## MIDA1: Recap

# Model Identification and Data Analysis I **Professor**: Sergio Bittanti

Authors: Simone Staffa

Based on the notes provided by Giulio Alizoni Appendix questions provided by Moreno and @idividebyzero



Released with Beerware License, Rev. 42 (https://spdx.org/licenses/Beerware.html) "As long as you retain this notice you can do whatever you want with this stuff. If we meet some day, and you think this stuff is worth it, you can buy me a beer in return"

June 17, 2020

## Contents

1	1 Random Variables Refresh			
2	Random Process Introduction	3		
3	Familes of SSP  3.1 Moving Average Process (MA)  3.2 Auto Regressive Process (AR)  3.3 ARMA Process	3 3 4 4		
4	Spectral Representation 4.1 Fundamental Theorem of Spectral Analysis	<b>5</b>		
5	Canonical Representation of a Stationary Process	5		
6	Prediction Problem           6.1 Fake Problem            6.2 True Problem            6.3 Prediction with eXogenous Variables            6.3.1 ARX Model            6.3.2 ARMAX Model	66 66 66 7		
7	Prediction Error minimization methods           7.1 Least Square method - LS	77 78 88 88 88 99		
8	Model Complexity Selection			
9	Durbin-Levinson algorithm9.1Time series analysis made easy9.1.1MA Models9.1.2AR Models9.1.3More on PARCOV	10 10 10 10 10		
10	Recursive Least Squares - RLS	11		
11	Appendix: Example Questions	12		

#### 1 Random Variables Refresh

A random variable v can be described by three main properties:

• Mean: E[v]. Key property of expected value  $E[\alpha_1 v_1 + \alpha_2 v_2] = \alpha_1 E[v_1] + \alpha_2 E[v_2]$ 

• Variance:  $E[(v - E[v])^2] \ge 0$ 

• Standard Deviation:  $\sqrt{Var[v]}$ 

 ${\bf Random~Vector}:$  a vector composed of random variables.

The Covariance (or Cross Variance) between two elements of it can be defined as:

$$\lambda_{12} = E[(v_1 - E[v_1])(v_2 - E[v_2])] \text{ with } V = |v_1 v_2|^T$$

We can define the **Variance Matrix** as  $Var[V] = \begin{bmatrix} \lambda_{1,1} & \lambda_{1,2} \\ \lambda_{2,1} & \lambda_{2,2} \end{bmatrix}$ . This matrix has some properties:

- Symmetric
- Positive Semi-Definite:  $det(Var[v]) \ge 0$

Defining the **covariance coefficient**  $\rho = \frac{\lambda_{12}}{\sqrt{\lambda_{11}} * \sqrt{\lambda_{22}}}$ , we obtain that  $|\rho| \leq 1$ 

#### 2 Random Process Introduction

A random process (also called **stochastic process**) is a sequence of random variables. We will focus on **Stationary Stochastic Process (SSP)** which have the following characteristics:

- Mean is constant (m)
- Variance is constant  $(\lambda^2)$
- $\gamma(t_1, t_2) = E[(v(t_1) m)(v(t_2) m)]$  the **covariance function**, it depends only on  $\tau = t_2 t_1$  and can therefore be indicated with  $\gamma(\tau)$ . The variance can be also defined as  $\gamma(0) = \lambda^2$

A White Noise (WN) is a SSP with  $\gamma(0) = 0 \ \forall \tau \neq 0$ . A WN is an unpredictable signal meaning that there is NO CORRELATION between  $\eta(t_1)$  and  $\eta(t_2)$ . Usually a WN is defined as  $\eta(t) \sim WN(0, \lambda^2)$ . Having them with zero-mean is not mandatory.

#### 3 Familes of SSP

#### 3.1 Moving Average Process (MA)

$$MA(n): v(t) = c_0 \eta(t) + c_1 \eta(t-1) + \dots + c_n \eta(t-n)$$

Where n is the order of the process,  $\eta(t) \sim WN(0, \lambda^2)$ .

- By definition, **MA processes are stationary** because they are the result of the sum of stationary processes (white-noise)
- Mean is  $E[v(t)] = (c_0 + c_1 + ... + c_n)E[\eta(t)] = 0$
- Covariance function:

$$\gamma(\tau) = \begin{cases} \lambda^2 * \sum_{i=0}^{n-\tau} c_i c_{i-\tau} & |\tau| \le n \\ 0 & otherwise \end{cases}$$
 (1)

• Transfer function:  $W(z) = c_0 + c_1 z^{-1} + ... + c_n z^{-n} = \frac{c_0 z^n + c_1 z^{n-1} + ... + c_n}{z^n}$ . All poles are then in the origin of the complex plane, while the zeroes depends on the values of the coefficient.

