



UNIVERSIDAD NACIONAL
AUTÓNOMA DE MÉXICO
FACULTAD DE INGENIERÍA
CIRCUITOS ELÉCTRICOS
SEMESTRE 2020 - 2

Tarea y Apuntes 4

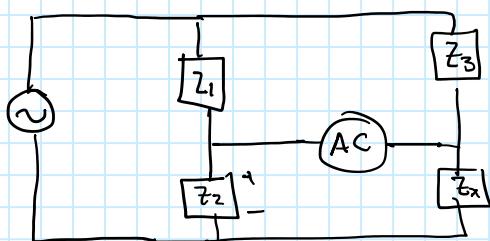
Profesor:
Dr. Juan Carlos Martínez Rosas

Integrantes:
Murrieta Villegas Alfonso

TAREA_4

Monday, 6 April 2020 4:24 PM

- Un circuito puente en A.C es utilizado en la medición de la Inductancia L de un inductor o la capacitancia de un capacitor



tal puente es un circuito balanceado, cuando no fluye ninguna corriente a través del medidor en A.C.

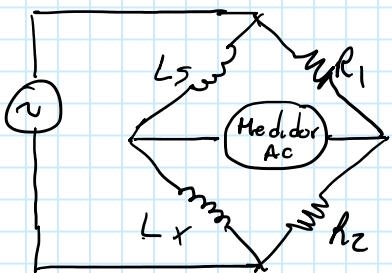
/ Divisor de voltaje

$$V_1 = \frac{Z_2}{Z_1 + Z_2} V_s = V_2 = \frac{Z_x}{Z_3 + Z_x} V_s$$

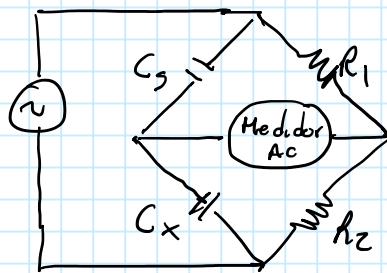
// Reduciendo

$$\frac{Z_2}{Z_1 + Z_2} = \frac{Z_x}{Z_3 + Z_x}; \\ Z_3 Z_2 = Z_x Z_1$$

$$\therefore Z_x = \frac{Z_3}{Z_1} Z_2 \quad \text{--- Equación Balanceada para un puente en A.C.}$$



$$L_x = \frac{R_2}{R_1} L_s$$



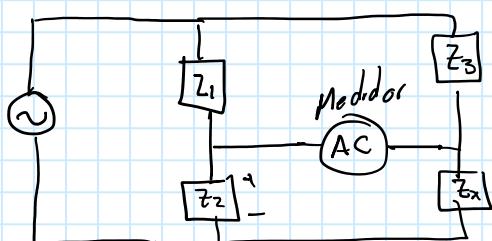
$$C_x = \frac{R_2}{R_1} C_s$$

- L_x y C_x son desconocidos mientras L_s y C_s si lo son, en ciertos casos R_1 y R_2 son variados hasta que el medidor tenga una lectura nula.

→ Ejemplo

El siguiente circuito está balanceada cuando Z_1 es un resistor de $1[\text{k}\Omega]$ y Z_2 de $4.2[\text{k}\Omega]$, mientras que Z_3 es una combinación paralelo de un resistor de $1.5[\text{M}\Omega]$ y un capacitor de $12[\mu\text{F}]$ y una $f = 2[\text{kHz}]$

- Las componentes en serie que hagan Z_X
- Las componentes en paralelo que hagan Z_X



Circuito B

Puesto que

$$R_3 = 1.5 \text{ M}\Omega \quad \text{y} \quad C_3 = 12 \mu\text{F}$$

$$\begin{aligned} Z_X &= \frac{Z_3}{Z_1} Z_2 \\ Z_X &= R_X + j X_X \\ Z_1 &= 1000 [\Omega] \\ Z_2 &= 4200 [\Omega] \\ Z_3 &= R_3 \parallel \frac{1}{j\omega C_3} = \\ &= \frac{R_3}{1 + j\omega R_3 C_3} = \frac{R_3}{1 + j\omega R_3 C_3} \end{aligned}$$

$$\therefore Z_3 = \frac{1.5 \times 10^6}{1 + (j)(2\pi)(2 \times 10^3)(1.5 \times 10^6)(12 \times 10^{-12})} \approx 1.427 - j(3228) \text{ [M}\Omega\text{]} //$$

