

Principle of Automatic Control II

(自动控制原理II)

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学时: 40学时

考试: 闭卷

References

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Chapter 3. Analyses in Time Domain for Linear Systems

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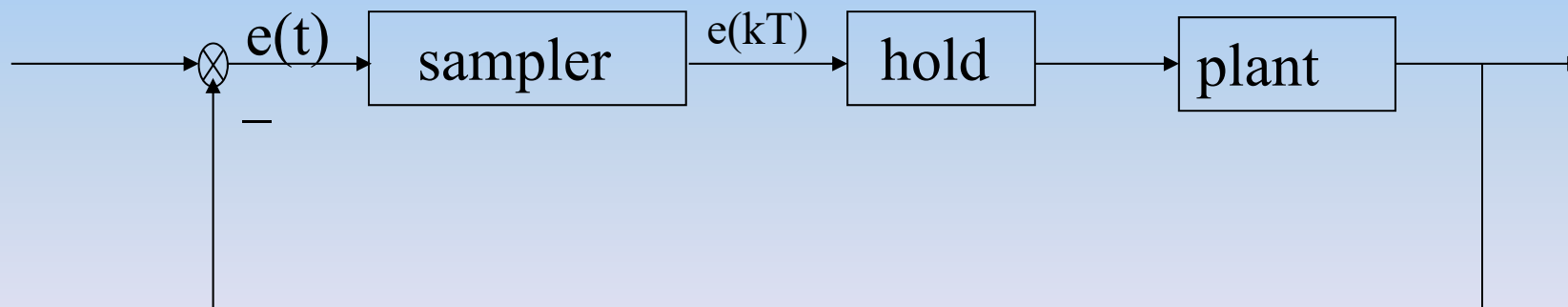
7.1 Introduction

Discrete-Time Systems:

Types: $\begin{cases} \text{Sampling systems:} & \text{Discrete Time, Continuous Value} \\ \text{Digital systems:} & \text{Discrete Time, Quantized Value} \end{cases}$

Digital System: There is one or more impulse series or digital signals in the system.

Sampled-Data System: a system that is continuous except for one or more sampling operations.



Sampled-data control system

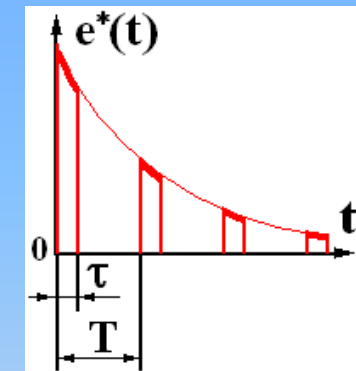
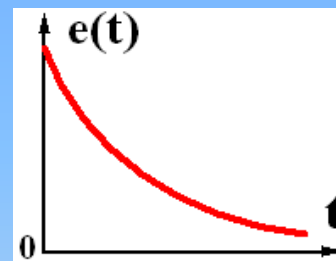
$e(kT)$ is obtained by sampling a continuous signal $e(t)$.

A/D : analog to digital converter

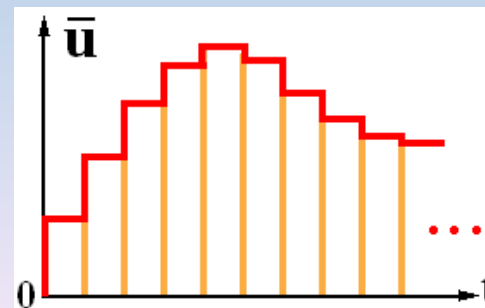
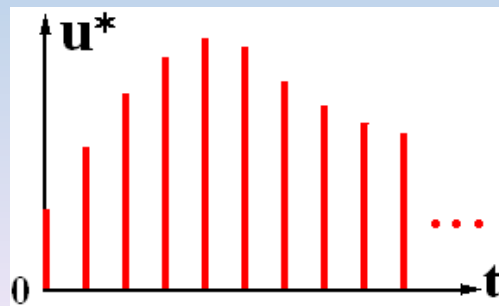
D/A : digital to analog converter

