

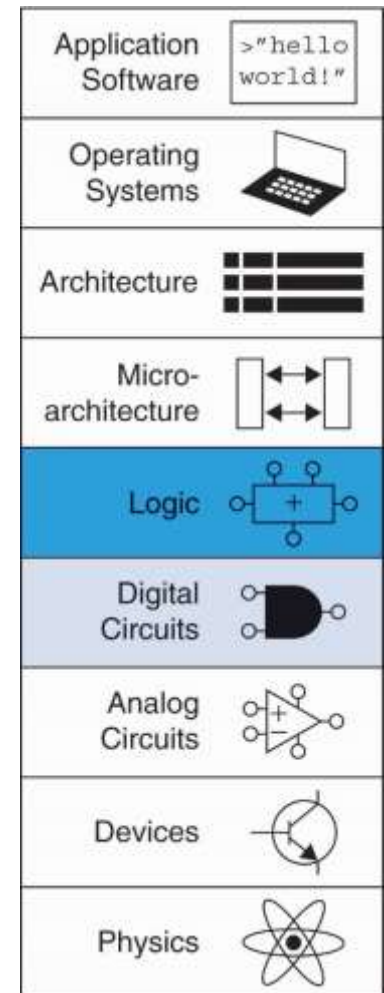
# Digital Design & Computer Architecture

Sarah Harris & David Harris

## Chapter 2: Combinational Logic Design

# Chapter 2 :: Topics

- **Combinational Circuits**
- **Boolean Equations**
- **Boolean Algebra**
- **From Logic to Gates**
- **X's and Z's, Oh My**
- **Karnaugh Maps**
- **Combinational Building Blocks**
- **Timing**



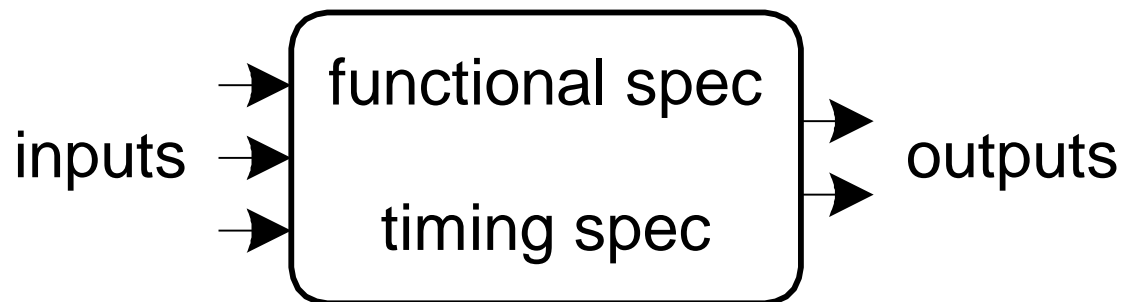
## Chapter 2: Combinational Logic

# Combinational Circuits

# Introduction

**A logic circuit is composed of:**

- Inputs
- Outputs
- Functional specification
- Timing specification



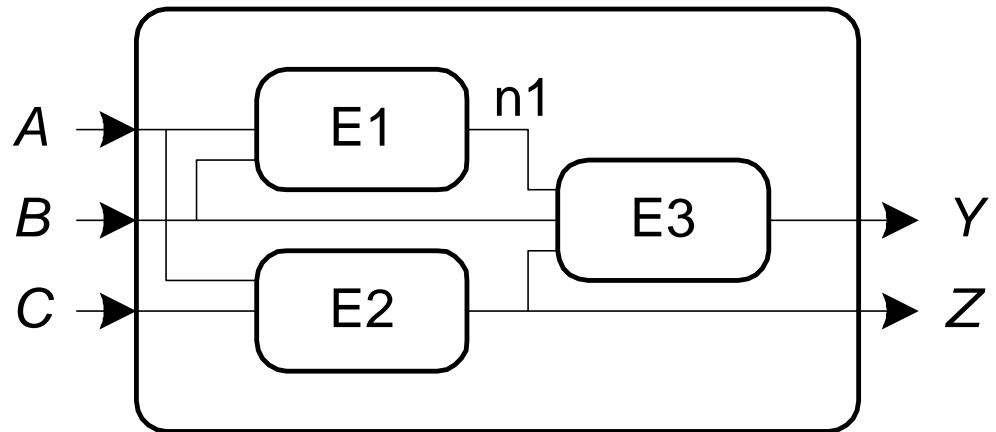
# Circuits

- **Nodes**

- Inputs:  $A, B, C$
- Outputs:  $Y, Z$
- Internal:  $n1$

- **Circuit elements**

- $E1, E2, E3$
- Each a circuit



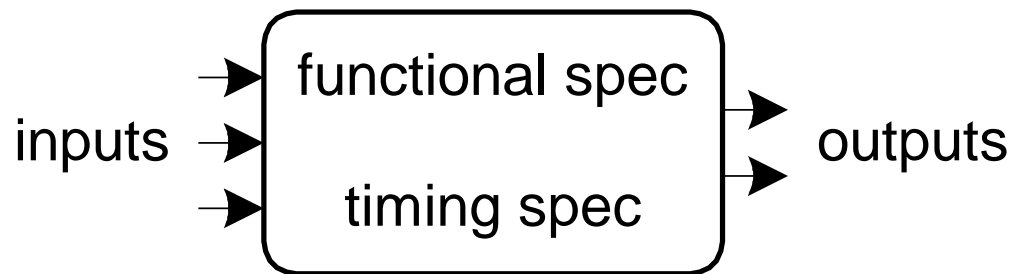
# Types of Logic Circuits

- **Combinational Logic**

- Memoryless
- Outputs determined by current values of inputs

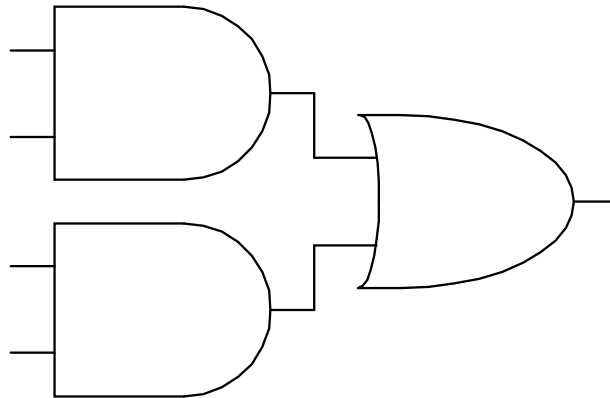
- **Sequential Logic**

- Has memory
- Outputs determined by previous and current values of inputs



# Rules of Combinational Composition

- Every element is combinational
- Every node is either an input or connects to *exactly one* output
- The circuit contains no cyclic paths
- **Example:**



## Chapter 2: Combinational Logic

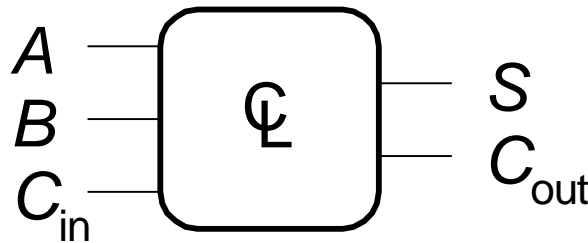
# Boolean Equations



# Boolean Equations

- Functional specification of outputs in terms of inputs

- Example:**  $S = F(A, B, C_{in})$   
 $C_{out} = F(A, B, C_{in})$



$$S = A \oplus B \oplus C_{in}$$
$$C_{out} = AB + AC_{in} + BC_{in}$$

# Some Definitions

- **Complement:** variable with a bar over it

$\bar{A}, \bar{B}, \bar{C}$

- **Literal:** variable or its complement

$A, \bar{A}, B, \bar{B}, C, \bar{C}$

- **Implicant:** product of literals

$AB\bar{C}, \bar{A}C, BC$

- **Minterm:** product that includes all input variables

$AB\bar{C}, \bar{A}\bar{B}\bar{C}, ABC$

- **Maxterm:** sum that includes all input variables

$(A+\bar{B}+C), (\bar{A}+B+\bar{C}), (\bar{A}+\bar{B}+C)$

# Sum-of-Products (SOP) Form

- All Boolean equations can be written in SOP form
- Each row has a **minterm**
- A minterm is a **product** (AND) of literals
- Each minterm is **TRUE** for that row (and only that row)
- Form function by **ORing minterms** where output is **1**
- Thus, a **sum** (OR) of **products** (AND terms)

<b>A</b>	<b>B</b>	<b>Y</b>	<b>minterm</b>	<b>minterm name</b>
0	0	0	$\overline{A} \overline{B}$	$m_0$
0	1	1	$\overline{A} B$	$m_1$
1	0	0	$A \overline{B}$	$m_2$
1	1	1	$A B$	$m_3$

$$Y = F(A, B) =$$

# Sum-of-Products (SOP) Form

- All Boolean equations can be written in SOP form
- Each row has a **minterm**
- A minterm is a **product** (AND) of literals
- Each minterm is **TRUE** for that row (and only that row)
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<b>A</b>	<b>B</b>	<b>Y</b>	<b>minterm</b>	<b>minterm name</b>
0	0	0	$\bar{A} \bar{B}$	$m_0$
0	1	1	$\bar{A} B$	$m_1$
1	0	0	$A \bar{B}$	$m_2$
1	1	1	$A B$	$m_3$

$$Y = F(A, B) = \bar{A}B + AB = \Sigma(\mathbf{1}, \mathbf{3})$$

Long-hand   Short-hand

# Product-of-Sums (POS) Form

- All Boolean equations can be written in POS form
- Each row has a **maxterm**
- A maxterm is a **sum** (OR) of literals
- Each maxterm is **FALSE** for that row (and only that row)
- Form function by **ANDing maxterms** where output is **0**
- Thus, a **product** (AND) of **sums** (OR terms)

<b>A</b>	<b>B</b>	<b>Y</b>	<b>maxterm</b>	<b>maxterm name</b>
0	0	0	$A + B$	$M_0$
0	1	1	$A + \overline{B}$	$M_1$
1	0	0	$\overline{A} + B$	$M_2$
1	1	1	$\overline{A} + \overline{B}$	$M_3$

$$Y = F(A, B) = (A + B) \bullet (\overline{A} + B) = \Pi(\mathbf{0}, \mathbf{2})$$

Long-handShort-hand

# Boolean Equations Example

- You are going to the cafeteria for lunch
  - You won't eat lunch ( $E = 0$ )
    - If it's not clean ( $C = 0$ ) or
    - If they only serve meatloaf ( $M = 1$ )
- Write a truth table for determining if you will eat lunch ( $E$ ).

