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System Representation in the z-Domain

The System Function

• The system function H(z) is given by

$$H(z)\square \sim Z[h(n)] = \sum_{-\infty}^{\infty} h(n)z^{-n}; \quad R_{h-} < |z| < R_{h+}$$

• Using the convolution property of the z-transform, the output transform Y(z) is given by

$$Y(z) = H(z)X(z)$$
 : $ROC_y = ROC_h \cap ROC_x$

The System Function

• Therefore a linear and timeinvariant system can be represented in the *z*-domain by

$$X(z) \rightarrow \overline{H(z)} \rightarrow Y(z) = H(z)X(z)$$

System Function From The Difference Equation Representation

• LTI systems are described by a difference equation

$$y(n) + \sum_{k=1}^{N} a_k y(n-k) = \sum_{\ell=1}^{M} b_{\ell} x(n-\ell)$$

System Function Representation

$$Y(z) + \sum_{k=1}^{N} a_k z^{-k} Y(z) = \sum_{\ell=0}^{M} b_{\ell} z^{-\ell} X(z)$$

$$H(z) \Box \frac{Y(z)}{X(z)} = \frac{\sum_{\ell=0}^{M} b_{\ell} z^{-\ell}}{1 + \sum_{k=1}^{N} a_{k} z^{-k}} = \frac{B(z)}{A(z)} = \frac{b_{o} z^{-M} \left(z^{M} + \ldots + \frac{b_{M}}{b_{o}}\right)}{z^{-N} \left(z^{N} + \ldots + a_{N}\right)}$$

System Function From The Difference Equation Representation

After factorization

$$H(z) = b_o z^{N-M} \frac{\prod_{\ell=1}^{N} (z - z_{\ell})}{\prod_{k=1}^{N} (z - p_k)}$$

- z_{ℓ} are the system zeros
- p_k are the system poles
- H(z) (an LTI system) can also be represented in the z-domain using a pole-zero plot

Transfer Function Representation

• A frequency response function or transfer function H(z) on the unit circle $z = e^{j\omega}$

$$H(e^{j\omega}) = b_o z e^{j(N-M)\omega} \frac{\prod_{\ell=1}^{N} (e^{j\omega} - z_{\ell})}{\prod_{k=1}^{N} (e^{j\omega} - p_k)}$$

- The factor $(e^{j\omega}-z_{\ell})$ can be interpreted as a *vector* in the complex *z*-plane from a zero *z* to the unit circle at $z=e^{j\omega}$
- The factor $(e^{j\omega} p_k)$ can be interpreted as a vector from a pole p_k to the unit circle at $z = e^{j\omega}$.

Transfer Function Representation

The magnitude response function

$$\left| H\left(e^{j\omega}\right) \right| = \left| b_o \right| \frac{\left| e^{j\omega} - z_1 \right| \dots \left| e^{j\omega} - z_M \right|}{\left| e^{j\omega} - p_1 \right| \dots \left| e^{j\omega} - p_N \right|}$$

• a product of the lengths of vectors from zeros to the unit circle divided by the lengths of vectors from poles to the unit circle and scaled by $|b_o|$.

Transfer Function Representation

The phase response function

$$\angle H(e^{j\omega}) = [0 \text{ or } \pi] + [(N-M)\omega] + \sum_{1}^{M} \angle (e^{j\omega} - z_k) - \sum_{1}^{N} \angle (e^{j\omega} - p_k)$$

• a sum of a constant factor, a linear-phase factor, and a nonlinear-phase factor (angles from the "zero vectors" *minus* the sum of angles from the "pole vectors").

Given a causal system

$$y(n) = 0.9y(n-1) + x(n)$$

- Determine H(z) and sketch its pole-zero plot.
- Plot $/H(e^{j\omega})/$ and $\angle H(e^{j\omega})$.
- Determine the impulse response h(n).

• The difference equation can be put in the form

$$y(n) - 0.9y(n - 1) = x(n)$$

$$H(z) = b_o z^{N-M} \frac{\prod_{\ell=1}^{N} (z - z_{\ell})}{\prod_{k=1}^{N} (z - p_k)}$$

$$H(z) = \frac{1}{1 - 0.9z^{-1}}; |z| > 0.9$$

• since the system is causal. There is one pole at 0.9 and one zero at the origin

• illustrate using the zplane function

$$>> b = [1, 0]; a = [1, -0.9]; zplane(b,a)$$

- determine the magnitude and phase of $H(ej\omega)$ using freqz function.
- take 100 points along the upper half of the unit circle

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\gg [H,w] = freqz(b,a,100); magH = abs(H); phaH = angle(H);
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- >> subplot(2,1,1);plot(w/pi,magH);grid
- >> xlabel('frequency in pi units'); ylabel('Magnitude');
- >> title('Magnitude Response')
- >> subplot(2,1,2);plot(w/pi,phaH/pi);grid
- >> xlabel('frequency in pi units'); ylabel('Phase in pi units');
- >> title('Phase Response')



• From the *z*-transform in Table

$$h(n) = Z^{-1} \left[\frac{1}{1 - 0.9z^{-1}}; \quad |z| > 0.9 \right]$$
$$= (0.9)^{n} u(n)$$





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