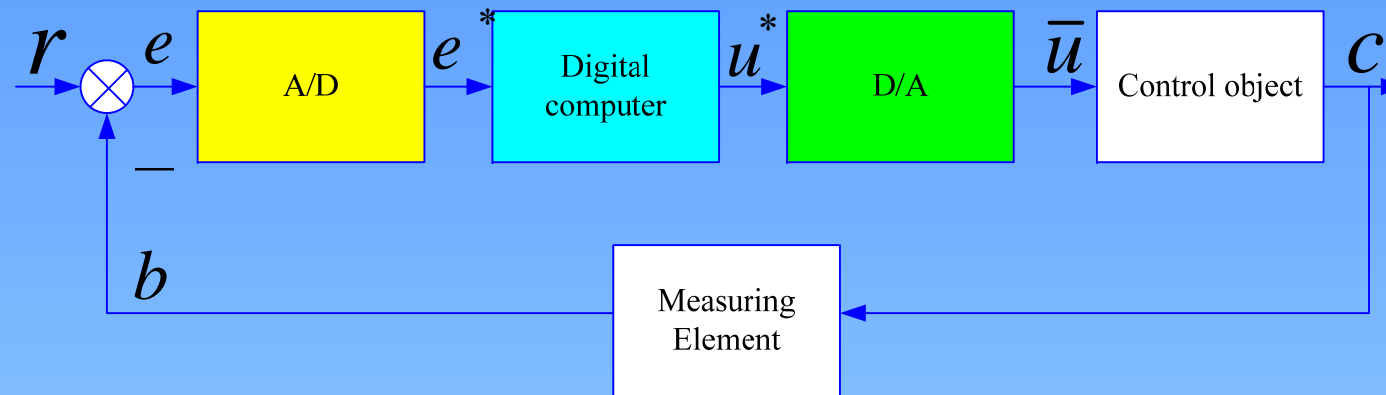
A/D process

- **Sampling** — Time sampled
- **Quantization** — Value quantized

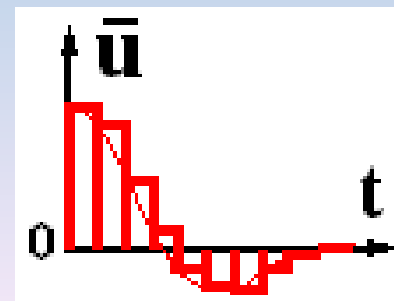
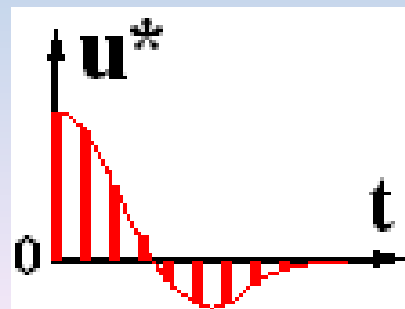
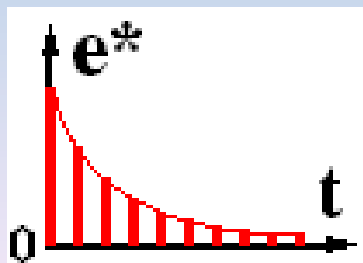
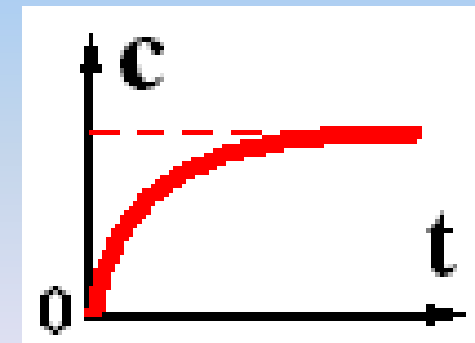
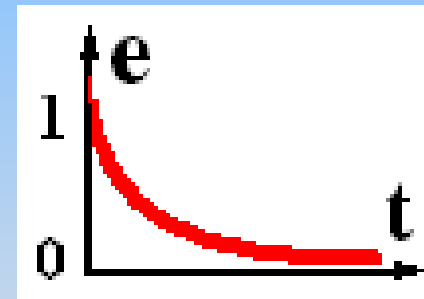
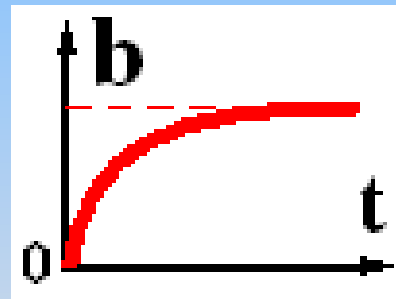
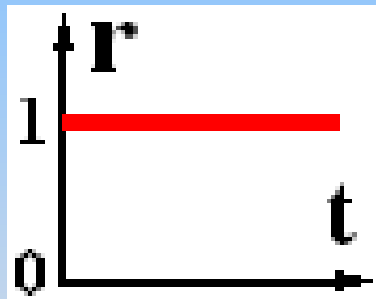


