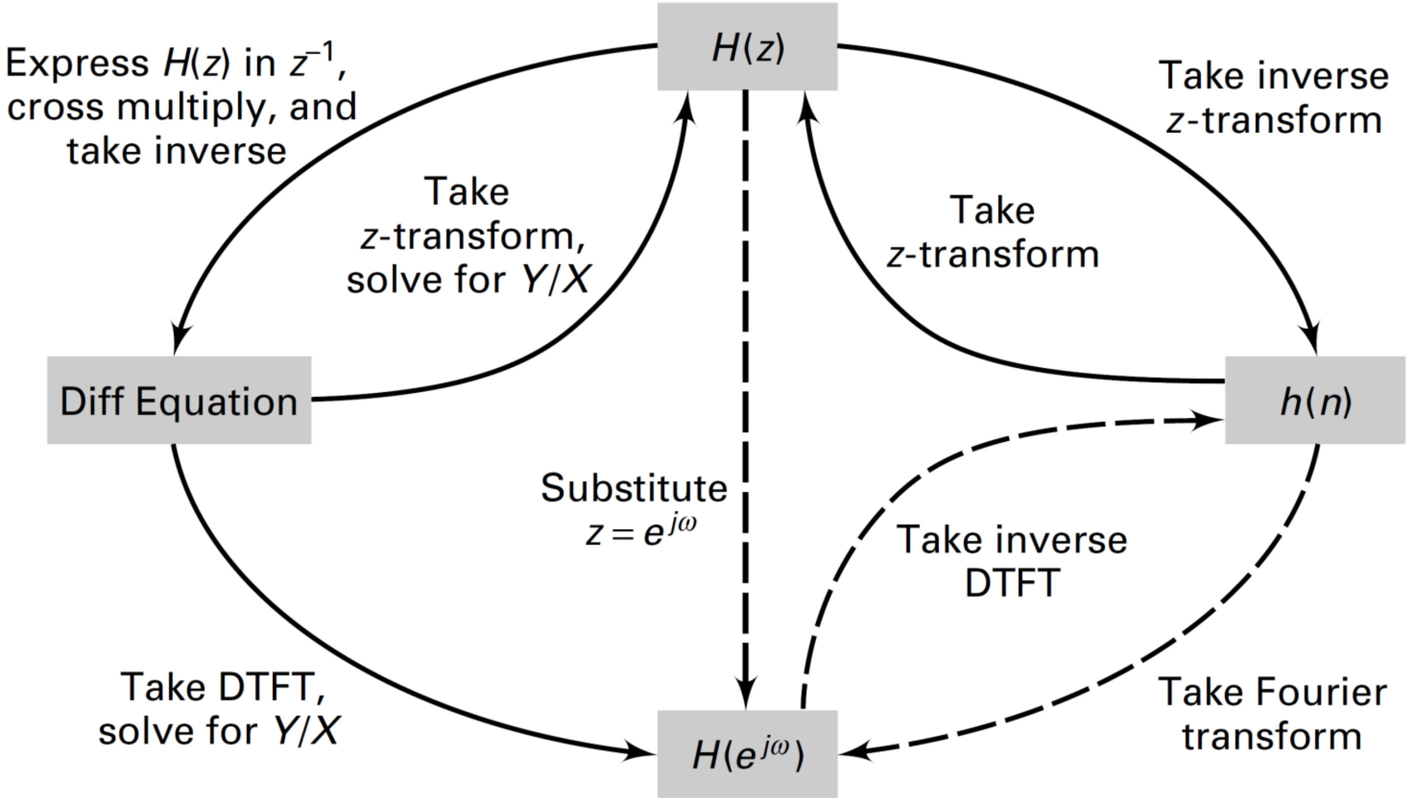
UNIVERSITAS INDONESIA
Veritas, Probitas, Iustitia

Pengolahan Sinyal Digital

Adhi Harmoko Saputro

System Representation in the z-Domain

Relationships between System Representations



z-Domain LTI Stability

- An LTI system is stable if and only if the unit circle is in the ROC of H(z).

z-Domain Causal LTI Stability

- A causal LTI system is stable if and only if the system function H(z) has all its poles inside the unit circle.

Example

- A causal LTI system is described by the following difference equation:

$$y(n) = 0.81y(n-2) + x(n) - x(n-2)$$

- Determine
  - the system function $H(z)$,
  - the unit impulse response $h(n)$,
  - the unit step response $v(n)$, that is, the response to the unit step $u(n)$, and
  - the frequency response function $H(e^{j\omega})$, and plot its magnitude and phase over $0 \leq \omega \leq \pi$.

