

## 5 Performance considerations

Performance criteria:

Performance	Noise
Disturbances	Robustness

- Increasing  $\gamma$ , we are unhappy with the **oscillations of our parameters  $\theta$**  and therefore with the oscillations of  $u(t)$ .
- We have no clue what the adaptive closed loop will do between  $t = 0$  and  $t = \infty$  other than boundedness

### 5.1 Adaptation with a closed loop reference model

**Now** Deal with transient response

**Idea** Adaptation changes with signals

$$\dot{\theta} = -\text{sgn}(\varepsilon) \varepsilon \gamma \phi$$

where the value of  $\varepsilon$  and  $\gamma$  are changeable.  
 $\Rightarrow$  we can alter the transient with  $\gamma$  (leads to oscillations), **or we can change  $\varepsilon(t)$** .

**So far** Open loop reference model (ORM)

$$\dot{x}_m^o(t) = a_m x_m^o(t) + k_m r(t) \quad (2)$$

**Now** Closed loop reference model (CRM)

$$\dot{x}_m^c(t) = a_m x_m^c(t) + k_m r(t) - l e^c(t) \quad (19)$$

ORM = CRM if  $l = 0$ .

"CRM is observer-like; M helps G by moving towards G and retreating to original position."  
**Through the movement, the reference model now has a different behaviour ( $M \rightarrow M'$ ) and the plant P is trying to follow  $M'$ .**

- $\gamma$  - learning effect
  - decreasing  $\gamma$  helps P follow  $M'$ ,
  - but the learning becomes slower
- $l$  - movement to P
  - increasing  $l$  helps P follow  $M'$

### 5.2 Stability proof

$$\dot{x}_m^c(t) = a_m x_m^c(t) + k_m r(t) - l e^c(t) \quad (19)$$

$$\dot{x}_p(t) = a_p x_p(t) + k_p u(t) \quad (1)$$

Input is

$$u = \begin{bmatrix} a(t) & k(t) \end{bmatrix} \begin{bmatrix} x_p(t) \\ r(t) \end{bmatrix} = \theta^T(t) \phi(t)$$

$$\dot{x}_p(t) = a_m x_p(t) + k_m r(t) + k_p \theta^T(t) \phi(t)$$

Tracking error

$$\begin{aligned} \dot{e}^c(t) &= \dot{x}_p(t) - \dot{x}_m^c(t) \\ &= (a_m + l) e^c(t) + k_p \theta^T(t) \phi(t) \end{aligned}$$

Lyapunov-like function

$$\begin{aligned} V(e^c, \tilde{\theta}) &= \frac{1}{2} (e^c)^2 + \frac{1}{2} \Gamma^{-1} |k_p| \tilde{\theta}^T(t) \tilde{\theta}(t) \\ \dot{V} &= e^c \dot{e}^c + \Gamma^{-1} |k_p| \tilde{\theta}^T \dot{\tilde{\theta}} = \dots \\ &= (a_m + l) (e^c)^2 \\ &\quad + \underbrace{e^c k_p \tilde{\theta} \phi + \Gamma^{-1} |k_p| \tilde{\theta}^T \dot{\tilde{\theta}}}_{\stackrel{!}{=} 0} \\ \dot{V} &= (a_m + l) (e^c)^2 \leq 0, \quad l < 0 \end{aligned}$$

Adaptive law

$$\dot{\theta} = -\Gamma \text{sgn } k_p e^c \phi$$

**Proof** as before.  $e^c(t) \rightarrow 0$  for  $t \rightarrow \infty$ <sup>8</sup>.

**Questions**

- How do we show increased performance? (using  $\|e^c(t)\|_{\mathcal{L}_2}$  as a performance criterion)
- How do we show that the oscillations decrease?

### 5.3 Analysing transient performance

Check the performance criterion  $\mathcal{L}_2$ -norm of  $e^c$

$$\begin{aligned} \int_0^\infty \dot{V}(e^c, \theta) d\tau &= V(\infty) - V(0) \\ -|a_m + l| \int_0^\infty e^{c^2} d\tau &= V(\infty) - V(0) \\ V(0) &= \underbrace{V(\infty)}_{\geq 0} + |a_m + l| \cdot \|e^c\|_2^2 \\ V(0) &\geq |a_m + l| \cdot \|e^c\|_2^2 \\ \|e^c\|_2 &\leq \sqrt{\frac{V(0)}{|a_m + l|}} \end{aligned}$$

$$\|e^c\|_2^2 \leq \frac{1}{2} \frac{(e^c(0))^2 + \frac{|k_p|}{\gamma} \theta^T(0) \theta(0)}{|a_m + l|} \quad (20)$$

<sup>8</sup>We assume here that  $e^c(t) \rightarrow 0$  follows from  $e^o(t) \rightarrow 0$ . In actuality, though,  $e^o(t)$  can't be proven for special functions. However, these cases are usually not relevant to engineering/industry. Therefore, **strictly speaking**, we can't actually assume that  $e^c(t) \rightarrow 0$

### Discussion

- Increasing  $\gamma$  reduces  $\|e^c\|_{\mathcal{L}_2}$  depending on the parameter errors  $\tilde{\theta}$
- Increasing the value of  $l$  reduces  $\|e^c\|_{\mathcal{L}_2}$  also from  $e^c(o)$

