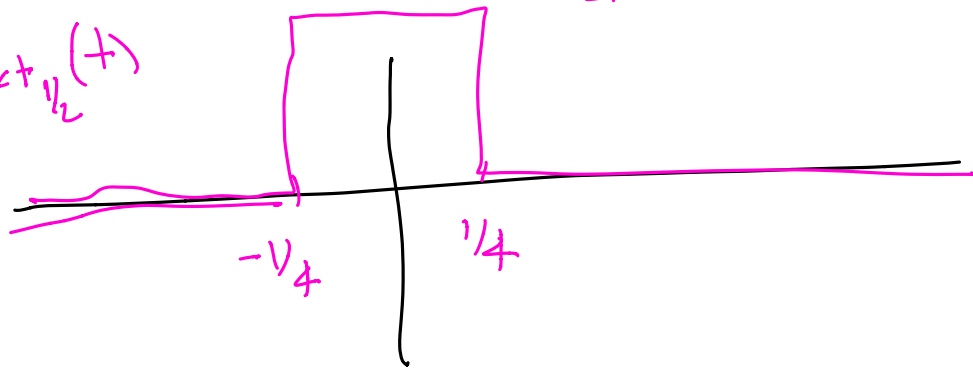


# Impulse Function (intuition)

$\text{rect}_\Delta(t)$



$\text{rect}_{1/2}(t)$



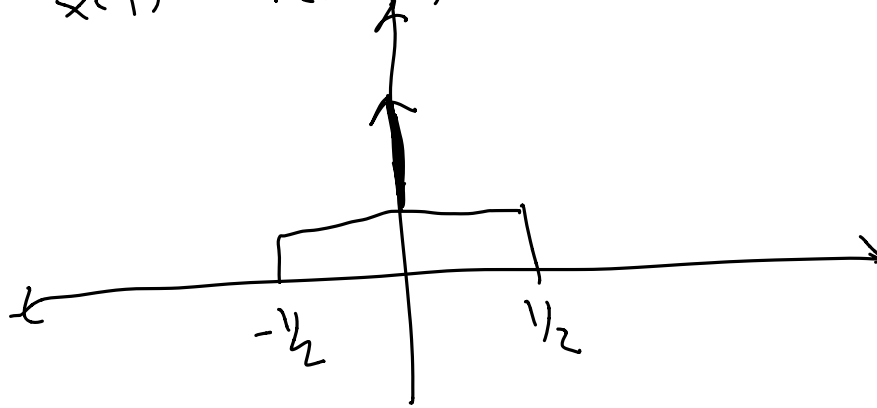
$f(t)$  as being  $\text{rect}_{\Delta \rightarrow 0}(t)$

$$\lim_{\Delta \rightarrow 0} \text{rect}_\Delta(t)$$

# Impulse Function Intuition

---

$$x(t) = \text{rect}\left(\frac{t}{\tau}\right) * \delta(t)$$

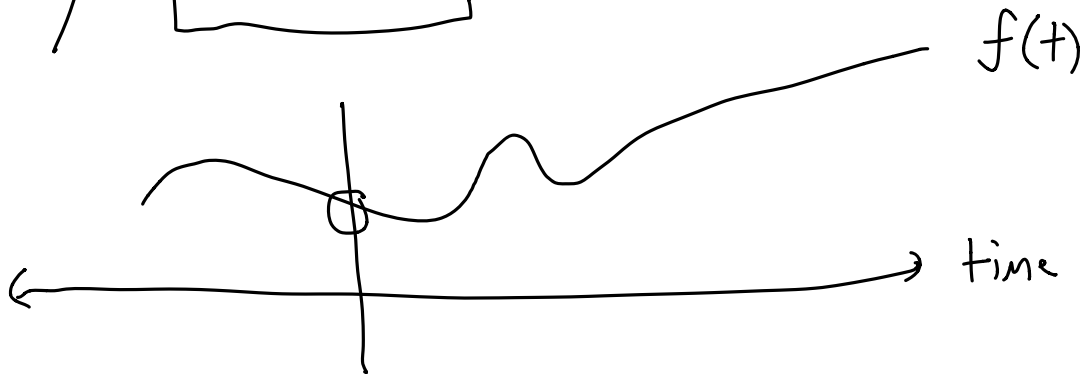




# Impulse Sampling Property

$$f(t=0)$$

$$f(t) \delta(t) = f(0) \delta(t)$$



$$f(t) \text{rect}_{0.001}(t) \approx f(0) \text{rect}_{0.001}(t)$$

# Impulse Sampling Property

$$\begin{aligned}\int_{-\infty}^{\infty} f(t) \delta(t) dt &= \int_{-\infty}^{\infty} f(t) \delta(t) dt \\ &= f(t) \int_{-\infty}^{\infty} \delta(t) dt \\ &= f(t)\end{aligned}$$