It exists also the  $MA(\infty)$  representation:

$$v(t) = c_0 \eta(t) + c_1 \eta(t-1) + \dots + c_i \eta(t-i) + \dots + infinitesum.$$

In this way the variance in particular becomes an infinite sum as well:

$$\gamma(0) = (c_0^2 + c_1^2 + \dots + c_i^2 + \dots)\lambda^2.$$

We have then to impose the <u>basic condition</u>:  $\sum_{i=0}^{\infty} c_i^2$  needs to be FINITE, because we need to have  $|\gamma(\tau)| \leq \gamma(0)$ . Under this condition,  $MA(\infty)$  is well defined and a stationary process.

#### 3.2 Auto Regressive Process (AR)

$$AR(n): v(t) = a_1v(t-1) + a_2v(t-2) + \dots + a_nv(t-n) + \eta(t)$$

Where n is the order of the process,  $\eta(t) \sim WN(0, \lambda^2)$ .

- Mean is computed applying E[v(t)].
- Covariance function with n = 1 (case of AR(1)):

$$\gamma(\tau) = \begin{cases} a\gamma(\tau - 1) & |\tau| \ge 0\\ \frac{1}{1 - a^2}\lambda^2 & \tau = 0 \end{cases}$$
 (2)

well defined if |a| < 1 (Yule Walker Equations)

• Transfer function:  $W(z) = \frac{z^n}{z^n - a_1 z^{n-1} - \dots - a_n}$ .

Note that poles' position depends on the value  $a_1, ..., a_n$ . If all poles are inside the unit circle, then the system is **stable**.

#### 3.3 ARMA Process

$$ARMA(n_a, n_c): v(t) = a_1v(t-1) + \ldots + a_{n_a}v(t-n_a) + c_0\eta(t) + \ldots + c_1\eta(t-1) + \ldots + c_{n_c}\eta t - n_c$$

It is composed by an AR and an MA part. The process is stationary if STABLE, meaning that all the poles needs to be inside the unit circle.

An ARMA process can be expressed as  $v(t) = \frac{C(z)}{A(z)} \eta(t)$  having:

$$\begin{split} C(z) &= c_0 + c_1 z^{-1} + \ldots + c_{n_c} z^{-n_c} \\ A(z) &= 1 - a_1 z^{-1} - \ldots - a_{n_a} z^{-n_a} \\ &\to W(z) = \frac{C(z)}{A(z)} \end{split}$$

Using the Long Division Algorithm one can obtain the **impulse response** representation of it, which is a sort of  $MA(\infty)$  expression of the process, where  $W(z) = w_0 + w_1 z^{-1} + w_2 z^{-2} + ...$  as the result of the long division between C(z) and A(z).

#### 4 Spectral Representation

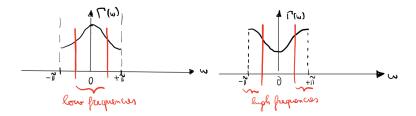
By the original definition, the spectrum of a stationary process is the  $fourier\ transform$  of its covariance function:

$$\Gamma(\omega) = F[\gamma(\tau)] = \sum_{\tau = -\infty}^{\infty} \gamma(\tau) e^{-j\omega\tau}$$

When computing the spectrum, remember the Eulero formula:  $\frac{e^{j\omega\tau}+e^{-j\omega\tau}}{2}=\cos(\omega\tau)$  Spectrum properties:

- $\Gamma(\omega)$  is a **real function** of the real variable  $\omega$
- $\Gamma(\omega)$  is **periodic** with  $T=2\pi$
- $\Gamma(\omega) = \Gamma(-\omega)$  (even function) since both  $\gamma(\bullet)$  and  $\cos(\bullet)$  are even functions
- $\Gamma(\omega) \geq 0 \ \forall \omega$

Remember that if the spectrum is higher with low frequencies, means that the signal moves very slowly. Vice versa, if the spectrum is higher with high frequencies the signal moves very quickly.



