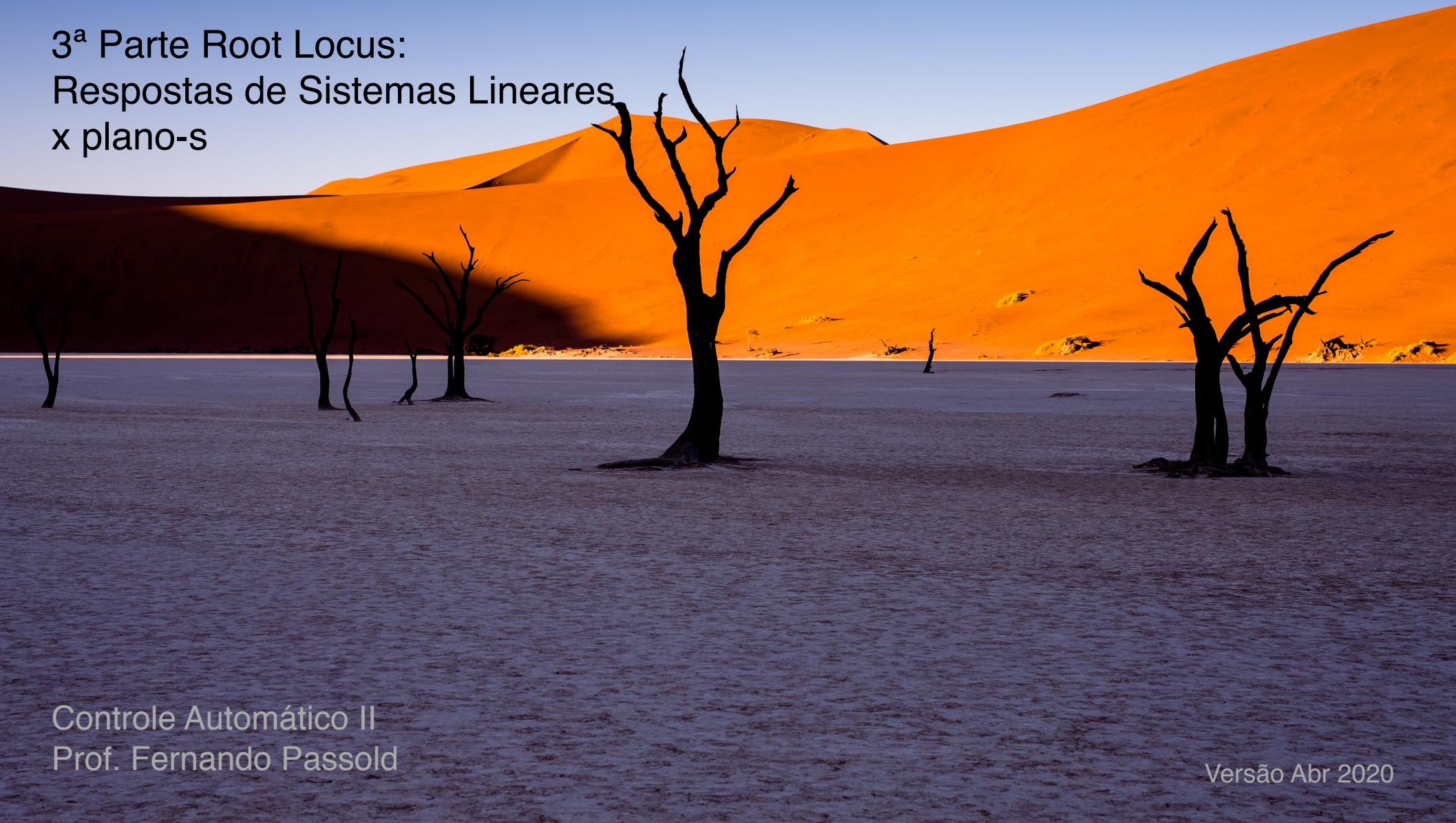


3^a Parte Root Locus: Respostas de Sistemas Lineares x plano-s

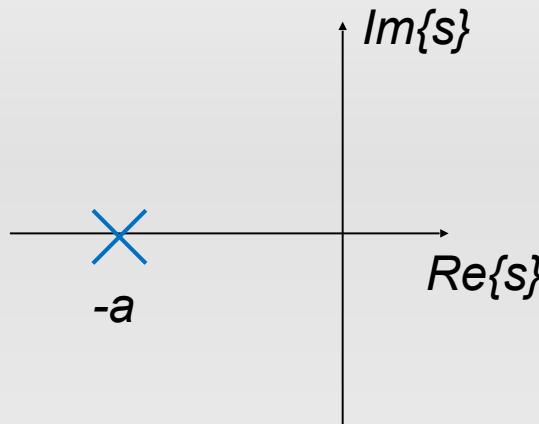
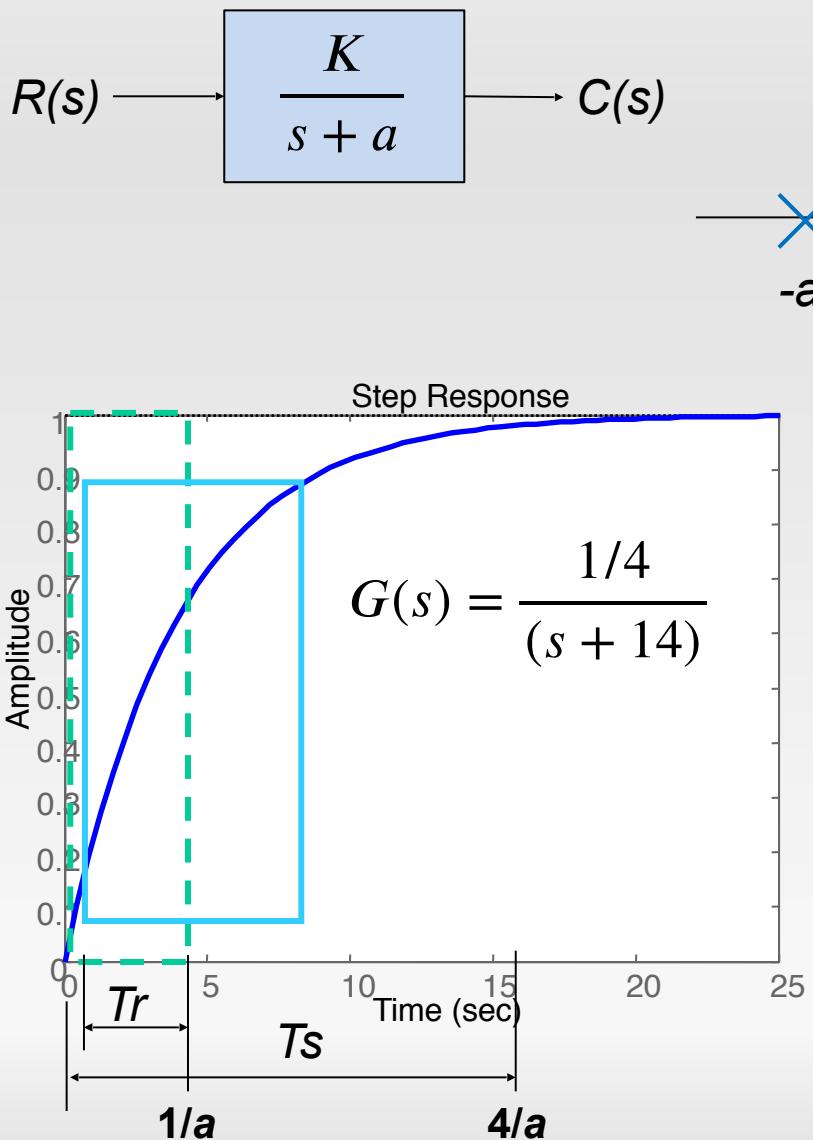


Procedimento Para Projetos

- 1) Desenhe o diagrama do lugar das raízes para o sistema;
- 2) Assuma que é um sistema de 2^a ordem com ou sem zeros e encontre o ganho que cumpre com as especificações de resposta transitória;
- 3) Justifique a aproximação para um sistema de 2^a ordem encontrando a posição de todos os polos mais rápidos e considerando o fato de que estão muito mais afastados do eixo jw que os pares dominantes de 2a-orden!
- 4) Também verifique que os zeros de laço fechado são aproximadamente cancelados pelos polos de maior ordem (os dominantes).
- 5) Se esta hipótese não pode ser justificada, sua solução terá que ser simulada com o fim de garantir de que se cumpre com a especificação para a resposta transitória.

