

Autonomous Mobile Robot Control Using Kinematics and Dynamics Based Approaches - An Experimental Analysis

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Abstract - Motion control of wheeled mobile robots is based on kinematic or dynamic models and has to take into account nonholonomic constraints resulting from ideal rolling requirement. In order to achieve autonomy in motion, the trajectory has to be generated in real time and has to be corrected in case of reaching a constraint limit, as for example servomotor torque saturation limit. In this paper, results of experimental verification of kinematics and dynamics based control strategies are reported and analysed from the point of view of computational complexity versus performance limitations.

I. INTRODUCTION

Wheeled mobile robots under ideal requirements are known to be difficult to control for centre of mass position and body orientation given that the nonholonomic constraints reduce the number of degrees of freedom. For example, a wheeled mobile robot moving on a flat surface has only two degrees of freedom and cannot be stabilized by a smooth state feedback [1]. Given the stabilization problem, discontinuous and time varying control laws have been proposed [1,2]. Also, for a tricycle with front wheel steering and driving, an artificial holonomic constraint approach was proposed such that smooth state feedback is used for front wheel angular speed and steering angle and an open loop control for body orientation [3]. The state feedback loop enclosed also an input-output linearization of the mobile robot dynamics [4].

The smoothness condition for input-output linearization requires verification of bounded variables and in particular of the servomotor torques versus their saturation limits [3,4]. Predictive control has been used for a holonomic system subject to feedback linearization with bounded inputs [5].

In this paper, these new results in dynamic-based control are verified experimentally and compared to the computationally less intensive approach based on kinematic model only. A short presentation of the experimental mobile robot is followed by the description of the control strategies and by the analysis of the experimental results.

II. EXPERIMENTAL MOBILE ROBOT

The mobile robot shown in fig. 1 has 0.51 x 0.51 x 0.53m and weights 24.5 Kg. It has a tricycle configuration with front wheel driving and steering. The DC servomotors have 75.1:1 planetary gears and have a torque output of maximum 10 Nm.

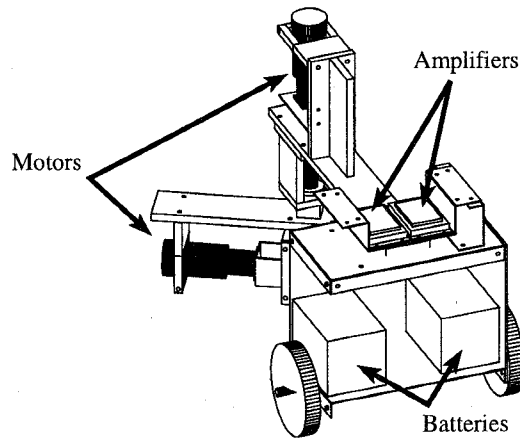


Fig.1 Major Components of the mobile robot

Power for the motors is supplied by two 12 V, 10 Ah batteries connected in series. The dSPACE controller (fig. 2) is DSP based(TMS 320C30) and contains D/A and incremental encoder boards [6]. Each optical encoder has 1000 steps per revolution at motor shaft resulting in 75100 steps per revolution for output shaft. The identification of dry friction torques gave approximately 1.1 Nm.

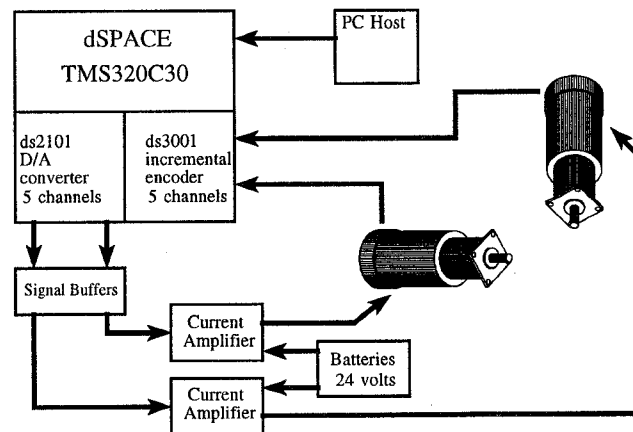


Fig.2 Electrical Components

III. DYNAMIC MODEL OF THE MOBILE ROBOT

The reference frames for the mobile robot are shown in fig.3 where δ, θ are the steering angle and the absolute orientation angle, respectively.