D/A process





Computer Controlled Systems



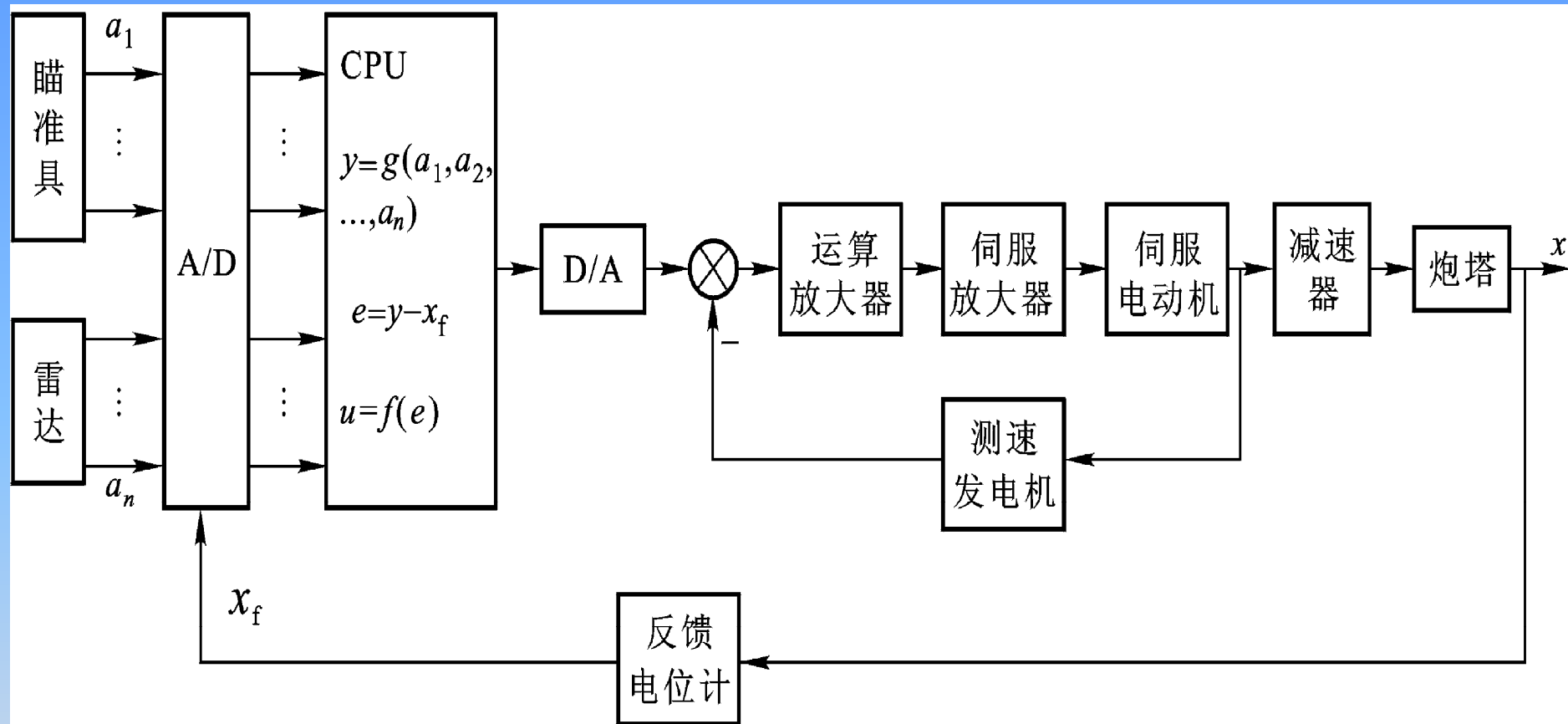


Fig 7-1 机载火力控制系统原理

History of Discrete-time system (p. 196-197)

DDC-Direct Digital Control (直接数字控制系统)

SCC- Surveillance Computer Control System(计算机监督控制系统)

**TDE- Total and Distributed Control(集散控制系统):
muti-agent robots**

Advantages and Disadvantages

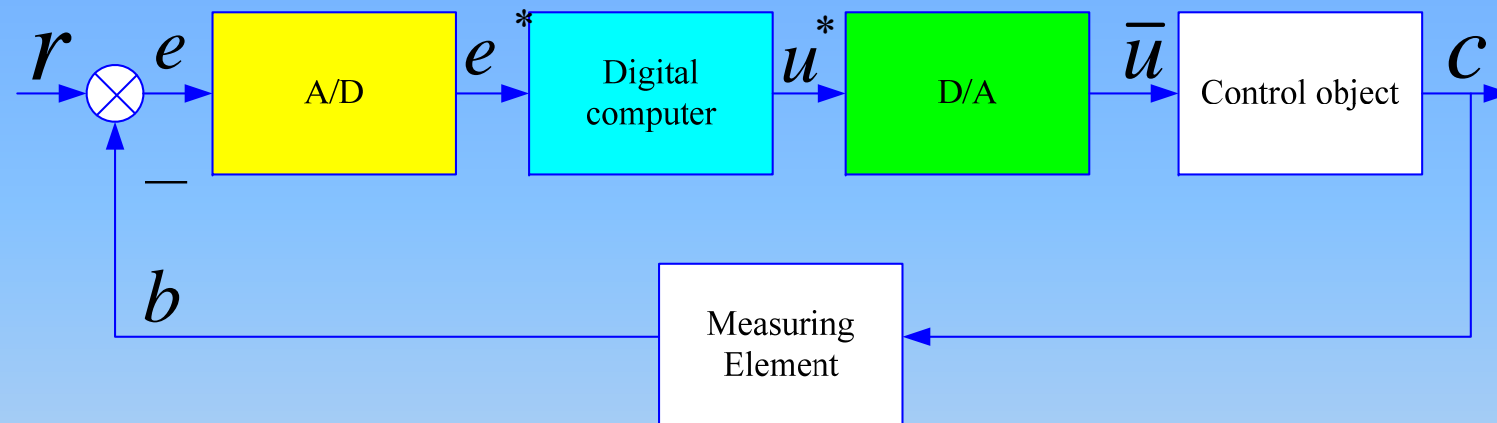
Computer Control System

- {
 - (1) Calculations are performed in the software. Easy for modification.
 - (2) Complex control laws easily realized;
 - (3) Reduced sensitivity to noise;
 - (4) One computer for multi-tasks, high utilization ;
 - (5) Network for process automation, macro-management and remote control.

- {
 - (1) Information between samples is lost. Compared with continuous system in the similar condition, the performance is reduced;
 - (2) Needs A/D and D/A conversion devices.

7.2 The Sampling Process and Sampling Theorem

7.2.1 The Sampling Process

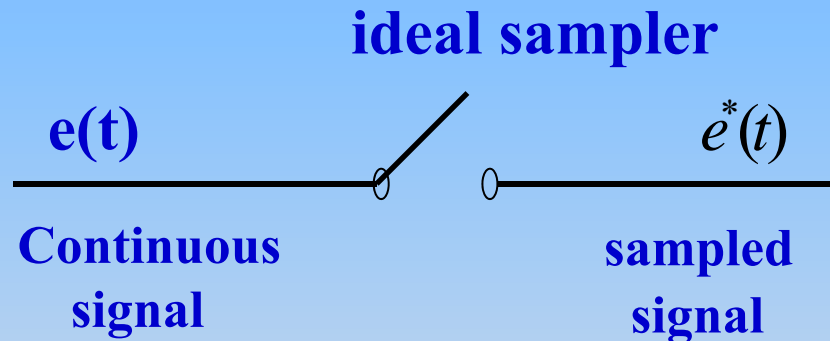


Computer Control System

Question: In the above computer control system, which signals are discrete, which signals are continuous?

- **Sampling Process: Continuous signal \rightarrow Discrete Signal**
- **Holding Process: Discrete Signal \rightarrow Continuous Signal.**
- **The two are inverse process to each other.**

Sampler: A switch which closes every T seconds for one instant of time.