Respostas transitórias

Sistema de 1^a-ordem



Resposta em malha fechada:

$$C(s) = R(s)G(s)$$

$$C(s) = \frac{1}{s} \cdot \frac{K}{(s+a)}$$

$$C(s) = \frac{K/a}{s} - \frac{K/a}{(s+a)}$$

$$c(t) = \frac{K}{a} - \frac{K}{a} e^{-at}$$

Constante de tempo \rightarrow

$$\tau = \frac{1}{a}$$

$$T_r = \frac{2.31}{a} - \frac{0.11}{a} = \frac{2.2}{a}$$

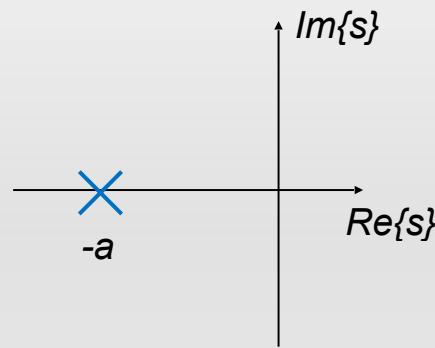
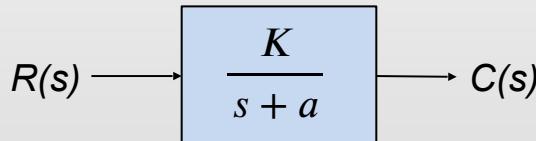
$$T_s = \frac{4}{a}$$

\leftarrow Tempo de subida:
 $K/a^*[0.1 \sim 0.9]$

\leftarrow Tempo de assentamento:
 $K/a^*0.98$

Respostas transitórias

Sistema de 1^a-ordem

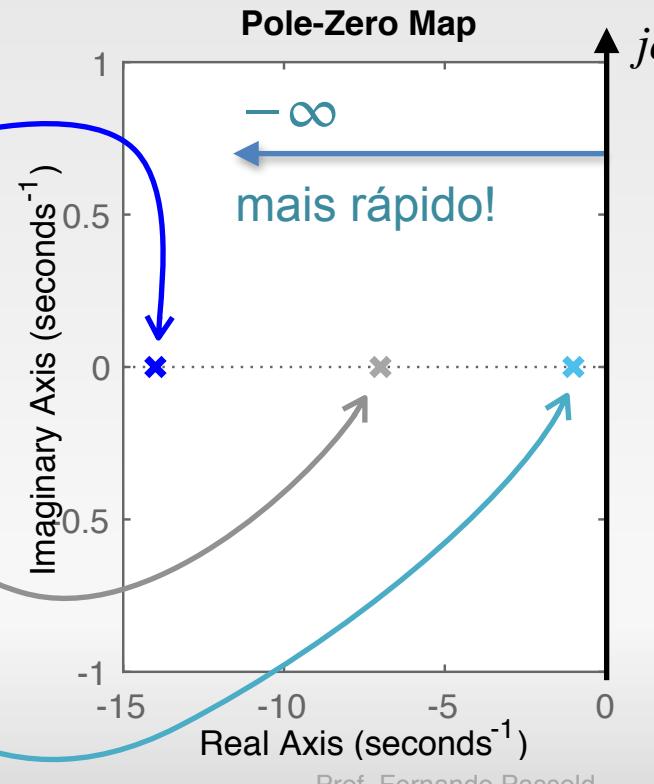


Comparando 3 sistemas de pólos simples reais:

$$G_1(s) = \frac{14}{s + 14}$$

$$G_2(s) = \frac{7}{s + 7}$$

$$G_3(s) = \frac{1}{s + 1}$$

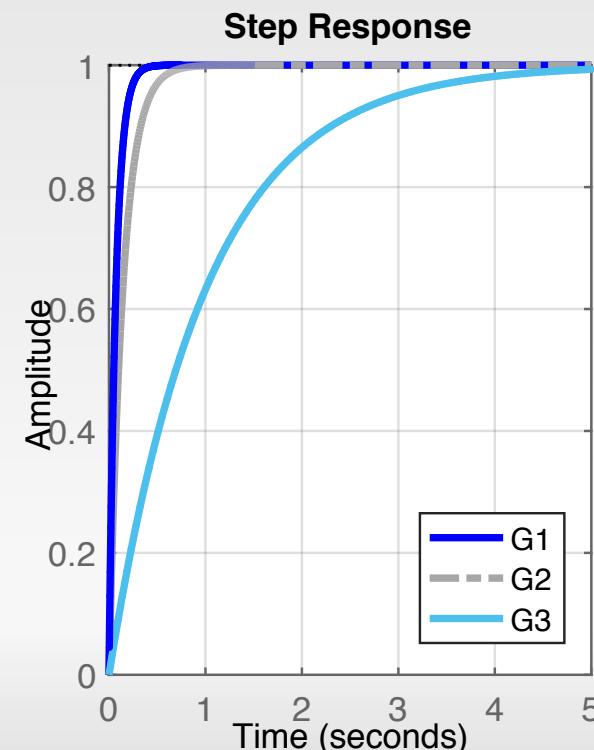


Resposta em malha fechada:

$$c(t) = \frac{K}{a} - \frac{K}{a} e^{-at}$$

$$\text{Constante de tempo} \rightarrow \tau = \frac{1}{a}$$

$$\text{Tempo de assentamento} \rightarrow t_s = \frac{4}{a} = 4\tau \\ (= K/a * 0,98)$$



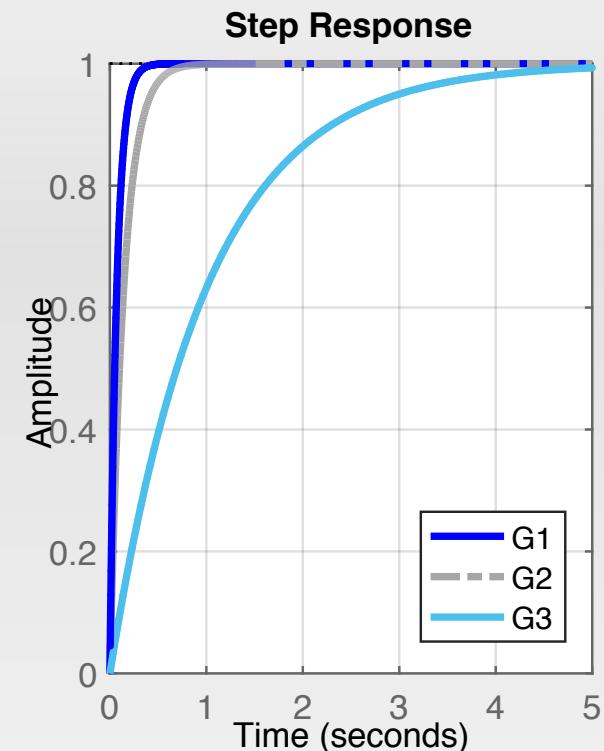
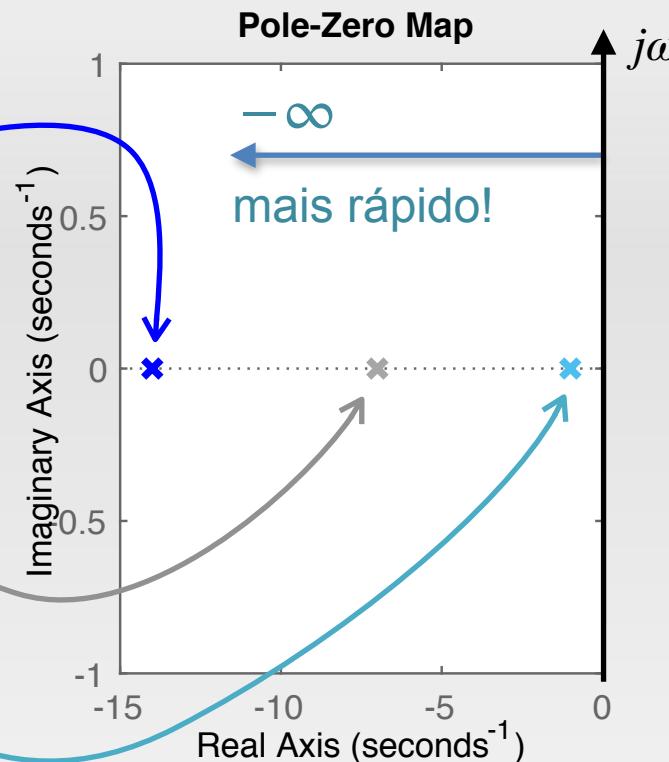
Respostas transitórias

