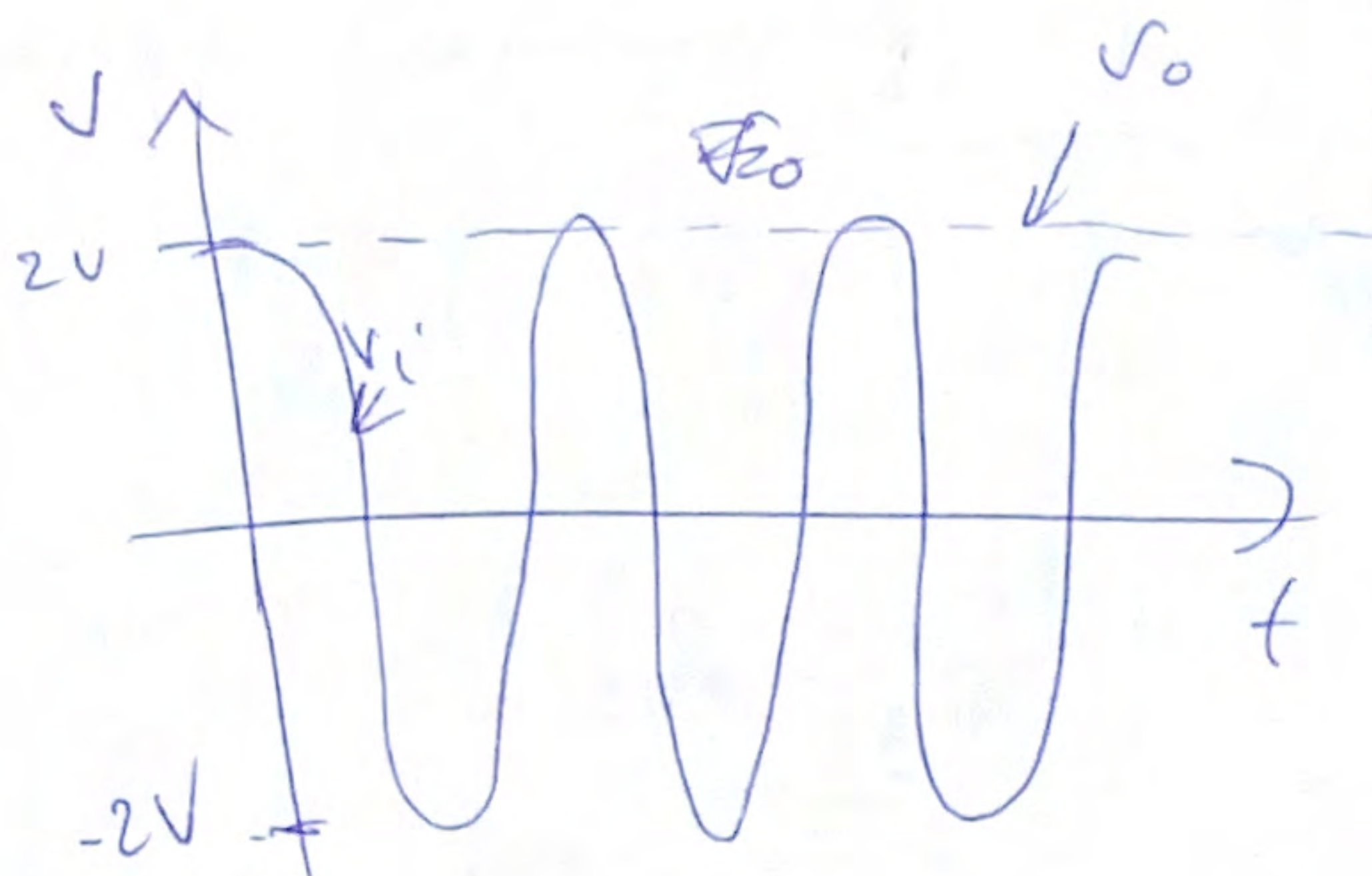