$C$	$M$	$E$
0	0	
0	1	
1	0	
1	1	

# SOP & POS Form

## SOP – sum-of-products

$C$	$M$	$E$	minterm
0	0	0	$\overline{C} \overline{M}$
0	1	0	$\overline{C} M$
1	0	1	$C \overline{M}$
1	1	0	$C M$

## POS – product-of-sums

$C$	$M$	$E$	maxterm
0	0	0	$C + M$
0	1	0	$C + \overline{M}$
1	0	1	$\overline{C} + M$
1	1	0	$\overline{C} + \overline{M}$

# SOP & POS Form

## SOP – sum-of-products

$C$	$M$	$E$	minterm
0	0	0	$\overline{C} \overline{M}$
0	1	0	$\overline{C} M$
1	0	1	$C \overline{M}$
1	1	0	$C M$

$$\begin{aligned} E &= C\overline{M} \\ &= \Sigma(2) \end{aligned}$$

## POS – product-of-sums

$C$	$M$	$E$	maxterm
0	0	0	$C + M$
0	1	0	$C + \overline{M}$
1	0	1	$\overline{C} + M$
1	1	0	$\overline{C} + \overline{M}$

$$\begin{aligned} E &= (C + M)(C + \overline{M})(\overline{C} + \overline{M}) \\ &= \Pi(0, 1, 3) \end{aligned}$$



# Forming Boolean Expressions

## Example 1:

We will go to the Park (***P*** is the output) if it's not Raining ( $\overline{R}$ ) and we have Sandwiches (***S***).

## Boolean Equation:

# Forming Boolean Expressions

## Example 2:

You will be considered a Winner (**W** is the output) if we send you a Million dollars (**M**) or a small Notepad (**N**).

## Boolean Equation:

# Forming Boolean Expressions

## Example 3:

You can Eat delicious food ( $E$  is the output) if you Make it yourself ( $M$ ) or you have a personal Chef ( $C$ ) and she/he is talented ( $T$ ) but not eXpensive ( $\bar{X}$ ).

## Boolean Equation:

# Forming Boolean Expressions

## Example 4:

You can Enter the building if you have a Hat and Shoes on or if you have a Hat on.

## Boolean Equation:

# Forming Boolean Expressions

## Example 5:

You can Enter the building if you have a Hat and Shoes on or if you have a Hat and no Shoes on.

## Boolean Equation:

## Chapter 2: Combinational Logic

# **Boolean Algebra: Axioms**

# Boolean Algebra

- Axioms and theorems to **simplify** Boolean equations
- Like regular algebra, but simpler: variables have only two values (1 or 0)
- **Duality** in axioms and theorems:
  - ANDs and ORs, 0's and 1's interchanged

# Boolean Axioms

Number	Axiom	Name
A1	$B = 0 \text{ if } B \neq 1$	Binary Field
A2	$\overline{0} = 1$	NOT
A3	$0 \bullet 0 = 0$	AND/OR
A4	$1 \bullet 1 = 1$	AND/OR
A5	$0 \bullet 1 = 1 \bullet 0 = 0$	AND/OR



# Boolean Axioms

Number	Axiom	Dual	Name
A1	$B = 0 \text{ if } B \neq 1$	$B = 1 \text{ if } B \neq 0$	Binary Field
A2	$\overline{0} = 1$	$\overline{1} = 0$	NOT
A3	$0 \bullet 0 = 0$	$1 + 1 = 1$	AND/OR
A4	$1 \bullet 1 = 1$	$0 + 0 = 0$	AND/OR
A5	$0 \bullet 1 = 1 \bullet 0 = 0$	$1 + 0 = 0 + 1 = 1$	AND/OR

**Dual:** Replace:  $\bullet$  with  $+$   
 $0$  with  $1$

## Chapter 2: Combinational Logic

# **Boolean Algebra: Theorems of One Variable**

# Boolean Theorems of One Variable

Number	Theorem	Name
T1	$B \bullet 1 = B$	Identity
T2	$B \bullet 0 = 0$	Null Element
T3	$B \bullet B = B$	Idempotency
T4	$\overline{\overline{B}} = B$	Involution
T5	$B \bullet \overline{B} = 0$	Complements

**Dual:** Replace:  $\bullet$  with  $+$   
 $0$  with  $1$

# Boolean Theorems of One Variable

Number	Theorem	Dual	Name
T1	$B \bullet 1 = B$	$B + 0 = B$	Identity
T2	$B \bullet 0 = 0$	$B + 1 = 1$	Null Element
T3	$B \bullet B = B$	$B + B = B$	Idempotency
T4	$\overline{\overline{B}} = B$		Involution
T5	$B \bullet \overline{B} = 0$	$B + \overline{B} = 1$	Complements

**Dual:** Replace:  $\bullet$  with  $+$   
 $0$  with  $1$

# T1: Identity Theorem

- $B \cdot 1 = B$
- $B + 0 = B$

# T2: Null Element Theorem

- $B \cdot 0 = 0$
- $B + 1 = 1$

# T3: Idempotency Theorem

- $B \cdot B = B$
- $B + B = B$

# T4: Involution Theorem

- $\overline{\overline{B}} = B$



# T5: Complement Theorem

- $B \cdot \overline{B} = 0$
- $B + \overline{B} = 1$

# Recap: Basic Boolean Theorems

Number	Theorem	Dual	Name
T1	$B \bullet 1 = B$	$B + 0 = B$	Identity
T2	$B \bullet 0 = 0$	$B + 1 = 1$	Null Element
T3	$B \bullet B = B$	$B + B = B$	Idempotency
T4	$\overline{\overline{B}} = B$		Involution
T5	$B \bullet \overline{B} = 0$	$B + \overline{B} = 1$	Complements

## Chapter 2: Combinational Logic

# **Boolean Algebra: Theorems of Several Variables**

# Boolean Theorems of Several Vars

#	Theorem	Dual	Name
T6	$B \bullet C = C \bullet B$	$B + C = C + B$	Commutativity
T7	$(B \bullet C) \bullet D = B \bullet (C \bullet D)$	$(B + C) + D = B + (C + D)$	Associativity
T8	$B \bullet (C + D) = (B \bullet C) + (B \bullet D)$	$B + (C \bullet D) = (B + C) (B + D)$	Distributivity
T9	$B \bullet (B + C) = B$	$B + (B \bullet C) = B$	Covering
T10	$(B \bullet C) + (B \bullet \bar{C}) = B$	$(B + C) \bullet (B + \bar{C}) = B$	Combining
T11	$(B \bullet C) + (\bar{B} \bullet D) + (C \bullet D) = (B \bullet C) + (\bar{B} \bullet D)$	$(B + C) \bullet (\bar{B} + D) \bullet (C + D) = (B + C) \bullet (\bar{B} + D)$	Consensus

**Warning:** T8' differs from traditional algebra:  
OR (+) distributes over AND ( $\bullet$ )

# How to Prove

- **Method 1:** Perfect induction
- **Method 2:** Use other theorems and axioms to simplify the equation
  - Make one side of the equation look like the other

# Proof by Perfect Induction

- Also called: **proof by exhaustion**
- Check every possible input value
- If the two expressions produce the same value for every possible input combination, the expressions are equal

# T9: Covering

Number	Theorem	Name
T9	$B \bullet (B + C) = B$	Covering

Prove true by:

- **Method 1:** Perfect induction
- **Method 2:** Using other theorems and axioms

# T9: Covering

Number	Theorem	Name
T9	$B \bullet (B+C) = B$	Covering

## Method 1: Perfect Induction

$B$	$C$	$(B+C)$	$B(B+C)$
0	0		
0	1		
1	0		
1	1		



# T9: Covering

Number	Theorem	Name
T9	$B \bullet (B+C) = B$	Covering

**Method 2:** Prove true using other axioms and theorems.

# T10: Combining

Number	Theorem	Name
T10	$(B \bullet C) + (B \bullet \overline{C}) = B$	Combining

Prove true using other axioms and theorems:

# De Morgan's Theorem: Dual

#	Theorem	Dual	Name
T12	$\overline{B \bullet C \bullet D \dots} = \bar{B} + \bar{C} + \bar{D} \dots$	$\overline{B + C + D \dots} = \bar{B} \bullet \bar{C} \bullet \bar{D} \dots$	De Morgan's Theorem

The **complement** of the **product** is the **sum** of the **complements**.

# Recap: Theorems of Several Vars

#	Theorem	Dual	Name
T6	$B \bullet C = C \bullet B$	$B + C = C + B$	Commutativity
T7	$(B \bullet C) \bullet D = B \bullet (C \bullet D)$	$(B + C) + D = B + (C + D)$	Associativity
T8	$B \bullet (C + D) = (B \bullet C) + (B \bullet D)$	$B + (C \bullet D) = (B + C) (B + D)$	Distributivity
T9	$B \bullet (B + C) = B$	$B + (B \bullet C) = B$	Covering
T10	$(B \bullet C) + (B \bullet \overline{C}) = B$	$(B + C) \bullet (B + \overline{C}) = B$	Combining
T11	$(B \bullet C) + (\overline{B} \bullet D) + (C \bullet D) = (B \bullet C) + (\overline{B} \bullet D)$	$(B + C) \bullet (\overline{B} + D) \bullet (C + D) = (B + C) \bullet (\overline{B} + D)$	Consensus
T12	$\overline{B \bullet C \bullet D \dots} = \overline{B} + \overline{C} + \overline{D} \dots$	$\overline{B + C + D \dots} = \overline{B} \bullet \overline{C} \bullet \overline{D} \dots$	De Morgan's

## Chapter 2: Combinational Logic

# **Boolean Algebra: Simplifying Equations**

# Simplifying an Equation

Simplifying may mean minimal sum of products form:

- SOP form that has the **fewest number of implicants**, where each implicant has the **fewest literals**
  - **Implicant:** product of literals  
 $\overline{A}\overline{B}C, A\overline{C}, B\overline{C}$
  - **Literal:** variable or its complement  
 $A, \overline{A}, B, \overline{B}, C, \overline{C}$

Simplifying could also mean fewest number of gates, lowest cost, lowest power, etc. For example,  $Y \equiv A \text{ xor } B$  is likely simpler than minimal Sum of Products  $Y = AB + \overline{A}\overline{B}$ . These depend on details of the technology.