Sistema de 1^a-ordem

$$G_1(s) = \frac{14}{s + 14}$$

$$G_2(s) = \frac{7}{s + 7}$$

$$G_3(s) = \frac{1}{s + 1}$$



Matlab:

```
>> G1=tf(14, [1 14])
```

G1 =

$$\frac{14}{s + 14}$$

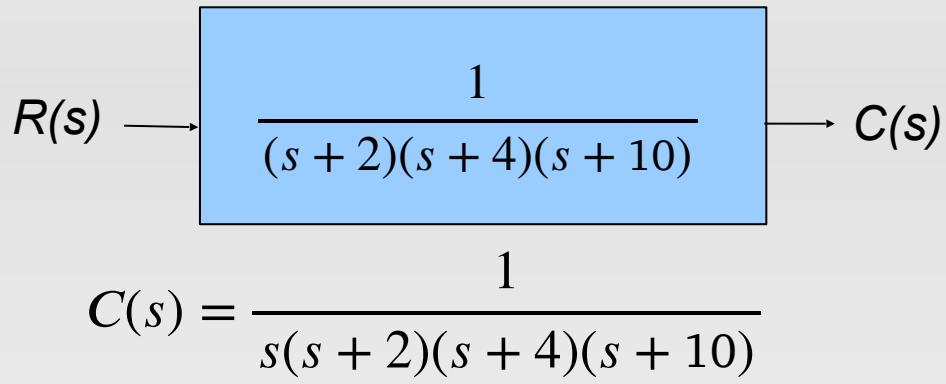
Continuous-time transfer function.

```
>> G2=tf(7, [1 7]);
>> G3=tf(1, [1 1]);
>> figure; subplot(121); pzmap(G1,G2,G3);
>> subplot(122); step(G1,G2,G3);
>> axis([0 5 0 1])
```

Respostas transitorias

Sistema com pólos reais

$$R(s) = \frac{1}{s} \leftarrow \text{Degrau (unitário)}$$



```

>> num=1;
>> den=poly([0 -2 -4
-10]);
>>
[r,p,k]=residue(num,den)
r =
-0.0021
 0.0208
-0.0313
 0.0125
p =
-10.0000
-4.0000
-2.0000
 0
k =
 []
>>

```

Resposta em malha aberta:

$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$K_1 = \frac{1 \cdot s}{s(s+2)(s+4)(s+10)}|_{s \rightarrow 0} = \frac{1}{80} = 0,0125$$

$$K_2 = \frac{1 \cdot (s+2)}{s(s+2)(s+4)(s+10)}|_{s \rightarrow -2}$$

$$K_2 = \frac{1}{s(s+4)(s+10)}|_{s \rightarrow -2} = -\frac{1}{32} = -0,0312$$

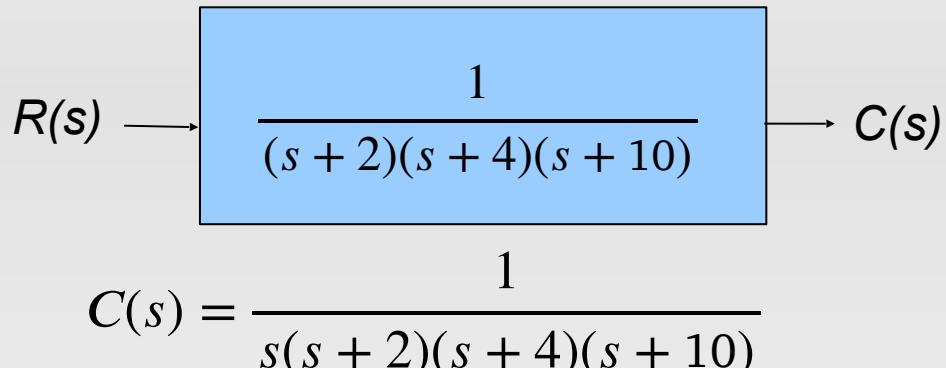
$$K_3 = \frac{1}{s(s+2)(s+10)}|_{s \rightarrow -4} = \frac{1}{48} = 0,0208$$

$$K_4 = \frac{1}{s(s+2)(s+4)}|_{s \rightarrow -10} = -\frac{1}{480} = -0,0021$$

Respostas transitorias

Sistema com pólos reais

$$R(s) = \frac{1}{s} \leftarrow \text{Degrau (unitário)}$$



```

>> num=1;
>> den=poly([0 -2 -4
-10]);
>>
[r,p,k]=residue(num,den)
r =
-0.0021
0.0208
-0.0313
0.0125
p =
-10.0000
-4.0000
-2.0000
0
k =
[]
>>

```

Resposta em malha aberta:

$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$K_1 = \frac{1 \cdot s}{s(s+2)(s+4)(s+10)}|_{s \rightarrow 0} = \frac{1}{80} = 0,0125$$

[r,p,k] = residue(b,a) finds the residues, poles, and direct term of a Partial Fraction Expansion of the ratio of two polynomials, where the expansion is of the form:

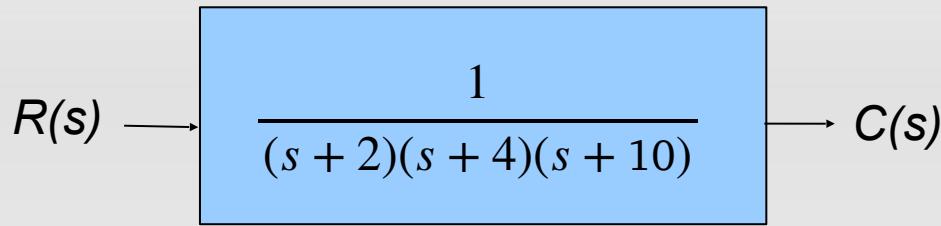
$$\frac{b(s)}{a(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}{a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0} = \frac{r_n}{s - p_n} + \dots + \frac{r_2}{s - p_2} + \frac{r_1}{s - p_1} + k(s).$$

$$K_3 = \frac{1}{s(s+2)(s+10)}|_{s \rightarrow -4} = \frac{1}{48} = 0,0208$$

$$K_4 = \frac{1}{s(s+2)(s+4)}|_{s \rightarrow -10} = -\frac{1}{480} = -0,0021$$

Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

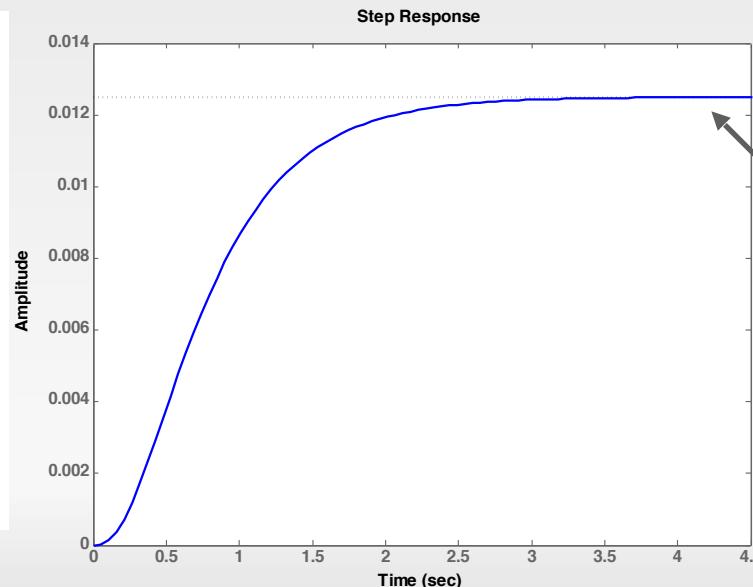
$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

Matlab:

```
>> num=1;
>> den=poly([-2 -4 -10]);
>> c=tf(num,den);
>> zpk(c)
Zero/pole/gain:
1
-----
(s+10) (s+4) (s+2)

>> dcgain(c)
ans =      0.0125
>>
```



Valor final:

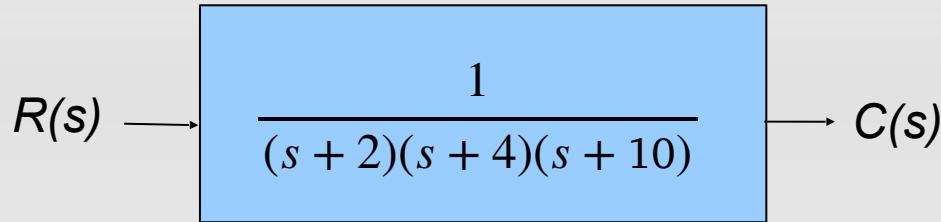
$$c(\infty) = \lim_{s \rightarrow 0} s \cdot C(s)$$

$$c(\infty) = \frac{1 \cdot s}{s(s+2)(s+4)(s+10)}$$

$$c(\infty) = \frac{1}{80} = 0,0125$$

Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

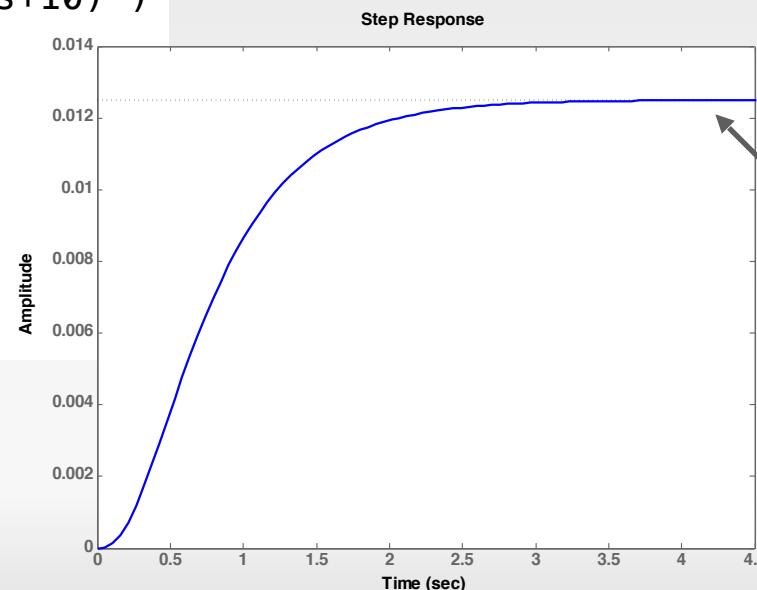
$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

Matlab: outra forma para $c(\infty)$:

```
>> den=sym('(s+2)*(s+4)*(s+10)')
den =
(s+2)*(s+4)*(s+10)

>> subs(den,0)
ans =
80

>>
```



Valor final:

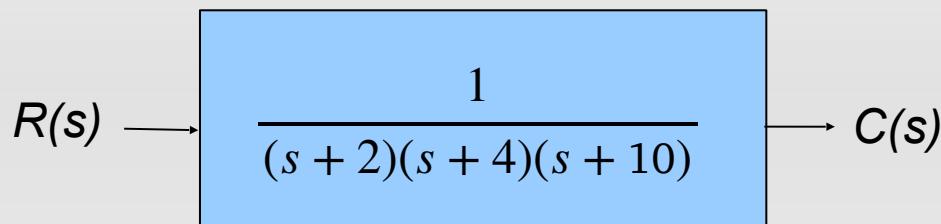
$$c(\infty) = \lim_{s \rightarrow 0} s \cdot C(s)$$

$$c(\infty) = \frac{1 \cdot s}{s(s+2)(s+4)(s+10)}$$

$$c(\infty) = \frac{1}{80} = 0,0125$$

Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

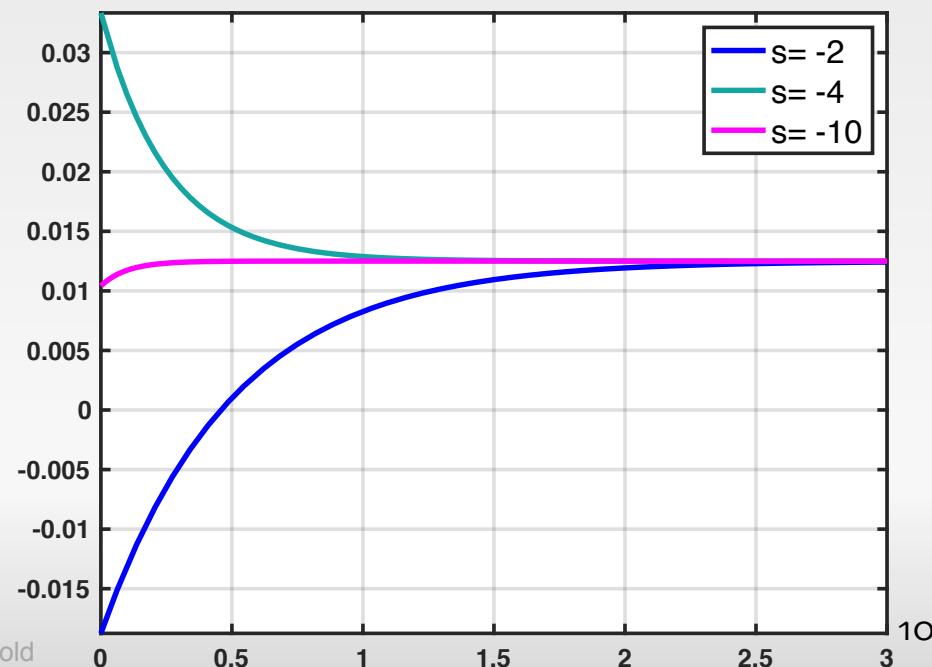
$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

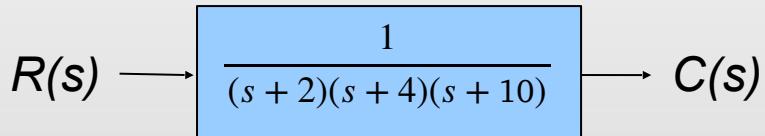
Matlab: verificando impacto de cada polo:

```
>> fplot(@(t)[(1/80)-(1/32)*exp(-2*t)], [0 3])
>> hold on
>> fplot(@(t)[(1/80)+(1/48)*exp(-4*t)], [0 3])
>> hold on
>> fplot(@(t)[(1/80)-(1/480)*exp(-10*t)], [0 3])
>> legend('s= -2', 's= -4', 's= -10')
```

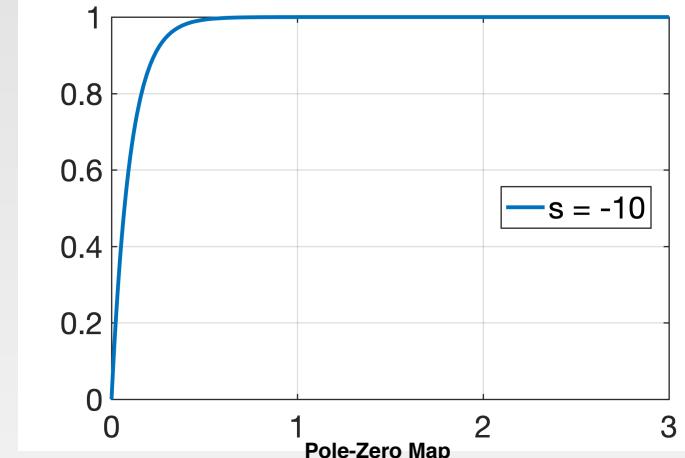
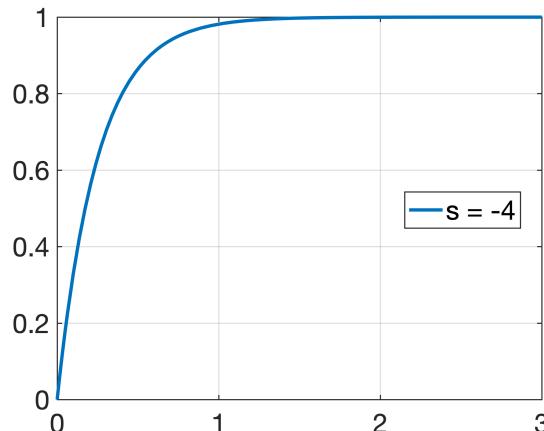
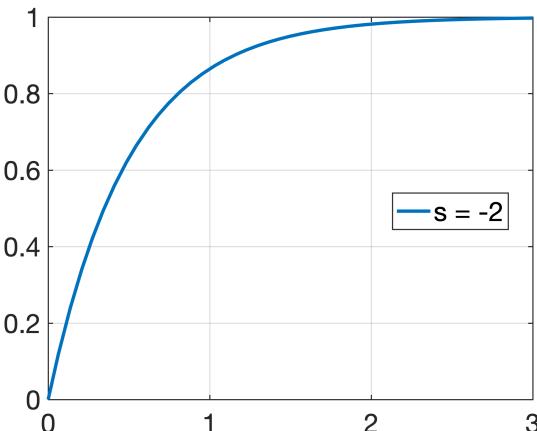


