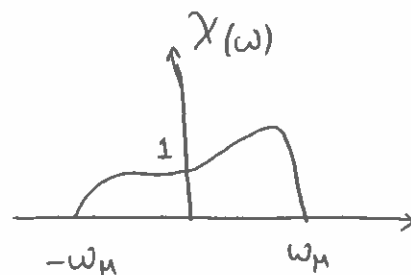
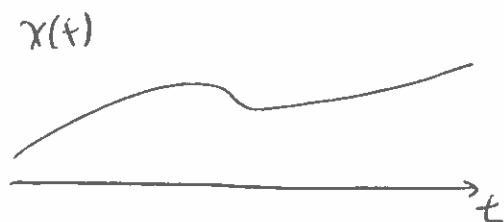
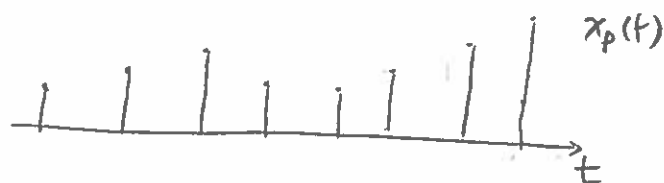


1.

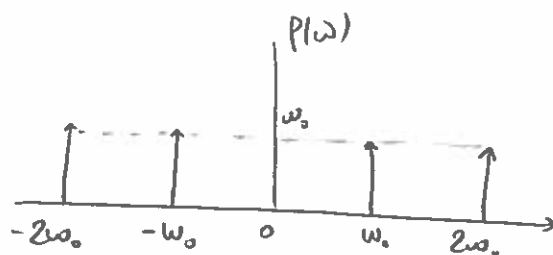


Moreover, we are given that  $p(t) = \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$  and  $x_p(t) = x(t)p(t)$ .

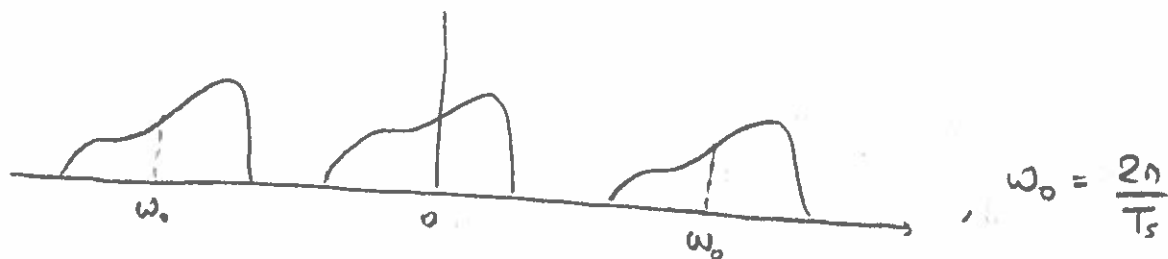
(a) Sketch of a representation of  $x_p(t)$ , is given below:



b) For us to sketch  $P(\omega)$ , we do the following:

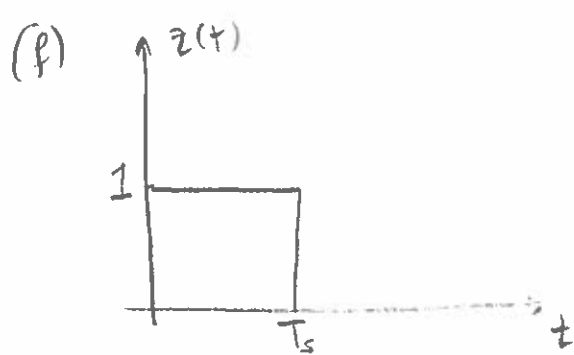


c) For us to sketch  $X_p(\omega)$ , we do the following:



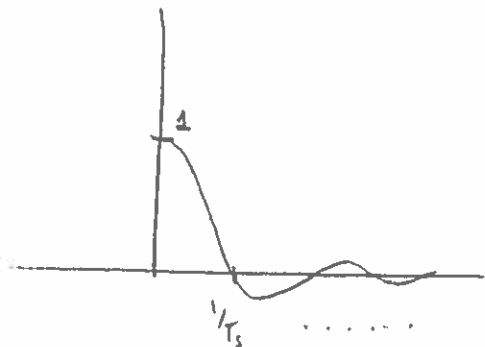
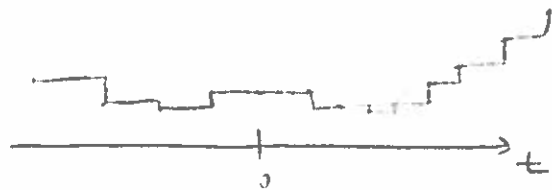
d) To ensure that  $X_p(\omega)$  contains all the information present in  $X(\omega)$ , we need to have that  $\frac{2\pi}{T_s} < \frac{\omega_m}{2}$ .

e) In order to recover  $x(t)$  from  $x_p(t)$  exactly, we can use a bandpass with frequency  $[-\omega_m, \omega_m]$ , which in the time domain is a sinc function.