There exist also the complex spectrum:

$$\Phi(z) = \sum_{\tau = -\infty}^{\infty} \gamma(\tau) z^{-\tau} \to \Gamma(\omega) = \Phi(z = e^{j\omega})$$

One can also anti-transform the spectrum to obtain the covariance function:

$$\gamma(\tau) = \int_{-\pi}^{+\pi} \Gamma(\omega) e^{j\omega\tau} d\omega$$

The spectrum of a white noise is constant and equal to its variance. Given  $\eta(t) \sim WN(0, \lambda^2)$ 

$$\Gamma_n(\omega) = \gamma(0) = Var[\eta(t)] = \lambda^2$$

#### 4.1 Fundamental Theorem of Spectral Analysis

The spectrum of a process y that takes in input a process u is:

$$\Gamma_{uu}(\omega) = |W(e^{j\omega})|^2 * \Gamma_{uu}(\omega)$$

## 5 Canonical Representation of a Stationary Process

To solve the multiplicity of ARMA models for a stationary process, in which there are many different ARMA representations for the same process, one can use the canonical representation. From a signal, one can build the spectrum (or equivalently the covariance function). Once we have the spectrum, we can derive the canonical spectral factor to solve the prediction problem. Indeed, given a rational process, there is one and only one ARMA representation which is canonical.

Properties of Canonical Representation (referring to the transfer function W(z)):

- Numerator and denominator have the same degree
- Numerator and denominator are monic (the term with the highest power has coefficient equal to 1)
- Numerator and denominator are coprime (no common factors to that can be simplified)
- Numerator and denominator are stable polynomials: all poles and zeroes of W(z) are inside the unit disk

#### 6 Prediction Problem

#### 6.1 Fake Problem

In the fake problem we want to find the predictor from the noise. We want to compute  $\hat{v}(t+r)$  given the past of  $\eta$ . We can write:

$$\begin{split} v(t+r) &= \hat{W(z)} \eta(t+r) = \\ \hat{w_0} \eta(t+r) &+ \hat{w}_1 \eta(t+r-1) + \ldots + \hat{w}_{r-1} \eta(t+1) + \hat{w}_r \eta(t) + \hat{w}_{r+1} \eta(t-1) + \ldots = \alpha(t) + \beta(t). \end{split}$$

Since  $\eta$  is a WN and we don't know anything of its future,  $\alpha(t)$  is FULLY UNPREDICTABLE. While  $\beta(t)$  is predictable, cause it depends on the past of  $\eta$ . Thus we can write:

$$\hat{v}(t+r|t) = \hat{v}(t|t-r) = \beta(t)$$

To evaluate the performance of our prediction we can use the **prediction error** and its variance:

$$\epsilon(t) = v(t) - \hat{v}(t+r|t) Var[\epsilon(t)] = \lambda^2(\hat{w}_0^2 + \hat{w}_1^2 + \dots + \hat{w}_{r-1}^2)$$

Note that the variance of the error increases with r, meaning that more distant prediction result to be less precise.

Practically  $\hat{v}(t+r|t) = \hat{W}_r(z)\eta(t)$ , where  $\hat{W}_r(z)$  is the result of the r-step LONG DIVISION of the numerator and denominator of W(z) in canonical form.

#### 6.2 True Problem

In the true problem we want to find the predictor from data, meaning the past values of  $v(\bullet)$ . The solution to this is simply:

$$W_r(z) = \hat{W}(z)^{-1} \hat{W}_r(z)$$

with  $\hat{W}_r(z)$  being the transfer function of the predictor, W(z) being the transfer function of the original system in canonical form and  $\hat{W}_r(z)$  being the solution to the fake problem. Here "r" are the number of steps of the predictor.

There is a shortcut that can be used with ARMA process, in general:

$$\hat{v}(t|t-1) = \frac{C(z) - A(z)}{C(z)} v(t),$$
 with  $\hat{W}(z) = \frac{C(z)}{A(z)}$ 

#### 6.3 Prediction with eXogenous Variables

An exogenous variable is another input variable u(t) for the system which differently from the WN is a deterministic variable.

#### 6.3.1 ARX Model

$$ARX(n_{a,b}): v(t) = a_1v(t-1) + ... + a_{n_a}v(t-n_a) + b_1u(t-1) + ... + b_{n_b}u(t-n_b) + \eta(t)$$
 or in operator form 
$$A(z)v(t) = B(z)u(t-1) + \eta(t)$$

meaning that the transfer function from u to v is  $\frac{B(z)}{A(z)}$  and the one from  $\eta$  to v is  $\frac{1}{A(z)}$ .

