



Iran University of Science & Technology

IUST

# Digital Logic Design

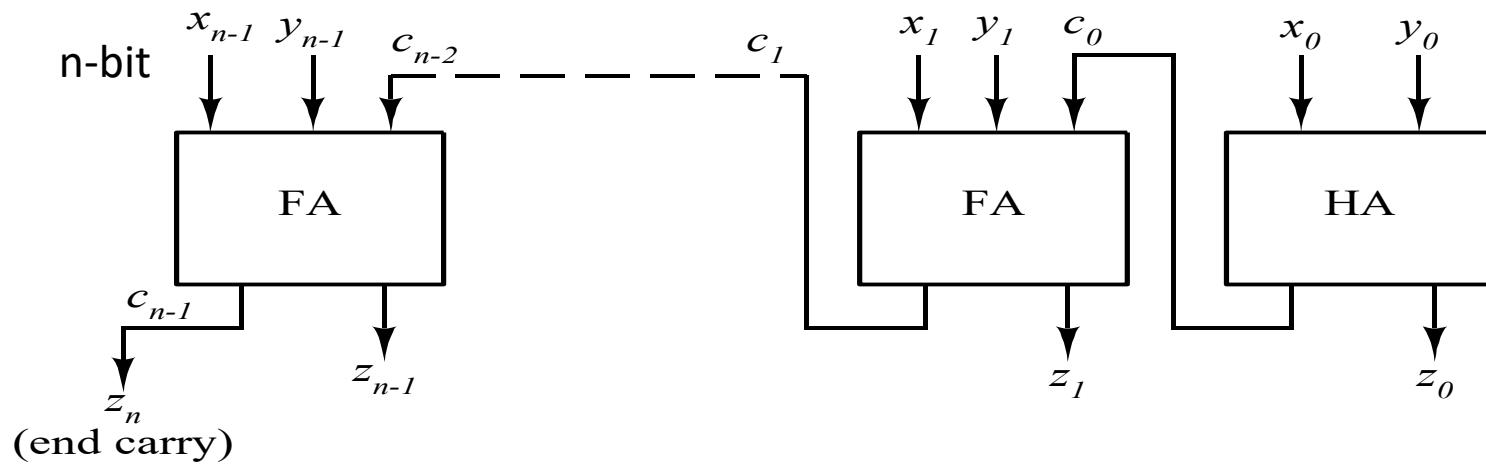
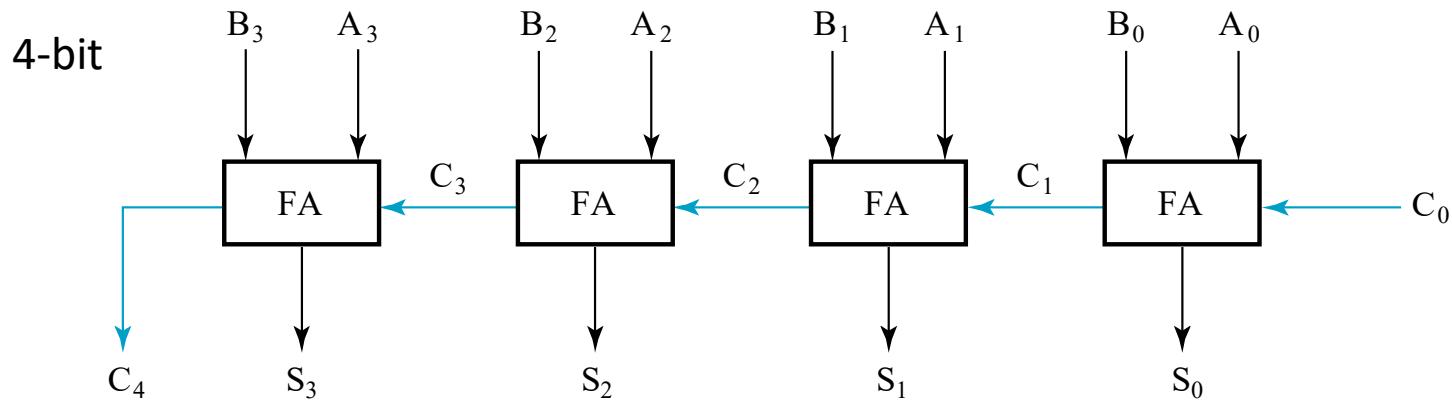
---

Hajar Falahati

Department of Computer Engineering  
IRAN University of Science and Technology

[hfalahati@iust.ac.ir](mailto:hfalahati@iust.ac.ir)

# Ripple-Carry-Adder (RCA)



# Carry Look Ahead (CLA)

$$s_i = p_i \oplus c_{i-1}$$

$$c_i = g_i + p_i \cdot c_{i-1}$$

$$c_0 = g_0 + p_0 c_{\text{in}}$$

$$s_0 = p_0 \oplus c_{\text{in}}$$

$$c_1 = g_1 + p_1 c_0$$

$$= g_1 + p_1 g_0$$

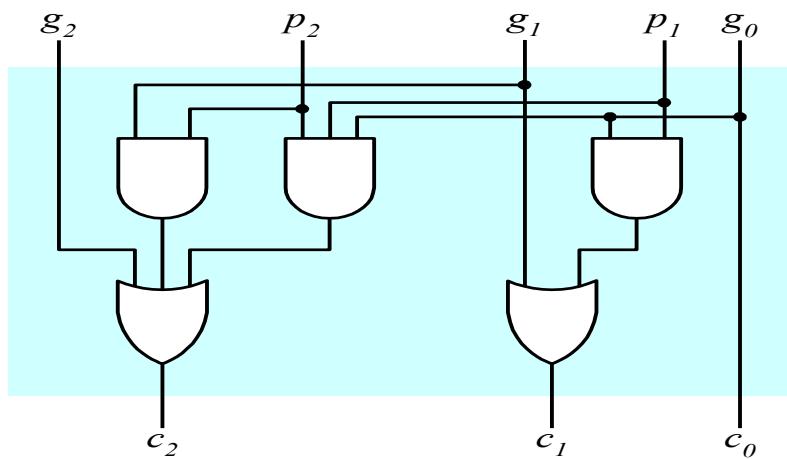
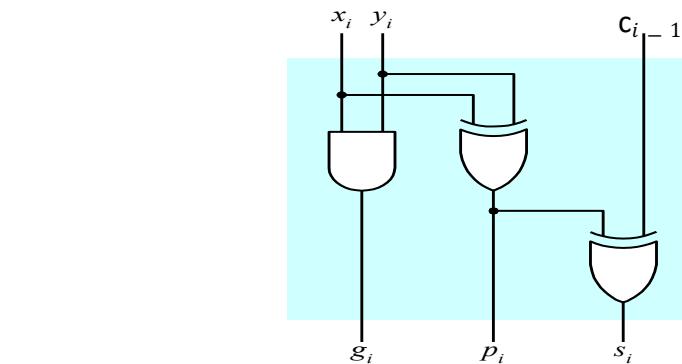
$$s_1 = p_1 \oplus c_0$$

$$c_2 = g_2 + p_2 c_1$$

$$= g_2 + p_2(g_1 + p_1 g_0)$$

$$= g_2 + p_2 g_1 + p_2 p_1 g_0$$

$$s_2 = p_2 \oplus c_1$$



# Outline

---

- Subtractor
- Comparator



# Subtractor

---

# Recall: 2's Complement Subtraction

---

- $A = B - C$ 
  - $(A)_2 = (B)_2 + (-(C)_2) = (B)_2 + [C]_2 = (B)_2 + 2^n - (C)_2 = 2^n + (B - C)_2$
  - If  $B \geq C$ 
    - $A \geq 2^n$  and the carry is **discarded**
    - $(A)_2 = (B)_2 + [C]_2$  | carry discarded
  - If  $B < C$ 
    - $A = 2^n - (C - B)_2 = [C - B]_2$  or  $A = -(C - B)_2$  (no carry in this case)
    - **No overflow for Case 2.**
- $(A)_2 = (B)_2 + (-(C)_2) = (B)_2 + [C]_2 = (B)_2 + (C')_2 + 1$