- Asumiendo que Z_X está compuesta de elementos en serie

$$R_X + j X_X = \frac{4200}{1000} (1.427 - j(3228)) \times 10^6 = 5.993 - j(3228) \text{ [M}\Omega\text{]}$$

Agrupando:

$$\begin{aligned} R_X &= 5.993 \text{ [M}\Omega\text{]} \\ X_X &= \frac{1}{j\omega} = 1.35 \times 10^6 \end{aligned}$$

$$C = \frac{1}{j\omega X_X} = \frac{1}{(2\pi)(2 \times 10^3)(1.35 \times 10^6)}$$

Agrupando:

$$K_x = 5.990 \text{ [MΩ]} \quad L - \overline{\omega_x} = \frac{(2\pi)(2 \times 10^3)}{(1.356 \times 10^6)} \text{ [H]}$$

$$X_x = \frac{1}{\omega C} = 1.35 \times 10^6 \quad C = 58.69 \text{ [pF]} //$$

b) Si Z_x está configurada de componentes en paralelo

$$Z_x = \frac{4200}{1000} R_3 \parallel \frac{1}{j\omega C_3} = 4.2 Z_3$$

Puesto que Z_3 consta de R_3 y $X_3 = \frac{1}{\omega C_3}$ existen varias formas de obtener $4.2 Z_3$

• Suponiendo que

$$R_x = 1.4 R_3 = 2.1 \text{ [kΩ]}$$

$$X_x = \frac{1}{\omega C_x} = 3 X_3 = \frac{3}{\omega C_3}; C_x = \frac{1}{3} C_3 = 4.67 \text{ [pF]}$$

Problema

En el circuito B, suponer que el balance se consigue cuando Z_1 es un resistor de 4.8 [kΩ] , $Z_2 = 10 \Omega$ en serie con un inductor de $.25 \mu\text{H}$, Z_3 es un resistor de 12 [kΩ] y $f = 6 \text{ [MHz]}$. Determinar los componentes en serie que hacen Z_x

Resaltado: 25Ω , en serie
un inductor de $.625 \mu\text{H}$

$$Z_1 = 4.8 \text{ [kΩ]}$$

$$Z_2 = 10 + (.25 \times 10^{-6})(2\pi)(6 \times 10^6)j = 10 + 3\pi j$$

$$Z_3 = 12 \text{ [kΩ]}$$

$$V_1 = \frac{Z_1}{Z_1 + Z_2} \quad V_2 = \frac{Z_x}{Z_3 + Z_x} \quad \frac{Z_1}{Z_1 + Z_2} = \frac{Z_x}{Z_3 + Z_x}$$

$$\therefore Z_x = \frac{Z_2}{Z_1} Z_3 = \frac{10 + 3\pi j}{4800} (12000) = \left(\frac{5}{2}\right) (10 + 3\pi j)$$

$$Z_x = 25 + \frac{15\pi}{2} j \text{ [Ω]}$$

$$\therefore R_X = 25 \text{ } [\Omega] \cancel{\text{ // }}$$

$$x_x = \frac{15\pi}{2} = \omega L ; \quad L = \frac{15\pi}{2\omega} \stackrel{?}{=} 6.25 \times 10^{-7}$$

\uparrow $(2\pi)(6 \times 10^6)$

$$\therefore L \approx 0.625 [\mu H]$$

TAREA_4

Monday, 6 April 2020 6:23 PM

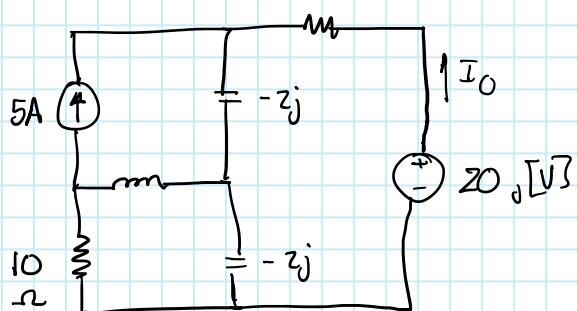
Teorema de Superposición

Puesto que el teorema aplica a Circuitos D.C., de la misma manera aplica a Circuito A.C.

Puesto que las impedancias dependen de la frecuencia entonces se debe tener un circuito en el dominio de la frecuencia para cada una frecuencia.