#### 6.3.2 ARMAX Model

Similarly to ARX models, it is defined as  $A(z)v(t) = C(z)\eta(t) + B(z)u(t-1)...$ 

It can be represented also with the Box & Jenkins model, in which the WN is considered as a disturb (or noise) and G(z) is the effect of the exogenous variable:

$$y(t) = G(z)u(t) + W(z)\eta(t)$$

The new predictor formula becomes:

$$\hat{y}(t|t-1) = \frac{C(z) - A(z)}{C(z)}y(t) + \frac{B(z)}{C(z)}u(t-1)$$

#### 7 Prediction Error minimization methods

**Identification** consists in estimating a model from data. We can define the prediction error as before:  $\epsilon(t) = v(t) - \hat{v}(t+r|t)$ . We want to minimize this prediction error, and representing it as a WN (fully unpredictable). Steps:

- 1. Data collection: u(1), ..., u(n) and y(1), ..., y(n)
- 2. Choice of the model family:  $M(\theta)$  and corresponding  $\epsilon_{\theta}(t)$ , where  $\theta$  is a vector of parameters
- 3. Choice of optimization criterion: for example we can have

$$J(\theta) = \frac{1}{N} \sum_{t=1}^{N} \epsilon_{\theta}(t)^2$$
 mean squared error  $J(\theta) = \frac{1}{N} \sum_{t=1}^{N} |\epsilon_{\theta}(t)|$  mean absolute error

**Optimization**: model parameters are obtained with  $\theta = minJ(\theta) \rightarrow \frac{dJ(\theta)}{d\theta} = 0$ 

Validation: final analysis of the results to evaluate if they satisfy our requirements

#### 7.1 Least Square method - LS

Consider the AR or ARX model:

$$M(\theta): y(t) = a_1(t-1) + ... + a_{n_a}y(t-n_a) + b_1u(t-1) + ... + u_{n_b}(t-n_b) + \xi(t) = \theta^T\phi(t) + \eta(t)$$
  
being  $\theta = |a_1, ... a_{n_a}, b_1, ..., b_{n_b}|$  and  $\phi(t) = |y(t-1), ...y(t-n_a), u(t-1), ..., u(t-n_b)|$ .  
Considering as optimization criterion the Mean Squared Error, we impose its derivative equal to zero and we obtain the **normal equations** and the parameters estimate  $(\hat{\theta})$ .

$$\sum_{t=1}^{N} \phi(t) \phi^{T}(t) \theta = \sum_{t=1}^{N} y(t) \phi^{T}(t) \to \hat{\theta} = \left[ \sum_{t=1}^{N} \phi(t) \phi^{T}(t) \right]^{-1} * \sum_{t=1}^{N} y(t) \phi^{T}(t)$$

Depending on the value of the second derivative, the matrix  $\frac{d^2J(\theta)}{d\theta^2}$ , we can have two situations:

- 1. Positive Definite Matrix: one point of minimum  $\hat{\theta}$  and  $J(\theta)$  is a bowl (a paraboloid with vertex in  $\hat{\theta}$
- 2. Positive Semidefinite Matrix (but not positive definite): infinite many solutions  $(J(\theta))$  is similar to a section of a pipe)

Let's now consider the R(N) matrix, defined as  $R(N) = \frac{1}{N} \sum_{t=1}^{N} \theta(t) \theta^{T}(t)$ . In an ARX(1,1) we have the following:

$$R(N) = \begin{bmatrix} \frac{1}{N}y(t-1)^2 & \frac{1}{N}y(t-1)u(t-1) \\ \frac{1}{N}u(t-1)y(t-1) & \frac{1}{N}u(t-1)^2 \end{bmatrix}$$

Note that the two elements in the main diagonal are respectively the sample variance of y and the sample variance of u.

7

Bringing 
$$N \to \infty$$
:  $\bar{R} = \begin{bmatrix} \bar{R}_{yy} & \bar{R}_{yu} \\ \bar{R}_{uy} & \bar{R}_{uu} \end{bmatrix} = \begin{bmatrix} \gamma_{yy}(0) & \gamma_{yu}(0) \\ \gamma_{uy}(0) & \gamma_{uu}(0) \end{bmatrix}$ 

In a general  $ARX(n_a,n_b)$  model, we obtain that:

$$\bar{R}_{uu} = \begin{bmatrix} \gamma_{uu}(0) & \gamma_{uu}(1) & \gamma_{uu}(2) & \dots \\ \gamma_{uu}(1) & \gamma_{uu}(0) & \dots \\ \gamma_{uu}(2) & \dots & \dots \end{bmatrix}$$

This matrix is called **Toeplitz matrix**: inside each diagonal there's the same element; the one inside the main diagonal is the variance. (Note that for  $\bar{R}_{yy}$  is the same but with  $y(\bullet)$ ). Our aim is to have  $\bar{R}_{uu}$  invertible (positive semi-definite) so that the normal equations will have a unique solution.