Respostas transitorias

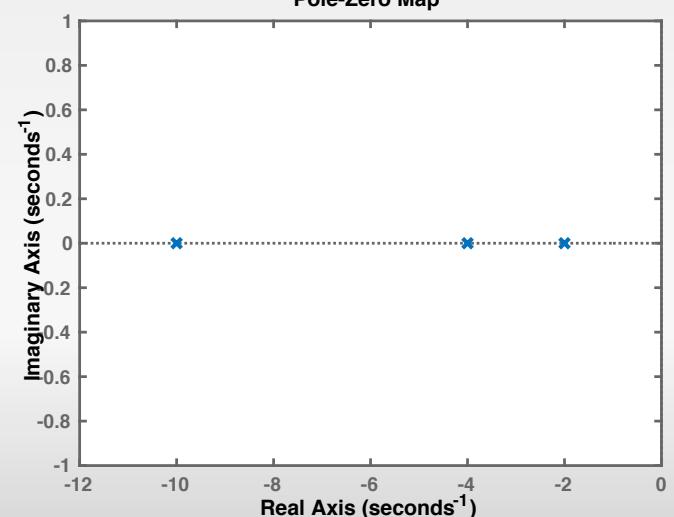
Sistema com pólos reais



Avaliando as respostas de cada pólo individualmente:

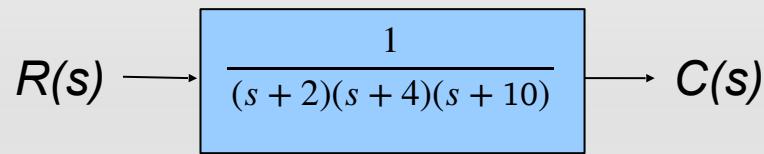


```
>> figure; fplot(@(t)[(1-exp(-2*t))], [0 3])
>> legend('s = -2')
>> figure; fplot(@(t)[(1-exp(-4*t))], [0 3])
>> legend('s = -4')
>> figure; fplot(@(t)[(1-exp(-10*t))], [0 3])
>> legend('s = -10')
>> figure; pzmap(c)
```



Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

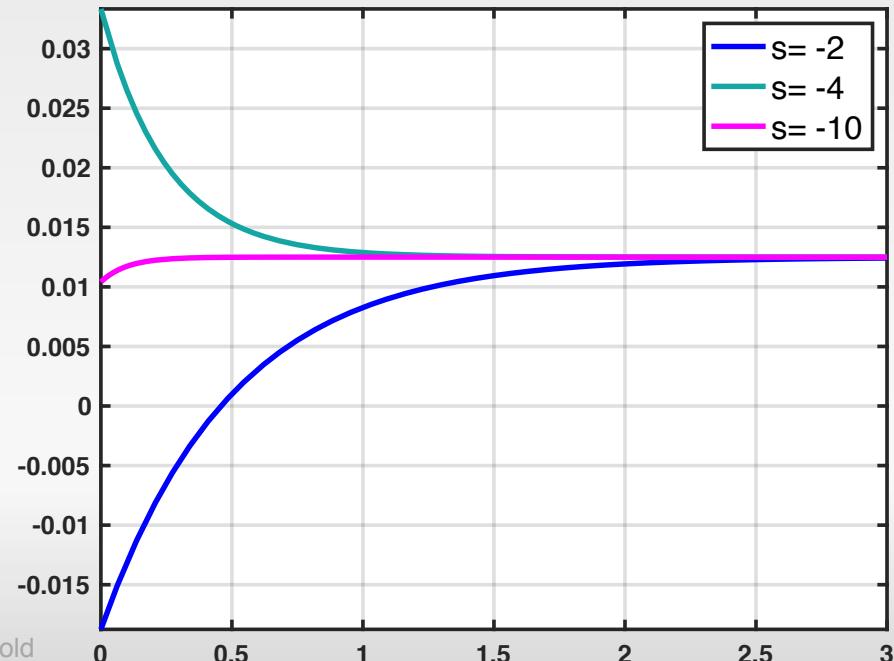
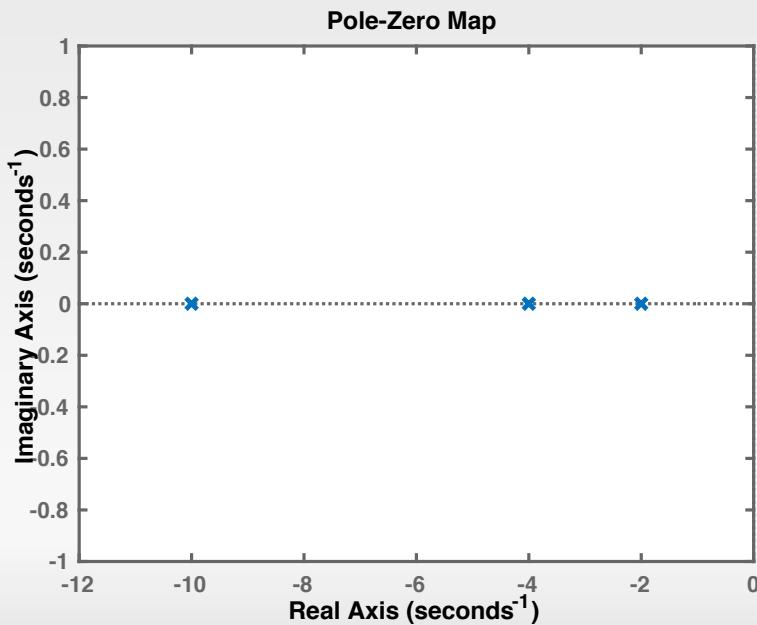
$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