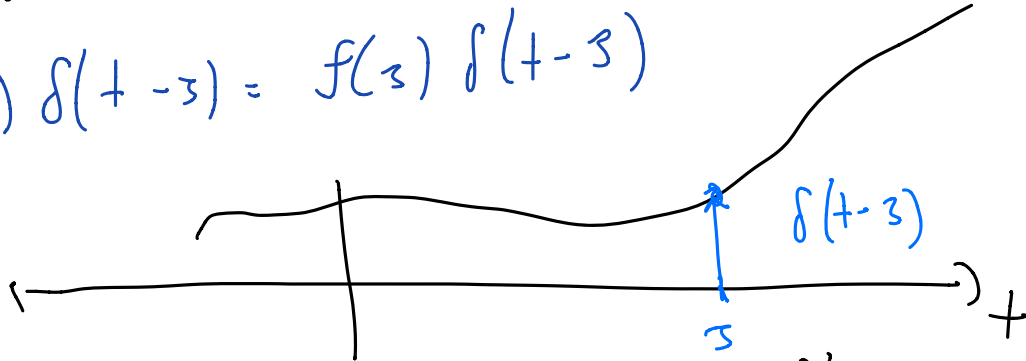
Dirac Delta has  
area of 1.

$\rightarrow f(t)$  to be continuous @  $t=0$ .

# Impulse Sifting Property

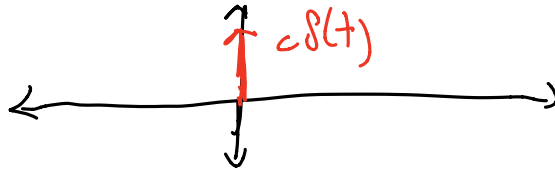
$$\int_{-\infty}^{\infty} f(t) \delta(t - \tau) dt = f(\tau)$$

$$f(t) \delta(t - \tau) = f(\tau) \delta(t - \tau)$$



//  $c \delta(t)$

$$\int_{-\infty}^{\infty} f(t) c \delta(t) dt = c f(0)$$



# Impulse Sifting Property

---

$$\int_{-\infty}^{\infty} f(t) dt = 1$$

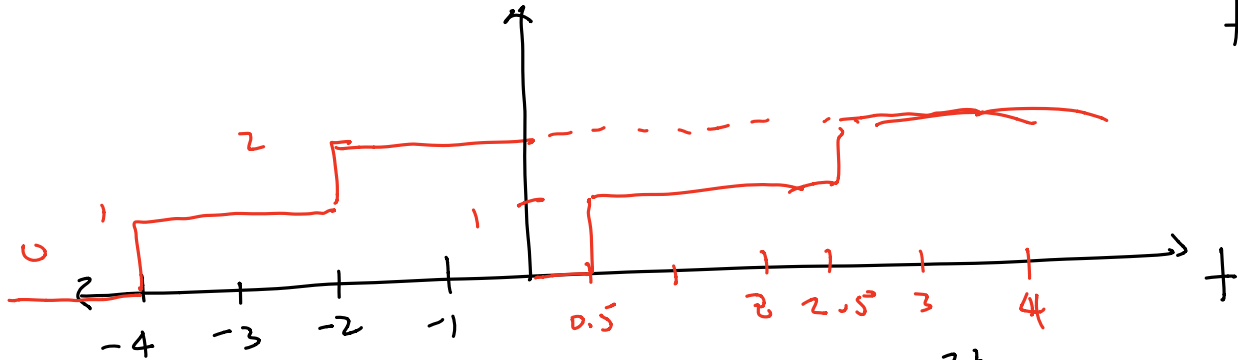
$$\int_{-\infty}^{0^-} f(t) dt = 0$$

$$\int_{-\infty}^{0^+} f(t) dt = 1$$

# CYU: Calculate

$$\int_{-2}^{3+} f(t) [1 + \delta(t+1) - 3\delta(t-1) + 2\delta(t+3)] dt$$

$$t = -3$$



$$\int_{-2}^{3+} f(t) dt + \underbrace{f(-1)}_2 - \underbrace{3f(1)}_3 + 2 \int_{-2}^{3+} f(t) \delta(t+3) dt$$

$\textcircled{7} + 2 - 3 + 0 = \textcircled{6}$

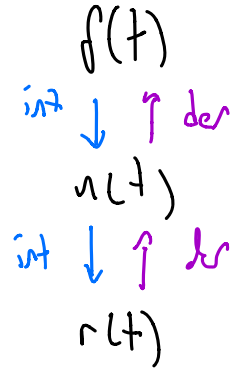
# CYU: Integral of an Impulse

$$\int_{-\infty}^t \delta(\tau) d\tau = u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & \text{o.w.} \end{cases}$$

Integral of Dirac  $\Rightarrow$  Step

Derivative of Step  $\Rightarrow$  Dirac

$$\frac{d u(t)}{dt} = \delta(t)$$



## CYU (Visual)

---

Suppose  $x(t) = 1 + \delta(t-1) - 2\delta(t-2)$  then what is  $y(t) = \int_0^t x(\tau) d\tau$

# Systems

---

A system transforms an input signal,  $x(t)$ , into an output signal,  $y(t)$ .

Systems, like signals, are similar to functions. However, they map a signal to another signal, so the term we might use is “operator”.

For EE102, we will not nitpick this distinction and focus on SISO systems.