The dynamic model is given by,[3,4]

$$\begin{bmatrix} \dot{\omega}_\delta \\ \dot{\omega}_1 \end{bmatrix} = \begin{bmatrix} f(\omega_1, \omega_\delta, \delta, \theta) \\ F(\omega_1, \omega_\delta, \delta, \theta) \end{bmatrix} + Ad \begin{bmatrix} \tau_d \\ \tau_s \end{bmatrix} \quad (1)$$

where, τ_d, τ_s are the driving and steering torques, respectively, and $\dot{\omega}_1, \dot{\omega}_\delta$ are the driving and steering angular accelerations, respectively (i.e. $\dot{\omega}_\delta = \ddot{\delta}$),

$$f(\omega_1, \omega_\delta, \delta, \theta) = \frac{r_1}{bD} \omega_1 \omega_\delta \left[r_1^2 \left\{ \frac{l}{2b} Q_1 (\sin^2(\delta) - \cos^2(\delta)) + (Q_3 - Q_4) \sin(\delta) \cos(\delta) \right\} \sin(\delta) - D \cos(\delta) \right] \quad (2)$$

$$F(\omega_1, \omega_\delta, \delta, \theta) = -\frac{r_1^2}{D} \omega_1 \omega_\delta \left[\frac{l}{2b} Q_1 (\sin^2(\delta) - \cos^2(\delta)) + (Q_3 - Q_4) \sin(\delta) \cos(\delta) \right] \quad (3)$$

$$Ad = \begin{bmatrix} -\frac{r_1}{bD} \sin(\delta) & \frac{1}{J_1} + \frac{r_1^2}{b^2 D} \sin^2(\delta) \\ \frac{1}{D} & -\frac{r_1}{bD} \sin(\delta) \end{bmatrix} \quad (4)$$

$$D = I_1 + r_1^2 \left[Q_4 \cos^2(\delta) + Q_3 \sin^2(\delta) - \frac{1}{b} Q_1 \cos(\delta) \sin(\delta) \right]$$

Complete equations used for deriving Q_1, Q_2, Q_3, Q_4 and definitions of r_1, b, c, J_1 , are given in [7].