# Recall:

## Sample 1

---

$$\begin{array}{r} 01001101 \\ - 00111010 \\ \hline \end{array}$$

$$\begin{array}{r} 01001101 \\ + 11000110 \\ \hline 100010011 \end{array}$$

Carry = 1 => Result is OK

# Recall:

## Sample 2

---

$$\begin{array}{r} 01000011 \\ - 01010100 \\ \hline \end{array}$$

$$\begin{array}{r} 01000011 \\ + 10101100 \\ \hline 11101111 \end{array}$$

Carry = 0 => Result is **not** OK

$$[11101111]_2 = 00010001$$

Result = - (00010001 )

# 2's Complement Subtractor

---

- Inputs

- 2 numbers
- $x_i, y_i$

$$\begin{array}{r} x_i \\ - y_i \\ \hline c_i s_i \end{array}$$

- Output

- 1 number
- 1 borrow

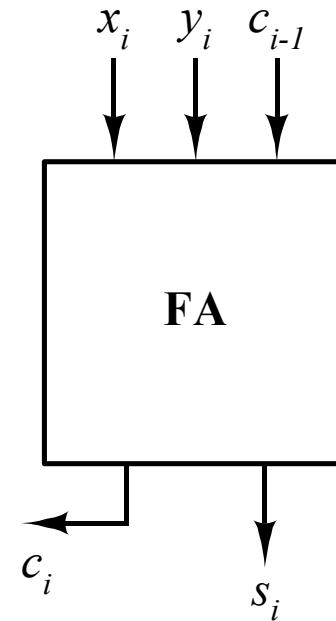
- Functionality

- Subtract two numbers

# 2's Complement Subtractor Vs. Full Adder

---

- $C_{i-1} = 0$ 
  - Input:  $x_i, y_i$
  - $S_i = x_i + y_i$
  
- $C_{i-1} = 1$ 
  - Input:  $x_i, (y_i)'$
  - $S_i = x_i - y_i$
  
- **Functionality**
  - Subtract two numbers



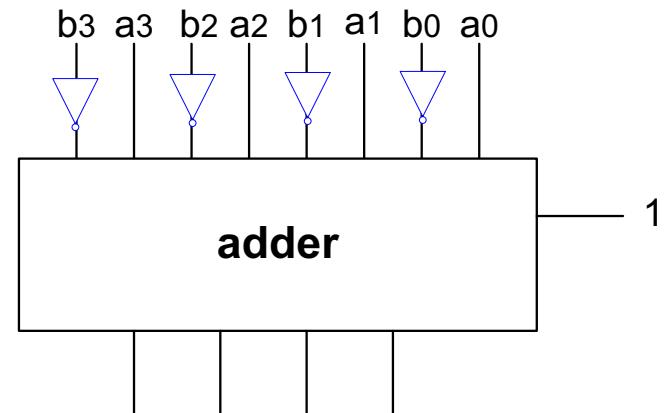
# 4-bit 2's Complement Subtractor

---

- **Inputs**
  - 2 4-bit numbers
  - $a, b$
  
- **Output**
  - $a-b$

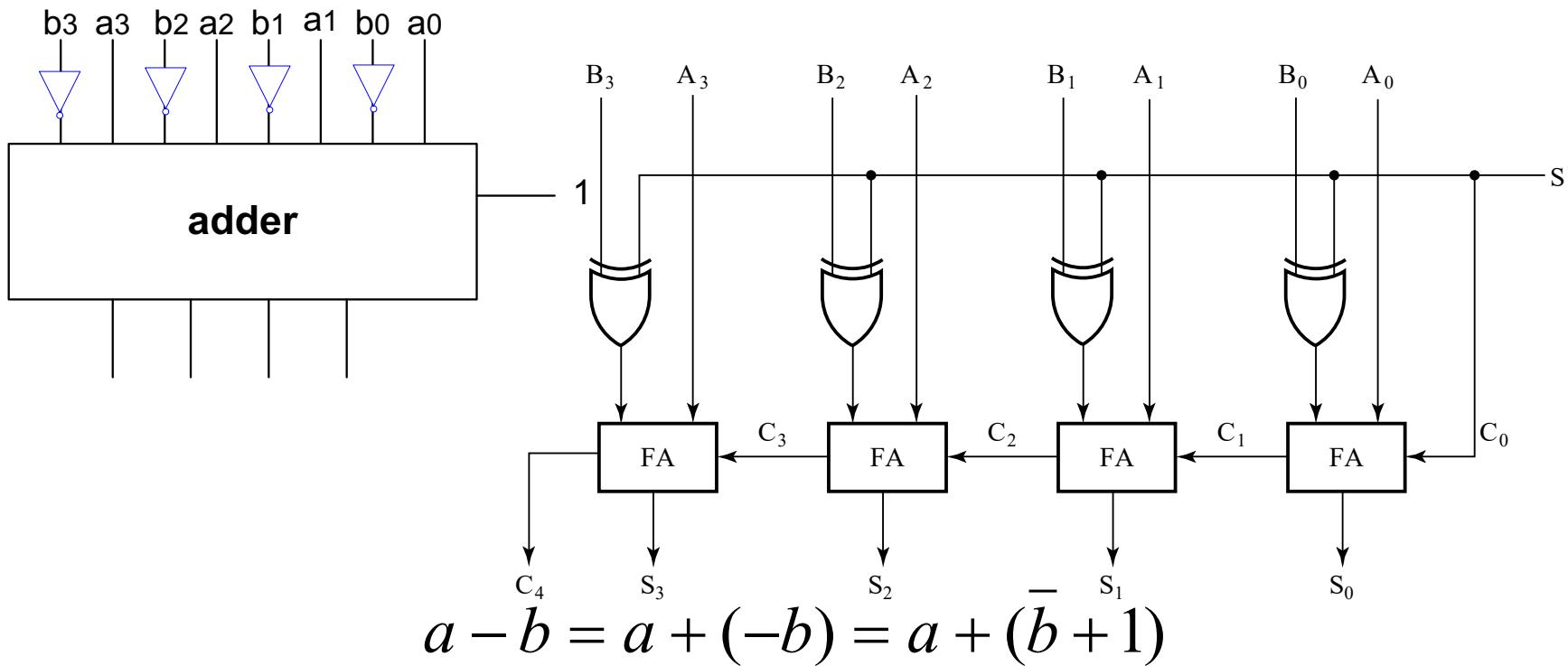
# 4-bit Subtractor: Realization

- **Inputs**
  - 2 4-bit numbers
  - $a, b$
- **Output**
  - $a - b$



$$a - b = a + (-b) = a + (\bar{b} + 1)$$

# 4-bit Subtractor: Detailed Realization



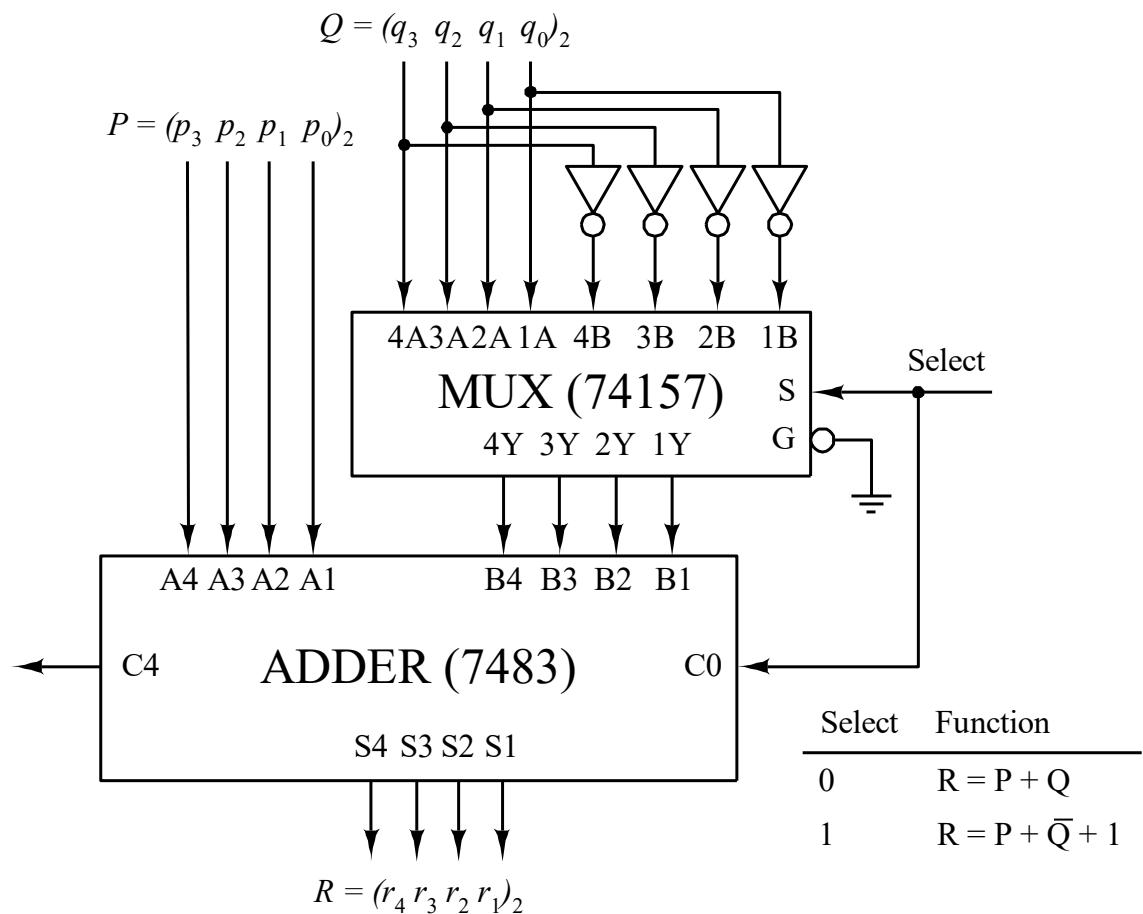