#### 7.2 Estimation of mean, covariance and spectrum

In this section we refer to a generic estimator from the N data (e.g., the sample mean or the sample covariance) as  $\hat{s}_N$ .

#### 7.2.1 Mean

#### Correctness:

 $\hat{s}_N$  is a correct estimator if  $E[\hat{s}_N] = \bar{s}$ . That is, the expected value of the estimator is equal to the probabilistic mean to be estimated.

#### Consistency:

 $\hat{s}_N$  is a consistent estimator if  $Var[\hat{s}_N] \to 0$  as  $N \to 0$ . That is, the estimate error variance tends to zero as the number of measured data tends to infinity.

#### 7.2.2 Spectrum

Consider  $\hat{\Gamma}$  the spectrum estimator. It is **asymptotically correct** because:

$$E[\hat{\Gamma}(\omega)] \to \Gamma(\omega) \text{ as } N \to \infty$$

but not consistent:

 $Var[\hat{\Gamma}(\omega)] \to \Gamma(\omega)^2 \text{ as } N \to \infty.$ 

The problem of inconsistency can be tackled using the Bartlett method.

#### 7.3 Maximum Likelihood method - ML

This method is based on ARMA or MA or ARMAX models instead of AR and ARX: no more linearity in the parameters and no normal equations.

$$M(\theta): A(z)y(t) = B(z)u(t-1) + C(z)\eta(t)$$

The procedure is still the same to find  $\hat{\theta}$  from data is the same: collect data, compute  $J(\theta)$  and then minimize it. The performance index  $J(\theta)$  based on the prediction error still can be the mean square error. Differently from the LS method, the function is now non-convex, thus we need an iterative method to solve the minimization problem.

#### 7.3.1 The Newton Method

Let's suppose without loss of generality, that  $\theta$  is a scalar. The basic idea of this iterative procedure is to **approximate**  $J(\theta)$  with a quadratic function  $V(\theta)$ . The minimum of this function for the  $r^{th}$  iteration will be considered as the estimated vector of parameters  $\hat{\theta}^{r+1}$  of the following iteration.

By letting  $r \to \infty$  we obtain the minimum  $\bar{\theta}$ . But there is no guarantee that the minimum found is a global minimum. One simple method to deal with this problem is to execute multiple times the algorithm with different initializations and take the best among the different runs. Procedure:

- 1. At iteration "r" we have to estimate  $\hat{\theta}^r$
- 2. From  $\hat{\theta}^r$  we can obtain  $A^r(z), B^r(z), C^r(z)$
- 3. We can obtain  $\epsilon^r(t)$  by means of those:  $C^r(z)\epsilon^r(t) = A^r(z)y(t) B^r(z)u(t-1)$

- 4. Filter data to obtain the gradient vector  $\Psi^r(t) = -\frac{d\epsilon^r(t)}{d\theta}$
- 5. Use Gauss-Newton formula to compute  $\hat{\theta}^{(r+1)} = \theta^{(r)} + (\sum_t \Psi(t) \Psi^T(t))^{-1} \sum_t \Psi(t) \epsilon(t)$
- 6. Repeat until convergence

#### 7.4 Performance of prediction error identification methods

If we construct the prediction error as usual:

$$\epsilon_{\theta}(t) = y(t) - \hat{y}_{\theta}(t)$$

Both y(t) and  $\hat{y}_{\theta}(t)$  are sequence of points. Thus the performance index  $J(\theta)$  depends on specific points that are provided. To highlight it, we add the subscript N:

$$J_N(\theta) = \frac{1}{N} \sum_{t=1}^{N} \epsilon_{\theta}(t)^2$$

Then also the estimated parameters  $\hat{\theta_N}$  depends on the data points.

If the prediction error can be seen as a stationary process, we expect that with  $N \to \infty$ :

$$J_N(\theta) \to \bar{J}(\theta) = Var[\epsilon_{\theta}(t)] \\ \hat{\theta}_N \to \bar{\theta}$$

 $\bar{J}(\theta)$  does not depend on the particular outcome of the random experiment and so will have a unique minimum.

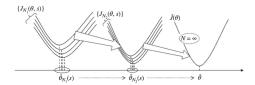


Figure 1: When  $N \to \infty$  we have a unique deterministic curve

The performance of a prediction error identification method can be studied asymptotically by analysing the properties of  $M(\bar{\theta})$ . Having a system  $S \in M(\theta)$ , we can define  $S = M(\theta^{\cdot})$  with  $\theta$  fixed and we can study the rate of convergence of  $\hat{\theta_N}$  to  $\theta^{\cdot}$  by performing the variance of their difference.