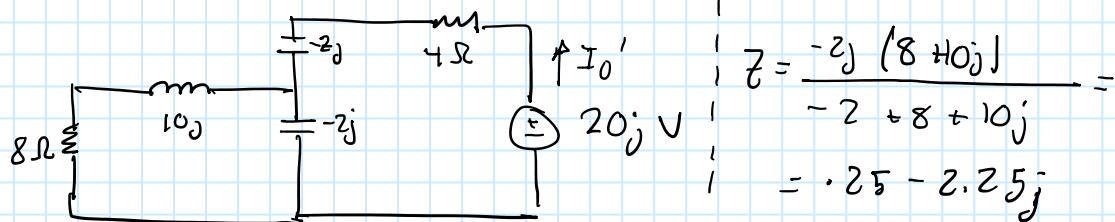
La respuesta total debe ser obtenido sumando los resultados individuales en el dominio del tiempo.

Ejemplo



$$\begin{aligned} I_O &= I_O' + I_O'' \\ \text{donde } I_O' \text{ y } I_O'' &\text{ son} \\ \text{debidos a las fuentes} &\text{ de voltaje y corriente} \end{aligned}$$

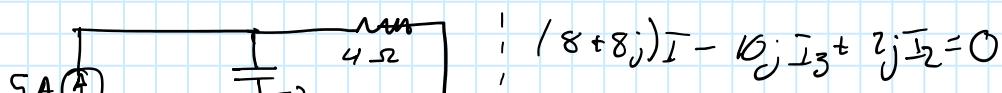
Para I_O' consideramos



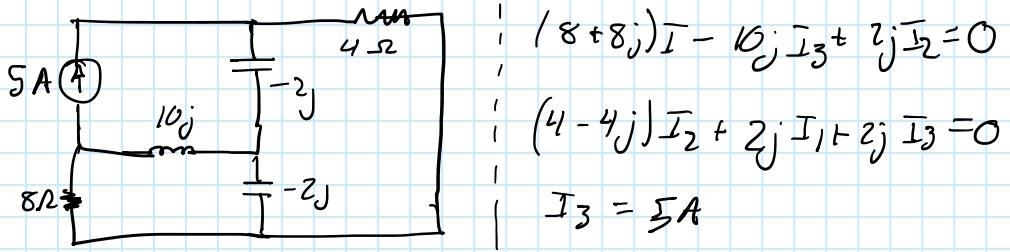
$$\begin{aligned} z &= \frac{-zj(8+10j)}{-2+8+10j} = \\ &= -2.353 - 2.353j \end{aligned}$$

$$\therefore I_O' = \frac{20j}{(4-zj) + z} = \frac{20j}{4.25 - 4.25j} \approx -2.353 + 2.353j$$

Para I_O'' consideramos el circuito siguiente



$$(8+8j)I - 10jI_3 + zjI_2 = 0$$



// Sustituyendo

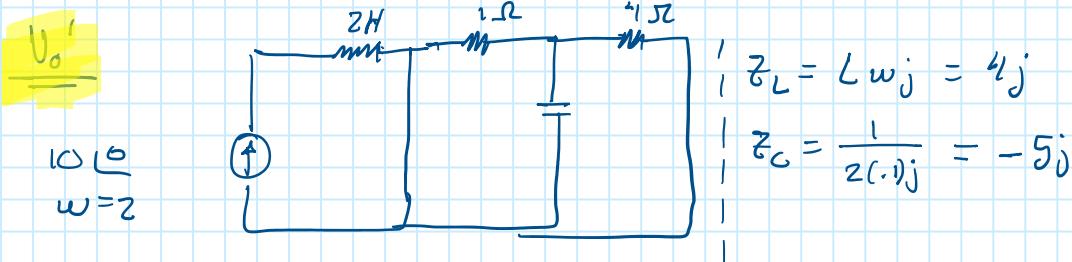
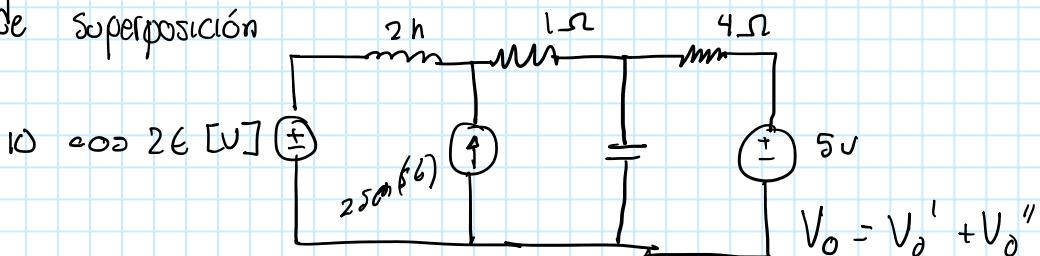
$$I_1 = (2 + 2j)I_2 - 5$$