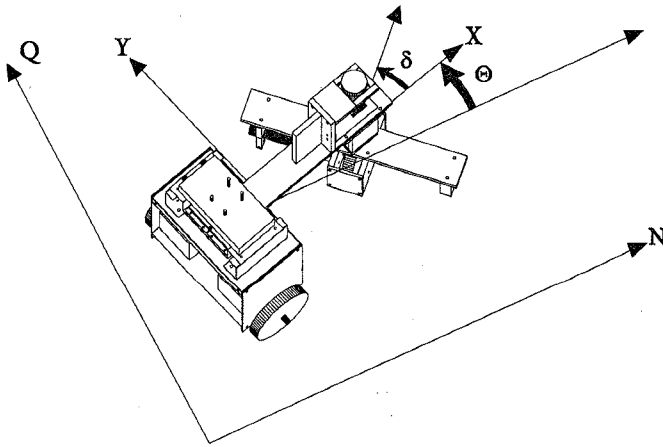


Fig. 3 Reference frames for the mobile robot

IV. TORQUE SATURATION AVOIDANCE FOR A MOBILE ROBOT MOVING ON A HORIZONTAL PLANE

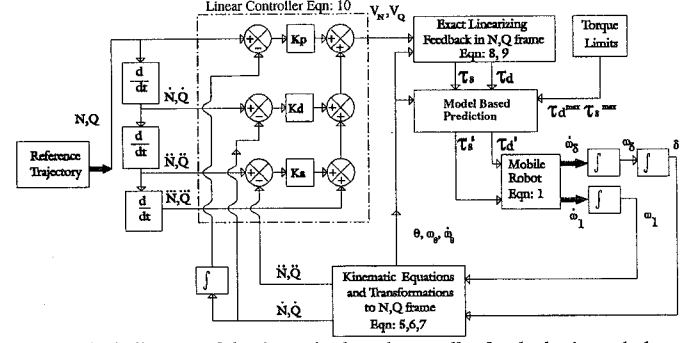


Fig. 4 Block diagram of the dynamics based controller for the horizontal plane motion of a mobile robot with torque saturation avoidance

The mobile robot represented in fig. 1 is analysed for the case of horizontal planar motion with actuator torque saturation avoidance. The block diagram of the dynamics based control system is shown in fig. 4 and contains an exact linearizing feedback controller, whose smoothness condition is achieved by a model based prediction.

The equations corresponding to the blocks of fig. 4 are the following:

Kinematic equations are:

$$\dot{N} = r_1 \omega_1 (\cos \delta \cos \theta - \frac{c}{b} \sin \delta \sin \theta) \quad (5)$$

$$\dot{Q} = r_1 \omega_1 (\cos \delta \sin \theta + \frac{c}{b} \sin \delta \cos \theta) \quad (6)$$

$$\omega_\theta = \dot{\theta} = \frac{1}{b} r_1 \omega_1 \sin \delta \quad (7)$$

Linearizing Feedback Controller is given by,

$$\tau = M^{-1} U \quad (8)$$

$$W = \Delta^{-1} (V - \Delta_0) \quad (9)$$

where:

$$\tau = \begin{bmatrix} \tau_d \\ \tau_s \end{bmatrix}$$

$$M = \begin{bmatrix} 1 & -\frac{r_1}{b} \sin(\delta) \\ 0 & 1 \end{bmatrix}$$

$$U = \begin{bmatrix} u_1 \\ w_2 \end{bmatrix} \quad W = \begin{bmatrix} \dot{u}_1 \\ \tau_s \end{bmatrix} \quad V = \begin{bmatrix} V_N \\ V_Q \end{bmatrix}$$

$$\Delta(\delta, \theta, \omega_1) = \begin{bmatrix} rI \frac{A1}{Den(\delta)} & - \left(A_3 + A_1 \frac{Num(\delta)}{Den(\delta)} \right) \frac{r_1 \omega_1}{J_1} \\ rI \frac{A_2}{Den(\delta)} & - \left(A_4 + A_2 \frac{Num(\delta)}{Den(\delta)} \right) \frac{r_1 \omega_1}{J_1} \end{bmatrix}$$

$$\Delta_O(\delta, \theta, \omega_1, \omega_\delta, \omega_\theta, \dot{\omega}_1) = \begin{bmatrix} \Delta_{ON}(\delta, \theta, \omega_1, \omega_\delta, \omega_\theta, \dot{\omega}_1) \\ \Delta_{OQ}(\delta, \theta, \omega_1, \omega_\delta, \omega_\theta, \dot{\omega}_1) \end{bmatrix}$$

$$\Delta_{ON} = r_1 A1 C2 - r_1 \omega_1 A3 C1 + 2r_1 \dot{\omega}_1 (-A2 \omega_\theta - A3 \omega_\delta) - r_1 \omega_1 \omega_\theta (A1 \omega_\theta - A4 \omega_\delta) - r_1 \omega_1 \omega_\delta (-A4 \omega_\theta + A1 \omega_\delta) - r_1 \omega_1 A3 \dot{\omega}_\theta$$