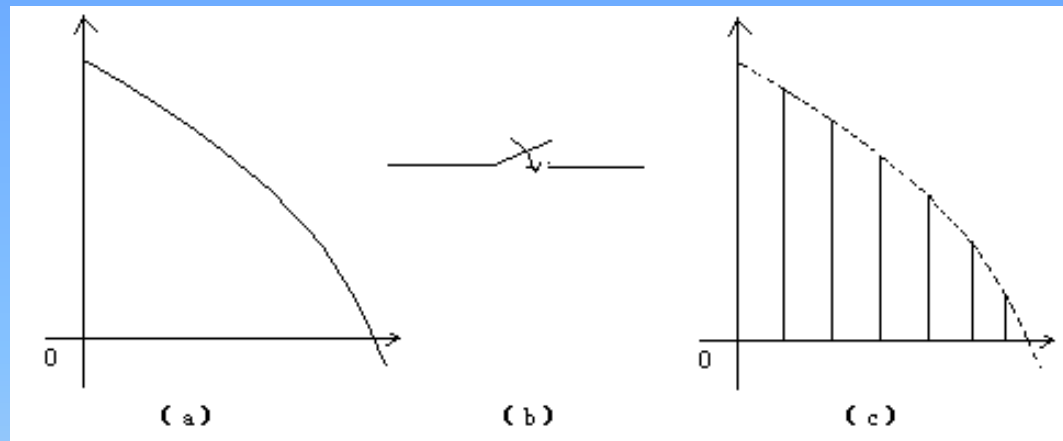


Where T is called the sampling period.

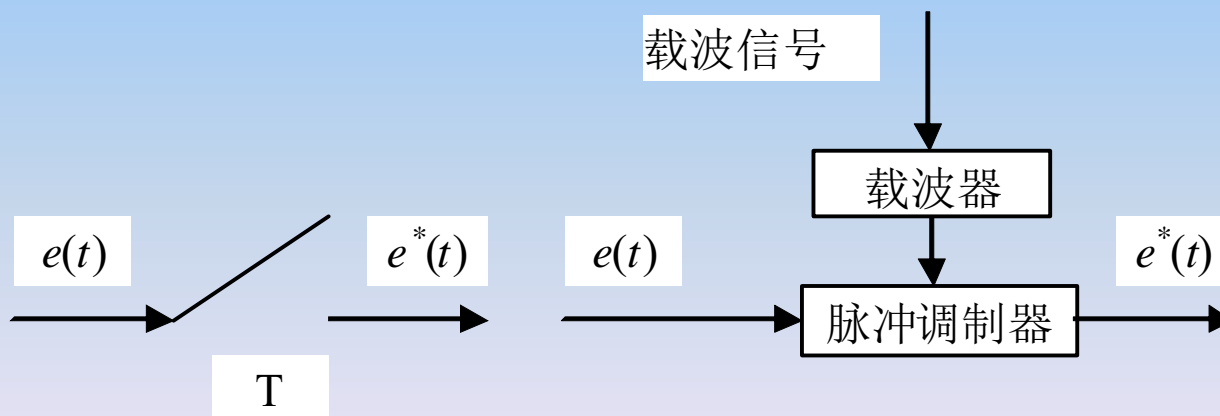
Ideal sampling process:

- (1) $t \ll T$. The sampling process is completed instantaneously
- (2) Word Length is enough, thus $e^*(Kt) = e(Kt)$

Types of Samplers: ideal, periodical, random,...



Sampling Process



7.2.2 Mathematical Model for sampling Signals

1、 Some ideal assumptions

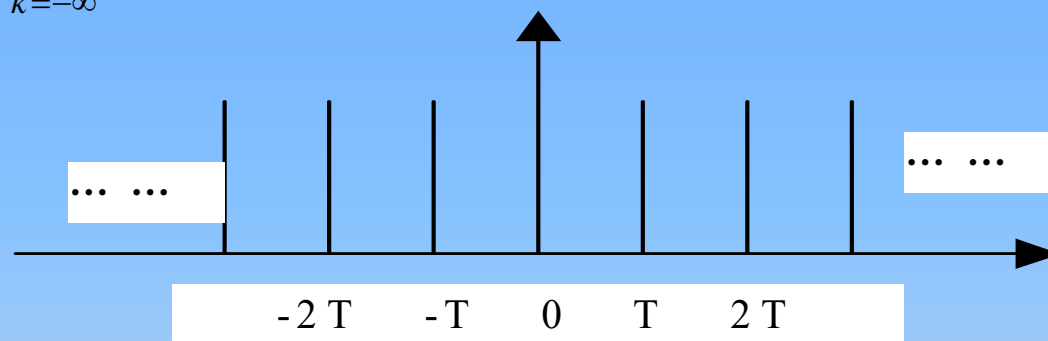
- The sampling process is completed instantaneously ;
- The signals in and out the sampler have no difference;
- $\tau \ll T$, that is $\tau \rightarrow 0$;
- The output is constant when sampler shuts down;
- Sample Period T is a constant.

2、 Unit Impulsive Signal $\delta(t)$

$$\delta(t) = \begin{cases} 1 & t = 0 \\ 0 & t \neq 0 \end{cases}$$

3、 Unit Impulse Sequence (Unit Impulse Train)

$$\delta_T(t) = \sum_{k=-\infty}^{\infty} \delta(t - kT) = \delta(t) + \delta(t - T) + \cdots + \delta(t - kT) + \cdots$$