## 5.4 Analysing the signal oscillations

$\mathcal{L}_2$ -norm of  $\dot{k}$

$$\begin{aligned}\dot{k} &= -\gamma \operatorname{sgn} k_p e^c r(t) \\ \int_0^\infty |\dot{k}|^2 d\tau &= \gamma^2 \int_0^\infty (e^c)^2 r^2 d\tau \\ (r(t) &\leq \|r\|_{\mathcal{L}_\infty}) \\ &\leq \gamma^2 \|r\|_{\mathcal{L}_\infty}^2 \int_0^\infty (e^c)^2 d\tau \\ &\leq \gamma^2 \|r\|_{\mathcal{L}_\infty}^2 \|e^c\|_{\mathcal{L}_2}^2\end{aligned}$$

$$\|\dot{k}\|_2 \leq \gamma \|r\|_{\mathcal{L}_\infty} \sqrt{\frac{V(o)}{|a_m + l|}}$$

Increasing  $\gamma$  or reducing  $l$  causes  $\dot{k}$  to decrease in magnitude.

$\mathcal{L}_2$ -norm of  $\dot{\theta}$

$$\begin{aligned}\dot{\theta} &= -\gamma \operatorname{sgn} k_p e^c x_p(t) \\ &= -\gamma \operatorname{sgn} k_p e^c (e^c + x_m(t)) \\ |\dot{\theta}|^2 &= \gamma^2 (e^c)^2 (e^c + x_m(t))^2 \\ &\quad (a+b)^2 \leq 2a^2 + 2b^2 \\ &\leq 2\gamma^2 (e^c)^2 [(e^c)^2 (e^c)^2 + x_m^2] \\ \int_0^\infty |\dot{\theta}|^2 d\tau &\leq 2\gamma^2 \left[ \int_0^\infty (e^c)^2 (e^c)^2 d\tau + \int_0^\infty (e^c)^2 x_m^2 d\tau \right] \\ &\vdots\end{aligned}$$

$$\begin{aligned}&\int_0^\infty |\dot{\theta}|^2 d\tau \\ &\leq 2\gamma^2 \frac{V(o)}{|a_m + l|} \left[ V(o) \left( 2 + \frac{l^2}{|a_m| \cdot |a_m + l|} \right) \right. \\ &\quad \left. + 2\|\dot{x}_m\|_{\mathcal{L}_\infty}^2 \right]\end{aligned}$$

### Discussion

- $l$  reduces contribution of the ORM  $\|\dot{x}_m(t)\|_{\mathcal{L}_\infty}$  on  $\|\dot{\theta}\|_{\mathcal{L}_2}$
- $l$  has no clear effect on the contributions of  $V(o)$ .
- $\gamma$  always increases the oscillations, s.  $\|\dot{\theta}\|_{\mathcal{L}_2}$

## 6 Output feedback adaptive control

Let  $M(s)$  be a linear time-invariant, asymptotically stable reference model, with I/O  $\{r(\cdot), y_m(\cdot)\}$ .  $r$  is uniform, bounded, piecewise continuous function of time. The plant  $G$  is defined such that

$$G(s) = k_p \frac{Z_p(s)}{R_p(s)}$$

- $Z_p, R_p$  are monic<sup>9</sup> polynomials of orders  $n-1$  and  $n$
- $G(s)$  is controllable, observable  $\Leftrightarrow Z_p, R_p$  are coprime<sup>10</sup> polynomials

We know

- (i) sign of high-frequency gain  $k_p$
- (ii) upper bound  $n$  on the order of  $G(s)$
- (iii) relative degree  $n^*$ <sup>11</sup>
- (iv) zeroes of  $Z_p(s)$  lie in  $\mathbb{C}^-$   
 $\Rightarrow$  min. phase<sup>12</sup>, no inverse response

**Minimal phase systems** “In minimal phase systems, we can predict the phase  $\varphi$  given the magnitude  $|G| \Leftarrow$  only if  $G$  stable with no time delay”.

e.g. a plane, or a forklift, are examples of non minimal phase systems (they have different inverse responses depending on where you measure, s. Figure 5)

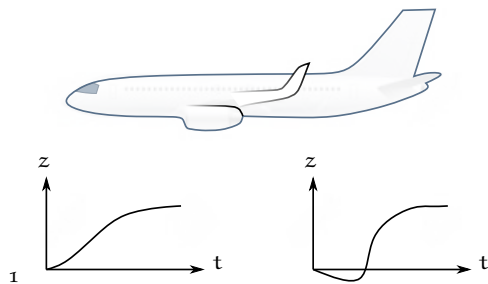


Figure 5: An ascending plane.

### Solution in three parts

- (i)  $k_p$  unknown
- (ii)  $Z_p$  unknown
- (iii)  $R_p$  unknown

In each case

- a) “Matching conditions”

Show that  $\exists$  parameters  $\theta^* = \text{const.}$ , such that the closed loop behaves exactly as the reference

- b) Error dynamics Derive error model as

$$e = \frac{1}{k^*} M(s) [\theta^T \phi]$$

- c) Use KY-lemma
- d) Show that  $e \rightarrow 0$  for  $t \rightarrow \infty$  (Barbalat’s lemma)

### 6.1 $k_p$ unknown

#### Matching conditions

<sup>9</sup>monic: coefficient of the highest order is 1, e.g.  $s^2 + 2s + 1$

<sup>10</sup>coprime: no cancellations, i.e. no common roots

<sup>11</sup>in this course, we only handle cases where  $n^* = 1$

<sup>12</sup>there is no standard definition for minimal phase