

Uplink SCMA with multiple receive antennas: ideas for TB-ESGA MPA receiver

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1 Preliminaries

$\mathcal{A}_q = \{a_1, \dots, a_q\}$ is an alphabet (set) of q unique complex projections

$\mathcal{C} = \{\mathbf{c}^{(1)}, \mathbf{c}^{(2)}, \dots, \mathbf{c}^{(M)}\}$ is the mother codebook with M symbols where each symbol is N -dimensional (complex dimensions), i.e. $\mathbf{c}^{(m)} \in \mathbb{C}^N$

the sparse codebook of user j is $\mathcal{X}_j = \{\mathbf{x}^{(1)}, \dots, \mathbf{x}^{(M)}\}$, where $\mathbf{x}_j^{(m)}$ is the sparse version (K complex dimensions) of $\mathbf{c}^{(m)}$, obtained with the signature \mathbf{f}_j of user j , and each symbol has $K - N$ null dimensions;

\mathbf{x}_j is the transmitted symbol of user j , drawn from \mathcal{X}_j using a uniform distribution such that $P[\mathbf{x}_j = \mathbf{x}^{(m)}] = 1/M$ for $m = 1, \dots, M$ (a *priori* pmf before decoder sends its estimated LLRs)

$P[\mathbf{x}_j[k] = a_i]$ is the probability that the sparse symbol of user j , \mathbf{x}_j , has projection a_m in FN k ; it is computed by adding up the probabilities of symbols $\mathbf{x}_j^{(m)}$ where $\mathbf{x}_j^{(m)}[k] = a_i$

$\omega_{j,k,n_r} = \{u : |h_{k,u,n_r}|^2 \geq r |h_{k,j,n_r}|^2, 0 \leq r \leq 1\}$ is the set of strong neighbors of user j in FN k for receive antenna n_r ; the set has d strong users, $|\omega_{j,k,n_r}| = d$, being users u_1, \dots, u_d

2 TB-ESGA MPA Algorithm for multiple Rx antennas

Algorithm 1 is a pseudo-code of the implementation of TB-ESGA-MPA with multiple receive antennas.

3 Simulation Scenario 1

Benchmark SCMA $J = 6$ users sharing $K = 4$ resources, using a codebook with $M = 4$ codewords of $N = 2$ complex-dimensions over an alphabet of size $q = 4$.

The base-station has $N_r = 1, 2, 3, 4$ antennas. MPA is used without threshold-approximation ($r = 0$).

LDPC code is short-block length with codeword length $n = 128$ bits, rate $1/2$. Each codeword is divided in packets of $k_b = \log_2(M) = 2$ bits, and each packet is mapped to an SCMA codeword. $N_s = 64$ SCMA symbols are sent through the Rayleigh channels $\mathbf{H}_j \in \mathbb{C}^{K \times N_r}$ for $j = 1, \dots, 6$.

Iterative detection and decoding is performed using $T_o = 3$ outer loops. During the first outer loop, $T_m = 5$ inner MPA iterations are performed. After, when the decoder provides soft-bits to the input of the MPA, only 1 inner MPA iteration is performed in each outer loop.

Algorithm 1 Pseudo-code of TB-ESGA-MPA with N_r receive antennas.

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procedure TB-ESGA-MPA-MANY-RX(...)
  for  $t = 1, \dots, T_m$  do                                     ▷ MPA iterations
    for  $k = 1, \dots, K$  do                                       ▷ For every FN
      for  $j$  in  $\phi_k$  do
        for  $i = 1, \dots, q$  do                                     ▷ For every VN connected to FN k
          for  $n_r = 1, \dots, N_r$  do                               ▷ For every codebook projection of VN j in FN k
            Find  $\omega_{j,k,n_r}$  and  $\bar{\omega}_{j,k,n_r}$  with threshold  $r |h_{k,j,n_r}|^2$ 
            Evaluate  $\mu_{j,k}$  and  $\sigma_{j,k}^2 \forall j \in \omega_{j,k,n_r}$ 
             $y' = y_{k,n_r} - h_{k,j,n_r} a_i - \mu_{j,k}$ 
            ▷ Sets of strong and weak neighbors
            ▷ Gaussian Approx.
            ▷ Result=Eff.Noise $_{k,n_r}$  + interference $_{k,n_r}$ 