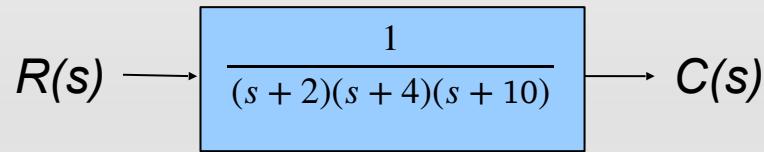
Matlab:

```
>> figure(3); pzmap(c)
```



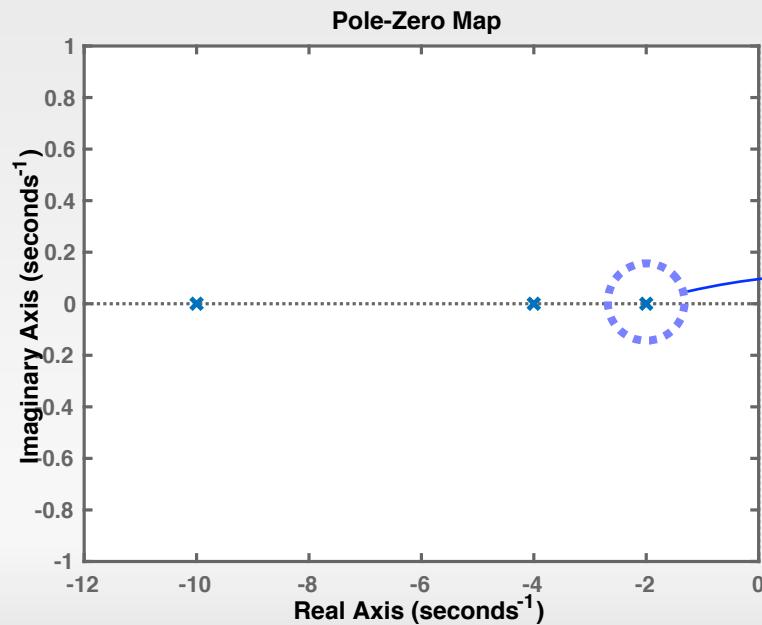
Respostas transitorias

Sistema com pólos reais



Matlab:

```
>> figure(3); pzmap(c)
```

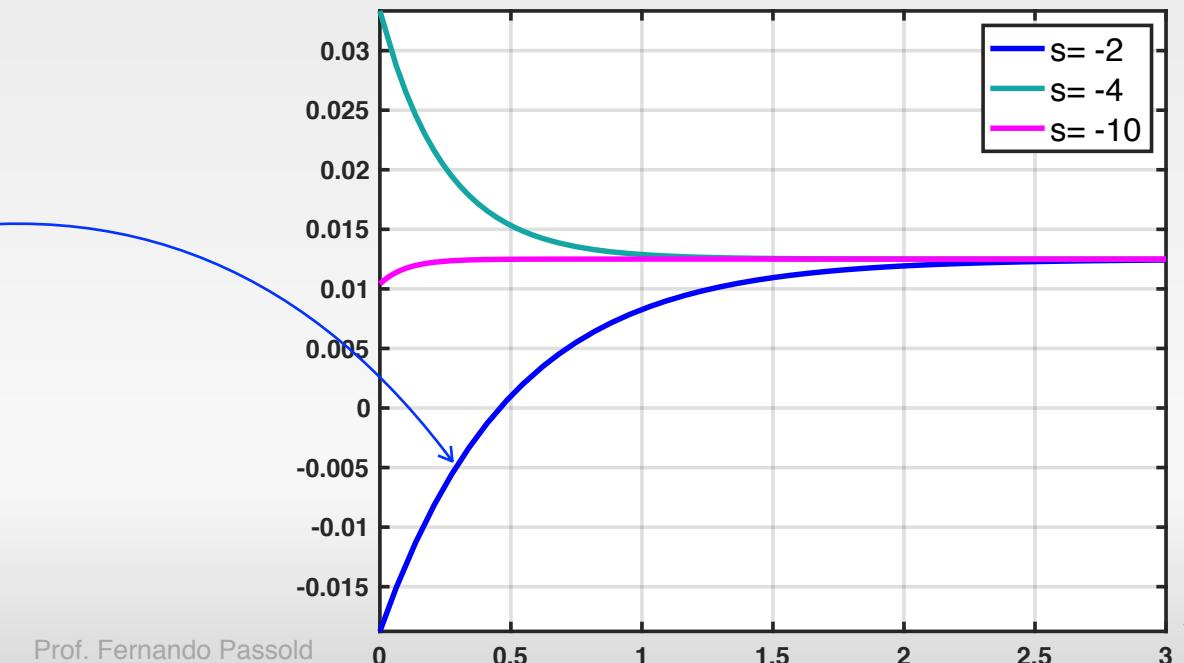


Resposta em malha aberta:

$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

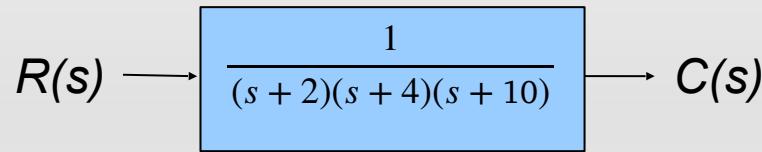
$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$



Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

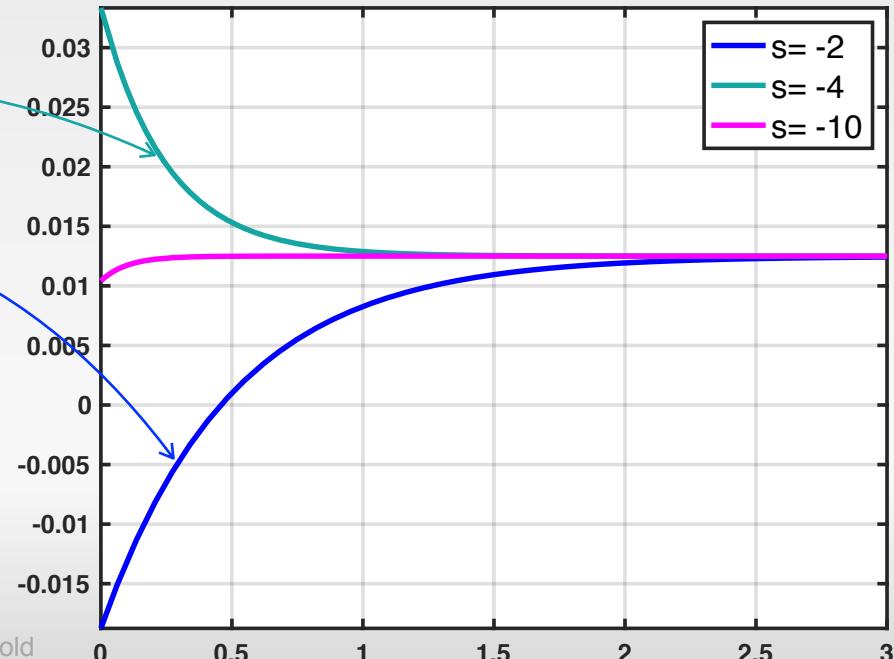
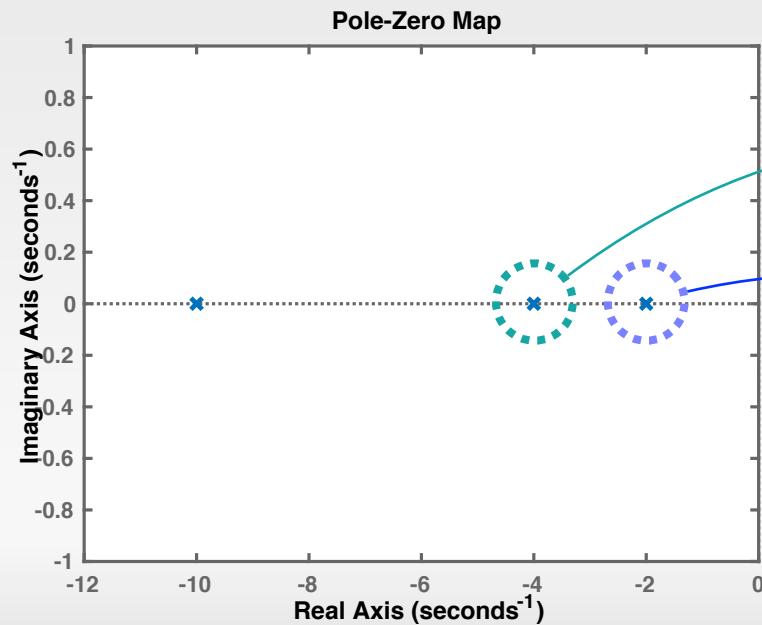
$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

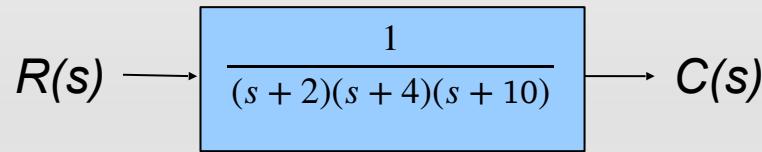
Matlab:

```
>> figure(3); pzmap(c)
```



Respostas transitorias

Sistema com pólos reais



Resposta em malha aberta:

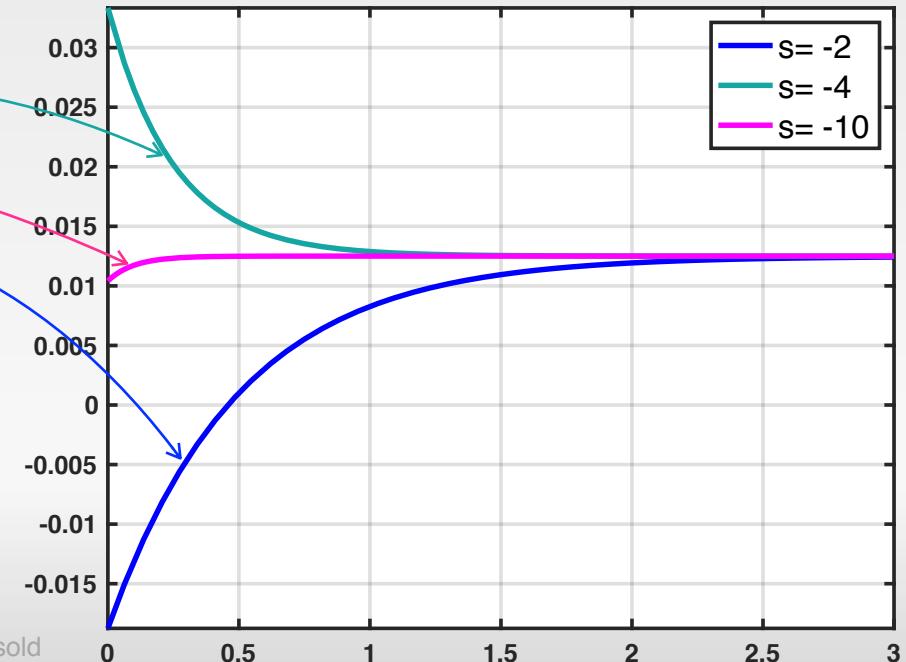
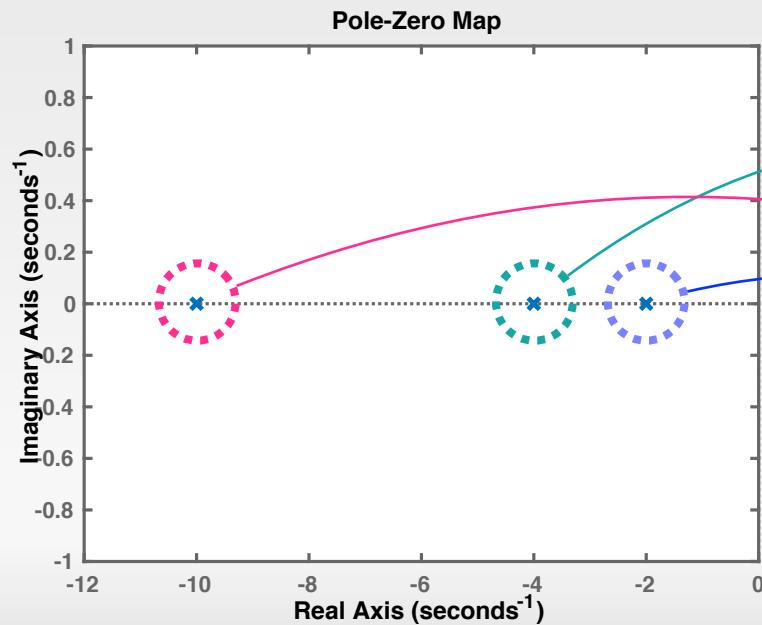
$$C(s) = \frac{K_1}{s} + \frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+10)}$$