## 8 Model Complexity Selection

After computing  $\hat{\theta}_N$ , we have  $J_N(\hat{\theta}_N)^{(n)}$  for a model of order n. How can we select the best value of n, namely the best model complexity?

- Naive approach: compute the performance index  $J_N$  for multiple increasing values of n until we find the best performance
- Cross Validation: divide the dataset with identification and validation set; use the former to build the model and evaluate the performance index using the validation set (we are wasting some of the data, because they canno be used in the identification process)
- Final Prediction Error:  $FPE = \frac{N+n}{N-n} J_N(\hat{\theta}_N)^{(n)}$  We are giving a penalty to the models with high complexity. The FPE functions is not monotonically decreasing, and the complexity corresponding to its minimum value can be chosen as complexity of the model
- Akaike Information Criterion:  $AIC = \frac{2n}{N} + \ln J_N(\hat{\theta}_N)^{(n)}$ For high values of N, this is equivalent to FPE. The first term is the one regarding the complexity of the model, while the second is the one regarding the fitting of the data
- Minimum Description Length:  $MDL = \frac{n}{N} \ln N + \ln J_N(\hat{\theta}_N)^{(n)}$ Asymptotically is similar to AIC but with lower penalization  $\to$  MDL leads to more parsimonious models

The Final Predictor Error – FPE 
$$n$$
 is the number of parameters to identify  $N$  is the number of samples of the realization  $N$  is the number of samples of the realization is the sampled variance of the prediction error. The Akaike Information Criterion – AIC 
$$AIC(n) = 2\frac{n}{N} + \ln[J_N(\hat{\theta}_N)] \qquad \qquad \text{If the number of samples $N$ is large, FPE and AIC tend to indicate the same order.}$$
 The Minimum Description Length - MDL 
$$MDL(n) = \ln(N)\frac{n}{N} + \ln[J_N(\hat{\theta}_N)] \qquad \qquad \text{is penalized with AIC and MDL.}$$

#### 9 Durbin-Levinson algorithm

A recursive algorithm to estimate the parameters of an AR(n+1) starting from an AR(n) model. Using this algorithm we avoid to invert several matrices, which is an expensive procedure. Procedure for  $AR(n) \to AR(n+1)$ :

$$\begin{split} a_{n+1}^{(n+1)} &= \frac{1}{\lambda_{(n)}^2} \left[ \gamma(n+1) - \sum_{1}^n a_i^{(n)} \gamma(n+1-i) \right] \\ a_i^{(n+1)} &= a_i^{(n)} - a_{n+1}^{(n+1)} a_{n+1-i}^{(n)}, \\ \lambda_{(n+1)}^2 &= \left[ 1 - (a_{n+1}^{(n+1)})^2 \right] \lambda_{(n)}^2. \end{split}$$

#### 9.1 Time series analysis made easy

Now we can see easily estimate the order of MA and AR processes.

#### 9.1.1 MA Models

We know that  $\gamma_y(\tau) = 0$  where  $\tau > n$ . From that we can determine the order of our MA model by finding the value of  $\tau$  for which the covariance function goes to zero.

#### 9.1.2 AR Models

In this case the  $PARCOV(\tau)$  function may be useful. Considering the two models:

$$\begin{split} AR(k-1):y(t) &= a_1^{(k-1)}y(t-1) + \ldots + a_{k-1}^{(k-1)}y(t-k+1) + \eta(t) \\ AR(k):y(t) &= a_1^{(k)}y(t-1) + \ldots + a_k^{(k)}y(t-k) + \eta(t) \\ PARCOV(\tau) &= a_{\tau}^{(\tau)} \end{split}$$

Namely, the PARtial COVariance function is the last identifier parameter  $a_{\tau}$  of an  $AR(\tau)$  mode. One then can estimate  $\gamma(\tau)$  and then  $PARCOV(\tau)$  to finally decide the order of the system. Indeed, if the order of the "true" AR model is n then:

$$PARCOV(\tau) = 0 \ \forall \tau > n$$

#### 9.1.3 More on PARCOV

In general the partial covariance function can be used to determine if the model to be used should be an AR or an MA:

- if, at a certain point, the covariance function goes to zero before the partial covariance does, the process is an MA
- if the partial covariance function goes to zero first, the process is an AR

### 10 Recursive Least Squares - RLS

All the algorithms seen so far are batch methods that work "offline", that use all the data at once. Now we will see a recursive method that is able to update the estimate by adding new data. The latter methods help to overcome the limitation of when the data are coming some at a time, thus this method can work "online".