Example

- Since the system is causal, the ROC will be outside a circle with radius equal to the largest pole magnitude.
- Taking the *z*-transform of both sides of the difference equation

$$H(z) = \frac{1 - z^{-2}}{1 - 0{,}81 z^{-2}} = \frac{1 - z^{-2}}{\left(1 + 0{,}9 z^{-1}\right)\left(1 - 0{,}9 z^{-1}\right)}, \quad |z| > 0{,}9$$

Example

- Using the MATLAB script for the partial fraction expansion

$$H(z)=\frac{1-z^{-2}}{1-0,81z^{-2}}=\frac{1-z^{-2}}{\left(1+0,9z^{-1}\right)\left(1-0,9z^{-1}\right)},\quad |z|>0,9$$

```
>> b = [1,0,-1]; a = [1,0,-0.81]; [R,p,C] = residuez(b,a);
R = -0.1173 -0.1173
p = -0.9000 0.9000
C = 1.2346
```

$$H(z)=1,2346-0,1173\frac{1}{1+0,9z^{-1}}-0,1173\frac{1}{1-0,9z^{-1}},\quad |z|>0,9$$

TABLE 4.1 *Some common z-transform pairs*

Sequence	Transform	ROC
$\delta(n)$	1	$\forall z$
$u(n)$	$\dfrac{1}{1-z^{-1}}$	$\|z\| > 1$
$-u(-n-1)$	$\dfrac{1}{1-z^{-1}}$	$\|z\| < 1$
$a^n u(n)$	$\dfrac{1}{1-az^{-1}}$	$\|z\| > \|a\|$
$-b^n u(-n-1)$	$\dfrac{1}{1-bz^{-1}}$	$\|z\| < \|b\|$
$[a^n \sin \omega_0 n]\, u(n)$	$\dfrac{(a \sin \omega_0) z^{-1}}{1-(2a \cos \omega_0) z^{-1} + a^2 z^{-2}}$	$\|z\| > \|a\|$
$[a^n \cos \omega_0 n]\, u(n)$	$\dfrac{1-(a \cos \omega_0) z^{-1}}{1-(2a \cos \omega_0) z^{-1} + a^2 z^{-2}}$	$\|z\| > \|a\|$
$na^n u(n)$	$\dfrac{az^{-1}}{(1-az^{-1})^2}$	$\|z\| > \|a\|$
$-nb^n u(-n-1)$	$\dfrac{bz^{-1}}{(1-bz^{-1})^2}$	$\|z\| < \|b\|$

Example

$$H(z) = 1,2346 - 0,1173\frac{1}{1+0,9z^{-1}} - 0,1173\frac{1}{1-0,9z^{-1}}, \quad |z| > 0,9$$

- from Table

$$h(n) = 1,2346\,\delta(n) - 0,1173\left\{1+(-1)^n\right\}(0,9)^n u(n)$$

Example

- From Table 4.1

$$Z\left[u(n)\right]=U(z)=\frac{1}{1-z^{-1}},\quad |z|>1$$

$$V(z)=H(z)U(z)$$

$$=\left[\frac{\left(1+z^{-1}\right)\left(1-z^{-1}\right)}{\left(1+0,9z^{-1}\right)\left(1-0,9z^{-1}\right)}\right]\left[\frac{1}{1-z^{-1}}\right],\quad |z|>0,9\cap|z|>1$$

$$=\frac{1+z^{-1}}{\left(1+0,9z^{-1}\right)\left(1-0,9z^{-1}\right)},\quad |z|>0,9$$

$$=1,0556\frac{1}{1-0,9z^{-1}}-0,0556\frac{1}{1+0,9z^{-1}},\quad |z|>0,9$$

Example

- Finally,

$$v(n) = \left[1{,}0556(0{,}9)^n - 0{,}556(-0{,}9)^n\right]u(n)$$

- There is a pole-zero cancellation at $z = 1$.
- This has two implications.
  - First, the ROC of $V(z)$ is still $\{|z| > 0.9\}$ and not $\{|z| > 0.9 \cap |z| > 1 = |z| > 1\}$
  - The step response $v(n)$ contains no steady-state term $u(n)$.

Example

- The frequency response function $H(e^{j\omega})$
- Substituting $z = e^{j\omega}$ in $H(z)$,

$$H\left(e^{j\omega}\right) = \frac{1-e^{-j2\omega}}{1-0,81e^{-j2\omega}}$$

Example

- MATLAB script to compute and plot responses

```
>> w = [0:1:500]*pi/500; H = freqz(b,a,w);
>> magH = abs(H); phaH = angle(H);
>> subplot(2,1,1); plot(w/pi,magH); grid
>> xlabel(‘frequency in pi units’); ylabel(’Magnitude’)
>> title(’Magnitude Response’)
>> subplot(2,1,2); plot(w/pi,phaH/pi); grid
>> xlabel(’frequency in pi units’); ylabel(’Phase in pi units’)
>> title(’Phase Response’)
```