# Simplifying Boolean Equations

## Example 1:

$$Y = \bar{A}B + AB$$

$$Y = B$$

T10: Combining

or

$$Y = B(A + \bar{A})$$

T8: Distributivity

$$= B(1)$$

T5': Complements

$$= B$$

T1: Identity

# Simplifying Boolean Equations

## Example 2:

$$Y = \bar{A}\bar{B}C + ABC + \bar{A}BC$$

$$= \bar{A}\bar{B}C + \textcolor{red}{ABC} + \textcolor{red}{ABC} + \bar{A}BC \quad \text{T3': Idempotency}$$

$$= (\textcolor{red}{\bar{A}\bar{B}C + ABC}) + (\textcolor{blue}{ABC + \bar{A}BC}) \quad \text{T7': Associativity}$$

$$= \textcolor{red}{AC} + \textcolor{blue}{BC} \quad \text{T10: Combining}$$



## Chapter 2: Combinational Logic

# Extra Examples

## **Boolean Algebra: Simplifying Equations**

# Simplification methods

- **Distributivity (T8, T8')**  
 $B(C+D) = BC + BD$   
 $B + CD = (B+C)(B+D)$
- **Covering (T9')**  
 $A + AP = A$
- **Combining (T10)**  
 $\overline{PA} + PA = P$
- **Expansion**  
 $P = \overline{PA} + PA$   
 $A = A + AP$
- **Idempotency (duplication)**  $A = A + A$
- **“Simplification” theorem**  
 $A + \overline{A}P = A + P$   
 $\overline{A} + AP = \overline{A} + P$

# Proving the “Simplification” Theorem

## “Simplification” theorem

$$A + \overline{A}P = A + P$$

Method 1:

Method 2:

# T11: Consensus

Number	Theorem	Name
T11	$(B \bullet C) + (\overline{B} \bullet D) + (C \bullet D) = (B \bullet C) + (\overline{B} \bullet D)$	Consensus

**Prove using other theorems and axioms:**

# Simplification methods

- **Distributivity (T8, T8')**  
 $B(C+D) = BC + BD$   
 $B + CD = (B+C)(B+D)$
- **Covering (T9')**  
 $A + AP = A$
- **Combining (T10)**  
 $\overline{P}A + PA = P$
- **Expansion**  
 $P = \overline{P}A + PA$   
 $A = A + AP$
- **Idempotency (duplication)**  $A = A + A$
- **“Simplification” theorem**  
 $A + \overline{A}P = A + P$   
 $\overline{A} + AP = \overline{A} + P$

# Simplification methods

- **Distributivity (T8, T8')**  
 $B(C+D) = BC + BD$   
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- **Idempotency (duplication)**  $A = A + A$
- **“Simplification” theorem**  
 $A + \overline{A}P = A + P$   
 $\overline{A} + AP = \overline{A} + P$

# Simplifying Boolean Equations

**Example 3:**

$$Y = A(AB + ABC)$$

# Simplification methods

- **Distributivity (T8, T8')**  
 $B(C+D) = BC + BD$   
 $B + CD = (B+C)(B+D)$
- **Covering (T9')**  
 $A + AP = A$
- **Combining (T10)**  
 $\overline{P}A + PA = P$
- **Expansion**  
 $P = \overline{P}A + PA$   
 $A = A + AP$
- **Idempotency (duplication)**  $A = A + A$
- **“Simplification” theorem**  
 $A + \overline{A}P = A + P$   
 $\overline{A} + AP = \overline{A} + P$



# Simplifying Boolean Equations

**Example 4:**

$$Y = A'BC + A'$$

Recall:  $A' = \bar{A}$

# Simplifying Boolean Equations

## Example 4:

$$Y = A'BC + A'$$

$$\text{Recall: } A' = \bar{A}$$

or

# Multiplying Out: SOP Form

An expression is in **sum-of-products (SOP)** form when all products contain literals only.

- **SOP form:**  $Y = AB + BC' + DE$
- **NOT SOP form:**  $Y = DF + E(A' + B)$
- **SOP form:**  $Z = A + BC + DE'F$

# Multiplying Out: SOP Form

## Example 5:

$$Y = (A + C + D + E)(A + B)$$

Apply T8' first when possible:  $W + XZ = (W + X)(W + Z)$

or

# Simplifying Boolean Equations

## Example 6:

$$Y = AB + BC + B'D' + AC'D'$$

Method 1:

Method 2:

# Literal and implicant ordering

- Variables within an implicant should be in alphabetical order.
- The order of implicants doesn't matter.

# Simplifying Boolean Equations

## Example 7:

$$Y = (A + BC)(A + DE)$$

Apply T8' first when possible:  $W + XZ = (W + X)(W + Z)$

or

# Review: Canonical SOP & POS Forms

**SOP – sum-of-products**  $E = \boxed{C\bar{M}}$

$C$	$M$	$E$	minterm
0	0	0	$\bar{C} \bar{M}$
0	1	0	$\bar{C} M$
1	0	1	$C \bar{M}$
1	1	0	$C M$

same

**POS – product-of-sums**  $E = (C + M)(C + \bar{M})(\bar{C} + \bar{M})$

$C$	$M$	$E$	maxterm
0	0	0	$C + M$
0	1	0	$C + \bar{M}$
1	0	1	$\bar{C} + M$
1	1	0	$\bar{C} + \bar{M}$



# Factoring: POS Form

An expression is in **product-of-sums (POS)** form when all sums contain literals only.

- **POS form:**  $Y = (A+B)(C+D)(E'+F)$
- **NOT POS form:**  $Y = (D+E)(F'+GH)$
- **POS form:**  $Z = A(B+C)(D+E')$

# Factoring: POS Form

## Example 8:

$$Y = (A + B' CDE)$$

Apply T8' first when possible:  $W + XZ = (W + X)(W + Z)$

# Factoring: POS Form

## Example 9:

$$Y = AB + C'DE + F$$

Apply T8' first when possible:  $W+XZ = (W+X)(W+Z)$

# De Morgan's Theorem

## Example 10:

$$Y = \overline{(A + BD)}\overline{C}$$

- Work from the **outside in** (i.e., top bar, then down)
- Use **involution** when possible

# De Morgan's Theorem

## Example 11:

$$Y = \overline{(\overline{A}\overline{C}\overline{E} + \overline{D})} + B$$

## Chapter 2: Combinational Logic

# Common Errors

## **Boolean Algebra: Simplifying Equations**

# Common Errors

- Using ticks ' instead of **bars** over variables when writing equations by hand – ticks are easy to lose
- **Not** keeping terms **aligned** from step to step
  - Alignment helps you see what changed from step-to-step.
  - It helps in both solving and double-checking the problem.
- Applying **multiple theorems** to the same term in one step
- Applying **your own personal theorems** – don't do it 😊
- And, on a related note: **almost** applying the correct theorem
- **Not** looking for opportunities to use combining, covering, and distributivity (especially the dual form).

# Common Errors

- **Losing bars** (alignment will help you avoid this)
- **Losing terms** (alignment will help you avoid this)
- **Trying to do multiple steps at once** – this is prone to errors!
- **Applying theorems incorrectly**, for example:
  - **Wrong:**  $ABC + \overline{A}BC = B$  **Correct:**  $ABC + \overline{A}BC = AC$ . Products may only **differ in a single term** when using the combining theorem.
  - **Wrong:**  $(A + \overline{A}) = 0$  **Correct:**  $A + \overline{A} = 1$
  - **Wrong:**  $(A \cdot \overline{A}) = 1$  **Correct:**  $A \cdot \overline{A} = 0$
  - **Wrong:**  $ABC = B$  **Correct:**  $B + ABC = B$ . In order to use the covering theorem, you must have a term that covers the other terms.
  - **Wrong:**  $\overline{AC} = \overline{A}\overline{C}$  **Correct:**  $\overline{AC} = \overline{A} + \overline{C}$  (De Morgan's)
  - **Wrong:**  $\overline{A + C} = \overline{A} + \overline{C}$  **Correct:**  $\overline{A + C} = \overline{A}\overline{C}$  (De Morgan's)



# Common Errors with De Morgan's

- Not starting from the outside parentheses and working in: this often causes additional steps.
- Trying to apply De Morgan's theorem to an entire **complex operation** (instead of just to terms ANDed under a bar or terms ORed under a bar)
- **Losing bars.** Remember that applying the De Morgan's Theorem is a 3 step process. For a product term under a bar:
  1. Change ANDs to ORs (or vice versa for a sum term under a bar)
  2. Bring down the terms
  3. Put bars over the individual terms
- Not keeping terms associated (i.e., **losing parentheses**)
  - For example,  $\overline{ABC} = (\overline{A} + \overline{B} + \overline{C})$
  - Example error:
    - **Wrong:**  $(ABC)'C + D' = A' + B' + C'C + D' = A' + B' + D'$
    - **Correct:**  $(ABC)'C + D' = (A' + B' + C')C + D' = A'C + B'C + D'$

# Chapter 2: Combinational Logic

## **From Logic to Gates**

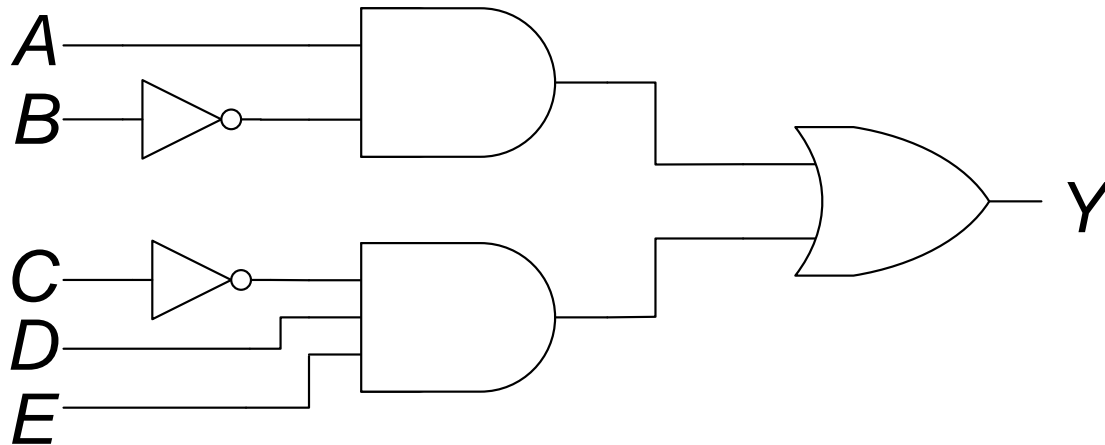
# From Logic to Gates

Build the following equation using logic gates:

$$Y = A\bar{B} + \bar{C}DE$$

# Circuit Schematics Rules

- Inputs on the left (or top)
- Outputs on right (or bottom)
- Gates flow from left to right
- Straight wires are best



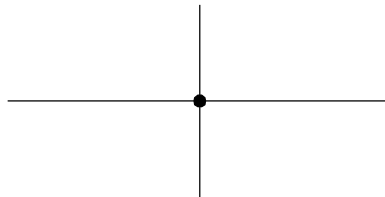
# Circuit Schematic Rules (cont.)