$$I_2 = \frac{90 - 40j}{39} \approx 2.697 - 1.176j = -\underline{\underline{I_0}}$$

$$\therefore I_0 = I_0' + I_0'' = -5 + 3.529j = 6.12 \angle 144.78^\circ [A]$$

Problema 1 * *Fue una aproximación realmente cercana*

Encuentre V_o en el siguiente circuito mediante teorema de superposición



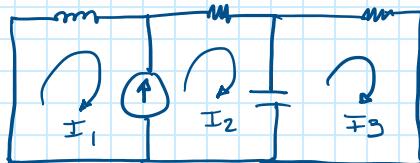
$$Z_{CR_1} = \left(\frac{1}{4} + \frac{1}{-5j} \right)^{-1} = \frac{1}{4} - \frac{j}{5} = \frac{1}{21} - \frac{j}{41}$$

$$Z_{CR_1R_2} = 1 + (Z_{CR_1}) \approx \underbrace{\frac{141}{41} - \frac{80}{41}j}_{\therefore Z = (\dots) + 4j} = \frac{141}{41} + \frac{84}{41}j$$

$$\therefore Z = (\dots) + q_j = \frac{141}{21} + \frac{84}{41} j \stackrel{!}{=} 2,146 - 1,278j$$

$$\therefore V = (2,146 - 1,278j)(1) = 2,146 - 1,278j \approx 2,49 \angle -30,78^\circ$$

// Definiendo sentidos y sistema de ecuaciones
en Mallas



$$\begin{cases} I_1, 10j + I_2 - 2j(I_2 - I_3) = 0 \\ \hookrightarrow 10j I_1 + (1 - 2j)I_2 + 2j I_3 = 0 \\ 4 I_3 - 2j(I_3 - I_2) = 0 \\ \hookrightarrow 2j I_2 + (4 - 2j)I_3 = 0 \end{cases}$$

$$\text{Extra} = -I_1 + I_2 = 2 \cos(5t - 90^\circ)$$

S.E

Forma Matricial

$$\begin{bmatrix} 0 & 2j & 4 - 2j \\ 10 & 1 - 2j & 2j \\ -1 & 1 & 0 \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \\ i_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 2 \cos(5t - 90^\circ) \end{bmatrix}$$

$$\det(A) = -44 + 30j$$

$$\therefore \Delta i_1 = \frac{1}{\det A} \begin{bmatrix} 0 & 2j & 4 - 2j \\ 0 & 1 - 2j & 2j \\ 2 \cos(5t - 90^\circ) & 1 & 0 \end{bmatrix} \stackrel{\sim}{=} .2256 + .3356j$$

$$\Delta i_2 = \frac{1}{\det A} \begin{bmatrix} 0 & 0 & 4 - 2j \\ 10 & 0 & 2j \\ -1 & 2 \cos(5t - 90^\circ) & 0 \end{bmatrix} \stackrel{\sim}{=} .2256 - 1.619j$$

$$\Delta i_3 = \frac{1}{\det A} \begin{bmatrix} 0 & 2 & 0 \\ 10 & -2j & 0 \\ -1 & 1 & 2 \cos(5t - 90^\circ) \end{bmatrix} \stackrel{\sim}{=} -.62059 - .4231j$$

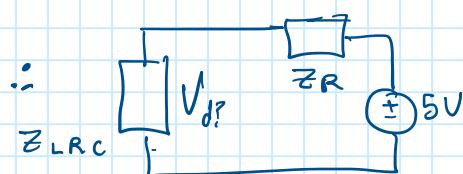
$$\det A = \begin{vmatrix} 1 & 1 & 1 \\ -1 & 1 & 1 \\ 1 & 1 & 2\cos(5t-90^\circ) \end{vmatrix} = -4231j$$

$$\therefore V_o = I_2 Z = (.2256 - 1.6643j)(1) \stackrel{\approx}{=} 1.679 \cos(5t - 82.8^\circ)$$

