## 电子学基础——第九次作业

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- 10.51 A student who has a single-ended voltage source constructs the circuit shown in Fig. 10-75, hoping to obtain differential outputs. Assume perfect symmetry but  $\lambda = 0$  for simplicity.
- (b) Viewing  $M_1$  as a common-source stage degenerated by the impedance seen at the source of  $M_2$ , calculate  $v_X$  in terms of  $v_{in}$ .
  - (b) Viewing  $M_1$  as a source follower and  $M_2$  as a common-gate stage, calculate  $v_Y$  in terms of  $v_{in}$ .
- (c) Add the results obtained in (a) and (b) with proper polarities. If the voltage gain is defined as  $(v_X v_Y)/v_{in}$ , how does it compare with the gain of differentially driven pairs?

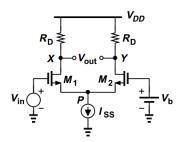


Figure 10-75

 $\mathbf{m}$  (a)  $M_2$ 从源端看入,输入电阻为 $\frac{1}{g_{m2}}$ ,而 $M_1$ 为源简并放大器。则

$$\frac{v_X}{v_{in}} = -\frac{g_{m1}R_D}{1 + g_{m1} \cdot \frac{1}{g_{m2}}} = -\frac{g_m R_D}{2}$$

(b)  $M_1$ 为源极跟随器,则 $M_1$ 源端电压即为 $v_{s1}=v_{in}$ ,而 $M_2$ 为共栅放大器,则

$$\frac{v_Y}{v_{s1}} = g_{m1}R_D$$

$$\therefore \frac{v_Y}{v_{in}} = g_{m1}R_D$$

$$\frac{v_X - v_Y}{v_{in}} = -\frac{3}{2}g_mR_D$$

这个增益是普通差分放大器增益的1.5倍。

(c)

10.70 Compute the common-mode rejection ratio of the stages illustrated in Fig. 10-89 and compare the results. For simplicity, neglect channel-length modulation in  $M_1$  and  $M_2$  but not in other transistors.

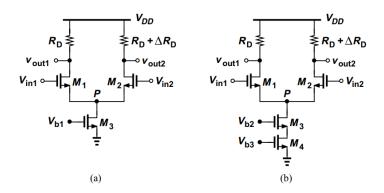


Figure 10-89

解

$$CMRR = 20 \log \left| \frac{A_{vd}}{A_{vc}} \right|$$

(a) 电路对称,可考虑半边电路。由交流小信号电路中P为虚地,则

$$A_{vd} = -g_{m1}R_D$$

考虑 $A_{vc}$ 时,可将 $M_3$ 视为 $r_{o3}$ ,进而再半边电路中视为 $2r_{o3}$ ,则 $M_1$ 为源简并放大器:

$$A_{vc} = \frac{-g_{m1}R_D}{1 + 2g_{m1}r_{o3}}$$

则

$$CMRR = 20\log(1 + 2g_{m1}r_{o3})$$

(b) 同(a), P再交流小信号电路中仍未虚地,则

$$A_{vd} = -g_{m1}R_d$$

再考虑 $A_{vc}$ ,将 $M_4$ 视为 $r_{o4}$ ,则 $M_3$ 为源简并放大器,可视为电阻 $r=(1+g_{m3}r_{o3})r_{o4}+r_{o3}$  从而半边电路中可将其视为 $2r=2[(1+g_{m3}r_{o3})r_{o4}+r_{o3}]$  此时 $M_1$ 仍为源简并放大器:

$$\therefore A_{vc} = \frac{-g_{m1}R_D}{1 + 2g_{m1}[(1 + g_{m3}r_{o3})r_{o4} + r_{o3}]}$$

:. CMRR = 
$$20 \log[1 + 2g_{m1}((1 + g_{m3}r_{o3})r_{o4} + r_{o3})]$$