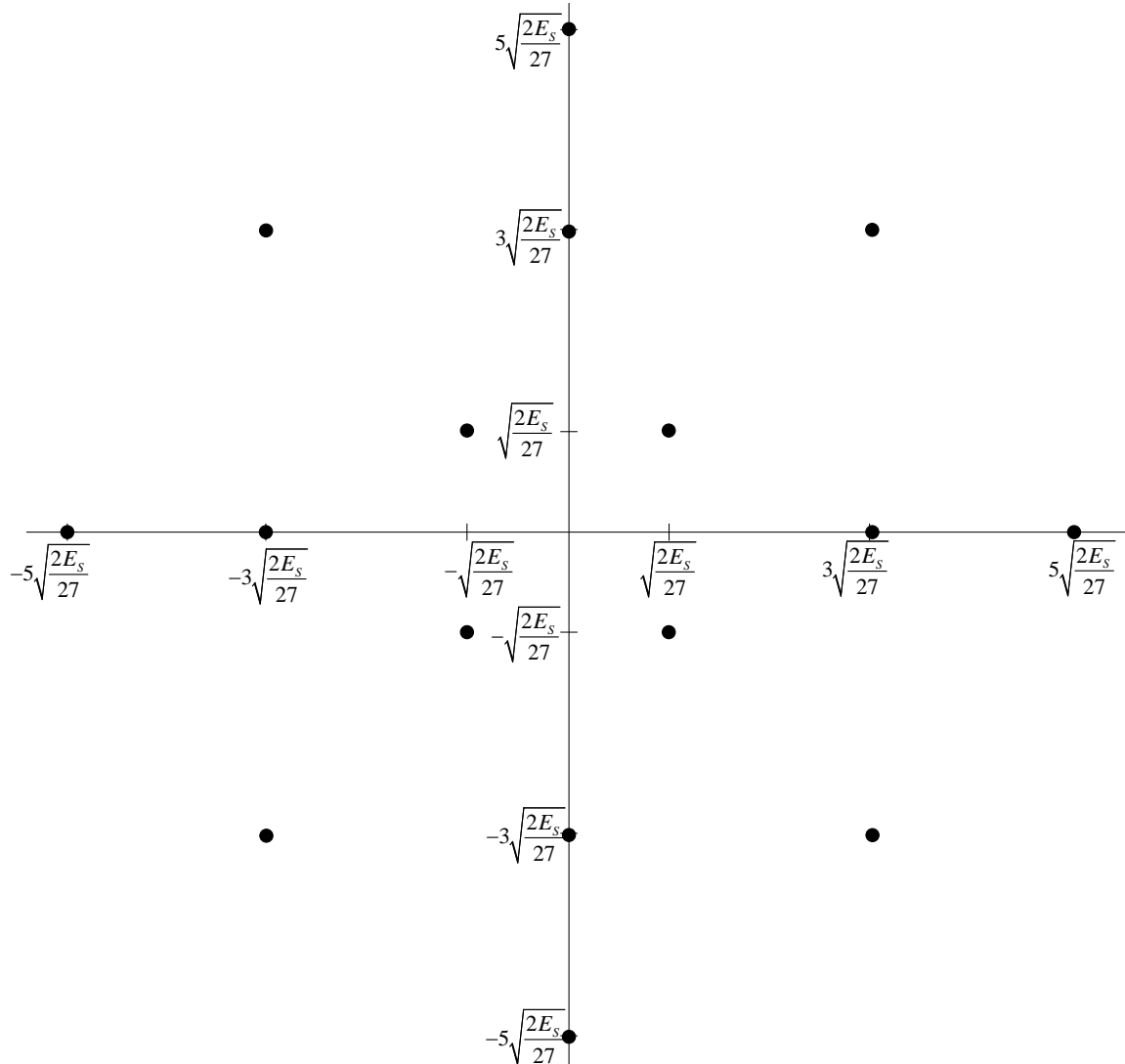


Wednesday 1 March 2017; Duration: 13:30 to 15:20

Two pages of documentation provided; a calculator is permitted.

Problem 1 (15 points out of 100)

Consider the following (non-square) 16QAM constellation shown in signal space.



- A. (5 points) Give the minimal distance.
- B. (8 points) Give the probability of error using the approximation from the union bound.