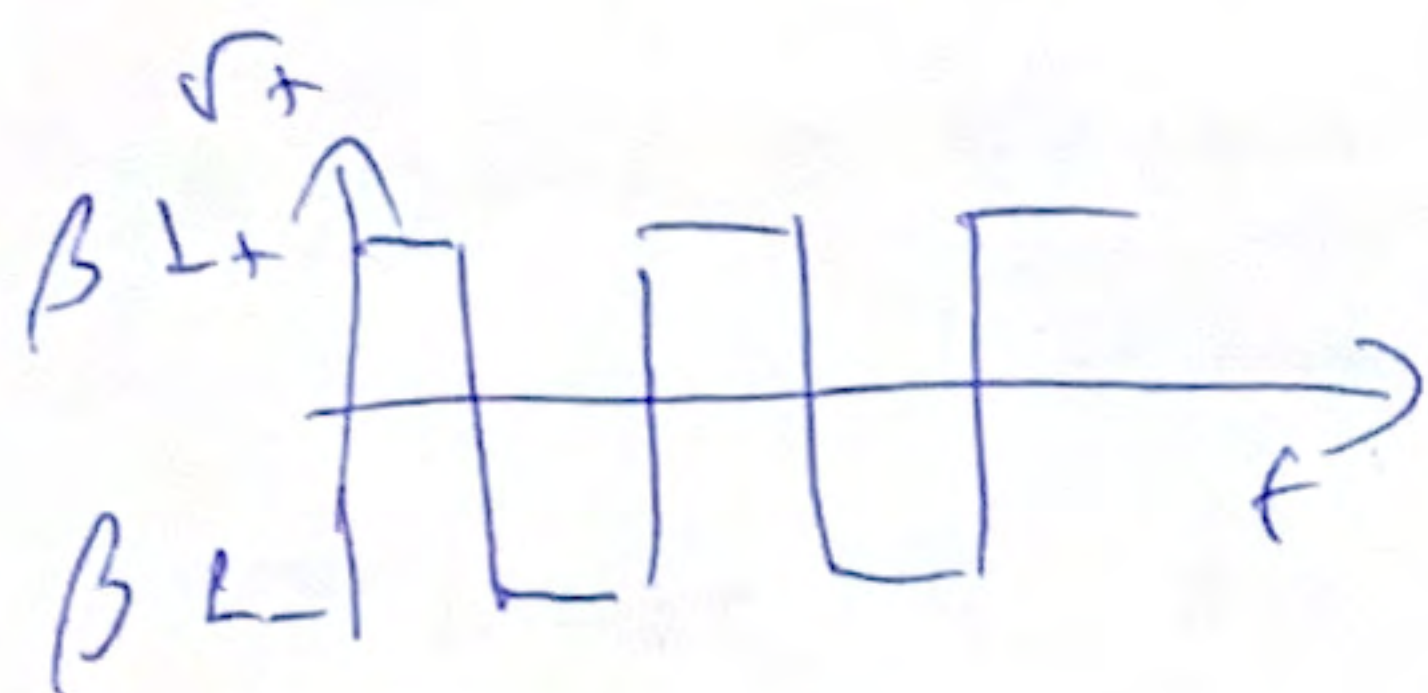