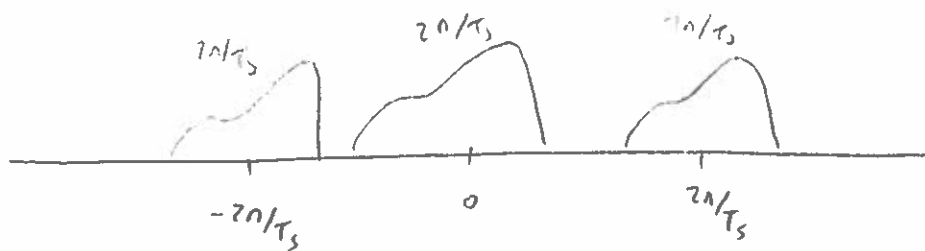


$$X_z(t) = X_p \cdot z(t)$$

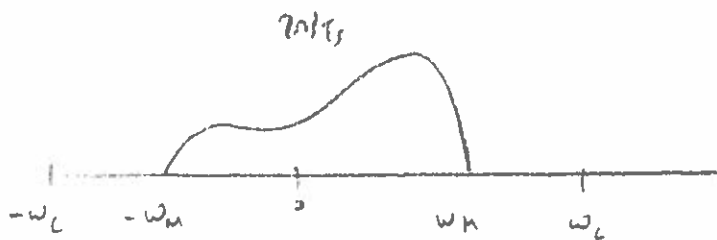
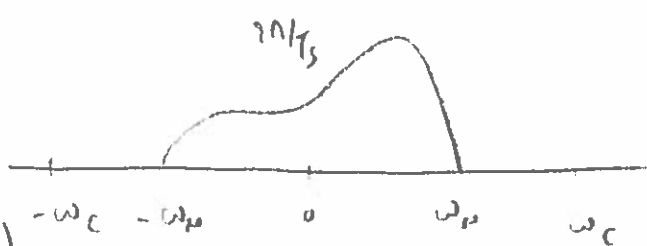
(g) It is in essence the zero hold of  $X_p(t)$  and it gives us the following result



(h) Now we need to sketch  $X_z(\omega)$ . It is known that  $Z(\omega) = \frac{2 \sin(\omega T_s/2) e^{j\omega T_s/2}}{\omega}$  since it is a sinc function. Hence,



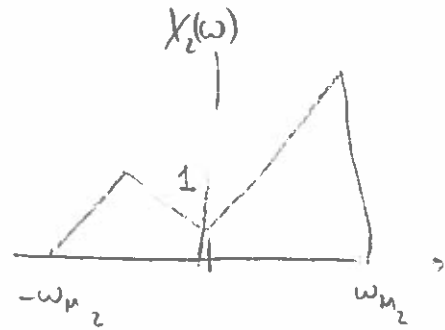
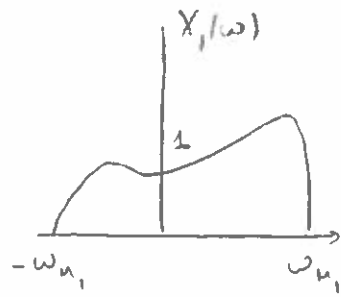
(i)  $\bar{X}(\omega) = X_z(\omega) H(\omega)$  ,  $\hat{X}(\omega) = X_p(\omega) H(\omega)$



j)  $\bar{X}(\omega)$  is going to decay since it contains a sinc function in its former transform, while they are both the same at 0,  $2n/T_s$ .

k) The ratio is 1, as pre-stated they are the same at the point  $\frac{\bar{X}(\omega)}{\hat{X}(\omega)} = 1$

2.  $y(t) = x_1(t)\cos(\omega_1 t) + x_2(t)\cos(\omega_2 t)$ , if  $X_1(\omega) = 0, X_2(\omega) = 0$ , if  $|\omega| > \omega_m$

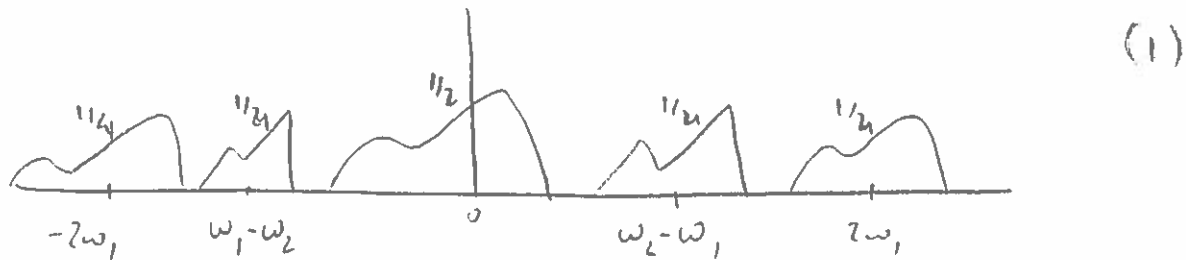


(a)