# 2's Complement Adder/ Subtractor

---

- **Inputs**
  - 2 4-bit numbers
  - **Select**
- **Output**
  - **Select** ==0 => Add
  - **Select** ==1 => Sub

# Adder/ Subtractor: Realization

- **Inputs**
  - 2 4-bit numbers
  - **Select**
- **Output**
  - **Select** ==0 => Add
  - **Select** ==1 => Sub



# Overflow

---

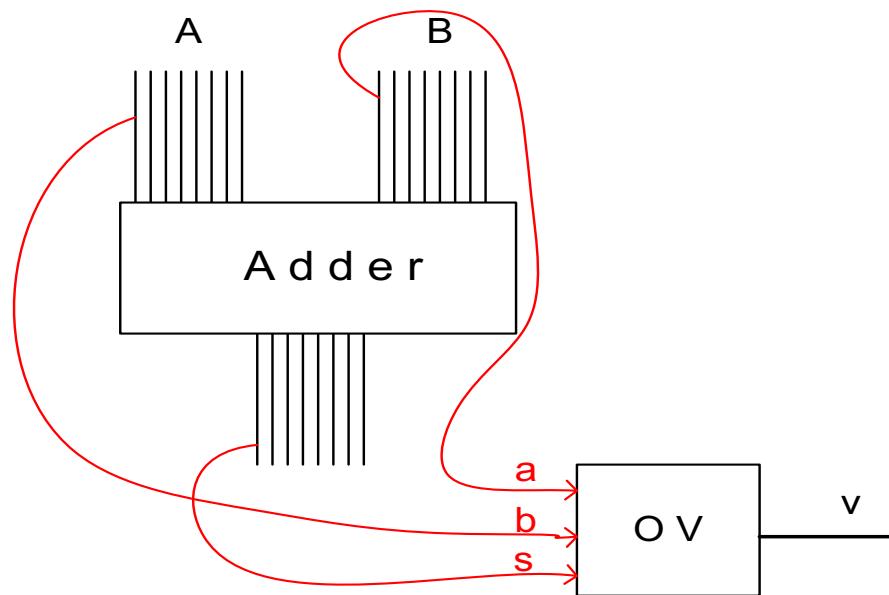
$$\begin{array}{r}
 + \quad 11000011 \\
 \hline
 10101100 \\
 \hline
 101101111
 \end{array}$$

$$\begin{array}{r}
 + \quad 00110100 \\
 \hline
 01101000 \\
 \hline
 10011100
 \end{array}$$

Case	Carry	Sign Bit	Condition	Overflow ?
B + C	0	0	$B + C \leq 2^{n-1} - 1$	No
	0	1	$B + C > 2^{n-1} - 1$	Yes
B - C	1	0	$B \leq C$	No
	0	1	$B > C$	No
-B - C	1	1	$-(B + C) \geq -2^{n-1}$	No
	1	0	$-(B + C) < -2^{n-1}$	Yes

# Overflow Detection

- Design a circuit to detect overflow



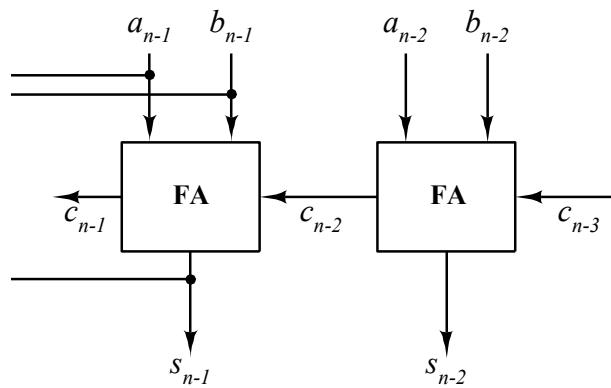
# Overflow Detection:

TT

---

$a_{n-1}$	$b_{n-1}$	$c_{n-2}$	$c_{n-1}$	$s_{n-1}$	$V$
0	0	0	0	0	0
0	0	1	0	1	1
0	1	0	0	1	0
0	1	1	1	0	0
1	0	0	0	1	0
1	0	1	1	0	0
1	1	0	1	0	1
1	1	1	1	1	0

Case	Carry	Sign Bit	Condition	Overflow ?
$B + C$	0	0	$B + C \leq 2^{n-1} - 1$	No
	0	1	$B + C > 2^{n-1} - 1$	Yes
$B - C$	1	0	$B \leq C$	No
	0	1	$B > C$	No
$-B - C$	1	1	$-(B + C) \geq -2^{n-1}$	No
	1	0	$-(B + C) < -2^{n-1}$	Yes