 $\mu$  is called **forgetting factor**, and is a parameters that allows to give more importance to the new data with respect to the old one. Indeed it is applied to the original LS formula in this way:

$$S(t) = \mu S(t-1) + \phi(t)\phi^{T}(t)$$

Note that if  $\mu=1$  the method is the normal LS estimation.

## 11 Appendix: Example Questions

The variance matrix is a Toeplitz matrix.	False. The Toeplitz matrix is used to guarantee that an even function $\gamma_y(\tau)$ is indeed a covariance function. Hence, $\gamma_y(\tau)$ is the covariance function of a stationary process if the associated Toeplitz matrix is positive semi-definite for each N. In such case, $\gamma_y(\tau)$ is a positive semi-definite function.
For a generic random process, its covariance matrix is a Toeplitz matrix?	<b>False</b> . It would be true if we assume the process to be stationary.
The covariance coefficient of a random vector of length 2 is: $\frac{\sqrt{\lambda_{11}*\sqrt{\lambda_{22}}}}{\lambda_{12}}$	False. $\rho = \frac{\lambda_{12}}{\sqrt{\lambda_{11}} * \sqrt{\lambda_{22}}}$
The variance matrix is positive definite.	False. It is positive semi-definite.
A random process depends only on time.	False. A random process is indexed with $t$ but depends also on the single outcome of a random experiment $s$ .
The Covariance concept is different from cross variance.	False. They are the same.
The whiteness property states that if a process has its real spectrum equal to a $K > 0$ in zero and 0 elsewhere, the it is a WN.	False. The whiteness property refers to the covariance of the process. The spectrum is instead a real and even function of $\omega$ , periodic of $2\pi$ , with values always greater than or equal to 0.
The whiteness property states that the covariance function of a $WN \sim (0, \lambda^2)$ is: $\gamma(\tau) = \begin{cases} \lambda^2 & \tau = 0\\ 0 & otherwise \end{cases}$	<b>True</b> . The covariance function of a WN process is an impulsive function, centered in zero.
The real spectrum of a WN is its variance	<b>True.</b> From the definition of spectrum: $\Gamma(\omega) = \sum_{\tau=-\infty}^{\infty} \gamma(\tau) e^{-j\omega\tau}$ . The only term remaining from the sum is the one in zero, namely $\gamma(0) = \lambda^2$ (because the others are null).
Given a spectrum with predominant high frequencies (high values among $\pi$ and $-\pi$ ), the (a) graph is the most likely signal from which it is obtained.	
2 1×(C)	<b>True</b> . Since the high frequencies are predominant, we see that a has very fast changes of the intensity that's why it is the most likely realization of the process.
Given $\eta(t) \sim WN(0, \lambda^2)$ and a process with the following behavior: $v(t) = c_0 \eta(t) + c_1 \eta(t-1), c_0, c_1 \in \mathbb{R}.$ Its complex spectrum is equal to $(c_0^2 + c_1^2)\lambda^2$ .	False. $v(t)$ is an MA(1) process. We know that the variance of such process is $Var[v(t)] = (c_0^2 + c_1^2)\lambda^2$ . Thus this is not the spectrum.