- Wires always connect at a T junction
- A dot where wires cross indicates a connection between the wires
- Wires crossing *without* a dot make no connection

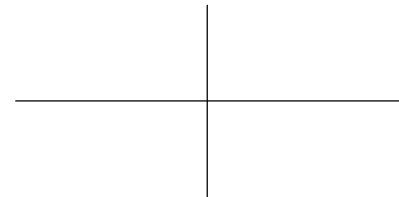
wires connect  
at a T junction



wires connect  
at a dot



wires crossing  
without a dot do  
not connect

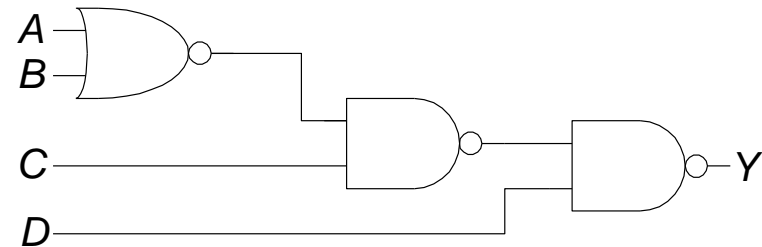
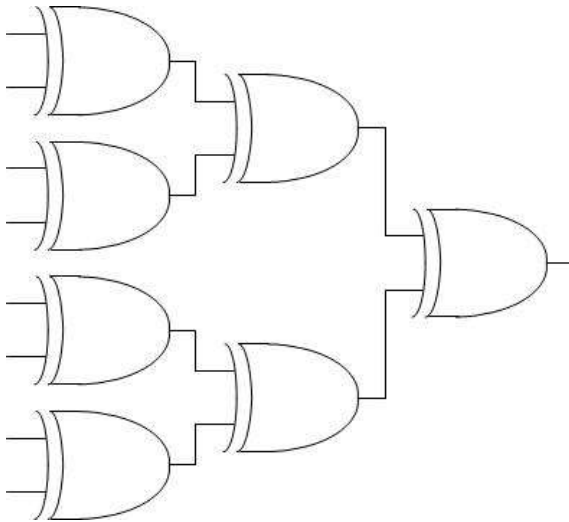


# Two-Level Logic

- Two-level logic: **ANDs** followed by **ORs**
- Example:  $Y = \overline{A}\overline{B}\overline{C} + A\overline{B}\overline{C} + A\overline{B}C$

# Multilevel Logic

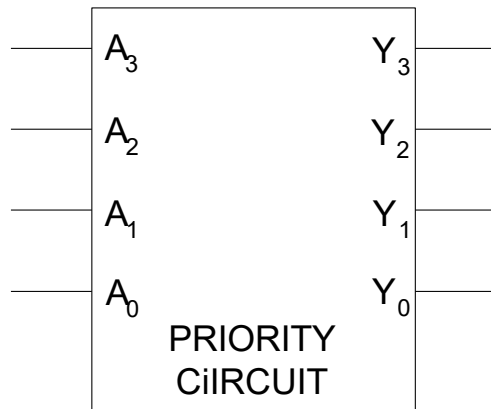
- Complex logic is often built from many stages of simpler gates.



# Multiple-Output Circuits

- Example: Priority Circuit**

Output asserted  
corresponding to most  
significant TRUE input



$A_3$	$A_2$	$A_1$	$A_0$	$Y_3$	$Y_2$	$Y_1$	$Y_0$
0	0	0	0				
0	0	0	1				1
0	0	1	0			1	
0	0	1	1			1	
0	1	0	0		1		
0	1	0	1		1		
0	1	1	0		1		
0	1	1	1		1		
1	0	0	0	1			
1	0	0	1	1			
1	0	1	0	1			
1	0	1	1	1			
1	1	0	0	1			
1	1	0	1	1			
1	1	1	0	1			
1	1	1	1	1			
1	1	1	1	1			



# Priority Circuit Hardware

$A_3$	$A_2$	$A_1$	$A_0$	$Y_3$	$Y_2$	$Y_1$	$Y_0$
0	0	0	0	0	0	0	0
0	0	0	1	0	0	0	1
0	0	1	0	0	0	1	0
0	0	1	1	0	0	1	0
0	1	0	0	0	1	0	0
0	1	0	1	0	1	0	0
0	1	1	0	0	1	0	0
0	1	1	1	0	1	0	0
1	0	0	0	1	0	0	0
1	0	0	1	1	0	0	0
1	0	1	0	1	0	0	0
1	0	1	1	1	0	0	0
1	1	0	0	1	0	0	0
1	1	0	1	1	0	0	0
1	1	1	0	1	0	0	0
1	1	1	1	1	0	0	0

# Don't Cares

$A_3$	$A_2$	$A_1$	$A_0$	$Y_3$	$Y_2$	$Y_1$	$Y_0$
0	0	0	0	0	0	0	0
0	0	0	1	0	0	0	1
0	0	1	0	0	0	1	0
0	0	1	1	0	0	1	0
0	1	0	0	0	1	0	0
0	1	0	1	0	1	0	0
0	1	1	0	0	1	0	0
0	1	1	1	0	1	0	0
1	0	0	0	1	0	0	0
1	0	0	1	1	0	0	0
1	0	1	0	1	0	0	0
1	0	1	1	1	0	0	0
1	1	0	0	1	0	0	0
1	1	0	1	1	0	0	0
1	1	1	0	1	0	0	0
1	1	1	1	1	0	0	0

$$Y_3 = A_3$$

$$Y_2 = \overline{A_3} A_2$$

$$Y_1 = \overline{A_3} \overline{A_2} A_1$$

$$Y_0 = \overline{A_3} \overline{A_2} \overline{A_1} A_0$$

$A_3$	$A_2$	$A_1$	$A_0$	$Y_3$	$Y_2$	$Y_1$	$Y_0$

# Chapter 2: Combinational Logic

## **Two-Level Logic Forms**

# Two-Level Logic Variations

- **ANDs** followed by **ORs**: **SOP** form
- **ORs** followed by **ANDs**: **POS** form
- Only **NAND** gates: **SOP** form
- Only **NOR** gates: **POS** form

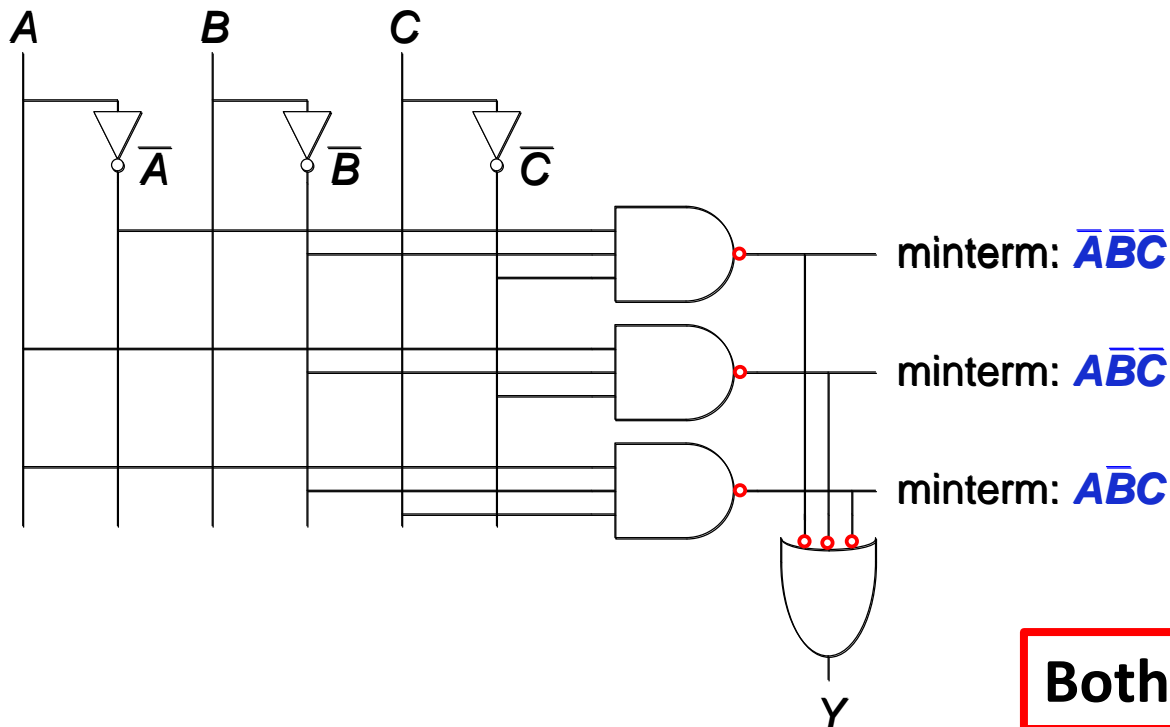
**Most common form of two-level logic**

# Two-Level Logic Variation

- Two-level logic variation: **ORs** followed by **ANDs**
- Example:  $Y = (\bar{A} + \bar{B})(A + B + \bar{C})$

# Two-Level Logic

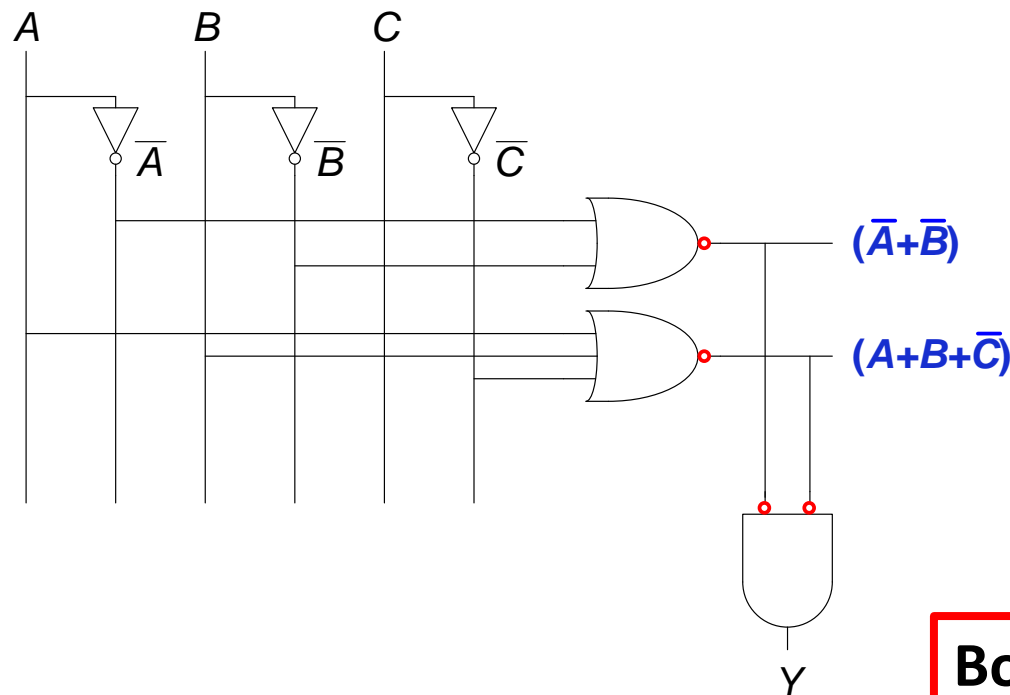
- Two-level logic: **ANDs** followed by **ORs** → **NANDs**
- Example:  $Y = \bar{A}\bar{B}\bar{C} + A\bar{B}\bar{C} + A\bar{B}C$



Both: **SOP** form

# Two-Level Logic Variation

- Two-level logic: **ORs** followed by **ANDs** → **NORs**
- Example:  $Y = (\bar{A} + \bar{B})(A + B + \bar{C})$