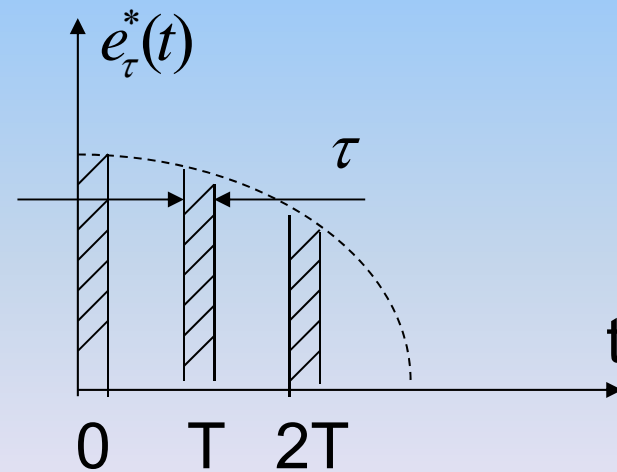
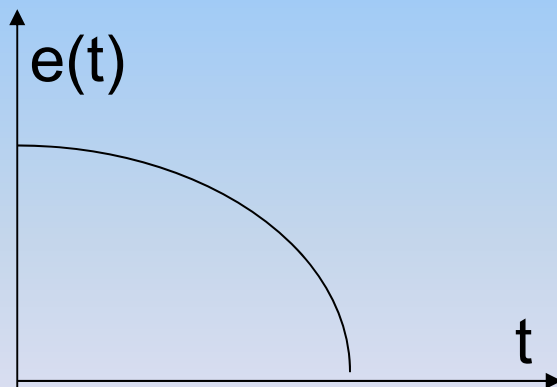
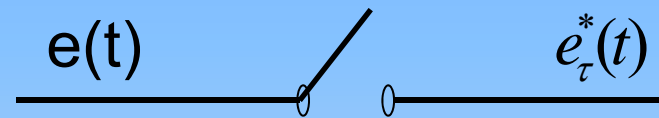
Unit Impulse sequence

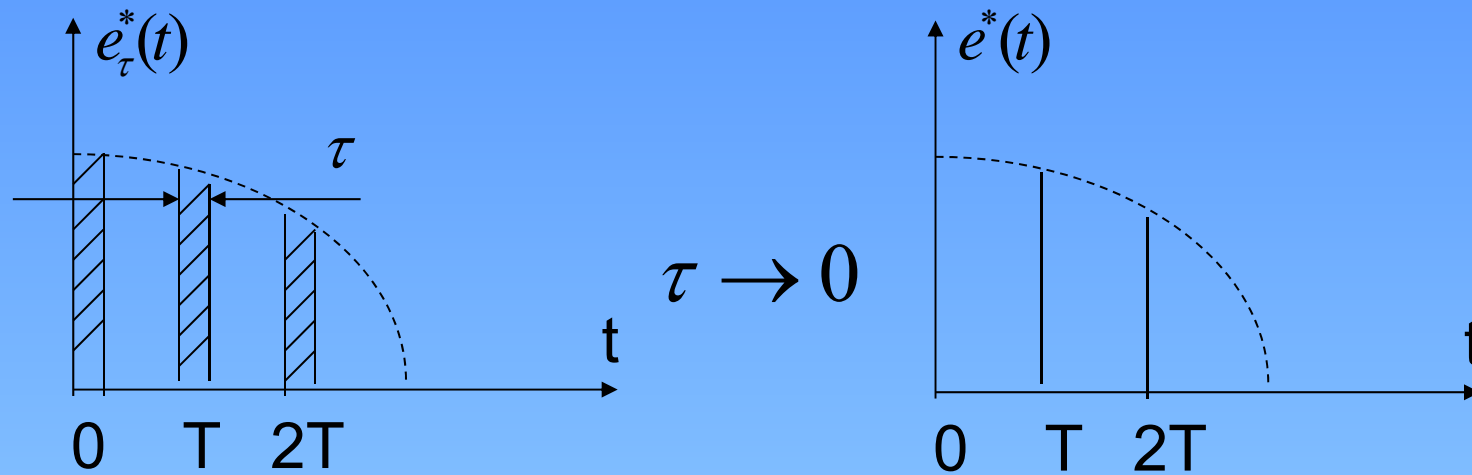
4、 Sampling Signal

$$e^*(t) = \sum_{k=-\infty}^{\infty} e(t) \delta(t - kT)$$

$$e^*(t) = \sum_{k=0}^{\infty} e(kT) \delta(t - kT)$$

real sampler





So the sampling operation can be expressed as

$$e^*(t) = \sum_{k=0}^{+\infty} e(kT) \cdot \delta(t - kT)$$

or

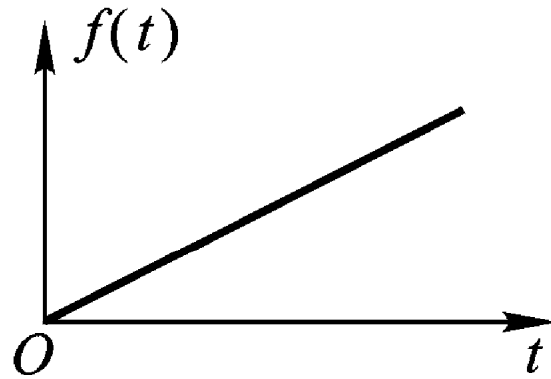
$$e^*(t) = e(t) \cdot \sum_{k=0}^{\infty} \delta(t - kT)$$