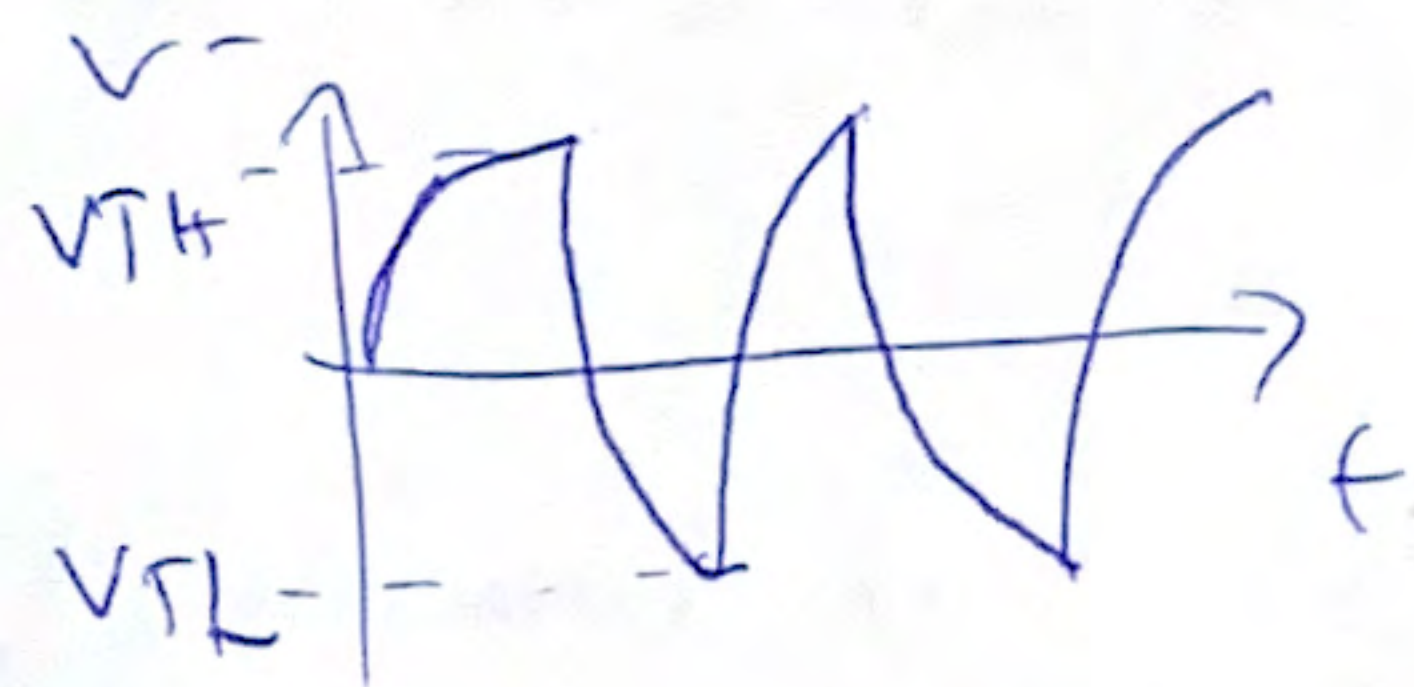
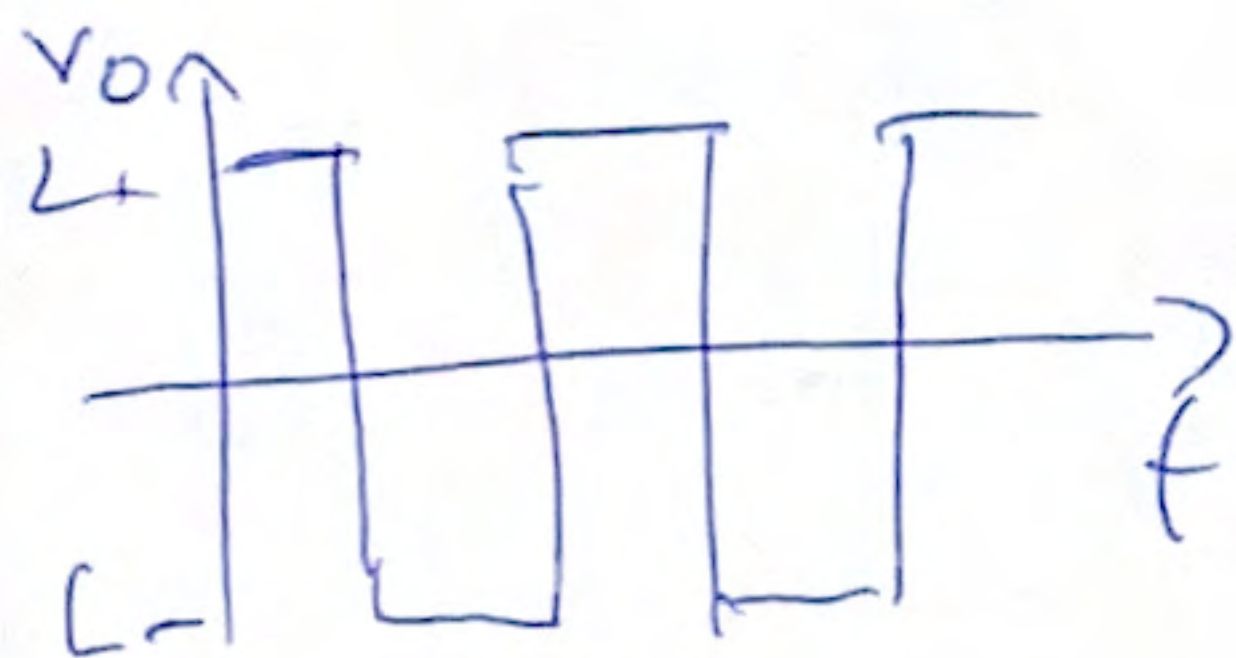
1-a)