Both: **POS** form

# Chapter 2: Combinational Logic

## **Bubble Pushing**



# De Morgan's Theorem

#	Theorem	Dual	Name
T12	$\overline{B \bullet C \bullet D \dots} = \bar{B} + \bar{C} + \bar{D} \dots$	$\overline{B + C + D \dots} = \bar{B} \bullet \bar{C} \bullet \bar{D} \dots$	De Morgan's Theorem

# De Morgan's Theorem

## Example D1:

$$\begin{aligned} Y &= \overline{A+BC} \\ &= \overline{A} \bullet \overline{\overline{BC}} \\ &= \overline{A} \bullet BC \\ &= \overline{A}BC \end{aligned}$$

- Work from the **outside in** (i.e., top bar, then down)
- Use **involution** when possible

# DeMorgan's Theorem

## Example D2:

$$\begin{aligned} Y &= \overline{A+BC+AB} \\ &= \overline{A} \bullet \overline{BC} \bullet \overline{AB} \\ &= \overline{A} \bullet BC \bullet (\overline{A} + \overline{B}) \\ &= \overline{A}BC \bullet (A + B) \\ &= \overline{A}BCA + \overline{A}BCB \\ &= \overline{A}BC \end{aligned}$$

- De Morgan **applies to**:
  - **Products** under a bar
  - **Sums** under a bar
- **Do not** try to apply DeMorgan's to a **mix of operations**

# De Morgan's Theorem

## Example D2:

$$Y = \overline{A + BC + \overline{A}\overline{B}}$$

$$= \overline{A} \cdot \overline{\overline{BC}} \cdot \overline{\overline{A}\overline{B}}$$

$$= \overline{A} \cdot BC \cdot (\overline{\overline{A}} + \overline{\overline{B}})$$

$$= \overline{A}BC \cdot (A + B)$$

$$= \overline{A}BCA + \overline{A}BCB$$

$$= \overline{A}BC$$

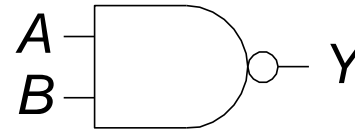
Don't forget these parentheses!

**Remember:**

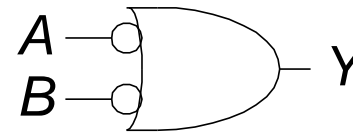
$$\overline{\overline{A}\overline{B}} = (\overline{\overline{A}} + \overline{\overline{B}})$$

# De Morgan's Theorem: Gates

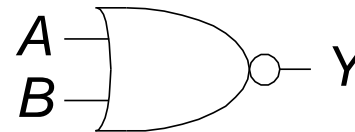
- $Y = \overline{AB} = \overline{A} + \overline{B}$



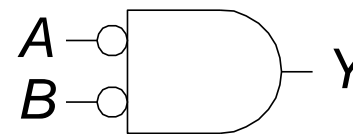
**NAND gate**  
two forms



- $Y = \overline{A + B} = \overline{A} \cdot \overline{B}$



**NOR gate**  
two forms



# Bubble Pushing

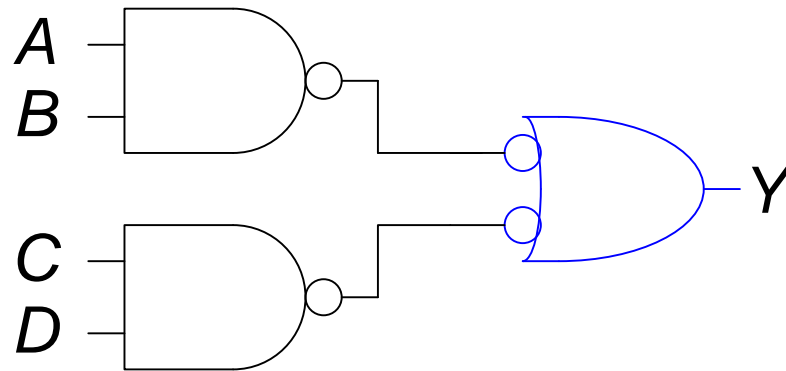
- **Backward:**

- Body changes
- Adds bubbles to inputs



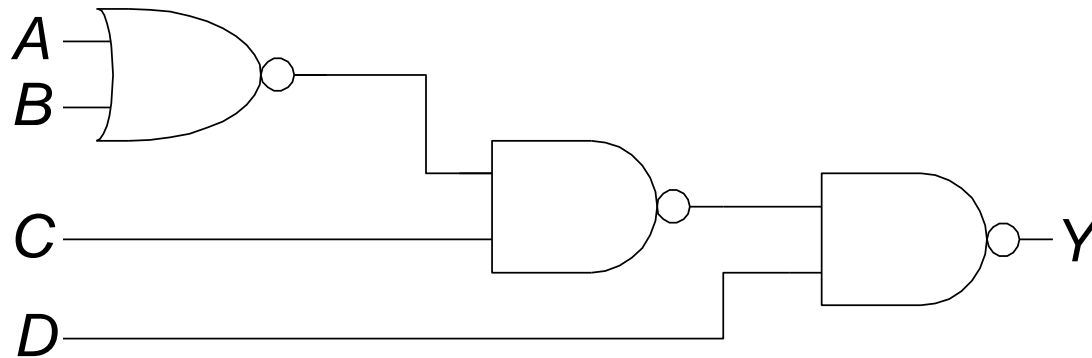
# Bubble Pushing

- What is the Boolean expression for this circuit?



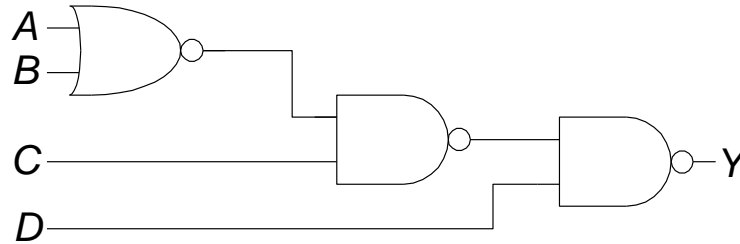
# Bubble Pushing Rules

- Begin at output, then work toward inputs
- Push bubbles on final output back
- Draw gates in a form so bubbles cancel





# Bubble Pushing Example

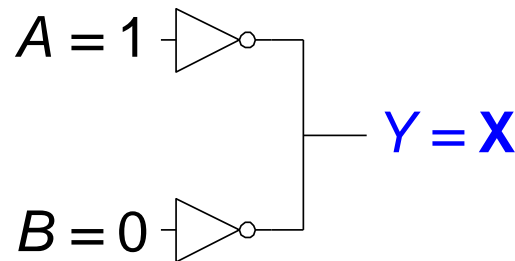


# Chapter 2: Combinational Logic

**X's and Z's, Oh My**

# Contention: X

- **Contention:** circuit tries to drive output to 1 and 0
  - Actual value somewhere in between
  - Could be 0, 1, or in forbidden zone
  - Might change with voltage, temperature, time, noise
  - Often causes excessive power dissipation

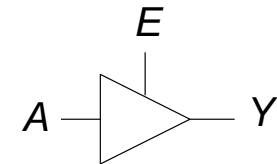


- **X is also used for:**
  - Uninitialized values
  - Don't Care
- **Warnings:**
  - Contention or uninitialized outputs usually indicate a **bug**.
  - Look at the context to tell meaning

# Floating: Z

- Floating, high impedance, open, high Z
- Floating output might be 0, 1, or somewhere in between
  - A voltmeter **won't** indicate whether a node is floating
  - But if you touch the node or your instructor walks over for a checkoff, it may change randomly

## Tristate Buffer

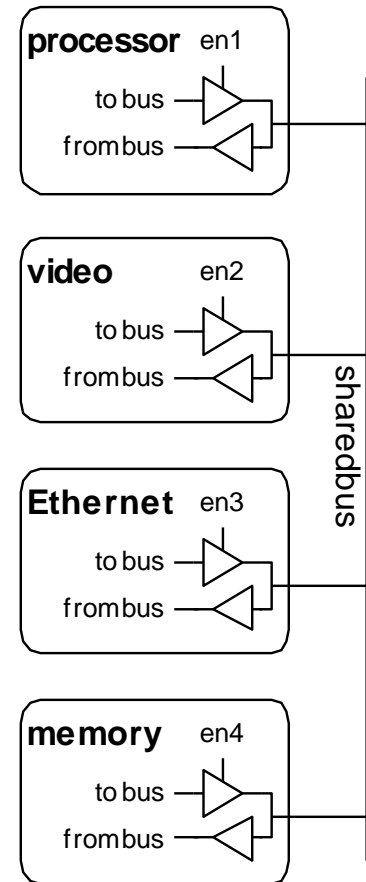


$E$	$A$	$Y$
0	0	Z
0	1	Z
1	0	0
1	1	1

# Tristate Busses

Floating nodes are used in tristate busses

- Many different drivers
- Exactly one is active at once



# Chapter 2: Combinational Logic

## **Karnaugh Maps**

# Karnaugh Maps (K-Maps)

- Boolean expressions can be minimized by combining terms
- K-maps minimize equations graphically
  - $PA + \overline{P}\overline{A} = P$

A	B	C	Y
0	0	0	1
0	0	1	1
0	1	0	0
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	0
1	1	1	0

Y C	AB			
	00	01	11	10
0				
1				

Y C	AB			
	00	01	11	10
0	$\overline{A}\overline{B}\overline{C}$	$\overline{A}B\overline{C}$	$AB\overline{C}$	$A\overline{B}\overline{C}$
1	$\overline{A}\overline{B}C$	$\overline{A}BC$	$ABC$	$A\overline{B}C$

# K-Map

- Circle 1's in adjacent squares
- In Boolean expression: include only literals whose true **and** complement form are **not** in the circle

A	B	C	Y
0	0	0	1
0	0	1	1
0	1	0	0
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	0
1	1	1	0

Y C	AB			
	00	01	11	10
0	1	0	0	0
1	1	0	0	0

Y C	AB			
	00	01	11	10
0	$\bar{A}\bar{B}\bar{C}$	$\bar{A}B\bar{C}$	$AB\bar{C}$	$A\bar{B}\bar{C}$
1	$\bar{A}\bar{B}C$	$\bar{A}BC$	$ABC$	$A\bar{B}C$

$$Y = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C = \bar{A}\bar{B}$$



# 3-Input K-Map

- Circle 1's in adjacent squares
- In Boolean expression: include only literals whose true **and** complement form are ***not*** in the circle

Truth Table			
<i>A</i>	<i>B</i>	<i>C</i>	<i>Y</i>
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	1
1	0	0	0
1	0	1	0
1	1	0	0
1	1	1	1

		K-Map			
<i>Y</i>	<i>C</i>	<i>AB</i>			
		00	01	11	10
	0				
	1				

$Y =$

# Some Definitions

- **Complement:** variable with a bar over it

$\bar{A}, \bar{B}, \bar{C}$

- **Literal:** variable or its complement

$A, \bar{A}, B, \bar{B}, C, \bar{C}$

- **Implicant:** product of literals

$AB\bar{C}, \bar{A}C, BC$

- **Prime implicant:** implicant corresponding to the **largest circle** in a K-map