$$\Delta_{OQ} = r1 A2 C2 - r_1 \omega_1 A4 C1 + 2r_1 \dot{\omega}_1 (A1 \omega_\theta - A4 \omega_\delta) + r_1 \omega_1 \omega_\theta (-A2 \omega_\theta - A3 \omega_\delta) - r_1 \omega_1 \omega_\delta (A3 \omega_\theta + A2 \omega_\delta)$$

$$C1 = \frac{r_1}{b_1} \sin(\delta) \dot{\omega}_1 - \frac{r_1}{b_1} \omega_1 \omega_\delta \cos(\delta)$$

$$C2 = \frac{Num(\delta)}{Den(\delta)} \omega_1 C1 - \frac{Num'(\delta) \omega_1 + Num(\delta) \dot{\omega}_1 \omega_\delta}{Den(\delta)} - \frac{Den'(\delta)}{Den(\delta)} \omega_1 - \omega_1 (\omega_\theta + \omega_\delta)^2$$

$$A1 = \cos(\delta) \cos(\theta) - \frac{c}{b} \sin(\delta) \sin(\theta)$$

$$A2 = \cos(\delta) \sin(\theta) + \frac{c}{b} \sin(\delta) \cos(\theta)$$

$$A3 = \sin(\delta) \cos(\theta) + \frac{c}{b} \cos(\delta) \sin(\theta)$$

$$A4 = \sin(\delta) \sin(\theta) - \frac{c}{b} \cos(\delta) \cos(\theta)$$

and C1, C2, Num(δ), Den(δ), Num'(δ), Den'(δ) are derived in [3,4].

Linear Controller is given by,

$$V = K_p(P_d - P) + K_d(P_d^{(1)} - P^{(1)}) + K_a(P_d^{(2)} - P^{(2)}) + P_d^{(3)} \quad (10)$$

where,

$$V = [V_N \ V_Q]^T \quad P = [N \ Q]^T$$

V_N, V_Q are the resolved jerks.

The one-step (Δt) model based prediction, shown in fig. 5, compares the input torque commands τ_d and τ_s with saturation torque limits and reduces any torque greater than the saturation limit by its saturation limit resulting in a new torque command vector τ_{safe} . For this τ_{safe} , the new (8),(9) written as,

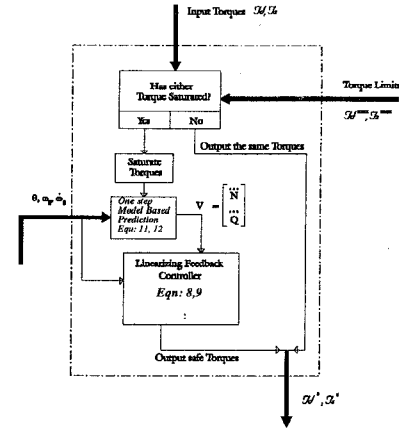


Fig. 5 Block diagram of the model based prediction and recalculation of the feasible torques for avoiding torque saturation

$$U = M \tau_{safe} \quad (11)$$

$$V_{new} = \Delta W + \Delta_0 \quad (12)$$

where M, Δ , Δ_0 were assumed equal to their values for previous resolved jerks given by the linear controller (10) (fig. 4). If M, Δ , Δ_0 vary significantly with the change from the old to the new V, i.e., functions of V, an iterative approach can be used to obtain the set of commands V and τ consistent with the path and within torque saturation limits. The new jerk commands vector V_{new} is used as input to the linearizing feedback controller for recalculating M, Δ , Δ_0 , W and τ using (8),(9) (fig.5). If the resulting torque commands are within saturation limits, they are used as τ_d^s, τ_s^s inputs to the mobile robot servomotors (fig. 4). If not, the iterations using the model based prediction of fig. 5 are continued. In simulations this approach converged fast. In experiments it is preferable to use very short steps Δt such that M, Δ , Δ_0 vary insignificantly with the changes in V. In this case the model based prediction is reduced to an algebraic dependence between τ and V in (10) and (11) and can be solved in one iteration. The block diagram for the model based prediction is given in fig. 5, shown for one iteration in order to keep the block diagram simple.