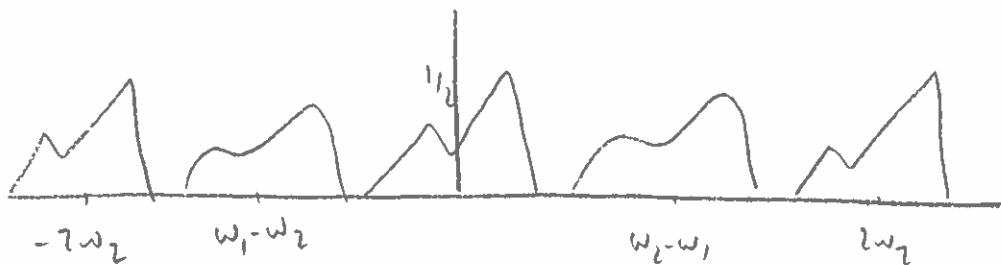


(b)  $y(t)\cos(\omega_1 t) = x_1(t)\cos^2(\omega_1 t) + x_2(t)\cos(\omega_1 t)\cos(\omega_2 t)$  (1)

$y(t)\cos(\omega_2 t) = x_1(t)\cos(\omega_1 t)\cos(\omega_2 t) + x_2(t)\cos^2(\omega_2 t)$  (2)



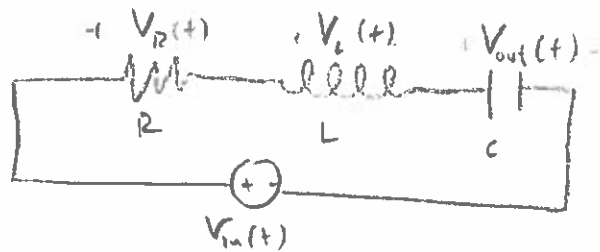
(1)



(2)

(c) We could multiply  $y(t)$  with  $\cos(\omega_1 t)$ , use a bandpass filter of  $[-\omega_m, \omega_m]$  and get back  $x_1(t)$ . The same process can be followed to deduce  $x_2(t)$ , multiplying by  $\cos(\omega_2 t)$ . It is important to mention that this will leave the signal scaled at  $1/2$ , so multiply by

$$3. (a) \left. \begin{aligned} i(t) &= C \frac{d}{dt} V_{out}(t) \\ V_L(t) &= L \frac{d}{dt} i(t) \end{aligned} \right\} \Rightarrow$$



$$\Rightarrow V_{in}(t) = V_R(t) + V_L(t) + V_{out}(t) \Rightarrow$$

$$\Rightarrow V_{in}(t) = RC \frac{d}{dt} V_{out}(t) + L \frac{d^2}{dt^2} V_{out}(t) + V_{out}(t)$$

$$(b) V_{in}(t) = e^{j\omega t}$$

$$V_{out}(t) = H(\omega) e^{j\omega t} \parallel \rightarrow e^{j\omega t} = [RCj\omega e^{j\omega t} - LC\omega^2 e^{j\omega t} + e^{j\omega t}] H(\omega)$$

$$\Rightarrow H(\omega) = \frac{1}{RCj\omega - L(\omega^2 + 1)}$$

$$(c) |H(\omega)| = \frac{1}{\sqrt{R^2 C^2 \omega^2 - (L(\omega^2 + 1))^2}}$$

$$(d) \frac{d}{d\omega} (|H(\omega)|) = \frac{1}{2\sqrt{R^2 C^2 \omega^2 - (L(\omega^2 + 1))^2}} (2\omega R^2 C^2 - 2(L(\omega^2 + 1) \cdot 2L\omega) =$$

$$= 2\omega R^2 C^2 - 4L(\omega(L(\omega^2 + 1))) = 2\omega R^2 C^2 - 4L^2 C^2 \omega^3 + 4LC\omega = 0$$

$$= 2R^2 C^2 - 4L^2 C^2 \omega^2 + 4LC = 0 \Rightarrow \omega = \pm \sqrt{\frac{4LC + 2R^2 C^2}{4L^2 C^2}} \Rightarrow$$

$$\Rightarrow \omega = \pm \sqrt{\frac{1}{LC} + \frac{R^2}{2L^2}}$$

$$\text{or } \omega = \frac{\sqrt{2}}{2LC} \cdot \sqrt{2LC + R^2 C^2} \quad \text{maximizes}$$

Bode Diagram