V_o

- $Z_L = 2j$
- $Z_C = \frac{-j}{(0.1)(1)} = -10j$

$$Z_{LRC} = \frac{1}{1+2j} - \frac{1}{-10j} \approx \frac{20}{13} + \frac{80}{13}j \Omega$$



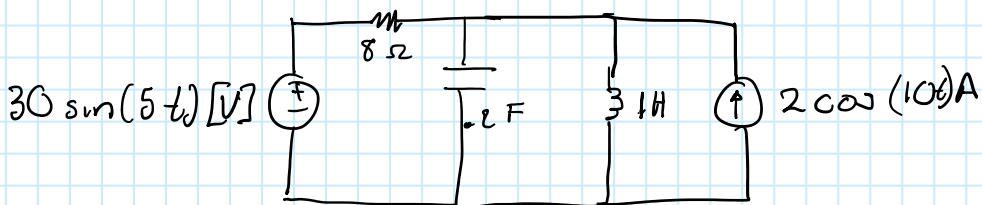
$$V = \frac{Z_{LRC}}{Z_{LRC} + Z_R}(5) = \approx \frac{25}{13} + \frac{50}{39}j$$

$$\therefore Z_R = 4$$

$$I = \frac{V_o}{Z_{LR}} = \frac{\frac{25}{13} + \frac{50}{39}j}{1+2j} = \frac{35}{39} - \frac{20}{39}j$$

$$\therefore V_o = \frac{35}{39} - \frac{20}{39}j \approx .8924 - .512820j = 1.033 \underline{[29.79]} \text{ [V]}$$

Problema 2



$$V_o' \quad Z_C = \frac{-j}{(0.5)(2)} = -j \quad Z_L = 5j \quad \therefore Z = 8 - \frac{5}{4}j$$

$$Z_{CL} = \left(\frac{1}{-j} + \frac{1}{5j} \right)^{-1} = -\frac{5}{9}j$$

$$\therefore V_0 = \frac{-\frac{5}{4}j}{8 - \frac{5}{4}j} (30 \angle -90^\circ) \approx 4.6313 \angle -171.119^\circ$$

$$= 4.6313 \cos(56 - 171.119^\circ)$$

Comprobando:

$$V_0' = 4.6313 \cos(56 - 171.119^\circ) =$$

$$= 4.6313 \cos(56 - 81.119^\circ) \cancel{\quad}$$

$$V_0'' | Z_C = \frac{-j}{1.2j \times 10} = -\frac{1}{2}j$$

$$| Z_L = 10j$$

$$| Z_{RC} = \left(\frac{1}{8} - \frac{1}{\frac{1}{2}j} \right)^{-1} = \frac{8}{25j} - \frac{128}{25j}j$$

$$Z_{eq} = \left(\frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3} \right)^{-1}$$

$$= \frac{Z_1 Z_2 Z_3}{Z_1 Z_2 + Z_1 Z_3 + Z_2 Z_3}$$

$$I = \frac{Z_{eq}}{Z_2} = \frac{12160}{580} + \frac{800}{580j}j \approx 2.096 + .1379j$$

$$\therefore V_0 = (2.096 + .1379j)(-\frac{1}{2}j) \approx 1.048 \angle -86.23^\circ$$

$$= 1.048 \cos(106 - 86.23^\circ) \cancel{\quad}$$

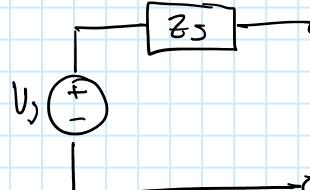
Total

$$V_0 = V_0' + V_0'' = 4.6313 \sin(56 - 81.119^\circ)$$

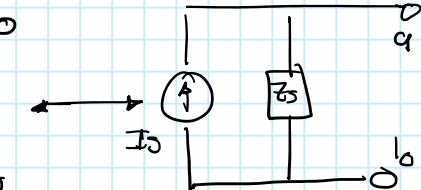
$$+ 1.048 \cos(106 - 86.23^\circ) \cancel{\quad}$$

Problema 3

Transformación en fuentes de A.C.