# Overflow Detection: Logic Equation

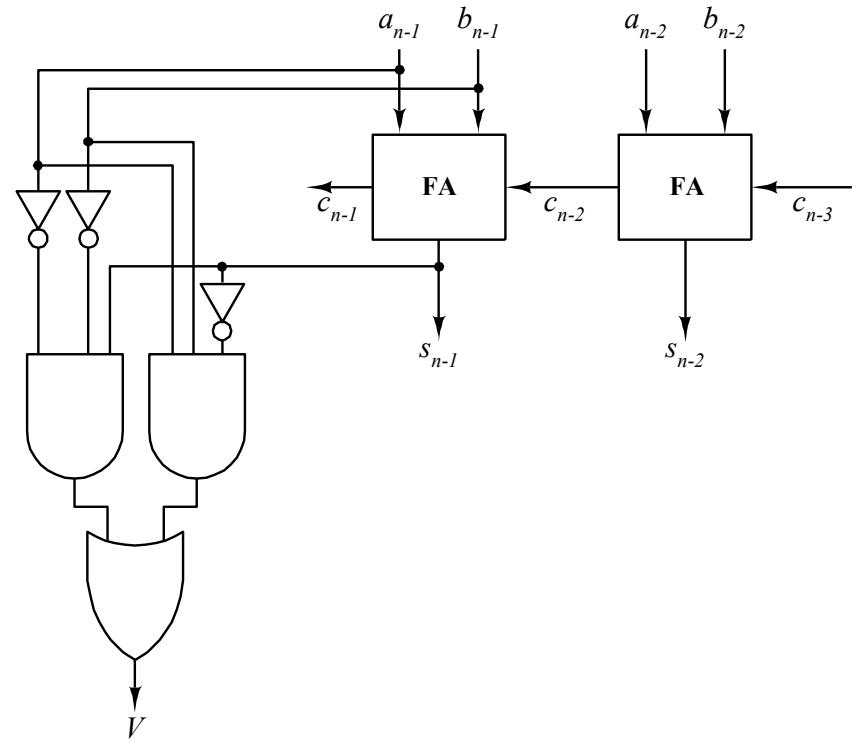
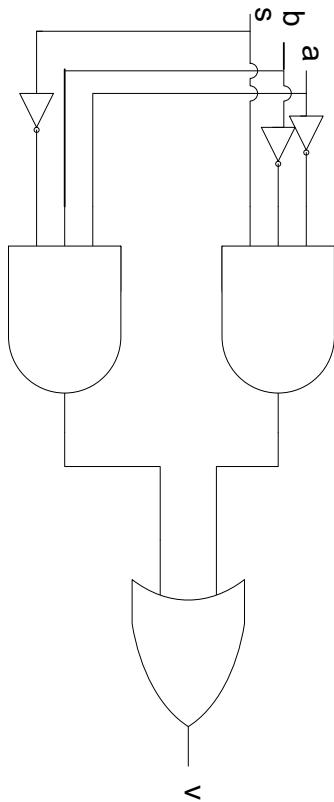
$a_{n-1}$	$b_{n-1}$	$c_{n-2}$	$c_{n-1}$	$s_{n-1}$	$V$
0	0	0	0	0	0
0	0	1	0	1	1
0	1	0	0	1	0
0	1	1	1	0	0
1	0	0	0	1	0
1	0	1	1	0	0
1	1	0	1	0	1
1	1	1	1	1	0

Case	Carry	Sign Bit	Condition	Overflow ?
$B + C$	0	0	$B + C \leq 2^{n-1} - 1$	No
	0	1	$B + C > 2^{n-1} - 1$	Yes
$B - C$	1	0	$B \leq C$	No
	0	1	$B > C$	No
$-B - C$	1	1	$-(B + C) \geq -2^{n-1}$	No
	1	0	$-(B + C) < -2^{n-1}$	Yes

$$V = a_{n-1} \cdot b_{n-1} \cdot (s_{n-1})' + (a_{n-1})' \cdot (b_{n-1})' \cdot s_{n-1}$$

# Overflow Detection: Realization

$$V = a_{n-1} \cdot b_{n-1} \cdot (s_{n-1})' + (a_{n-1})' \cdot (b_{n-1})' \cdot s_{n-1}$$

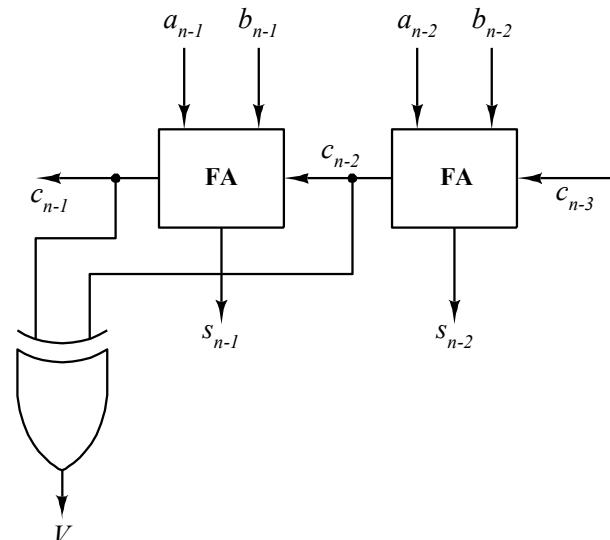


# Overflow Detection: Realization 2

$a_{n-1}$	$b_{n-1}$	$c_{n-2}$	$c_{n-1}$	$s_{n-1}$	$V$
0	0	0	0	0	0
0	0	1	0	1	1
0	1	0	0	1	0
0	1	1	1	0	0
1	0	0	0	1	0
1	0	1	1	0	0
1	1	0	1	0	1
1	1	1	1	1	0

$$V = a_{n-1} \cdot b_{n-1} \cdot (s_{n-1})' + (a_{n-1})' \cdot (b_{n-1})' \cdot s_{n-1}$$