# K-Map Rules

- **Every 1 must be circled** at least once
- Each circle must span a **power of 2** (i.e. 1, 2, 4) squares in each direction
- Each circle must be as **large** as possible
- A circle may **wrap around the edges**

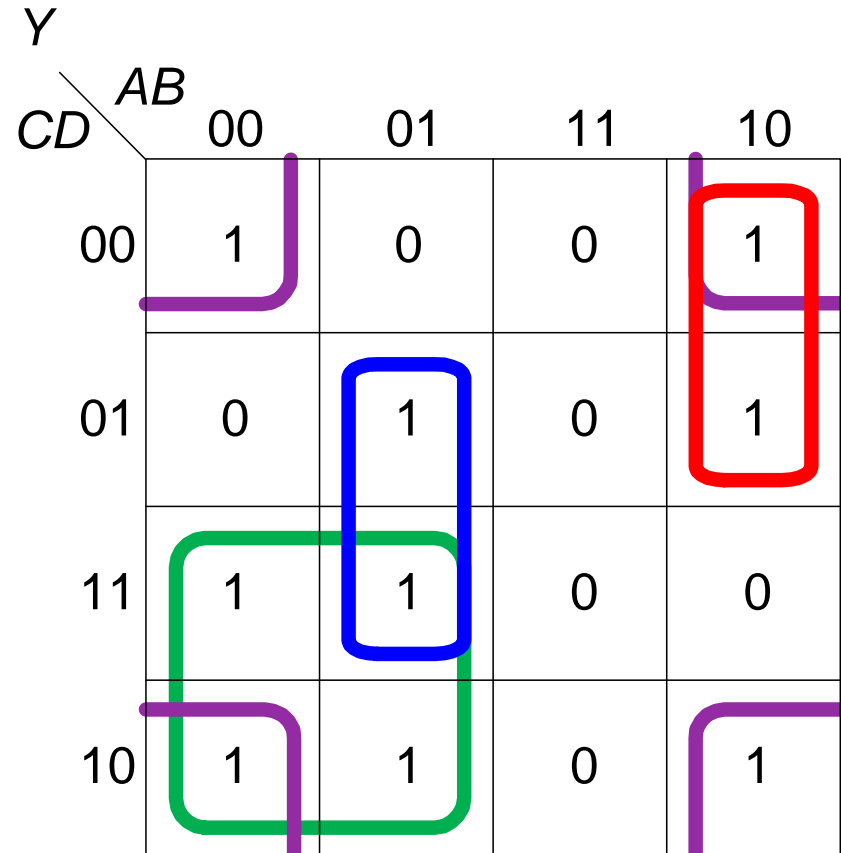
# 4-Input K-Map

<i>A</i>	<i>B</i>	<i>C</i>	<i>D</i>	<i>Y</i>
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	1
1	0	1	1	0
1	1	0	0	0
1	1	0	1	0
1	1	1	0	0
1	1	1	1	0

<i>Y</i>		<i>AB</i>			
<i>CD</i>		00	01	11	10
	00				
01					
11					
10					

# 4-Input K-Map

<i>A</i>	<i>B</i>	<i>C</i>	<i>D</i>	<i>Y</i>
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	1
1	0	1	1	0
1	1	0	0	0
1	1	0	1	0
1	1	1	0	0
1	1	1	1	0



$Y =$

## Chapter 2: Combinational Logic

# **Karnaugh Maps with Don't Cares**

# K-Map Rules

- **Every 1 must be circled** at least once
- Each circle must span a **power of 2** (i.e. 1, 2, 4) squares in each direction
- Each circle must be as **large** as possible
- A circle may **wrap around the edges**
- Circle a “**don't care**” (X) **only if it helps** minimize the equation

# K-Maps with Don't Cares

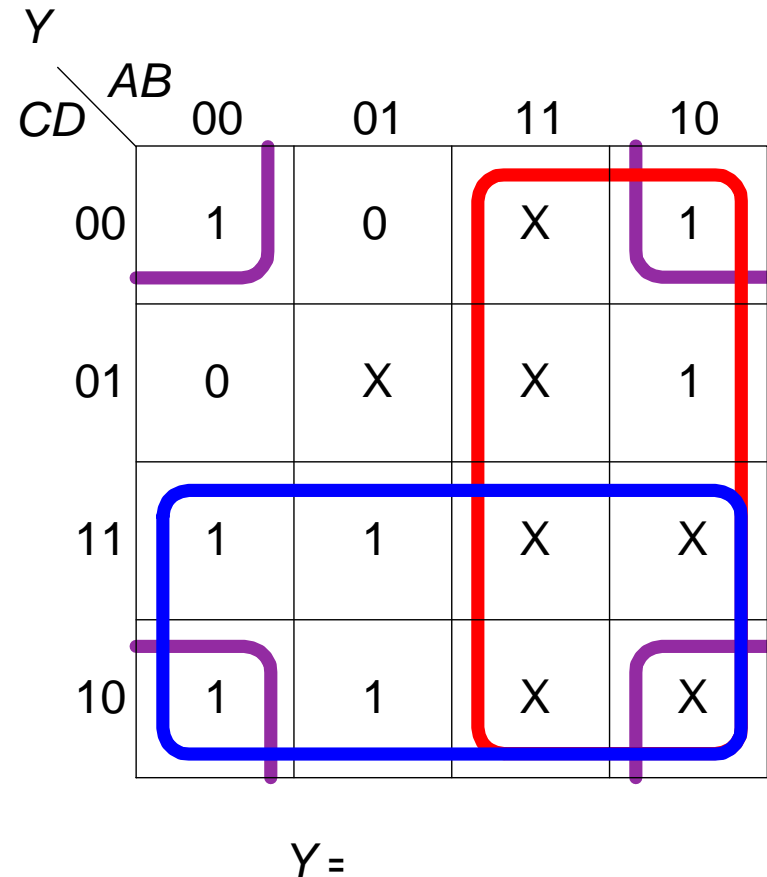
A	B	C	D	Y
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	X
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	X

		Y			
CD	AB				
		00	01	11	10
00					
01					
11					
10					



# K-Maps with Don't Cares

<i>A</i>	<i>B</i>	<i>C</i>	<i>D</i>	<i>Y</i>
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	X
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	X



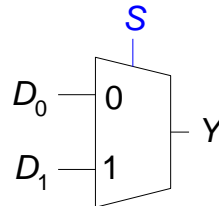
## Chapter 2: Combinational Logic

# **Combinational Building Blocks: Multiplexers**

# Multiplexer (Mux)

- Selects between one of  $N$  inputs to connect to output
- **Select** input is  $\log_2 N$  bits – control input
- Example:

2:1 Mux



$S$	$D_1$	$D_0$	$Y$	$S$	$Y$
0	0	0	0	0	$D_0$
0	0	1	1	1	$D_1$
0	1	0	0		
0	1	1	1		
1	0	0	0		
1	0	1	0		
1	1	0	1		
1	1	1	1		

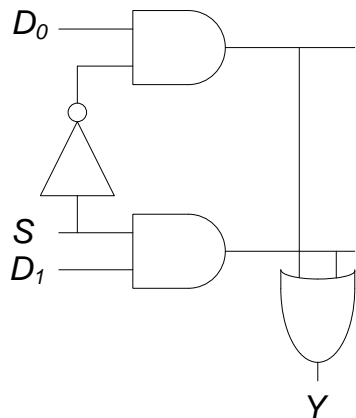
# 2:1 Multiplexer Implementations

- **Logic gates**

- Sum-of-products form

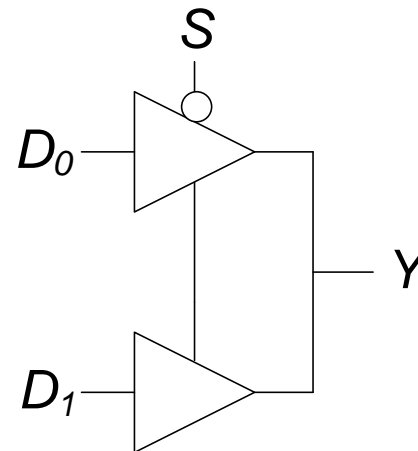
Y S	$D_0 D_1$		00	01	11	10
	0	1	0	0	1	1
1	0	1	0	1	1	0