Problem 2 (10 points out of 100)

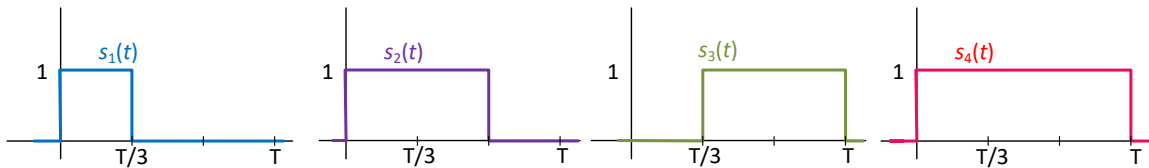
Consider the “Bandwidth Efficiency Plane”. Find the coordinates for 16QAM (square) et 8FSK (coherent) for a probability of error of 10^{-6} .

Problem 3 (20 points out of 100)

- A. (10 points) Describe the phenomenon known as ISI, intersymbol interference. Describe circumstances where ISI might be present. How does a Nyquist pulse combat ISI?
- B. (10 points) What is *a priori* probability in a communications system? How can knowledge of the *a priori* probability be used to improve system performance?

Problem 4 (30 points out of 100)

For the following set of four signals



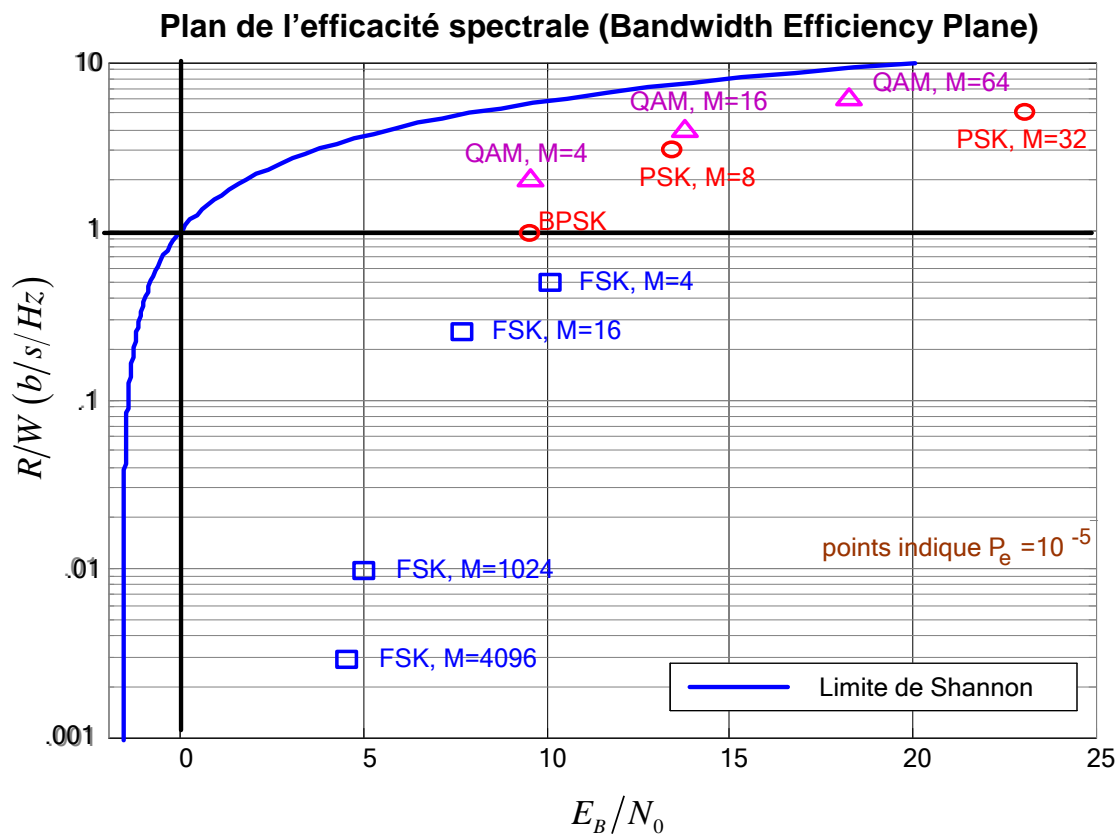
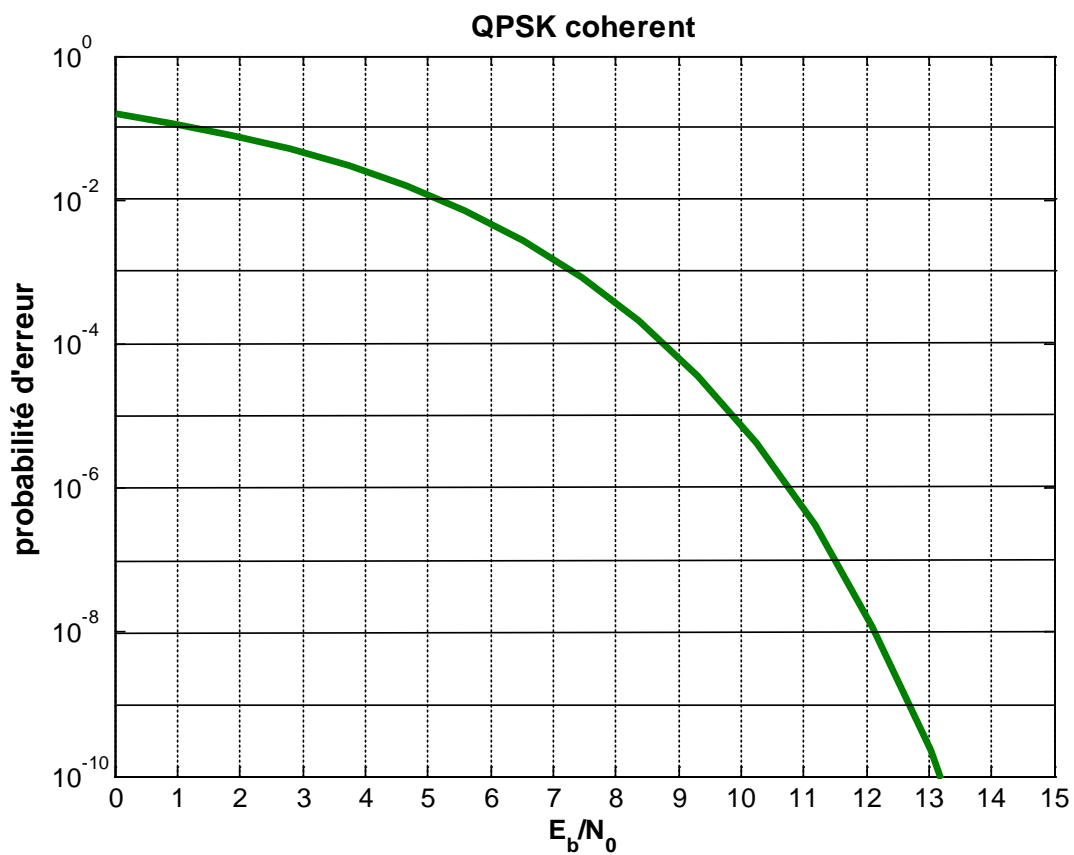
- A. (20 points) Find a set of orthonormal basis vectors.
- B. (10 points) Give the signal constellation coordinates in this basis for the signal space. The coordinates should be in terms of the average energy per bit, E_b .