or

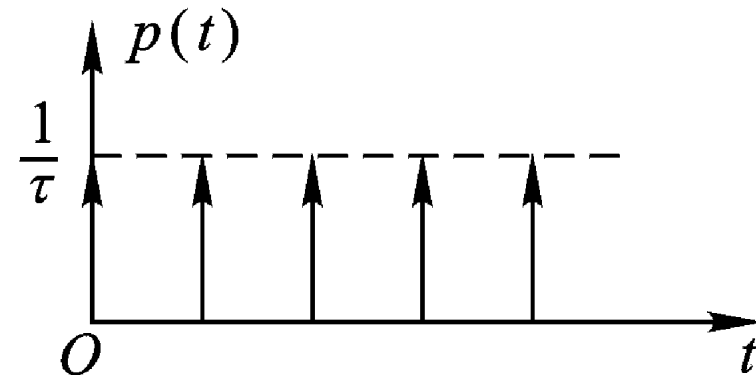
$$e^*(t) = e(t) \cdot \delta_T(t)$$

where

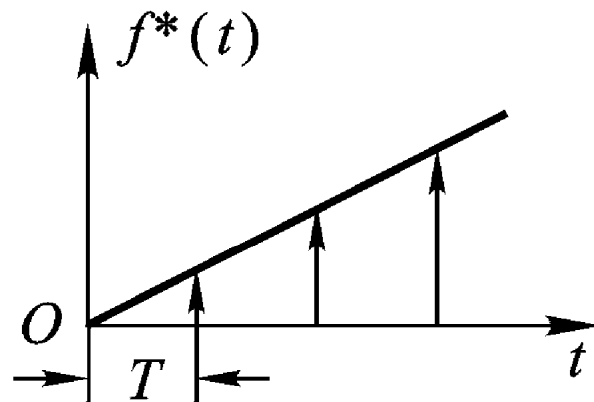
$$\delta_T(t) = \sum_{k=0}^{\infty} \delta(t - kT)$$



(a)



(b)



(c)



$$f_{\tau}^*(t) = p(t) \cdot f(t)$$

Fig 7 - 3 Sampling Process

Laplace Transformation

(6) 位移定理:

a. 实域中的位移定理, 若原函数在时间上延迟 τ , 则其象函数应乘以 $e^{-\tau \cdot s}$

$$L[f(t - \tau)] = e^{-\tau \cdot s} F(s)$$

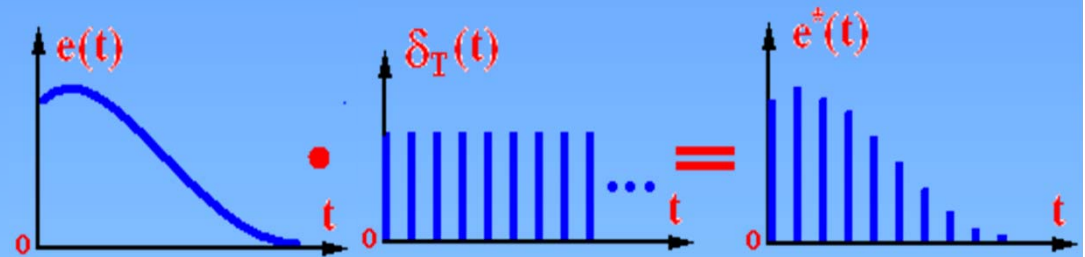
b. 复域中的位移定理, 象函数的自变量延迟 a , 原函数应乘以 e^{at} 即

$$L[e^{at} f(t)] = F(s - a)$$

Ideal sampling sequence

$$\delta_T(t) = \sum_{k=0}^{\infty} \delta(t - kT)$$

$$e^*(t) = e(t) \cdot \delta_T(t)$$



$$= e(t) \cdot \sum_{k=0}^{\infty} \delta(t - kT) = \sum_{k=0}^{\infty} e(kT) \cdot \delta(t - kT)$$

(2) L : $E^*(s) = L[e^*(t)]$

$$= L \left[\sum_{k=0}^{\infty} e(kT) \cdot \delta(t - kT) \right] = \sum_{k=0}^{\infty} e(kT) \cdot e^{-kTs}$$

Example 7-1 $e(t) = 1(t)$ Obtain $E^*(s)$

$$E^*(s) = \sum_{k=0}^{\infty} e(kT) \cdot e^{-kTs}$$

Solution $E^*(s) = \sum_{k=0}^{\infty} 1 \cdot e^{-kTs}$

$$= 1 + e^{-Ts} + e^{-2Ts} + \dots = \frac{1}{1 - e^{-Ts}} = \frac{e^{Ts}}{e^{Ts} - 1}$$

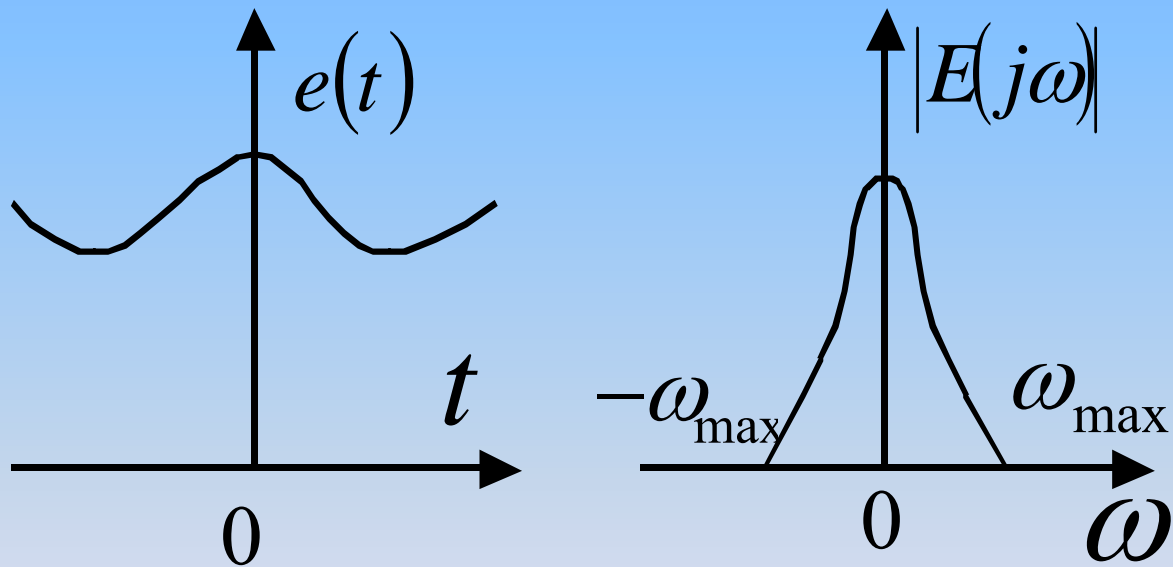
Example 7-2 $e(t) = e^{-at}$ Obtain $E^*(s)$