An $AR(2)$ process has 2 poles.	True. $AR(2): v(t) = a_1v(t-1) + a_2v(t-2) + \eta(t)$ . Its transfer function is $W(z) = \frac{z^2}{z^2 - a_1z - a_2}$ . Indeed, its denominator has 2 solutions, that corresponds to the poles. Moreover such an AR(2) will be stable if $ a_2  < 1$ , since that parameter is the product of the roots of the denominator.
Given $\eta(t) \sim WN(0, \lambda^2)$ and a process with	
the following behavior:	
$v(t)=c_0\eta(t)+c_1\eta(t-1),\ c_0,c_1\in {\rm I\!R}.$ Its real spectrum will never have zeroes.	False. That process is an MA(1). The spectrum of such process is: $\gamma(\omega) = [(c_0^2 + c_1^2 + 2c_0c_1cos(\omega))]\lambda^2$ . Thus, it can be null for some values of $c_0$ and $c_1$ . For example, when $c_0 = c_1$ and $\omega = \pi$ , the spectrum is null.
Given $\eta(t) \sim WN(0, \lambda^2)$ and a process with the following behavior:	
$v(t) = c_0 \eta(t) + c_1 \eta(t-1), c_0, c_1 \in \mathbb{R}.$ Its complex spectrum is: $(c_0^2 + c_1^2)\lambda^2 + 2c_0c_1\lambda^2cos(\omega)$	False. That is the formula to compute the real spectrum for an MA(1) process.
An AR(2) process has always 2 zeroes.	True. It has 2 zeroes both in zero.
An $MA(\infty)$ process is the same as an $AR(1)$ ?	False. Actually, an $AR(n)$ can be reduced to an $MA(\infty)$ applying some constraints on the coefficients $( a  < 1)$ .
To obtain Yule Walker Equations from a process $v(t)$ , you can sum to both parts of the equation $v(t-\tau)$ and then compute the covariance function.	False. You should multiply, not sum.
An $AR(10)$ process has no zeroes in its transfer function.	False. It has 10 zeroes, all in zero.
An $MA(10)$ process may have no zeroes or poles in its transfer function.	False. It always have both.
An $ARMA(1,1)$ may have no zeroes or poles.	True. If they can be simplified or if it would have been in canonical form it could have been true.
The complex spectrum is the same thing as the real spectrum.	False. The real spectrum is equivalent to the complex spectrum evaluated for $z = e^{j\omega}$ .
Given any transfer function with input $u$ and output $y$ , is it possible to invert the transfer function?	False. In order to invert it, it should be in canonical form or at least NUM and DEN should have same degree, with poles and zeroes inside unit circle.
Every process has a canonical representation.	False. Processes that are not stochastic and stationary don't have it.
Every stationary sthocastic process has a canonical representation.	True. In particular, given a rational process there is one and only one ARMA representation of it which is canonical. This special representation is used to solve the problem of multiplicity of ARMA models for which there are many different ARMA models for the same process.
For every possible process, we can write a predictor.	True.
Yule-Walker equations can be applied only to $AR(N)$ processes.	False. They can be used on $ARMA$ and $MA$ too, but could be much less useful (especially in $MA$ ).
Is FPE monotonically decreasing?  AIC achieves the minimum for lower orders w.r.t. MDL.	False. None of those methods are. False. Is the contrary. MDL achieves the minimum for lower order w.r.t. AIC.
dictor.	<ul> <li>False. They can be used on ARMA and MA too, but could be much less useful (especially in MA).</li> <li>False. None of those methods are.</li> <li>False. Is the contrary. MDL achieves the min-</li> </ul>

Anderson Whiteness Test allows you to con-	
clude that the prediction error is certainly a	False. It only provides a probability about it.
WN.	raise. It only provides a probability about it.
In the end, to see if the estimated model is	True. We want to reduce our prediction er-
valid, we need to detect if the prediction error	ror to a white noise because to improve our
is a WN.	estimation, it should be fully unpredictable.
To model systems, we use AR, MA and ARMA	False. We use them for time series analysis.
model families.	V
To model time series, we use ARX and AR-	False. We use them for systems.
MAX.	
Least squares method can be used with	False. LS method is only for AR/ARX mod-
ARMA, ARMAX and MA.	els.
Let's consider the matrix $R(N) =$	False. Only if the matrix is positive semi-
$\frac{1}{N}\sum_{t=1}^{N}\phi(t)\phi^{T}(t)$ . It always have only	definite and also invertible, then the normal
one solution.	equations have a unique solution.
An estimator is consistent if its variance is the	False. An estimator is consistent if the error
same of the real WN.	variance tends to 0 as $N \to \infty$ .
	False. An estimator is correct if its expected
An estimator is correct when its expected	value is the probabilistic mean to be esti-
value is the same of the white noise.	mated.
Maximum Likelihood has non-linear parame-	True. Instead the Least Squares method is
ters.	linear in parameters.
ARMA/ARMAX are better to be estimated	False. Only AR/ARX can be used with Least
with Least Squares method.	Squares.
To estimate the covariance, we apply the	False. With Bartlett we estimate the spec-
Bartlett method.	-
	trum.
The reason why Bartlett method uses large	TD.
amount of data is because it requires to build	True.
many subseries of data.	7.1
The Newton method guarantees to find global	False. There is no guarantee that the mini-
minimum.	mum found is a global minimum.
A model of a process is reliable if its error	True. The error should be a fully unpre-
w.r.t. the measured data is a WN.	dictable signal.
FPE penalizes model complexity.	True.
AIC penalizes model complexity as much as	
FPE with N big, while it penalizes more than	True.
FPE with small N.	
MDL penalizes model complexity more than	The state of the s
AIC.	True.
The variance of a correct and consistent esti-	False. The variance of its error must be zero
mator is 0.	when $N \to \infty$ .
110001 10 01	