$$V = c_{n-2} \oplus c_{n-1}$$



# Comparator

---

# Comparator

---

- **Inputs**
  - 2 numbers
  - $x_i, y_i$
  
- **Output**
  - 3 1-bit
  - Equal
  - Greater than
  - Lower than

# Comparator: Check Equality

---

- Design a circuit to check the equality of two numbers

# Comparator: Check Inequality

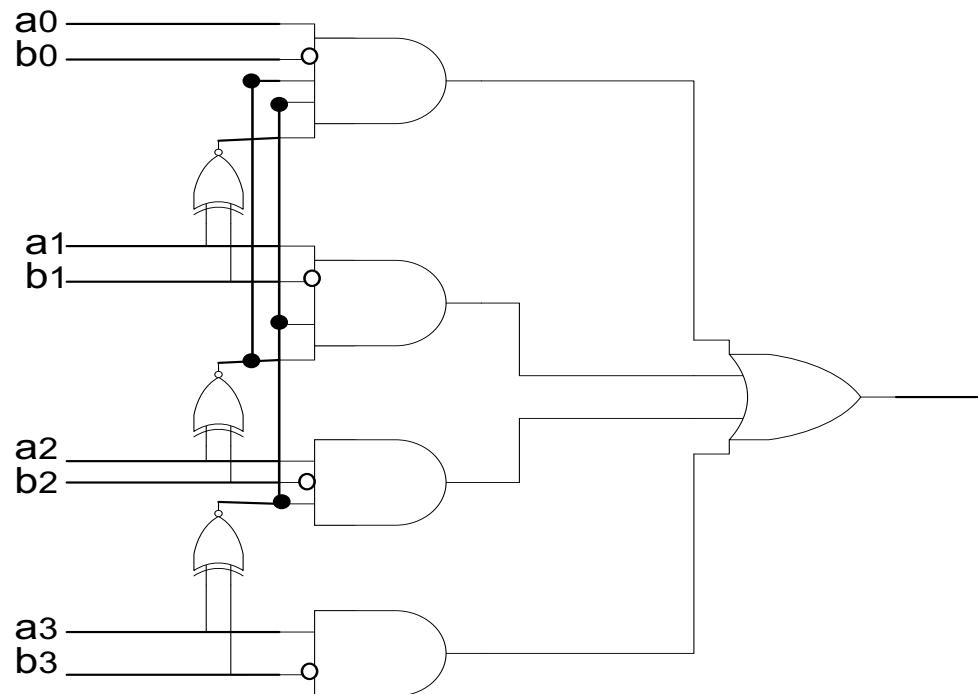
---

- Design a circuit to check  $a>b$

# Check Inequality: Realization

- Check MSB of a

- If  $a_3 = 1$  and  $b_3 = 0 \Rightarrow a > b$
- If  $a_3 = 0$  and  $b_3 = 1 \Rightarrow a < b$
- Else If  $a_2 = 1$  and  $b_2 = 0 \Rightarrow a > b$
- Else If  $a_2 = 0$  and  $b_2 = 1 \Rightarrow a < b$
- ( $a_3 = b_3$ )
- Else If  $a_1 = 1$  and  $b_1 = 0 \Rightarrow a > b$
- Else If  $a_1 = 0$  and  $b_1 = 1 \Rightarrow a < b$
- ( $a_3 = b_3, a_2 = b_2$ )



# Comparator Circuit

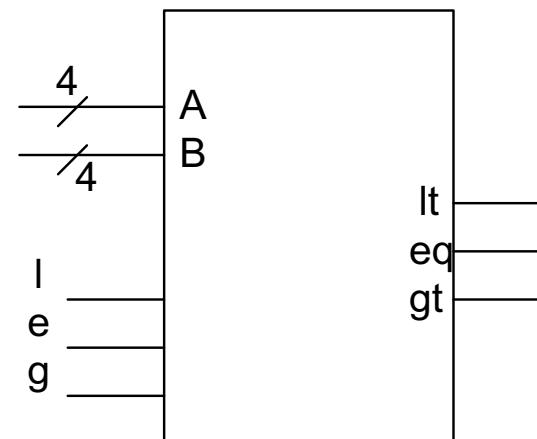
- Design a comparator circuit (7485)

- Inputs

- 2 4-bit numbers
- l, e and g (cascading)

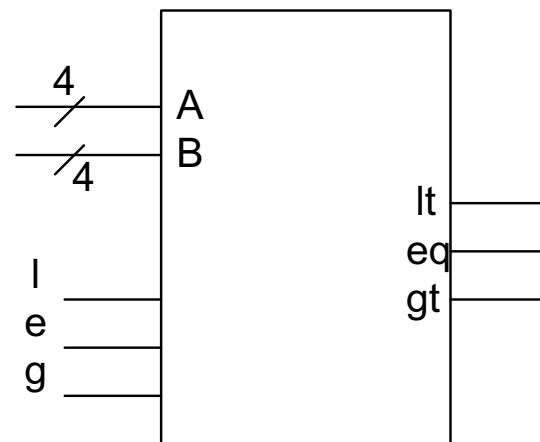
- Outputs

- If ( A > B ) => lt=0, eq=0, gt=1
- if ( A < B ) => lt=1, eq=0, gt=0
- if ( A = B ) => lt=l, eq=e, gt=g



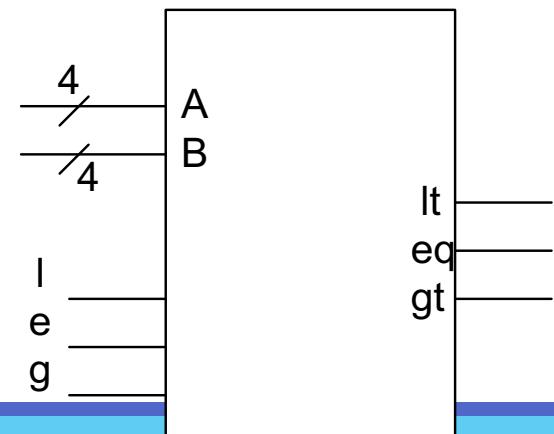
# Comparator: 16 bit

- Compare two 16-bit numbers



# 16-bit Comparator: Steps

1. Compare the **4 most significant bits** of the two inputs
  - ‘A’ is **less** than ‘B’ if its 4 MSB are **smaller** => STOP
  - ‘A’ is **greater** than ‘B’ if its 4 MSB are **larger** => STOP
  - If these 4 MSB are the **last ones**, two numbers are equal => STOP
  - If they are **equal** the **result** can be found in **this first stage** => GO Step2
2. Comparator will **recursively** do the same comparison on **less significant bits four by four**



# Comparator: Signed numbers

---

- Design a circuit to compare two signed numbers

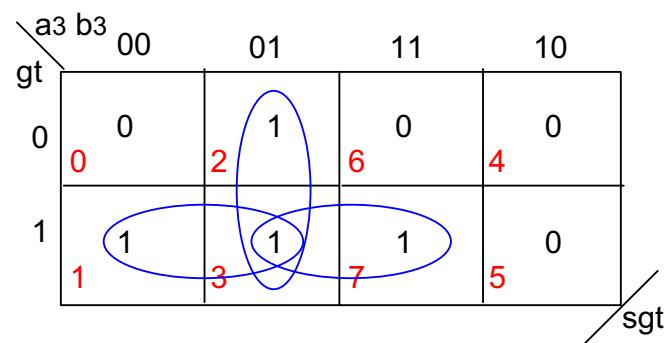
# Signed Numbers ( $A > B$ )

---

- $A > B$ 
  - If **A** is **positive** and **B** is **negative**  $\Rightarrow A > B$
  - If **Both** are **positive** or **negative**  $\Rightarrow$  consider gt
- **Example:** compare +3 and +2 by a 4 bit unsigned comparator
  - $3 = (0010)_2$
  - $2 = (0011)_2$
  - In unsigned comparator  $\Rightarrow 3 > 2$
- **Example:** compare -2 and -3 by a 4 bit unsigned comparator
  - $(-2)_{2s} = 2^n - 2 \Rightarrow (1110)_{2s}$
  - $(-3)_{2s} = 2^n - 3 \Rightarrow (1101)_{2s}$
  - In unsigned comparator  $\Rightarrow 2^n - 2 > 2^n - 3$

# Signed Numbers ( $A > B$ ): K-map

---



$$sgt = \overline{a_3} \cdot b_3 + \overline{a_3} \cdot gt + b_3 \cdot gt$$

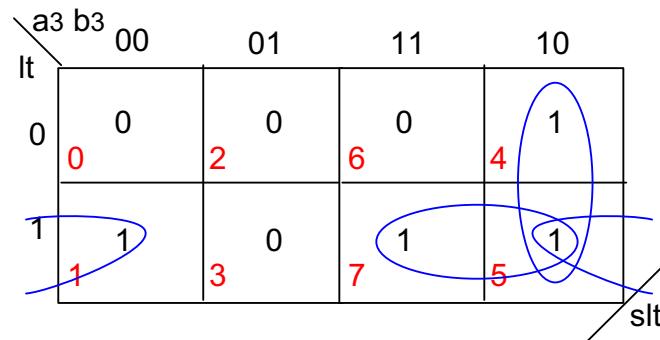
# Signed Numbers ( $A < B$ )

---

- $A < B$ 
  - If **A** is **negative** and **B** is **positive**  $\Rightarrow A < B$
  - If **Both** are **positive** or **negative**  $\Rightarrow$  consider It
  
- **Example:** compare +2 and +3 by a 4 bit unsigned comparator
  - $2 = (0010)_2$
  - $3 = (0011)_2$
  - In unsigned comparator  $\Rightarrow 2 < 3$
  