Solution $E^*(s) = \sum_{k=0}^{\infty} e^{-akT} \cdot e^{-kTs} = \sum_{n=0}^{\infty} e^{-(s+a)kT}$

$$= \frac{1}{1 - e^{-(s+a)T}} = \frac{e^{Ts}}{e^{Ts} - e^{-aT}}$$

7.2.3 Frequency Spectrum Analysis of Sampled Signal

Consider a continuous signal and its amplitude spectrum are:



The Fourier-series expansion of $\delta_T(t)$ is

$$\delta_T(t) = \frac{1}{T} \sum_{k=-\infty}^{\infty} e^{jk\omega_s t}$$

So the sampled signal is

$$e^*(t) = \frac{1}{T} \cdot \sum_{k=-\infty}^{\infty} e(t) \cdot e^{jk\omega_s t}$$

which Laplace transform is

$$E^*(s) = \frac{1}{T} \sum_{k=-\infty}^{\infty} E(s + jk\omega_s)$$

$$E^*(j\omega) = \frac{1}{T} \sum_{k=-\infty}^{+\infty} E[j(\omega + k\omega_s)]$$

where the operator s is replaced by $j\omega$

$$E^*(s) = \sum_{n=0}^{\infty} e(nT) \cdot e^{-nTs}$$

① shows the value relation of $E^*(s)$ and $e(t)$ on the sampling point;

② can be written into the closed form;

③ can be used to obtain the time response and the Z transform of $e^*(t)$

$$E^*(s) = \frac{1}{T} \sum_{n=-\infty}^{\infty} E(s + jn\omega_s)$$

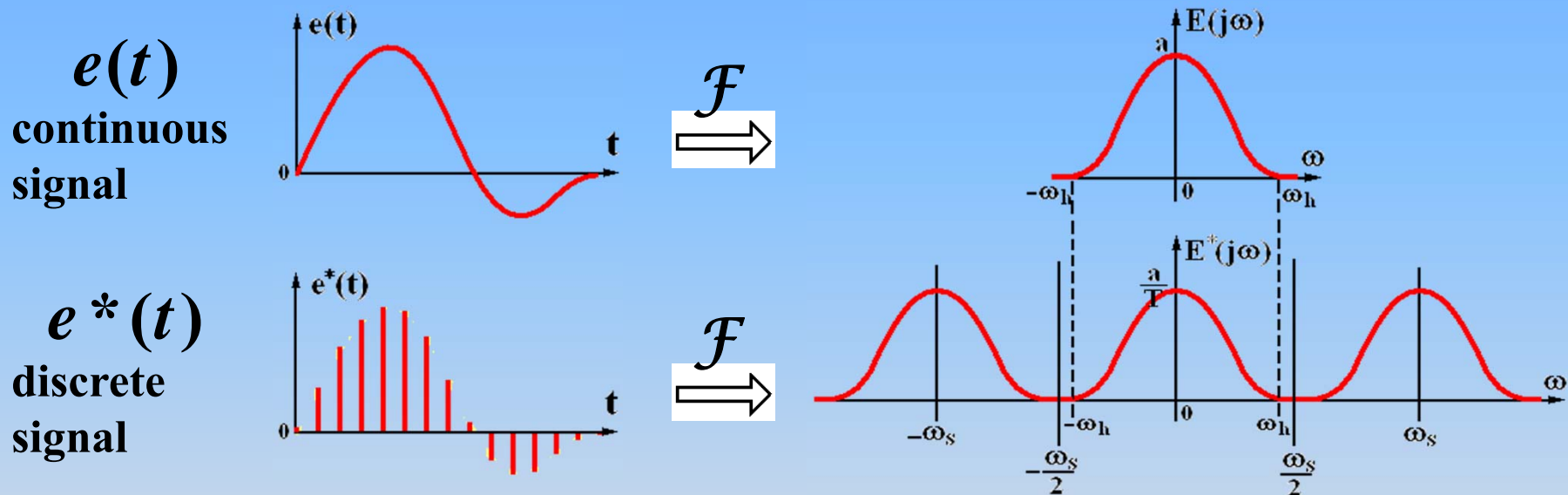
① shows relationship of $E^*(s)$ and $E(s)$;

② can not be written as close form;

③ can be used for the frequency spectrum analysis of $e^*(t)$.

The frequency spectrum analysis of continuous signal $e(t)$ and discrete signal $e^*(t)$