Solutions of the Difference Equations

One-sided z-transform

- The one-sided z-transform of a sequence x(n) is given by

$$Z^{+}\left[x(n)\right] \triangleq Z\left[x(n)u(n)\right] \triangleq X^{+}[z] = \sum_{n=0}^{\infty} x(n) z^{-n}$$

  - Difference equations generally evolve in the positive n direction.
  - Time frame for these solutions will be $n \geq 0$
  - One form involved finding the particular and the homogeneous solutions
  - The other form involved finding the zero-input (initial condition) and the zero-state responses

One-sided z-transform

- The sample shifting property is given by

$$Z^{+}\left[x(n-k)\right]=x(-1)z^{1-k}+x(-2)z^{2-k}+\ldots+x(-k)+z^{-k}X^{+}(z)$$

- The result can now be used to solve difference equations with nonzero initial conditions or with changing inputs

$$1+\sum_{k=1}^{N}a_k y(n-k)=\sum_{m=0}^{M}b_m x(n-m),\quad n\geq 0$$

- subject to these initial conditions:

$$\left\{y(i),i=-1,\ldots,-N\right\}\qquad\qquad\left\{x(i),i=-1,\ldots,-M\right\}$$

Example

- Solve

$$y(n)-\frac{3}{2}y(n-1)+\frac{1}{2}y(n-2)=x(n), \quad n\geq 0$$

- where

$$x(n)=\left(\frac{1}{4}\right)^n u(n)$$

- subject to $y(-1) = 4$ and $y(-2) = 10$.

Example

$$y(n)-\frac{3}{2}y(n-1)+\frac{1}{2}y(n-2)=x(n),\quad n\geq 0$$

- Taking the one-sided *z*-transform of both sides of the difference equation

$$Y^{+}(z)-\frac{3}{2}\left[y(-1)+z^{-1}Y^{+}(z)\right]+\frac{1}{2}\left[y(-2)+z^{-1}y(-1)+z^{-2}Y^{+}(z)\right]=\frac{1}{1-\frac{1}{4}z^{-1}}$$

- Substituting the initial conditions and rearranging

$$Y^{+}(z)\left[1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}\right]=\frac{1}{1-\frac{1}{4}z^{-1}}+\left(1-2z^{-1}\right)$$

Example

$$Y^{+}(z)\left[1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}\right]=\frac{1}{1-\frac{1}{4}z^{-1}}+\left(1-2z^{-1}\right)$$

$$Y^{+}(z)=\frac{\frac{1}{1-\frac{1}{4}z^{-1}}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}+\frac{1-2z^{-1}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}$$

Using the partial fraction expansion

$$Y^{+}(z)=\frac{1}{1-\frac{1}{2}z^{-1}}+\frac{\frac{2}{3}}{1-z^{-1}}+\frac{\frac{1}{3}}{1-\frac{1}{4}z^{-1}}$$

Example

- After inverse transformation the solution is

$$y(n)=\left[\left(\frac{1}{2}\right)^n+\frac{2}{3}+\frac{1}{2}\left(\frac{1}{4}\right)^n\right]u(n)$$

Example

- Homogeneous and particular parts

$$y(n)=\underbrace{\left[\left(\frac{1}{2}\right)^{n}+\frac{2}{3}\right]u(n)}_{\text{Homogeneous part}}+\underbrace{\frac{1}{2}\left(\frac{1}{4}\right)^{n}u(n)}_{\text{Particular part}}$$

  - The homogeneous part is due to the *system poles*, and the particular part is due to the *input poles*.

Example

- Transient and steady-state responses

$$y(n)=\underbrace{\left[\frac{1}{3}\left(\frac{1}{4}\right)^n\left(\frac{1}{2}\right)^n\right]u(n)}_{\text{Transient response}}+\underbrace{\frac{2}{3}u(n)}_{\text{Steady-state response}}$$

- The transient response is due to poles that are *inside* the unit circle, whereas the steady-state response is due to poles that are *on* the unit circle.