$$c(t) = K_1 + K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-10t}$$

$$c(t) = \frac{1}{80} - \frac{1}{32}e^{-2t} + \frac{1}{48}e^{-4t} - \frac{1}{480}e^{-10t}$$

Matlab:

```
>> figure(3); pzmap(c)
```



Respostas de Sistemas de 2^a ordem



Desenho do RL e encontrando pontos chaves

Exemplo: $G(s) = \frac{K(s^2 - 4s + 20)}{(s + 2)(s + 4)}$

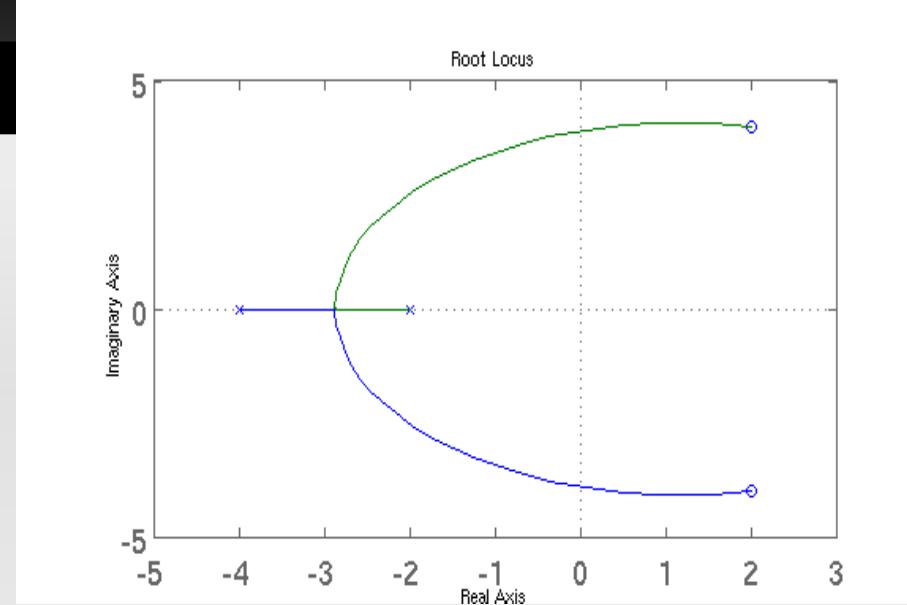
$$\zeta = 0.45 \quad \theta = 180^\circ - \arccos(0.45) = 116.7^\circ$$

```
>> num=[1 -4 20];
>> den=conv([1 2],[1 4]);
>> g=tf(num,den)
Transfer function:
s^2 - 4 s + 20
-----
s^2 + 6 s + 8

>> zpk(g)
Zero/pole/gain:
(s^2 - 4s + 20)
-----
(s+4) (s+2)

>> roots(num)
ans =
2.0000 + 4.0000i
2.0000 - 4.0000i

>> rlocus(g)
>>
```

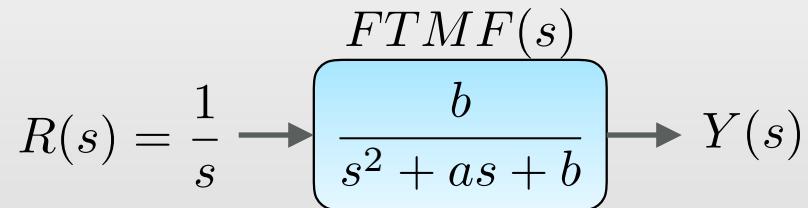


Questões:

- Cruzamento com eixo $j\omega$
- Ponto de quebra
- Faixa de K (ganho):

Respostas transitorias

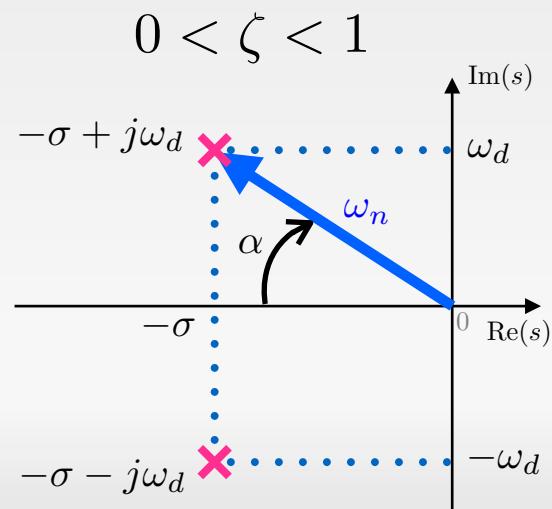
Sistemas de 2a-orden:



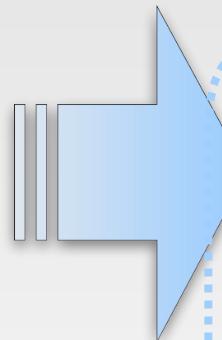
$$FTMF(s) = \frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{b}{s^2 + as + b}$$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = K \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right)$$

pólos em: $s = \sigma \pm j\omega_d$ ou: $s = -\zeta\omega_n \pm j\omega_n\sqrt{1 - \zeta^2}$



$$\begin{aligned}\sigma &= \omega_n \cos(\alpha) = \omega_n \zeta; \\ \omega_d &= \omega_n \sin(\alpha) = \omega_n \sqrt{1 - \zeta^2}; \\ \zeta &= \cos(\alpha); \\ \sin(\alpha) &= \sqrt{1 - \zeta^2};\end{aligned}$$

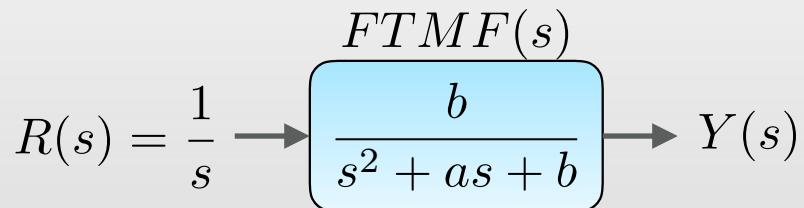


3 casos:

- 1) Raízes reais:
 - a) Distintas;
 - b) Iguais;
- 2) Raízes complexas:
 - a) Distante de $j\omega$;
 - b) Sobre $j\omega$.