- **Example:** compare -3 and -2 by a 4 bit unsigned comparator
  - $(-3)_{2s} = 2^n - 3 \Rightarrow (1101)_{2s}$
  - $(-2)_{2s} = 2^n - 2 \Rightarrow (1110)_{2s}$
  - In unsigned comparator  $\Rightarrow 2^n - 3 < 2^n - 2$

# Signed Numbers (A < B): K-map

---



$$slt = a_3 \cdot \overline{b_3} + lt \cdot a_3 + lt \cdot \overline{b_3}$$

# Sample Logic1: BCD Adder

---

- Design a circuit to add two BCD numbers

# BCD Adder: Truth Table

- Compare **BCD sum** with **binary sum**

Binary Sum K Z <sub>8</sub> Z <sub>4</sub> Z <sub>2</sub> Z <sub>1</sub>	BCD Sum C S <sub>8</sub> S <sub>4</sub> S <sub>2</sub> S <sub>1</sub>	Decimal
0 0 0 0 0	0 0 0 0 0	0
0 0 0 0 1	0 0 0 0 1	1
0 0 0 1 0	0 0 0 1 0	2
0 0 0 1 1	0 0 0 1 1	3
0 0 1 0 0	0 0 1 0 0	4
0 0 1 0 1	0 0 1 0 1	5
0 0 1 1 0	0 0 1 1 0	6
0 0 1 1 1	0 0 1 1 1	7
0 1 0 0 0	0 1 0 0 0	8
0 1 0 0 1	0 1 0 0 1	9
0 1 0 1 0	1 0 0 0 0	10
0 1 0 1 1	1 0 0 0 1	11
0 1 1 0 0	1 0 0 1 0	12
0 1 1 0 1	1 0 0 1 1	13
0 1 1 1 0	1 0 1 0 0	14
0 1 1 1 1	1 0 1 0 1	15
1 0 0 0 0	1 0 1 1 0	16
1 0 0 0 1	1 0 1 1 1	17
1 0 0 1 0	1 1 0 0 0	18
1 0 0 1 1	1 1 0 0 1	19

# BCD Adder Vs. Binary Adder

- When are they different?

Binary Sum K Z <sub>8</sub> Z <sub>4</sub> Z <sub>2</sub> Z <sub>1</sub>	BCD Sum C S <sub>8</sub> S <sub>4</sub> S <sub>2</sub> S <sub>1</sub>	Decimal
0 0 0 0 0	0 0 0 0 0	0
0 0 0 0 1	0 0 0 0 1	1
0 0 0 1 0	0 0 0 1 0	2
0 0 0 1 1	0 0 0 1 1	3
0 0 1 0 0	0 0 1 0 0	4
0 0 1 0 1	0 0 1 0 1	5
0 0 1 1 0	0 0 1 1 0	6
0 0 1 1 1	0 0 1 1 1	7
0 1 0 0 0	0 1 0 0 0	8
0 1 0 0 1	0 1 0 0 1	9
0 1 0 1 0	1 0 0 0 0	10
0 1 0 1 1	1 0 0 0 1	11
0 1 1 0 0	1 0 0 1 0	12
0 1 1 0 1	1 0 0 1 1	13
0 1 1 1 0	1 0 1 0 0	14
0 1 1 1 1	1 0 1 0 1	15
1 0 0 0 0	1 0 1 1 0	16
1 0 0 0 1	1 0 1 1 1	17
1 0 0 1 0	1 1 0 0 0	18
1 0 0 1 1	1 1 0 0 1	19

# BCD Adder Vs. Binary Adder: Difference

- When are they different?
  - $F = K + Z_8 Z_4 + Z_8 Z_2$

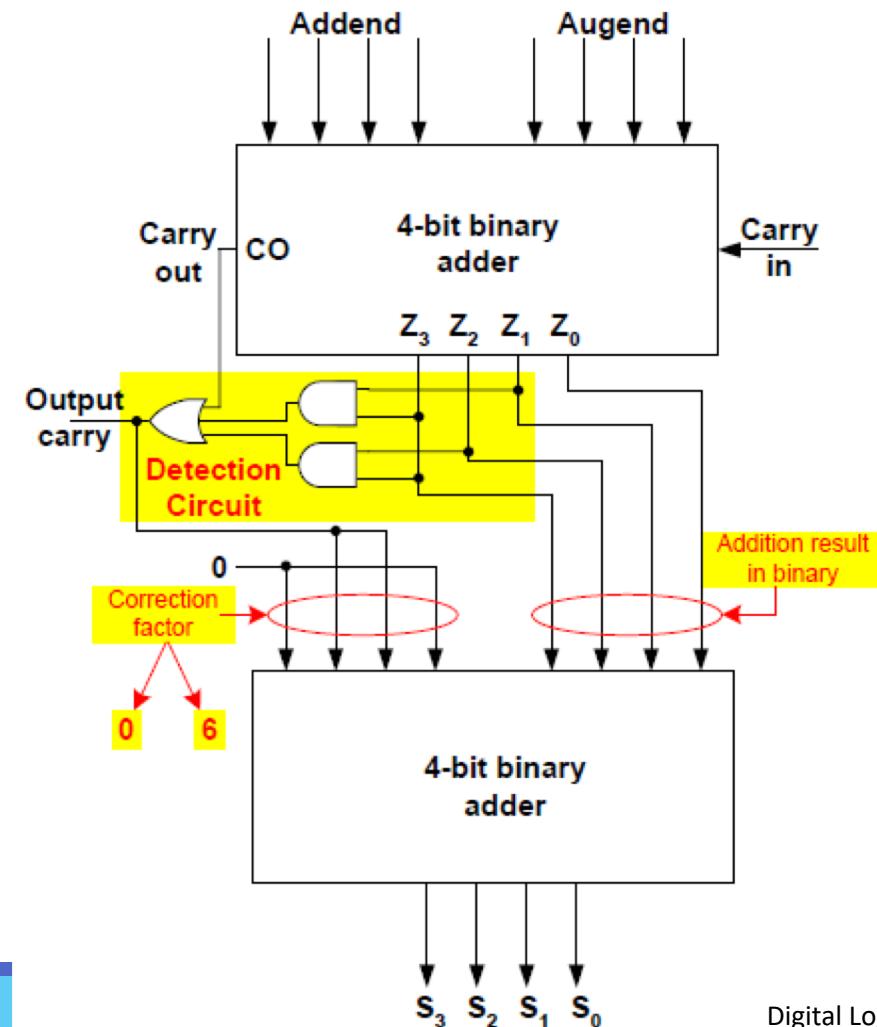
Binary Sum K Z <sub>8</sub> Z <sub>4</sub> Z <sub>2</sub> Z <sub>1</sub>	BCD Sum C S <sub>8</sub> S <sub>4</sub> S <sub>2</sub> S <sub>1</sub>	Decimal
0 0 0 0 0	0 0 0 0 0	0
0 0 0 0 1	0 0 0 0 1	1
0 0 0 1 0	0 0 0 1 0	2
0 0 0 1 1	0 0 0 1 1	3
0 0 1 0 0	0 0 1 0 0	4
0 0 1 0 1	0 0 1 0 1	5
0 0 1 1 0	0 0 1 1 0	6
0 0 1 1 1	0 0 1 1 1	7
0 1 0 0 0	0 1 0 0 0	8
0 1 0 0 1	0 1 0 0 1	9
0 1 0 1 0	1 0 0 0 0	10
0 1 0 1 1	1 0 0 0 1	11
0 1 1 0 0	1 0 0 1 0	12
0 1 1 0 1	1 0 0 1 1	13
0 1 1 1 0	1 0 1 0 0	14
0 1 1 1 1	1 0 1 0 1	15
1 0 0 0 0	1 0 1 1 0	16
1 0 0 0 1	1 0 1 1 1	17
1 0 0 1 0	1 1 0 0 0	18
1 0 0 1 1	1 1 0 0 1	19