Example

- Zero-input (or initial condition) and zero-state responses

$$Y^{+}(z)=\frac{\dfrac{1}{1-\frac{1}{4}z^{-1}}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}+\frac{1-2z^{-1}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}$$

$$Y_{ZS}(z)=H(z)X(z) \qquad\qquad Y_{ZI}(z)=H(z)X_{IC}(z)$$

- $X_{IC}(z)$ can be thought of as an equivalent *initial-condition input* that generates the same output Y_{ZI} as generated by the initial conditions.

$$x_{IC}(n)=\left\{\underset{\uparrow}{1},-2\right\}$$

Example

- taking the inverse z-transform of each part of

$$Y^{+}(z)=\frac{\dfrac{1}{1-\frac{1}{4}z^{-1}}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}+\frac{1-2z^{-1}}{1-\frac{3}{2}z^{-1}+\frac{1}{2}z^{-2}}$$

- The complete response as

$$y(n)=\underbrace{\left[\frac{1}{3}\left(\frac{1}{4}\right)^{n}-2\left(\frac{1}{2}\right)^{n}+\frac{8}{3}\right]u(n)}_{\text{Zero-state response}}+\underbrace{\left[3\left(\frac{1}{2}\right)^{n}-2\right]u(n)}_{\text{Zero-input response}}$$

Matlab Implementation

y = filter(b,a,x,xic)

- xic is an equivalent initial-condition input array

Matlab Implementation

```
>> n = [0:7]; x = (1/4).^n; xic = [1, -2];
>> format long; y1 = filter(b,a,x,xic)
>> y2 = (1/3)*(1/4).^n+(1/2).^n+(2/3)*ones(1,8)
```

Matlab Implementation

xic = filtic(b,a,Y,X)

- to determine $x_{IC}(n)$ analytically
- b and a are the filter coefficient arrays and Y and X are the initialcondition arrays from the initial conditions on $y(n)$ and $x(n)$

```
>> Y = [4, 10]; xic = filtic(b,a,Y)
```

Example

- Solve the difference equation

$$y(n)=\frac{1}{3}\left[x(n)+x(n-1)+x(n-2)+0,95y(n-1)-0,9025y(n-2)\right]$$

- where $x(n) = \cos(\pi n/3)u(n)$ and $y(-1) = -2$, $y(-2) = -3$; $x(-1) = 1$, $x(-2) = 1$
- First determine the solution analytically and then by using MATLAB

Example

- Taking a one-sided *z*-transform of the difference equation

$$Y^{+}(z)=\frac{1}{3}\left[X^{+}(z)x(-1)+z^{-1}X^{+}(z)+x(-2)+z^{-1}x(-1)+z^{-2}X^{+}(z)\right]$$

$$+0,95\left[y(-1)+z^{-1}Y^{+}(z)\right]-0,9025\left[y(-2)+z^{-1}y(-1)+z^{-2}Y^{+}(z)\right]$$

- and substituting the initial conditions

$$Y^{+}(z)=\frac{\frac{1}{3}+\frac{1}{3}z^{-1}+\frac{1}{3}z^{-2}}{1-0,95z^{-1}+0,9025z^{-2}}X^{+}(z)+\frac{1,4742+2,1383z^{-1}}{1-0,95z^{-1}+0,9025z^{-2}}$$

- Clearly, $x_{IC}(n)$ = [1,4742, 2,1383].

Example

- This simplification and further partial fraction expansion can be done using MATLAB.

```
>> b = [1,1,1]/3; a = [1,-0.95,0.9025];
>> Y = [-2,-3]; X = [1,1]; xic=filtic(b,a,Y,X)
>> bxplus = [1,-0.5]; axplus = [1,-1,1]; % X(z) transform coeff.
>> ayplus = conv(a,axplus) % Denominator of Yplus(z)
>> byplus = conv(b,bxplus)+conv(xic,axplus)
>> [R,p,C] = residuez(byplus,ayplus)
>> Mp = abs(p), Ap = angle(p)/pi
```

Example

- Substituting $X^{+}(z)$

$$X^{+}(z)=\frac{1-0,5z^{-1}}{1-z^{-1}+z^{-2}}$$

- obtain $Y^{+}(z)$ as a rational function

$$Y^{+}(z)=\frac{0,0584+j3,9468}{1-e^{-j\pi/3}z^{-1}}+\frac{0,0584-j3,9468}{1-e^{j\pi/3}z^{-1}}$$
$$+\frac{0,8453+j2,0311}{1-0,95e^{j\pi/3}z^{-1}}+\frac{0,8453-j2,0311}{1-0,95e^{-j\pi/3}z^{-1}}$$

Example

- From Table

$$y(n) = 0{,}1169\cos(\pi n/3) + 7{,}8937\sin(\pi n/3)$$
$$+(0{,}95)^n\left[1{,}6906\cos(\pi n/3) - 4{,}0623\sin(\pi n/3)\right], \quad n \geq 0$$

Example – Matlab Verification

```
>> n = [0:7]; x = cos(pi*n/3); y = filter(b,a,x,xic)

% Matlab Verification

>> A=real(2*R(1)); B=imag(2*R(1)); C=real(2*R(3));
D=imag(2*R(4));

>> y=A*cos(pi*n/3)+B*sin(pi*n/3)+((0.95).^n).*(C*cos(pi*n/3)+
D*sin(pi*n/3))
```

Terima Kasih