Frequency spectrum — Frequency expansion of the signal



$$E^*(s) = \sum_{n=0}^{\infty} e(nT) \cdot e^{-nTs}$$

$$E^*(s) = \frac{1}{T} \sum_{n=-\infty}^{\infty} E(s + jn\omega_s)$$

Shannon Sampling Theorem—

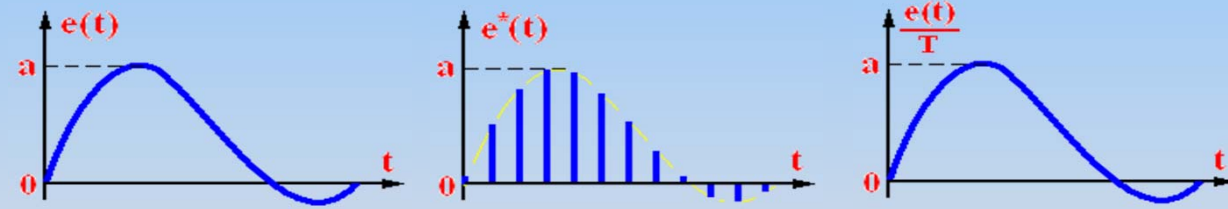
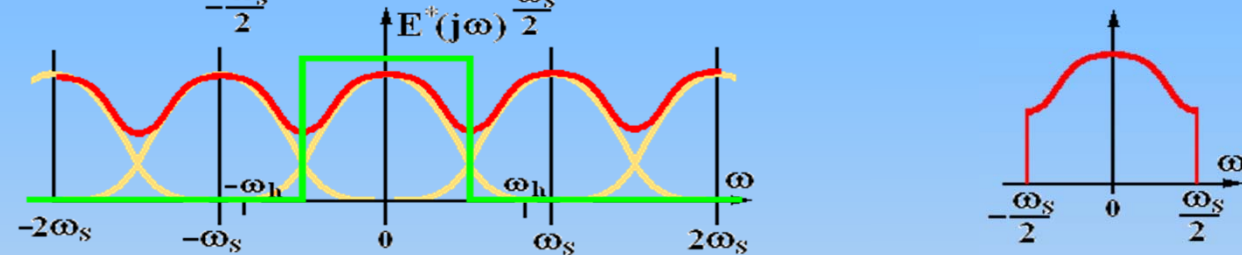
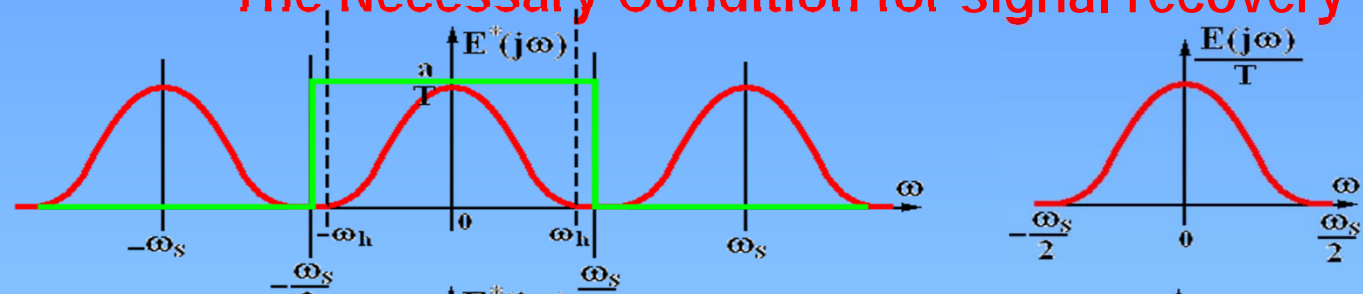
The Necessary Condition for signal recovery

$$\omega_s = \frac{2\pi}{T} > 2\omega_h$$

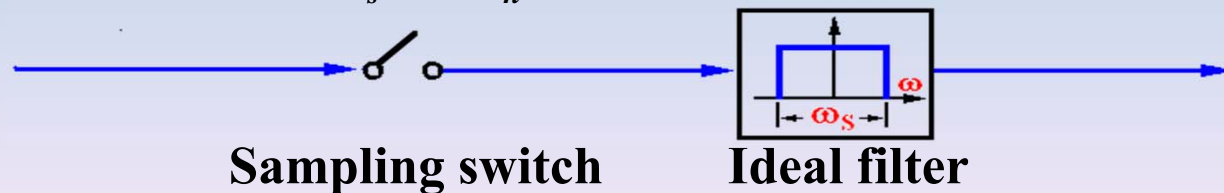
$$\omega_s = \frac{2\pi}{T} < 2\omega_h$$

$$\omega_s = \frac{2\pi}{T} > 2\omega_h$$

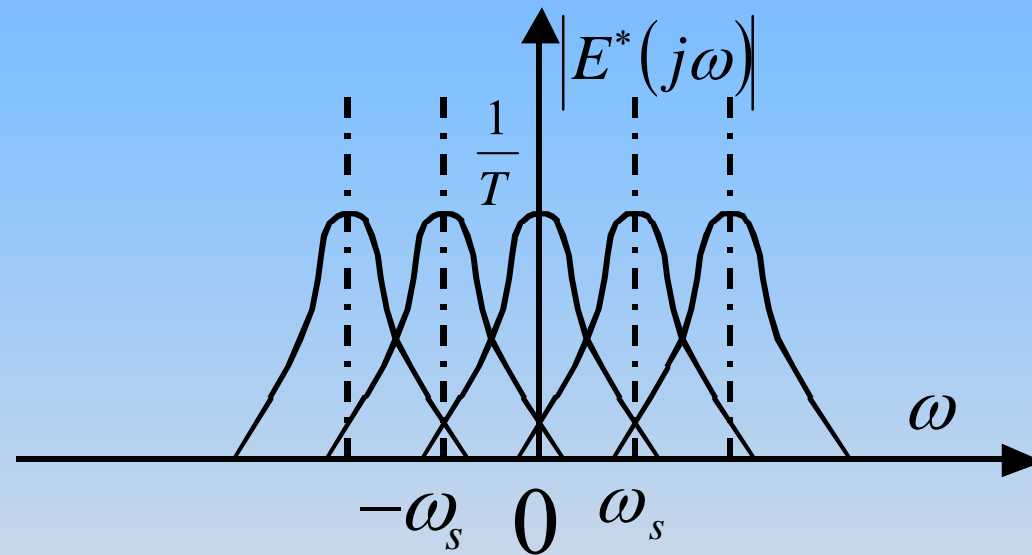
$$T < \frac{\pi}{\omega_h}$$



$$\omega_s > 2\omega_h$$



there are no overlap of each component, so the input signal can be recovered approximately. This is called sampling theorem or **Shannon's Theorem**



In the figure the input signal can't be recovered.