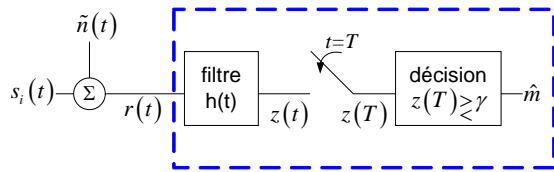
Problem 5 (25 points out of 100)

Identify a communications system from the table that is band limited. Justify your choice, that is, explain the qualities of a band limited system. Suggest an appropriate modulation format.

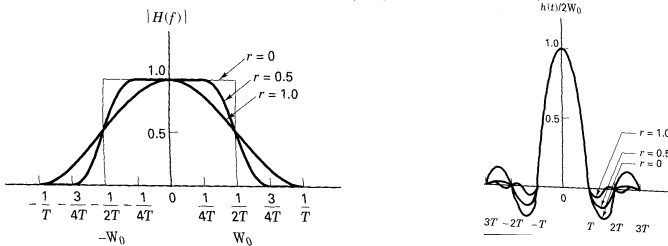
Identify a communications system from the table that is power limited. Justify your choice, that is, explain the qualities of a power limited system. Suggest an appropriate modulation format.

	Bandwidth	Bit rate	Power source	Signal strength
Twisted pair	3 kHz	10 kb/s	grid	~30 dB
Car door opener	20 MHz	10 b/s	battery	varies greatly with distance; requires line of sight
Microwave link	30 MHz	100 Mb/s	grid	good; distance chosen for good signal strength and line of sight
Mars lander	4 GHz	20 kb/s	solar	very weak, even with very high gain antennas
Cellular data	100 kHz	1 Mb/s	battery	varies greatly with distance
Cable internet	10 MHz	50 Mb/s	grid	~25 dB



Récepteur d'échantillonnage**MAP:** i qui maximise $p(z|s_i) p(s_i)$ i qui minimise $\|\mathbf{r} - \mathbf{s}_i\|^2 - N_0 \ln P(\mathbf{s}_i)$ $P(\mathbf{s}_i)$ = probabilité a priori de symbole \mathbf{s}_i **ML:** i qui maximise $p(z|s_i)$ i qui minimise $\|\mathbf{r} - \mathbf{s}_i\|^2$

Raised cosine $v(t) = \frac{\sin(\pi t/T_s)}{\pi t/T_s} \frac{\cos(r\pi t/T_s)}{1 - 4r^2 t^2/T_s^2}$

**Énergie moyenne**

$$E_{\text{moy}} = \frac{1}{M} \sum_{i=1}^M \|\mathbf{s}_i\|^2$$

$$= \frac{1}{M} \sum_{i=1}^M [\text{énergie du signal } i]$$

Énergie par bit v. énergie par symbole

$$E_b \log_2 M = E_s$$

Conversion de l'espace I/Q vers espace du signal

$$(\tilde{a}_n^I, \tilde{a}_n^Q) = \sqrt{\frac{M \cdot E_s}{\sum_{i=1}^M [(a_n^I)^2 + (a_n^Q)^2]}} (a_n^I, a_n^Q)$$

coordonnées, espace du signal (pointing to $\tilde{a}_n^I, \tilde{a}_n^Q$)

coordonnées, espace I/Q (pointing to a_n^I, a_n^Q)

QAM cas rectangulaire (carrée) $M=L^2$

$$P_e = 2 \left(1 - \frac{1}{\sqrt{M}} \right) Q \left(\sqrt{\frac{3 \log_2 M}{(M-1)} \frac{E_b}{N_0}} \right) \quad d_{\min} = \sqrt{\frac{6 \log_2 L}{L^2 - 1}}$$

Borne d'union

$$P_e \approx \frac{2K}{M} Q \left(\frac{D_{\min}}{\sqrt{2N_0}} \right) = \frac{2K}{M} Q \left(d_{\min} \sqrt{\frac{E_b}{N_0}} \right)$$

 K est le nombre des paires des signaux séparés par la distance minimale D_{\min} **Distance minimale** dans l'espace du signal

$$D_{\min} = \min_{i \neq k} \|\mathbf{s}_i - \mathbf{s}_k\| \quad \text{et} \quad d_{\min} = \frac{D_{\min}}{\sqrt{2E_b}}$$

$$P_e(BPSK) = Q \left(\sqrt{\frac{2E_b}{N_0}} \right)$$

$$P_e(OOK) = Q \left(\sqrt{\frac{E_b}{N_0}} \right)$$

$$P_e(QPSK) \approx 2Q \left(\sqrt{\frac{2E_b}{N_0}} \right)$$

Perte par rapport à QPSK

$$d_{\min} = \sqrt{x} \sqrt{2} \quad \text{perte} = -10 \log_{10} x$$

Pour une modulation orthogonale

$$P_e(\text{bit}) = P_b = P_e(\text{symbol}) \frac{M/2}{M-1}$$

Pour une modulation non-orthogonale avec codage de gray

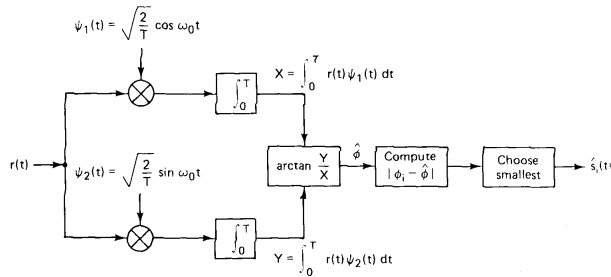
$$P_e(\text{bit}) = P_b = \frac{P_e(\text{symbol})}{\log_2 M}$$

Efficacité spectrale

$$\eta = \frac{R_b}{W} = \frac{1}{T_b} \frac{1}{W} \text{ bits/s}$$

MPSK cohérent

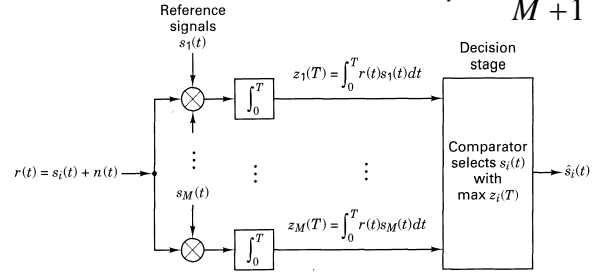
$$\eta = \log_2 M^{\dagger}$$



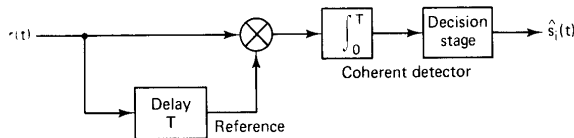
$$P_e(M) \approx 2Q\left(\sqrt{\frac{2E_s}{N_0}} \sin \frac{\pi}{M}\right) = 2Q\left(\sqrt{\frac{2E_b \log_2 M}{N_0}} \sin \frac{\pi}{M}\right)$$