$$Y = D_0 \bar{S} + D_1 S$$



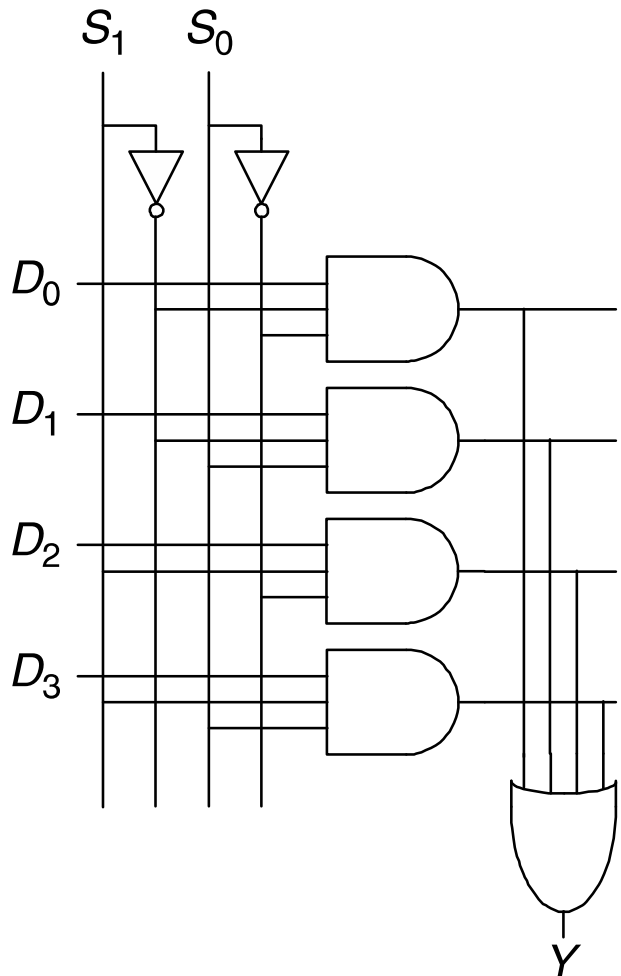
- **Tristates**

- Two tristates
- Turn on exactly one to select the appropriate input

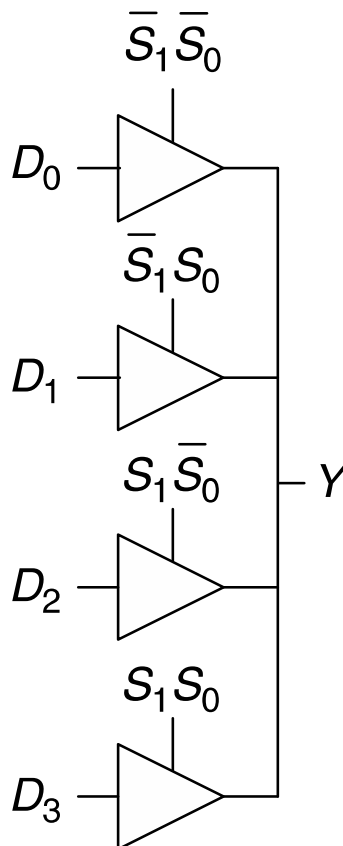


# 4:1 Multiplexer Implementations

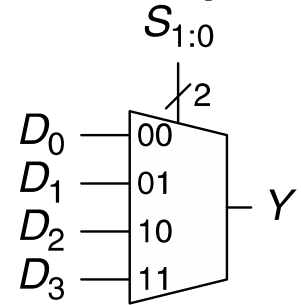
## 2-Level Logic



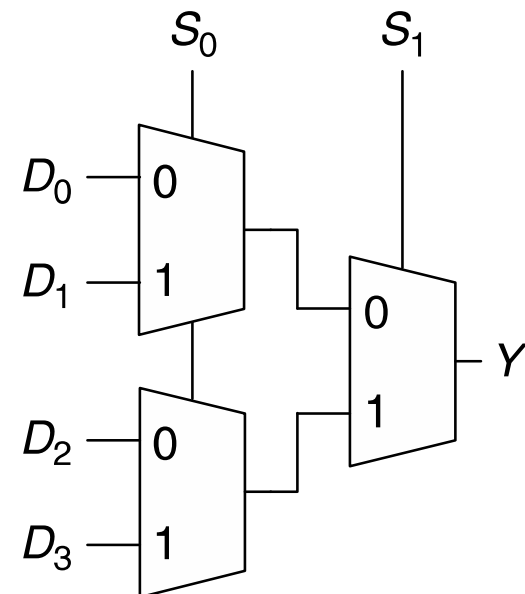
## Tristates



## 4:1 Mux Symbol



## Hierarchical

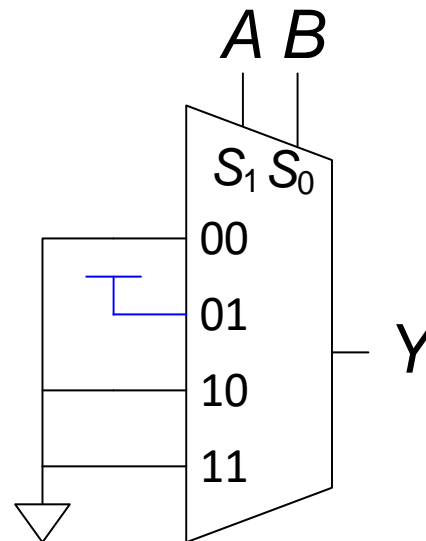


# Logic using Multiplexers

Using mux as a **lookup table**

<i>A</i>	<i>B</i>	<i>Y</i>
0	0	0
0	1	1
1	0	0
1	1	0

$$Y = \overline{A}B$$

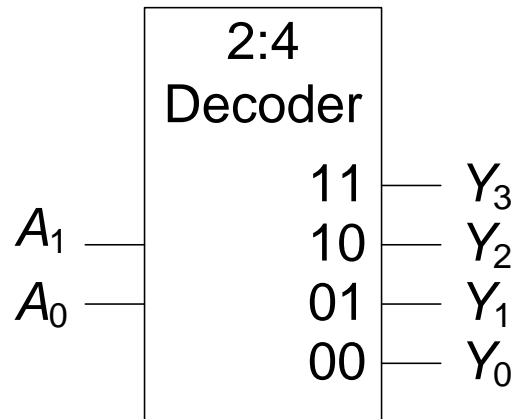


## Chapter 2: Combinational Logic

# **Combinational Building Blocks: Decoders**

# Decoders

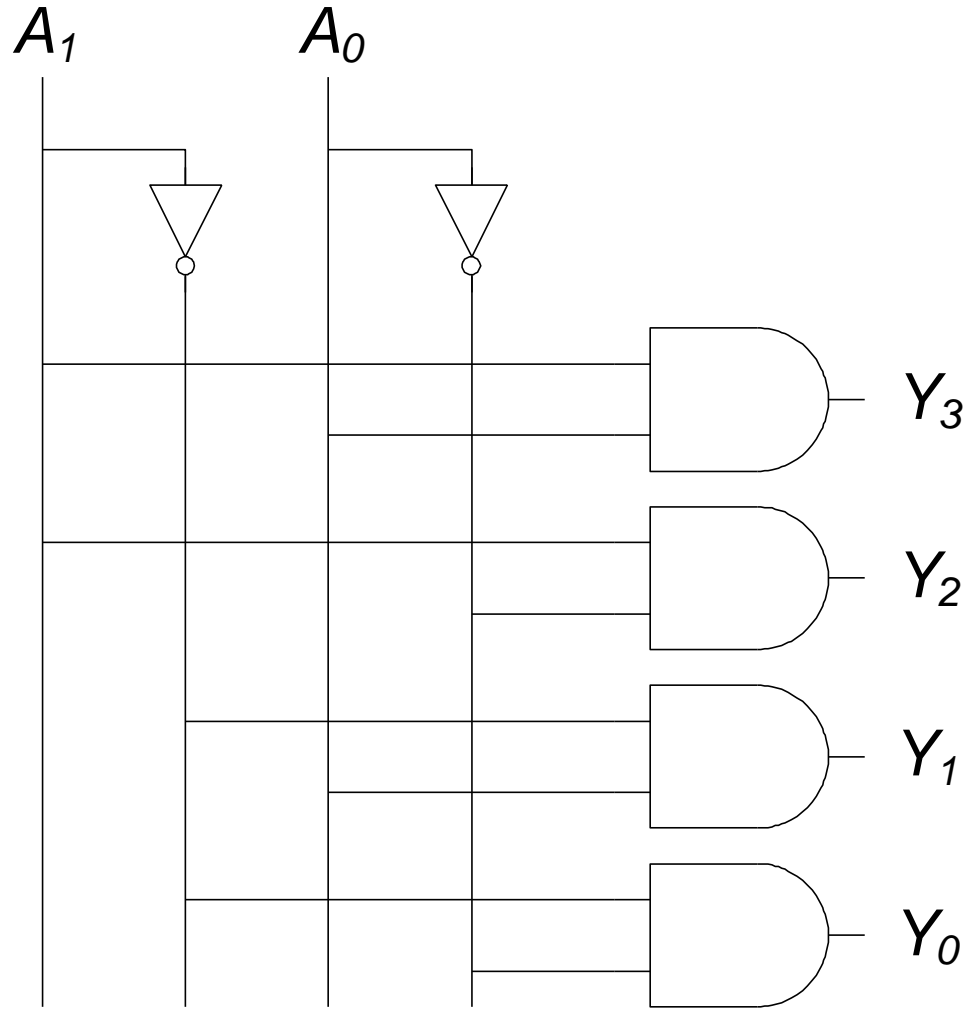
- $N$  inputs,  $2^N$  outputs
- **One-hot outputs:** only one output **HIGH** at once



$A_1$	$A_0$	$Y_3$	$Y_2$	$Y_1$	$Y_0$
0	0	0	0	0	<b>1</b>
0	1	0	0	<b>1</b>	0
1	0	0	<b>1</b>	0	0
1	1	<b>1</b>	0	0	0

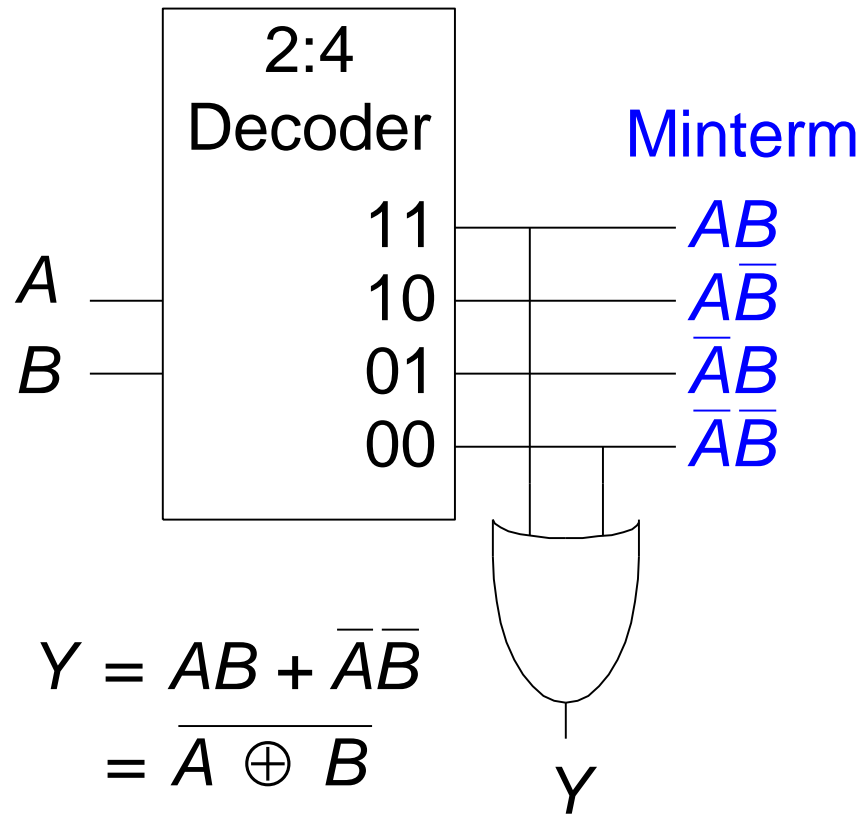


# Decoder Implementation



# Logic Using Decoders

OR the minterms:

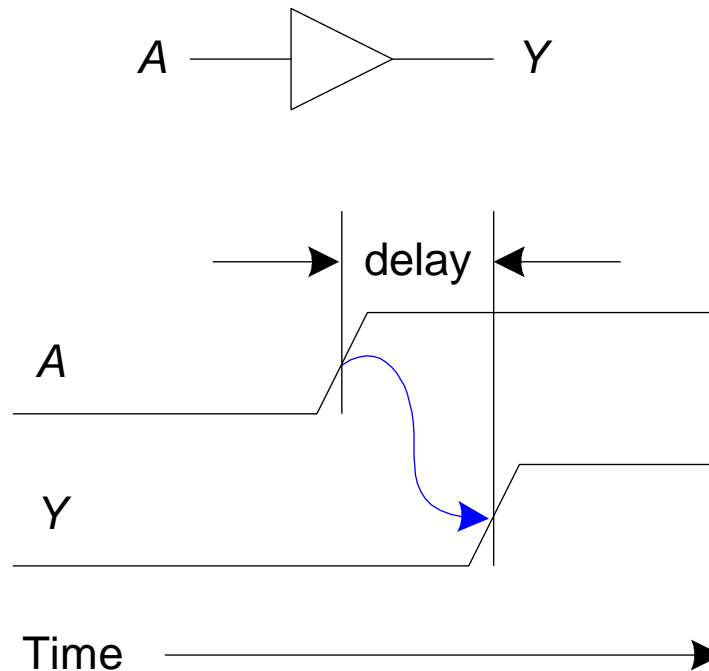


# Chapter 2: Combinational Logic

## **Timing**

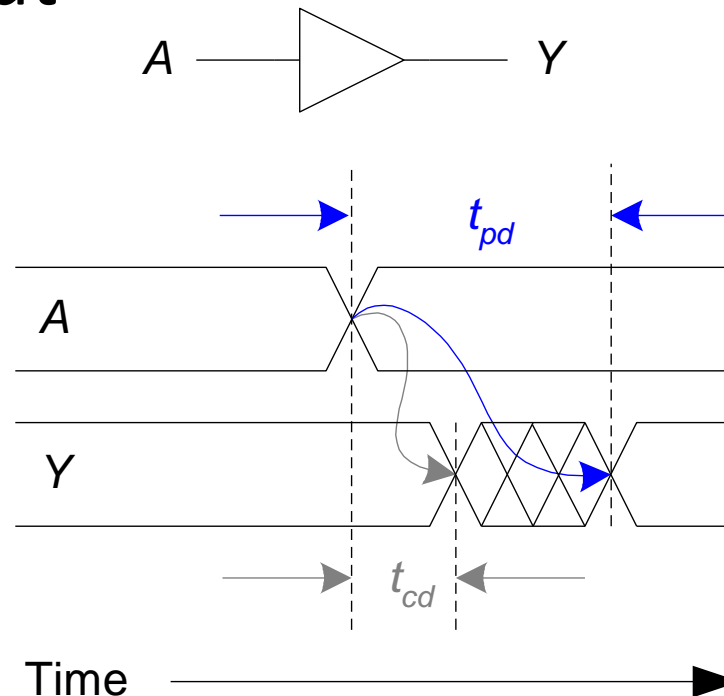
# Timing

- **Delay:** time between input change and output changing
- How to build fast circuits?



# Propagation & Contamination Delay

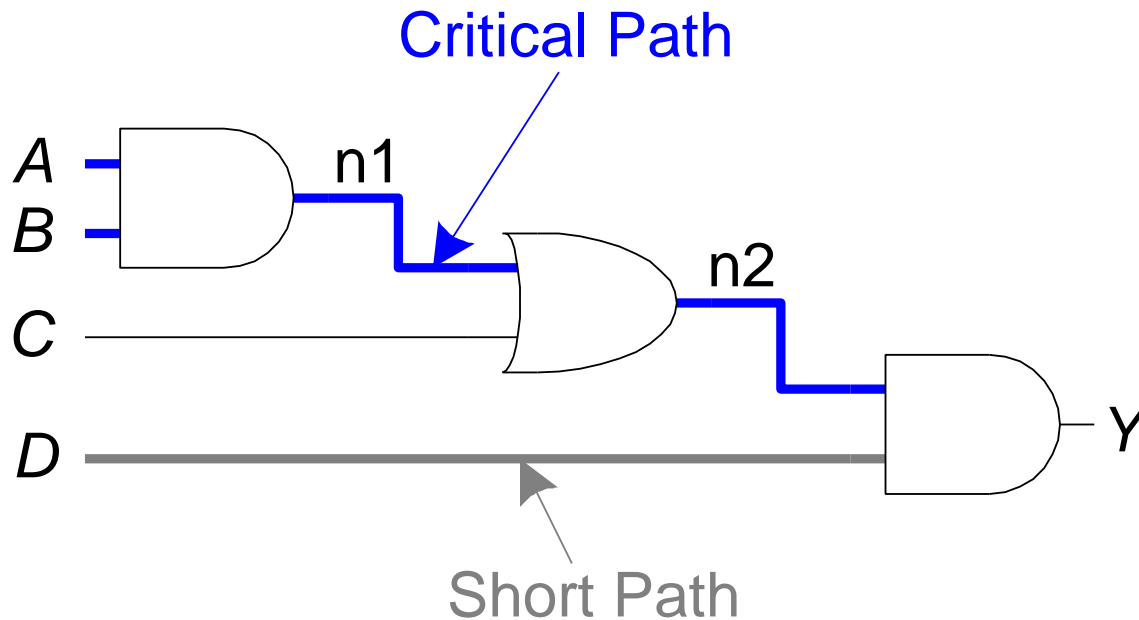
- **Propagation delay:**  $t_{pd} = \mathbf{max}$  delay from input to output
- **Contamination delay:**  $t_{cd} = \mathbf{min}$  delay from input to output



# Propagation & Contamination Delay

- **Delay is caused by**
  - Capacitance and resistance in a circuit
  - Speed of light limitation
- **Reasons why  $t_{pd}$  and  $t_{cd}$  may be different:**
  - Different rising and falling delays
  - Multiple inputs and outputs, some of which are faster than others
  - Circuits slow down when hot and speed up when cold

# Critical (Long) & Short Paths



**Critical (Long) Path:**  $t_{pd} = 2t_{pd\_AND} + t_{pd\_OR}$  (max delay)

**Short Path:**  $t_{cd} = t_{cd\_AND}$  (min delay)

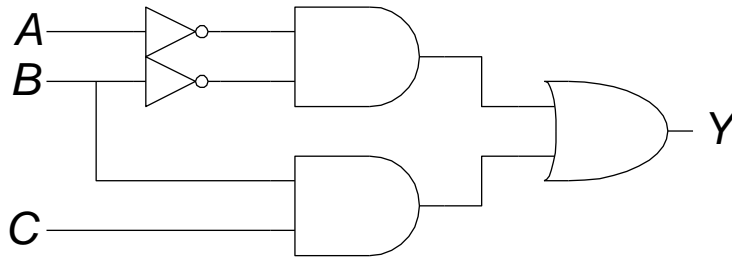
# Glitches

When a single input change causes an output to change multiple times



# Glitch Example

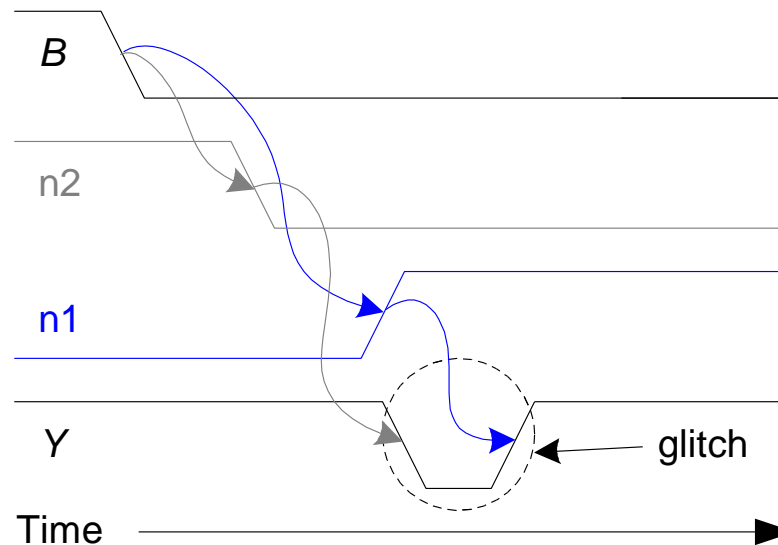
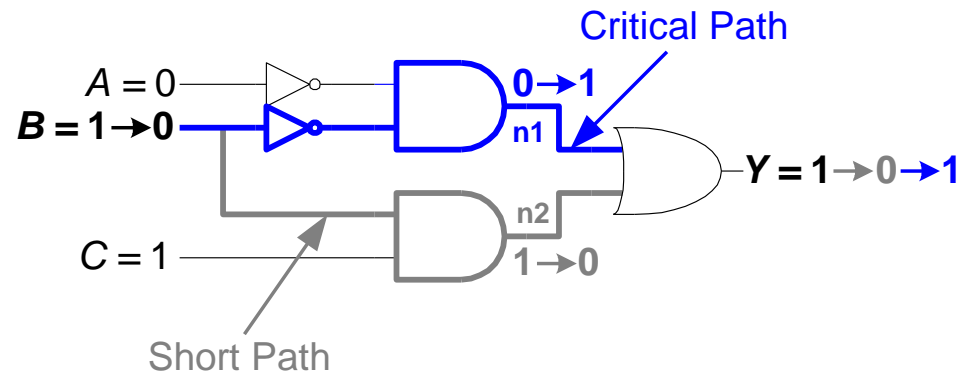
What happens when  $A = 0$ ,  $C = 1$ ,  $B$  falls?



		AB			
		00	01	11	10
C	0	1	0	0	0
	1	1	1	1	0

$$Y = \bar{A}\bar{B} + BC$$

# Glitch Example (cont.)

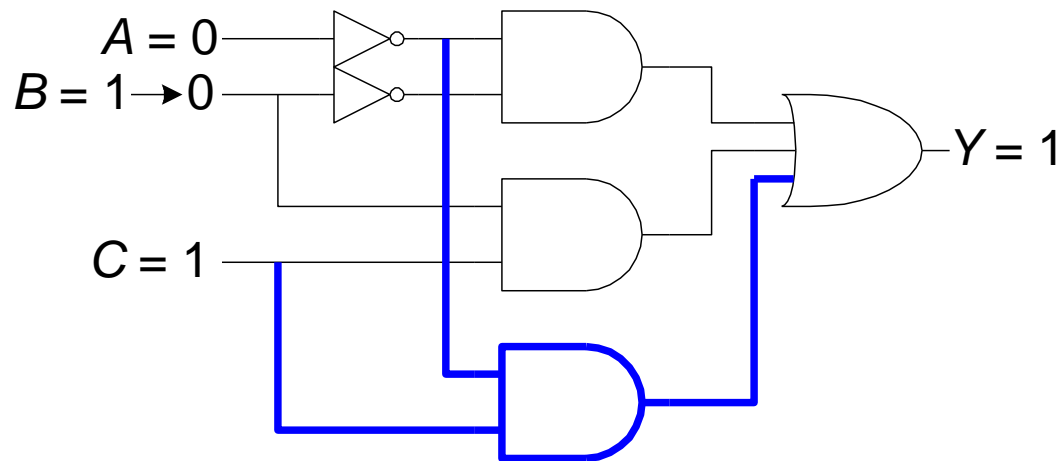


# Fixing the Glitch

		AB			
C	Y	00	01	11	10
	0	1	0	0	0
1	1	1	1	1	0

$\bar{A}C$  →

$$Y = \bar{A}\bar{B} + BC + \bar{A}C$$



# Why Understand Glitches?

- Because of **synchronous design** conventions (see Chapter 3), glitches don't cause problems.
- It's important to **recognize** a glitch: in simulations or on oscilloscope.
- We **can't get rid of all glitches** – simultaneous transitions on multiple inputs can also cause glitches.

# About these Notes

**Digital Design and Computer Architecture Lecture Notes**

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