

Time-Domain Analysis of LTIC Systems

The first step in analyzing any system is to model the system using a mathematical expression or a rule that satisfactorily approximates the dynamical behaviour of the system.

Suppose we have a linear, time-invariant, continuous-time (LTIC) system for which the input $x(t)$ and the output $y(t)$ are related by linear differential equations of the form:

$$\frac{d^n y}{dt^n} + a_{n-1} \frac{d^{n-1} y}{dt^{n-1}} + \dots + a_1 \frac{dy}{dt} + a_0 y(t) =$$

$$b_m \frac{d^m x}{dt^m} + b_{m-1} \frac{d^{m-1} x}{dt^{m-1}} + \dots + b_1 \frac{dx}{dt} + b_0 x(t)$$

where all the coefficients a_i and b_i are constants.

Using a compact notation D for $\frac{d}{dt}$, we can write:

$$(D^n + a_{n-1} D^{n-1} + \dots + a_1 D + a_0) y(t) =$$

$$(b_m D^m + b_{m-1} D^{m-1} + \dots + b_1 D + b_0) x(t)$$

substituting $Q(D) = D^n + a_{n-1}D^{n-1} + \dots + a_1D + a_0$

$$P(D) = b_mD^m + b_{m-1}D^{m-1} + \dots + b_1D + b_0$$

$$\Rightarrow Q(D)y(t) = P(D)x(t)$$

Note: Noise considerations restrict practical systems to have $m \leq n$.

Since the system is linear, the total response is

$$\text{Total response} = \begin{array}{c} \text{Zero-input} \\ \text{response} \end{array} + \begin{array}{c} \text{zero-state} \\ \text{response} \end{array}$$

↙
result of internal system conditions alone (such as energy storages, initial conditions)

↓
result of external input $x(t)$ when the system is in absence of all internal energy storages, initial conditions.

Zero-Input Response:

The zero-input response $y_0(t)$ is the solution to the equation:

$$Q(D)y_0(t) = 0$$

$$\text{or } (D^n + a_{n-1}D^{n-1} + \dots + a_1D + a_0)y_0(t) = 0$$

A linear combination of $y_0(t)$ and its n successive derivatives is zero for all $t \iff y_0(t)$ and all its n successive derivatives are of the same form. $e^{\lambda t}$ has this property.

Let's assume $y_0(t) = ce^{\lambda t}$

$$\Rightarrow c(\lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0)e^{\lambda t} = 0$$

$$\Rightarrow \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0 = 0$$

$$\Rightarrow Q(\lambda) = 0 = (\lambda - \lambda_1)(\lambda - \lambda_2)\dots(\lambda - \lambda_n)$$

Hence, λ has n solutions: $\lambda_1, \lambda_2, \dots, \lambda_n$

$$\Rightarrow Q(D)y_0(t) = 0 \text{ has } n \text{ solutions } c_1 e^{\lambda_1 t}, \dots, c_n e^{\lambda_n t}$$

Hence, a general solution for $y_0(t)$ is:

$$y_0(t) = c_1 e^{\lambda_1 t} + c_2 e^{\lambda_2 t} + \dots + c_n e^{\lambda_n t}$$

Note: The polynomial $Q(\lambda)$ is called the characteristic polynomial of the system and $Q(\lambda) = 0$ is called the characteristic equation.

$\lambda_1, \lambda_2, \dots, \lambda_n$ are called the characteristic roots or eigenvalues. The exponentials $e^{\lambda_i t}$ are called the characteristic modes of the system.

Repeated Roots: If our characteristic equation has repeated roots, then the form of the solution $y_0(t)$ is slightly modified.

For the differential equation $(D - \lambda)^n y_0(t) = 0$, the solution is:

$$y_0(t) = (c_1 + c_2 t + \dots + c_n t^{n-1}) e^{\lambda t}$$

Complex Roots: If our characteristic equation has complex roots, they must occur in conjugate pairs if the coefficients of $Q(\lambda)$ are to be real. Hence, if $\alpha + j\beta$ is a root, then $\alpha - j\beta$ must also be a root.

$$\Rightarrow y_0(t) = c_1 e^{(\alpha + j\beta)t} + c_2 e^{(\alpha - j\beta)t}$$

For a real system, $y_0(t)$ must also be real,

$\Rightarrow c_1$ and c_2 must be conjugates

$$\text{If } c_1 = \frac{c}{2} e^{j\theta} \Rightarrow c_2 = \frac{c}{2} e^{-j\theta}$$

$$\Rightarrow y_0(t) = c e^{\alpha t} \cos(\beta t + \theta)$$