b) c'est un oscillateur astable. R_1 et R_2 permettent d'établir les seuils V_{TH} et V_{TL} du bistable inverseur.



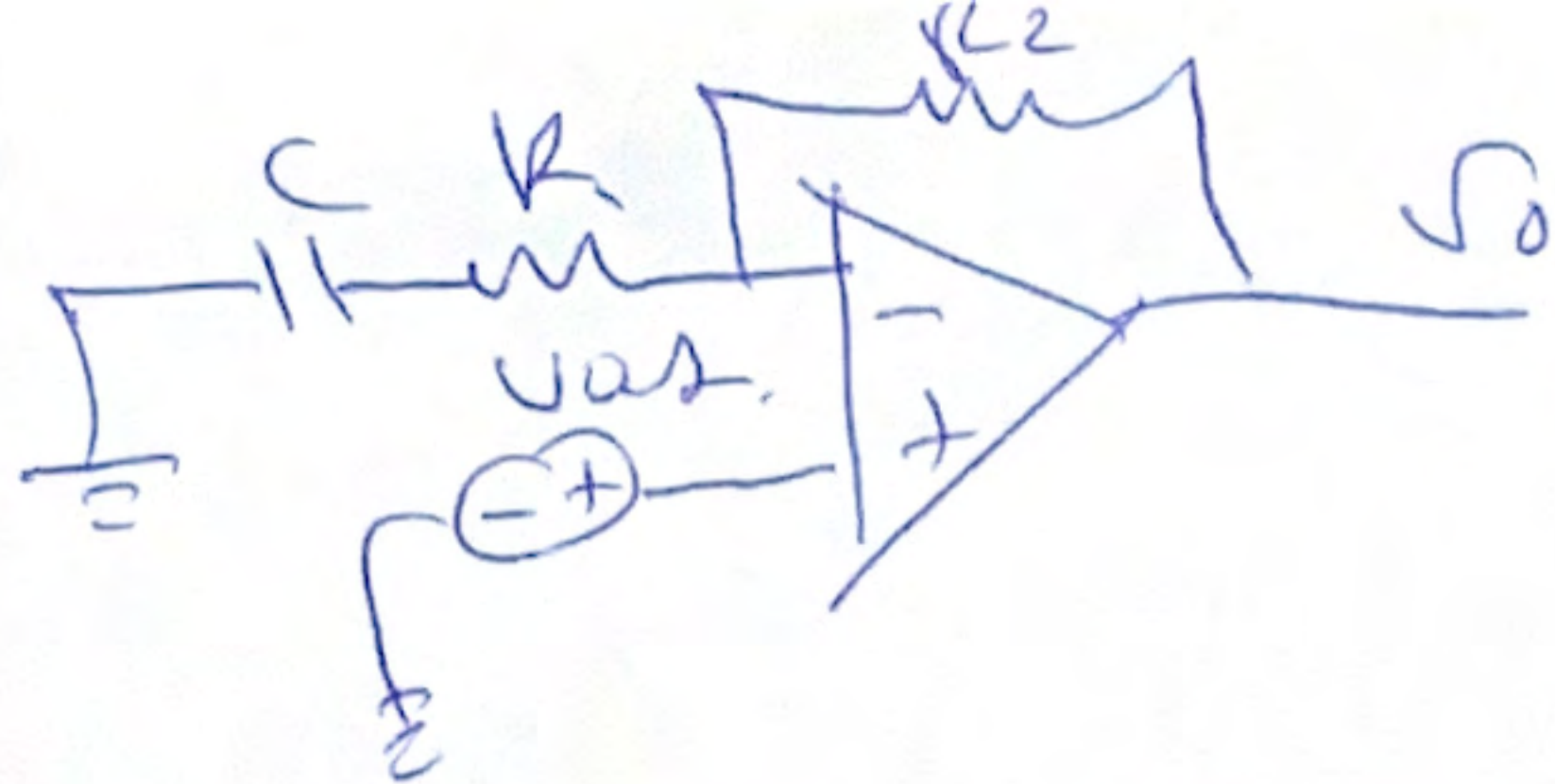
$$\beta = \frac{R_1}{R_1 + R_2}$$



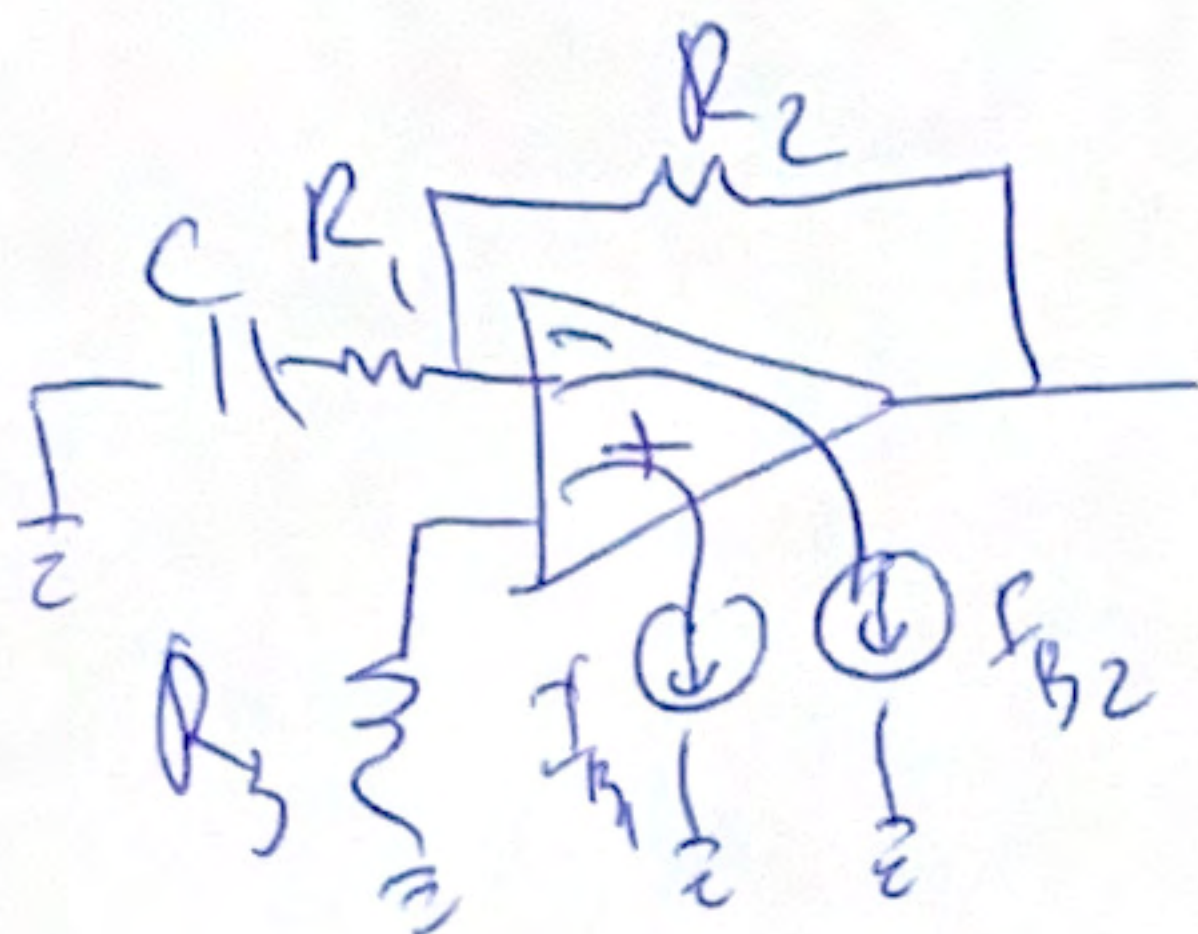
$$c) f_{-3dB} = f_t / A_{BP} = 10 \text{ MHz} / 10 = 1 \text{ MHz}$$

$$A_{BP} = 1 + \frac{900 \Omega}{100 \Omega} = 10 \text{ V/V}$$

d)



e)



$$\underline{R_3 = R_2}$$

2- a) ampli d'instrumentation

$$\frac{v_o}{v_{id}} = \frac{R_4}{R_3} \left(1 + \frac{R_2}{R_1} \right)$$

étage 1 : $1 + \frac{R_2}{R_1} = 10$: $R_2/R_1 = 9$

étage 2 : $\frac{R_4}{R_3} = 10$

b) $v_{od} = 500 \text{ mV}$ ($v_{od} = A_d \cdot v_{id}$)