# BCD Adder: How to correct?

- When are they different?
  - $F = K + Z_8 Z_4 + Z_8 Z_2$
- How to correct it?
  - Add by 6 (0110)

Binary Sum K Z <sub>8</sub> Z <sub>4</sub> Z <sub>2</sub> Z <sub>1</sub>	BCD Sum C S <sub>8</sub> S <sub>4</sub> S <sub>2</sub> S <sub>1</sub>	Decimal
0 0 0 0 0	0 0 0 0 0	0
0 0 0 0 1	0 0 0 0 1	1
0 0 0 1 0	0 0 0 1 0	2
0 0 0 1 1	0 0 0 1 1	3
0 0 1 0 0	0 0 1 0 0	4
0 0 1 0 1	0 0 1 0 1	5
0 0 1 1 0	0 0 1 1 0	6
0 0 1 1 1	0 0 1 1 1	7
0 1 0 0 0	0 1 0 0 0	8
0 1 0 0 1	0 1 0 0 1	9
0 1 0 1 0	1 0 0 0 0	10
0 1 0 1 1	1 0 0 0 1	11
0 1 1 0 0	1 0 0 1 0	12
0 1 1 0 1	1 0 0 1 1	13
0 1 1 1 0	1 0 1 0 0	14
0 1 1 1 1	1 0 1 0 1	15
1 0 0 0 0	1 0 1 1 0	16
1 0 0 0 1	1 0 1 1 1	17
1 0 0 1 0	1 1 0 0 0	18
1 0 0 1 1	1 1 0 0 1	19

# BCD Adder: Realization

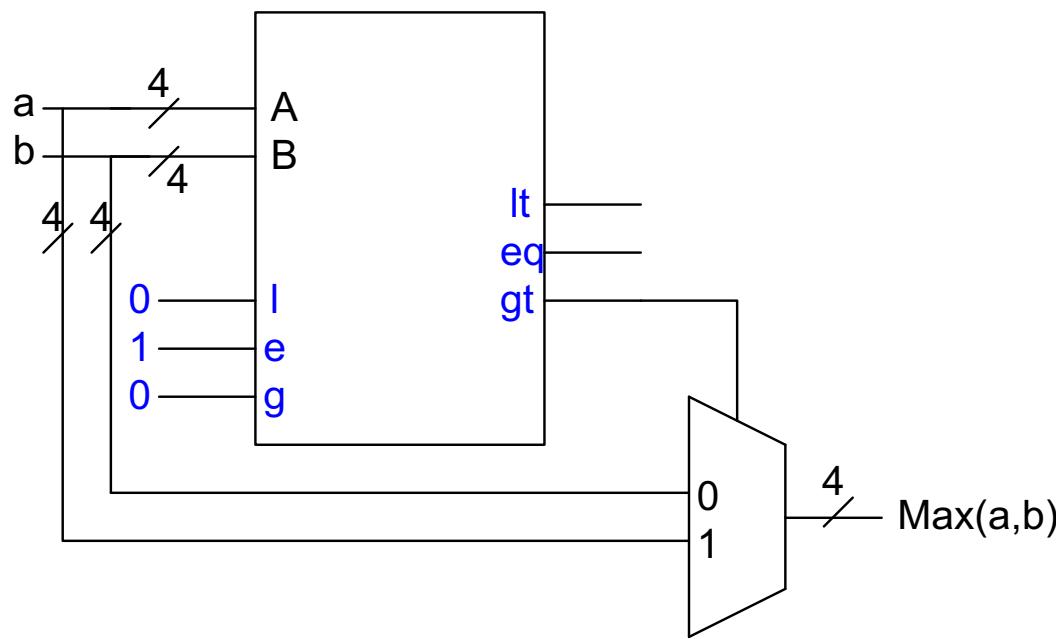


# Sample Logic2: Max Logic

---

- Design a circuit to find the maximum number

# Max Logic



# Sample Logic3: Multiplier

- Design a circuit to multiply two 2-bit numbers

$$X = \sum_{i=0}^{M-1} X_i 2^i$$

↗  
Multiplicand

$$Y = \sum_{i=0}^{N-1} Y_i 2^i$$

↖  
Multiplier

Product

$$Z = X * Y$$

# Multiplier: Recall

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline \end{array}$$

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline 1100 \end{array}$$

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline 1100 \\ 0000 \end{array}$$

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline 1100 \\ 0000 \\ 1100 \end{array}$$

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline 1100 \\ 0000 \\ 1100 \\ 0000 \end{array}$$

$$\begin{array}{r} 1100 : 12_{10} \\ 0101 : 5_{10} \\ \hline 1100 \\ 0000 \\ 1100 \\ 0000 \\ \hline 00111100 : 60_{10} \end{array}$$

# Multiplier: Partial Products

$$X = \sum_{i=0}^{M-1} X_i 2^i$$

Multiplicand

$$Y = \sum_{i=0}^{N-1} Y_i 2^i$$

Multiplier

$$\begin{array}{r} 1100 \\ 0101 \\ \hline 1100 \end{array} : 12_{10}$$

$$\begin{array}{r} 0000 \\ 1100 \\ \hline \end{array}$$

$$\begin{array}{r} 0000 \\ \hline 00111100 \end{array} : 60_{10}$$

multiplicand  
multiplier

partial  
products

product

$$Z = X * Y$$

$$= \sum_{i=0}^{N-1} \left( \sum_{j=0}^{M-1} X_i Y_j 2^{i+j} \right)$$

Partial products

# Multiplier:

## Binary Representation

Multiplicand:  $Y = (y_{M-1}, y_{M-2}, \dots, y_1, y_0)$

Multiplier:  $X = (x_{N-1}, x_{N-2}, \dots, x_1, x_0)$

Product:  $P = \left( \sum_{j=0}^{M-1} y_j 2^j \right) \left( \sum_{i=0}^{N-1} x_i 2^i \right) = \sum_{i=0}^{N-1} \sum_{j=0}^{M-1} x_i y_j 2^{i+j}$