$$V_s = Z_S I_o$$

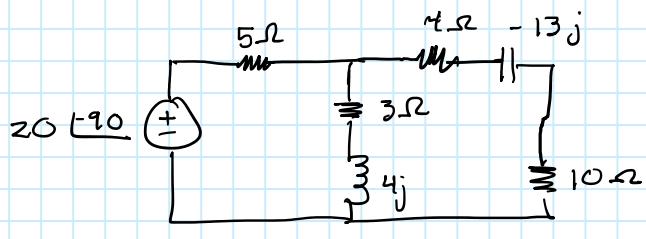


$$I_o = \frac{V_o}{Z_S}$$

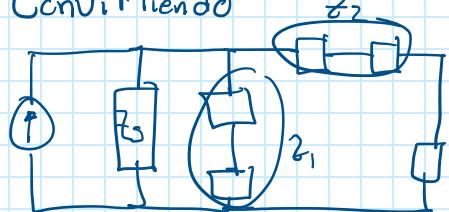
Calcular V_x en el



Calcular U_x en el
siguiente circuito



Convertiendo



$$I_s = -4j$$

$$Z_3 = 5$$

$$Z_1 = 3 + 4j$$

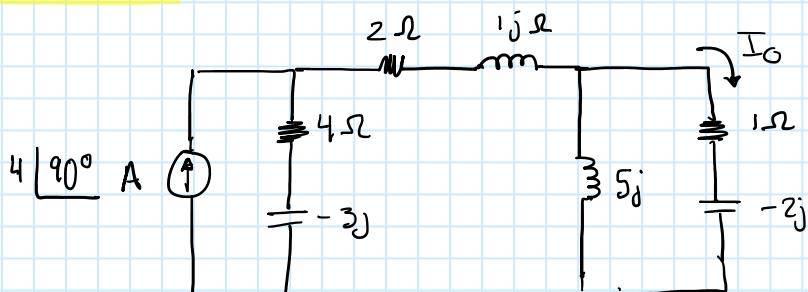
$$Z_2 = 14 - 13j$$

$$\therefore Z_{eq} = \left(\frac{1}{5} + \frac{1}{3 + 4j} + \frac{1}{14 - 13j} \right)^{-1} = \frac{3270}{1513} + \frac{1135}{1513}j$$

$$I = \frac{Z_{eq}}{Z_2} \left(\frac{1}{j} \right) = \left(\frac{1}{\frac{3270}{1513} + \frac{1135}{1513}j} \right) \left(-4j \right) = .4874 + .8644j$$

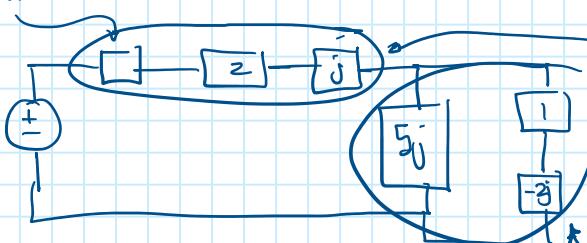
$$U_x = (.4874 + .8644j)(10) = 5.5194 \angle -27.97^\circ [V] \quad \cancel{\text{X}}$$

Problema 4



$$R_{aux} = 4 - 3j$$

// Conversión



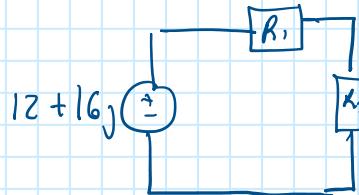
$$R_1 = 4 - 3j + 2 + j = 6 - 2j$$

$$R_2 = \left(\frac{1}{2j} + \frac{1}{-2j} \right)^{-1}$$

$$= \frac{5}{2} - \frac{5}{2}j$$

// Resuelto

$$V = (4 - 3j)(4 \angle 20^\circ) = 12 + 16j$$



$$\begin{aligned}
 V &= \frac{R_2}{R_1 + R_2} V = \frac{\frac{5}{2} - \frac{5}{2}j}{(6 - 2j) + \left(\frac{5}{2} - \frac{5}{2}j\right)} (12 + 16j) = \\
 &= \frac{220}{37} + \frac{160}{37}j = 5.9459 + 4.324j
 \end{aligned}$$

$$V = 12; I_o = \frac{5.9459 + 4.324}{1 - 2j} \approx -0.54054 + 3.2432j$$

$$\therefore I_o \approx 3.2879 \angle 99.46^\circ [A] \cancel{\text{}}$$