c) $A_{cm} = \frac{v_{ocm}}{v_{icm}} = \frac{1 \text{ mV}}{100 \text{ mV}} = 0,01$

$$T_{RMC} = 20 \log \left(\frac{A_d}{A_{cm}} \right) = 20 \log \left(\frac{100}{0,01} \right) = \underline{80 \text{ dB}}$$

d) $Z_{in} = \infty$ car $i_{in} = 0$

e) $A_{cm}(\text{étage 1}) = 1 \text{ V/V}$ car $V_1 = V_2 = v_{icm}$
ne crée pas de courant dans $\underline{2R_1}$.

f) (bonus) $v_1 = -\frac{0,005}{2} \cos \omega_1 t + 0,1 \cos \omega_2 t$

$$v_2 = \frac{0,005}{2} \cos \omega_1 t + 0,1 \cos \omega_2 t$$

$$v_{o1} = -\frac{0,05}{2} \cos \omega_1 t + 0,1 \cos \omega_2 t$$

$$v_{o2} = \frac{0,05}{2} \cos \omega_1 t + 0,1 \cos \omega_2 t$$

$$\cancel{v_o} = \cancel{0,5} \cos \omega_1 t + \cancel{0,001} \cos \omega_2 t$$
$$v_o = 0,5 \cos \omega_1 t + 0,001 \cos \omega_2 t$$

$$3- \omega_p = 2\pi \cdot 5 \text{ kHz}, \quad \omega_s = 2\pi \cdot 15 \text{ kHz}$$

$$\epsilon = \sqrt{10^{0.5/10} - 1} = 0.34931$$

$$A_{\text{min}} = 10 \text{ dB} \leq A(\omega_s = 15 \text{ kHz}) =$$

$$10 \log \left(1 + \epsilon^2 \left(\omega_s / \omega_p \right)^{2N} \right)$$

$$a) \underline{N \geq 4}$$

$$b) \overline{T}(s) = \frac{K_1}{s^2 + 0.765s + 1} \cdot \frac{K_2}{s^2 + 1.848s + 1}$$

$$H(s) = \overline{T}(s) \Big|_{s = \frac{j\omega}{\omega_0}} = \frac{K_1 \omega_0^2}{s^2 + 0.765 \omega_0 s + \omega_0^2}$$

$$\begin{aligned} \omega_0 &= \omega_p (1/\epsilon)^{1/N} \\ &= 2\pi \cdot 5 \text{ kHz} (1/0.34931)^{1/4} = \\ &= 2\pi \cdot \underline{6.5 \text{ kHz}} \end{aligned}$$

$$\frac{K_2 \omega_0^2}{s^2 + 1.848 \omega_0 s + \omega_0^2}$$

$$Q_1 = \frac{1}{0.765}, \quad Q_2 = \frac{1}{1.848}$$

c) active Miller key.

$$\overline{T}_1(s) = \frac{a K'_1 \omega_0^2}{s^2 + s(\omega_0/Q) + \omega_0^2}$$

$$Q_1 = 1/0.765$$

$$C = 1 \text{ nF}$$

$$R_1 = R_2 = R_A = 1/\omega_0 C = 24,485 \Omega$$

$$R_B = (2 - 1/Q_1) R_A = 30,239 \Omega$$

$$K'_1 = 1 + R_B/R_A = 2.235$$

$$a = 1/K'_1 = 1/2.235$$

$$K_1 = a \cdot K'_1 = 1$$

section inductance nulle :

on donne $C_6 = C_4 = C = 1 \text{ nF}$ et $R_1 = R_2 = R_3 = R_5 = R$

$$R = \frac{1}{\omega_0 C} = \frac{1}{(2\pi \cdot 6,5 \text{ kHz})(1 \text{ nF})} = 24,485 \Omega$$

$$\frac{\omega_0}{Q_2} = \frac{1}{C_6 R_6} \quad \text{d'où } R_6 = \frac{1}{2\pi \cdot 6,5 \text{ kHz} \cdot 1 \text{ nF}} = 13250 \Omega$$

$$Q_2 = \frac{1}{1,048}$$

$$K_2 = 1$$