V. KINEMATICS AND DYNAMICS BASED CONTROLLER

Kinematics based controller is shown in fig. 6. Front wheel angular speed ω_1 and steering angle δ are controlled by a proportional and, respectively, proportional derivative control laws in order to follow the reference trajectory given in operational space N-Q. The position and orientation of the mobile robot are obtained by odometry using front wheel angular speed and steering angle measurements. Servomotor torques τ_d and τ_s are limited by software to conventional saturation torques chosen lower than actual saturation torques.

VII. CONCLUSIONS

The experimental results confirm that, while more intensive computationally, dynamics based controller is more suitable in achieving autonomous motion for mobile robots given that the controller gains do not have to be retuned for various reference paths and that servomotor torque saturation limits are not reached as a result of the model based prediction used in conjunction with the linearizing feedback controller.

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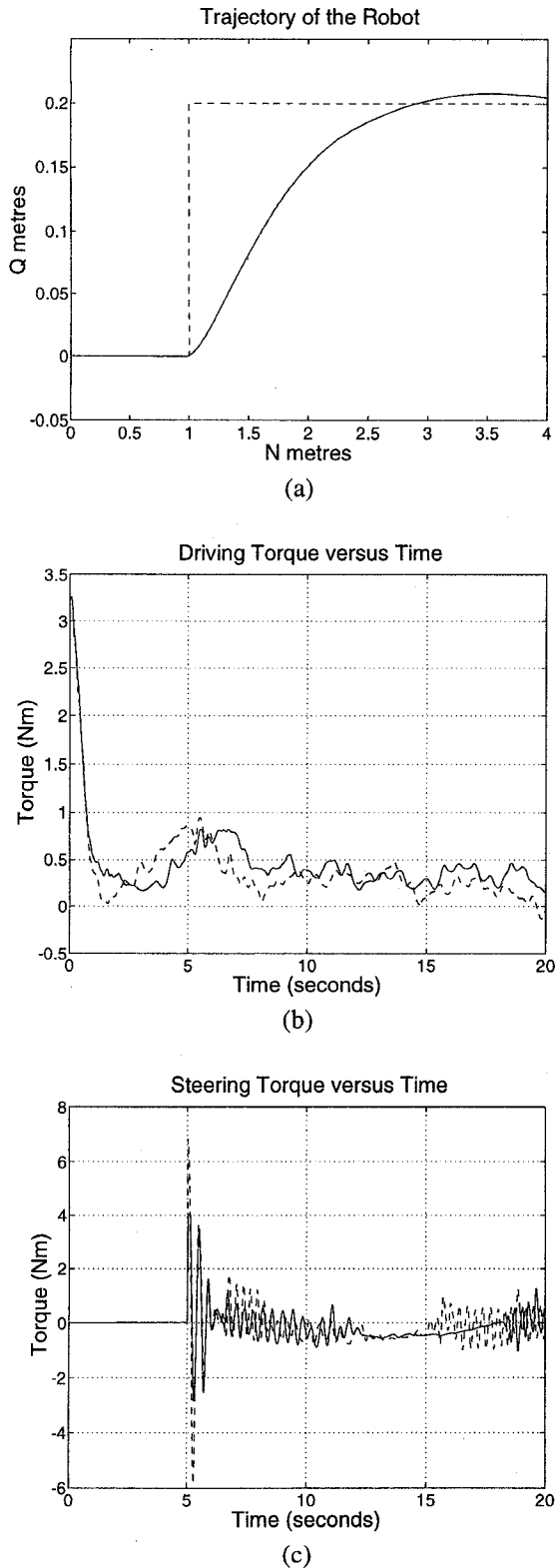


Fig. 8 Experimental Results using a Dynamics based controller
a) Trajectory with torque saturation
b) Driving Torque versus time
c) Steering Torque versus time