            Proj Prob $_{k \rightarrow j}(i) = \max_{m_{u_1}, \dots, m_{u_d}} \left[ \log \left( \frac{1}{2\pi \sqrt{\sigma^2 + \sigma_{j,k}^2}} \right) - \frac{\left| y' - \sum_{i=u_1}^{u_d} h_{k,i,n_r} \mathcal{A}_q(m_i) \right|^2}{2(\sigma^2 + \sigma_{j,k}^2)} + \sum_{l=u_1}^{u_d} P[\mathbf{x}_l[k] = a_{m_l}] \right]$  (1)

          end for
           $I_{k \rightarrow j}(m) = \text{Proj Prob}_{k \rightarrow j}(i)$  for every symbol  $m$  where  $\mathbf{x}_j^{(m)}[k] = a_i$ 
        end for
        Normalize messages  $I_{k \rightarrow j}(m)$  such that  $\sum_{m=1}^M \exp[I_{k \rightarrow j}(m)] = 1$ 
      end for
    end for
    for  $j = 1, \dots, J$  do                                       ▷ For every VN
      for  $k$  in  $\varphi_j$  do                                           ▷ For every FN connected to VN j
        if  $t < T_m$  then
           $\varphi = \varphi_j - k$                                      ▷ remove previous belief of  $k$  to compute the message to  $k$ 
        else
           $\varphi = \varphi_j$                                          ▷ Last iteration ( $t = T_m$ ): use messages from all connected FNs to compute final beliefs
        end if
        for  $m = 1, \dots, M$  do                                     ▷ For every symbol of the codebook
           $I_{j \rightarrow k}(m) = \sum_{k \in \varphi} I_{k \rightarrow j}(m) + \log(P[\mathbf{x}_j^{(m)}])$ 
        end for
      end for
      Normalize messages  $I_{j \rightarrow k}(m)$  such that  $\sum_{m=1}^M \exp[I_{j \rightarrow k}(m)] = 1$ 
      if  $t = T_m$  then
        Use  $I_{j \rightarrow k}$  as the a posteriori pmf of  $\mathcal{X}_j$  to calculate LLRs
      end if
    end for
  end for
end procedure

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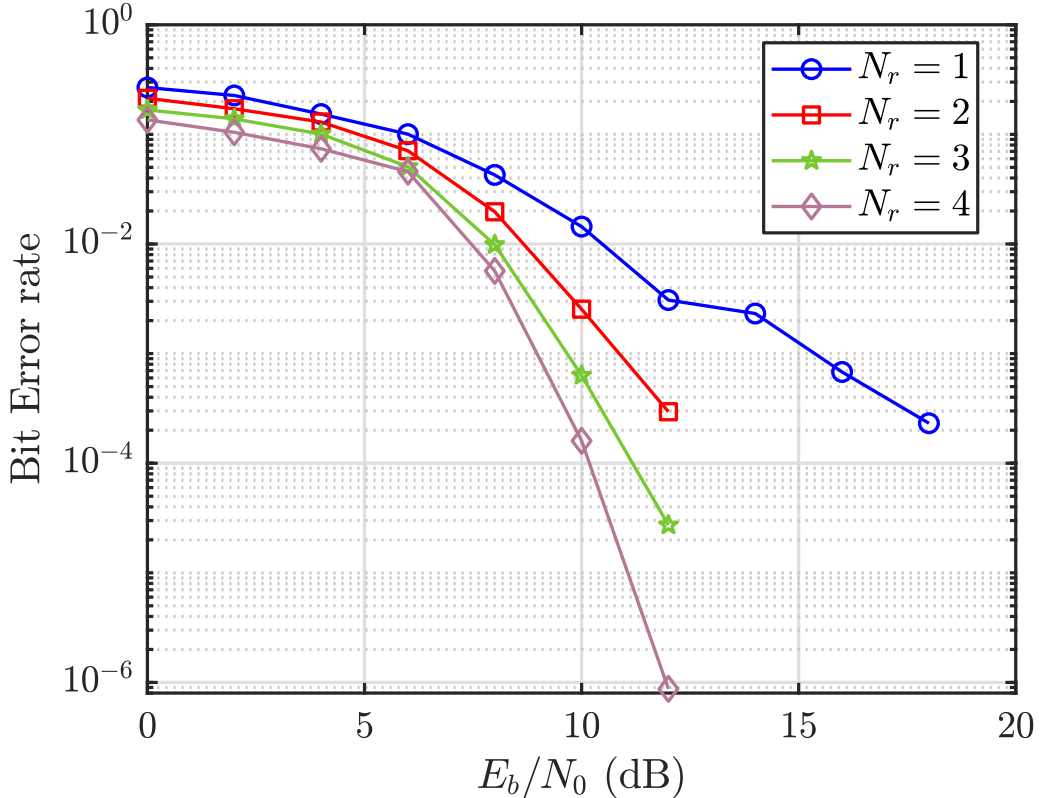


Figure 1: Results for simulation scenario 1.