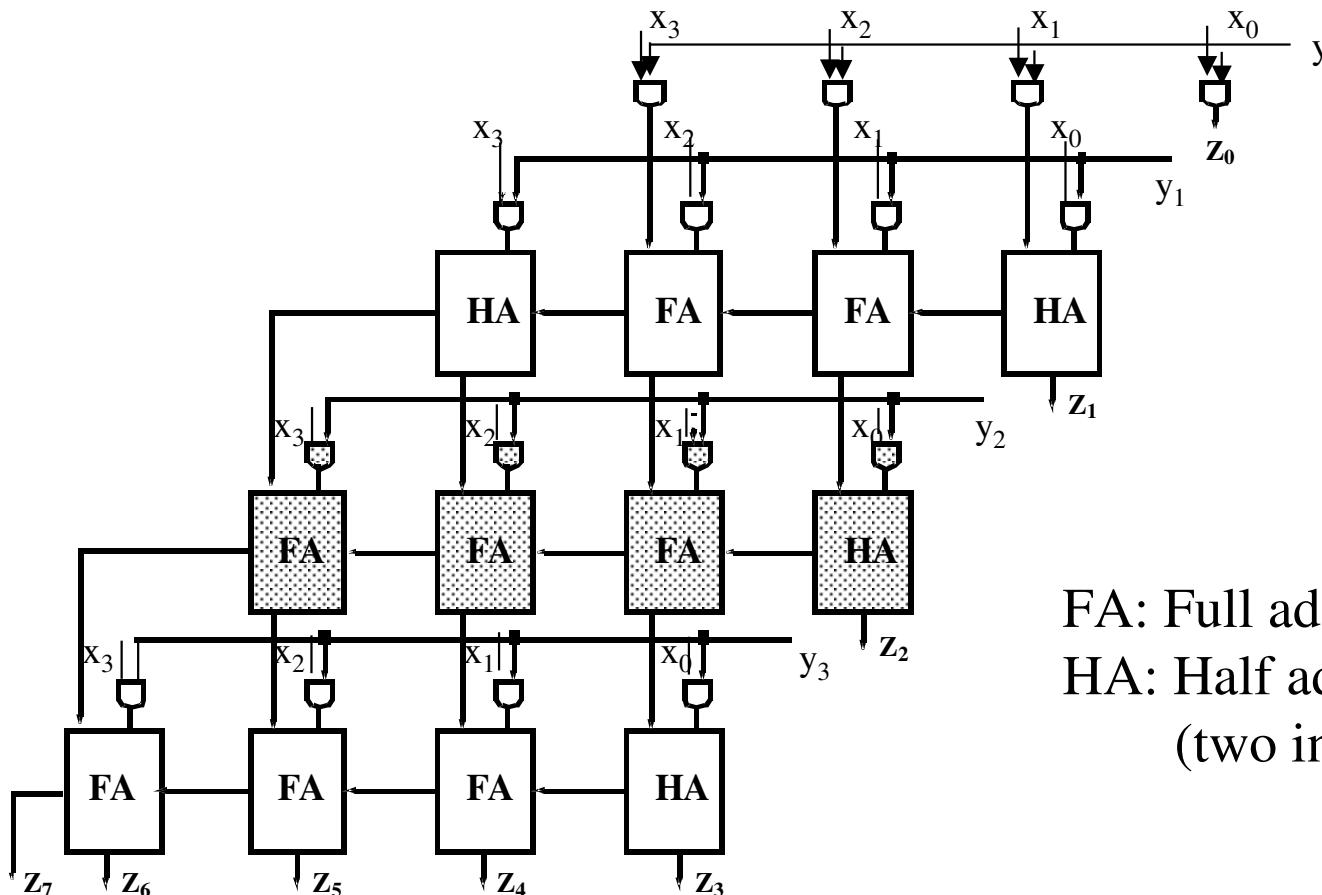
		$y_5$	$y_4$	$y_3$	$y_2$	$y_1$	$y_0$					
		$x_5$	$x_4$	$x_3$	$x_2$	$x_1$	$x_0$					
		$x_0y_5$	$x_0y_4$	$x_0y_3$	$x_0y_2$	$x_0y_1$	$x_0y_0$					
		$x_1y_5$	$x_1y_4$	$x_1y_3$	$x_1y_2$	$x_1y_1$	$x_1y_0$					
		$x_2y_5$	$x_2y_4$	$x_2y_3$	$x_2y_2$	$x_2y_1$	$x_2y_0$					
		$x_3y_5$	$x_3y_4$	$x_3y_3$	$x_3y_2$	$x_3y_1$	$x_3y_0$					
		$x_4y_5$	$x_4y_4$	$x_4y_3$	$x_4y_2$	$x_4y_1$	$x_4y_0$					
		$x_5y_5$	$x_5y_4$	$x_5y_3$	$x_5y_2$	$x_5y_1$	$x_5y_0$					
$p_{11}$	$p_{10}$	$p_9$	$p_8$	$p_7$	$p_6$	$p_5$	$p_4$	$p_3$	$p_2$	$p_1$	$p_0$	

multiplicand  
multiplier

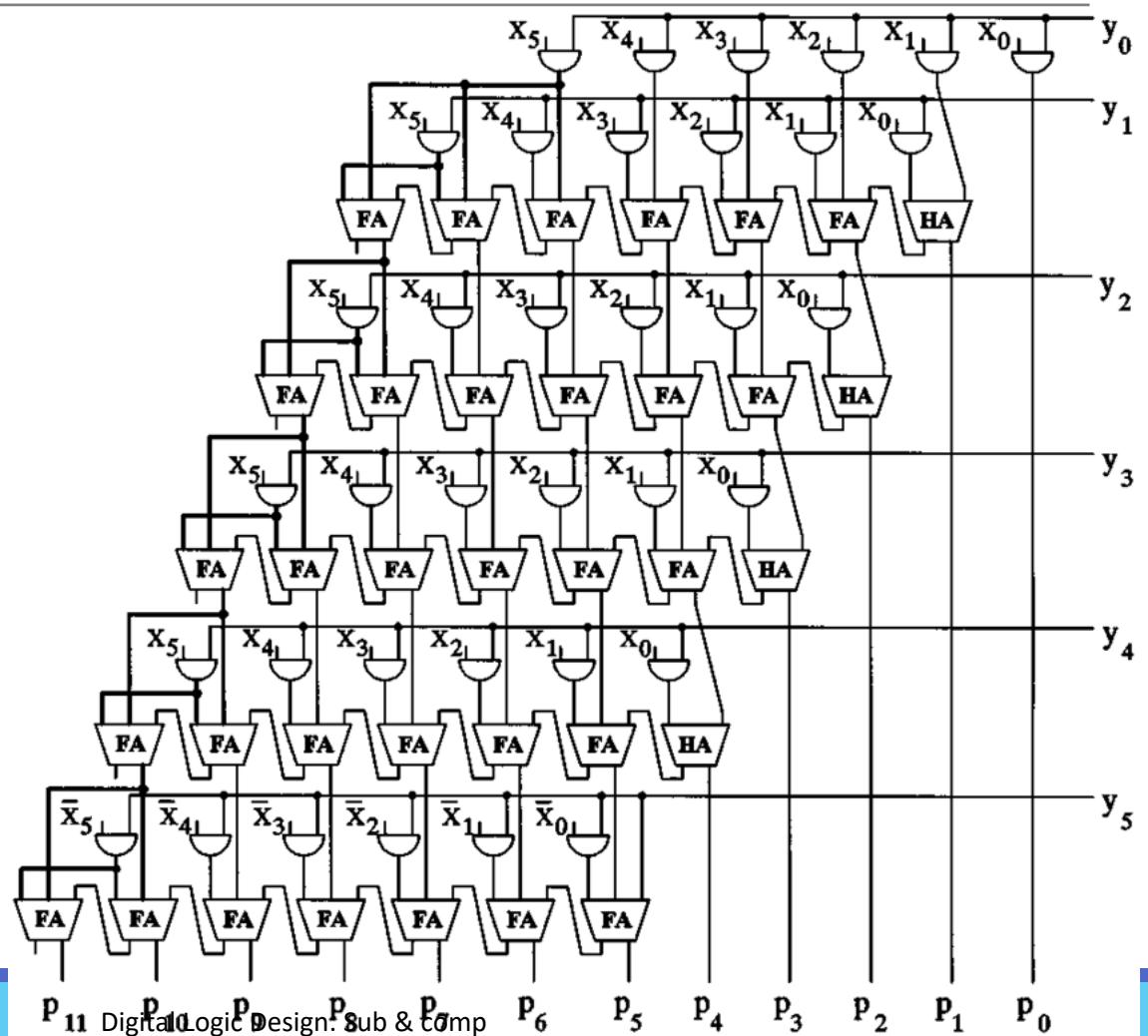
partial products

product

# 4-bit Multiplier: Realization



# 6-bit Multiplier: Realization



# Thank You

---