MFSK cohérent

$$\eta = \frac{2 \log_2 M}{M+1}^{\dagger}$$



$$P_e = (M-1)Q\left(\sqrt{\frac{E_s}{N_0}}\right) = (M-1)Q\left(\sqrt{\frac{E_b \log_2 M}{N_0}}\right)$$

Séparation minimale $1/2T_s$ **DPSK incohérent**

~1 dB de perte entre DPSK et BPSK

$$P_e = \frac{1}{2} e^{-E_b/N_0}$$

Processus Gram Schmidt

$$\psi_1(t) = \frac{1}{\sqrt{E_1}} s_1(t) \text{ où } E_1 \triangleq \int_0^T s_1^2(t) dt$$

$$\theta_2(t) \triangleq s_2(t) - \langle s_2(t), \psi_1(t) \rangle \psi_1(t)$$

$$E_2 \triangleq \int_0^T \theta_2^2(t) dt \quad \psi_2(t) = \frac{\theta_2(t)}{\sqrt{E_2}}$$

$$i. \quad \theta_i(t) = s_i(t) - \sum_{k=1}^{i-1} \langle s_i(t), \psi_k(t) \rangle \psi_k(t)$$

$$E_i \triangleq \int_0^T \theta_i^2(t) dt \quad \psi_i(t) = \frac{\theta_i(t)}{\sqrt{E_i}}$$

Relations trigonométriques

$$\sin(\alpha \pm \beta) = \sin \alpha \cos \beta \pm \cos \alpha \sin \beta$$

$$\cos(\alpha \pm \beta) = \cos \alpha \cos \beta \mp \sin \alpha \sin \beta$$

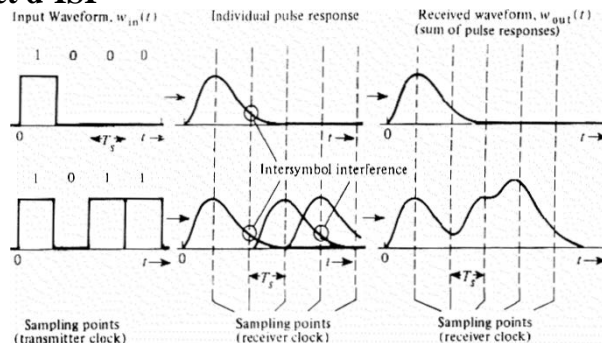
$$\cos 2\theta = 2 \cos^2 \theta - 1 = 1 - \sin^2 \theta \quad \tan \theta = \sin \theta / \cos \theta$$

Loi de Shannon

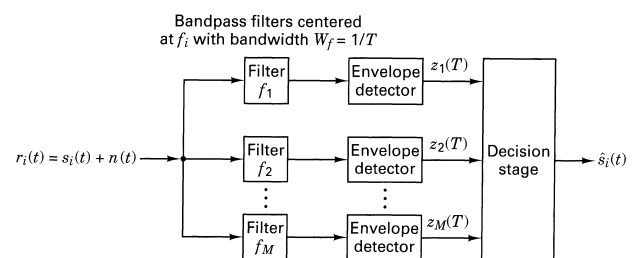
$$C = W \log_2(1 + SNR)$$

$$SNR = \frac{E_b}{N_0} \frac{R_b}{W}$$

$$\frac{E_b}{N_0} = \frac{W}{C} (2^{C/W} - 1) \quad \frac{C}{W} \rightarrow 0 \Rightarrow \frac{E_b}{N_0} \rightarrow -1.6 \text{ dB}$$

L'effet d'ISI**MFSK incohérent**

$$\eta = \frac{\log_2 M}{M}^{\dagger}$$



$$P_e(BFSK) = \frac{1}{2} e^{-E_b/2N_0}$$

~1 dB de perte entre BFSK cohérente et incohérente

Séparation minimale $1/T_s$

† en supposant une impulsion Nyquist idéale