Respuestas transitorias

Sistemas de 2a-orden:



3 casos:

- 1) Raízes reais:
 - a) Distintas;
 - b) Iguais
- 2) Raízes complexas:
 - a) Distante de $j\omega$;
 - b) Sobre $j\omega$.

$$G_{1a}(s) = \frac{9}{s^2 + 9s + 9}$$

2 pólos reais em $-p_1$ e $-p_2$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = k \frac{w_n^2}{s^2 + 2\zeta w_n s + w_n^2}$$

$$w_n^2 = 9 \rightarrow w_n = 3$$

$$\zeta = \frac{9}{6} = 1,5$$

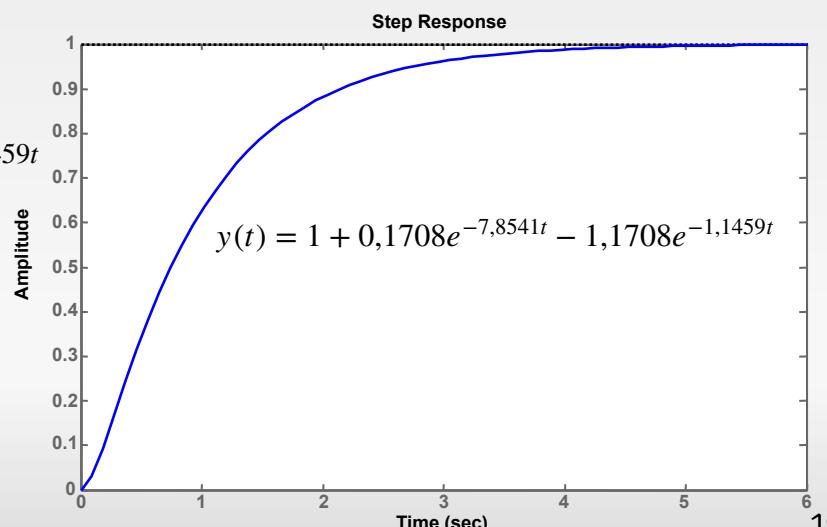
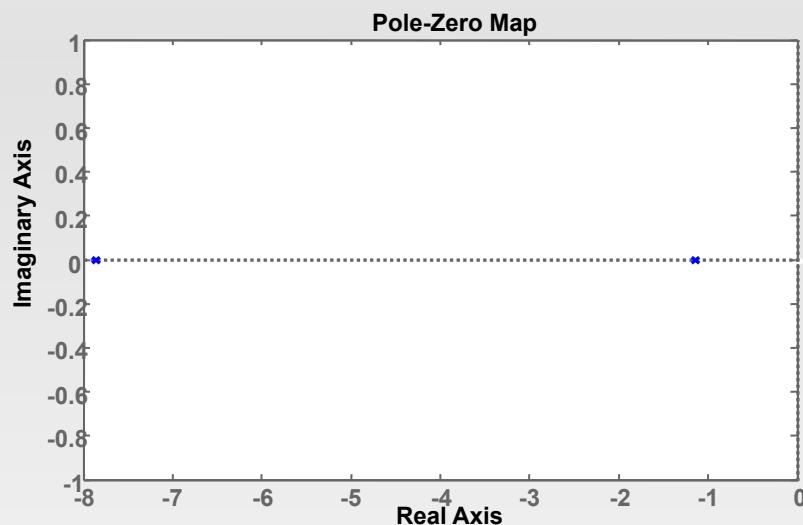
$$y(t) = K_1 e^{-p_1 t} + K_2 e^{-p_2 t}$$

$$y(t) = 1 + 0,1708e^{-7,8541t} - 1,1708e^{-1,1459t}$$

Sistema superamortecido \Rightarrow
($\zeta > 1$)

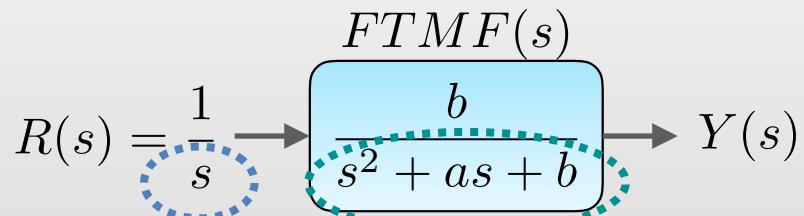
Overdamped response \Rightarrow

Sistema superamortecido \Rightarrow ($\zeta > 1$)



Respuestas transitorias

Sistemas de 2a-orden:



3 casos:
 1) Raízes reais:
 a) Distintas;
 b) Igualas

Matlab:

```
>> num=9;
>> den=conv([1 0],[1 9
9]);
>>
[r,p,k]=residue(num,den)
r =
    0.1708
   -1.1708
    1.0000
p =
   -7.8541
   -1.1459
    0
k =
    []
>> ftmf=tf(9,[1 9 9]);
>> pzmap(ftmf)
>> step(ftmf)
```

$$G_{1a}(s) = \frac{9}{s^2 + 9s + 9}$$

2 pólos reais em $-p_1$ e $-p_2$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = k \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n^2 = 9 \rightarrow \omega_n = 3$$

$$\zeta = \frac{9}{6} = 1,5$$

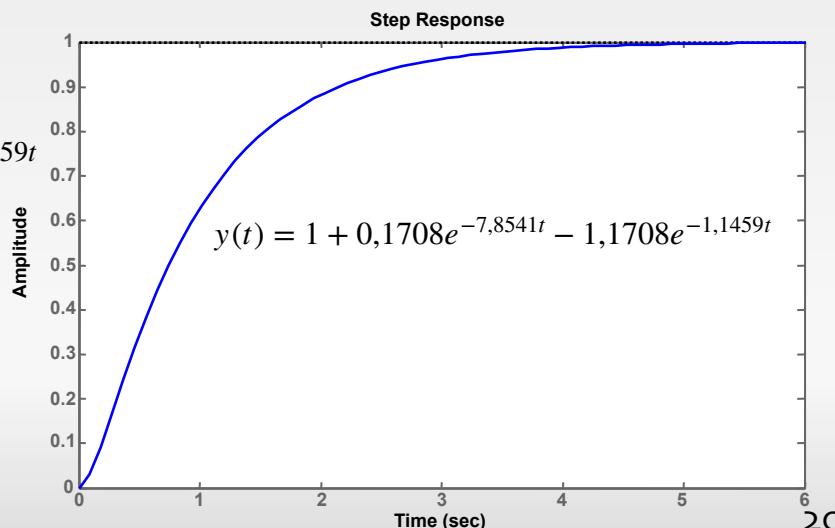
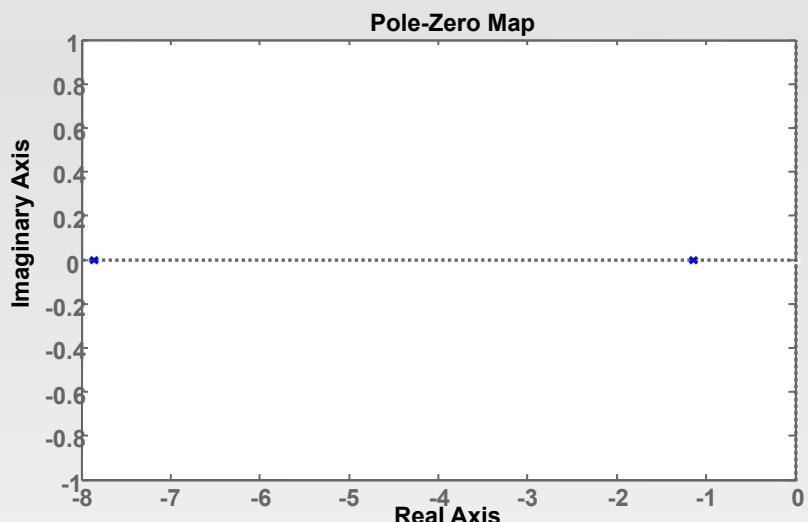
$$y(t) = K_1 e^{-p_1 t} + K_2 e^{-p_2 t}$$

$$y(t) = 1 + 0,1708e^{-7,8541t} - 1,1708e^{-1,1459t}$$

Sistema superamortecido $\rightarrow (\zeta > 1)$

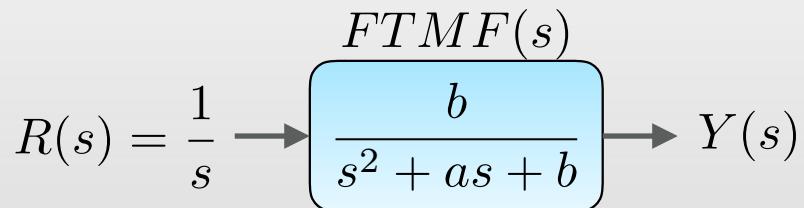
Overdamped response \rightarrow

Sistema superamortecido $\rightarrow (\zeta > 1)$



Respuestas transitorias

Sistemas de 2a-orden:



3 casos:

1) Raízes reais:

- a) Distintas;
- b) Iguais

2) Raízes complexas:

- a) Distante de $j\omega$;
- b) Sobre $j\omega$.

$$G(s) = \frac{9}{s^2 + 6s + 9}$$

2 pólos reais em $-p_1$ e $-p_2$

$$p_{1,2} = -3$$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = k \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n^2 = 9 \rightarrow \omega_n = 3$$

$$\zeta = \frac{6}{6} = 1$$

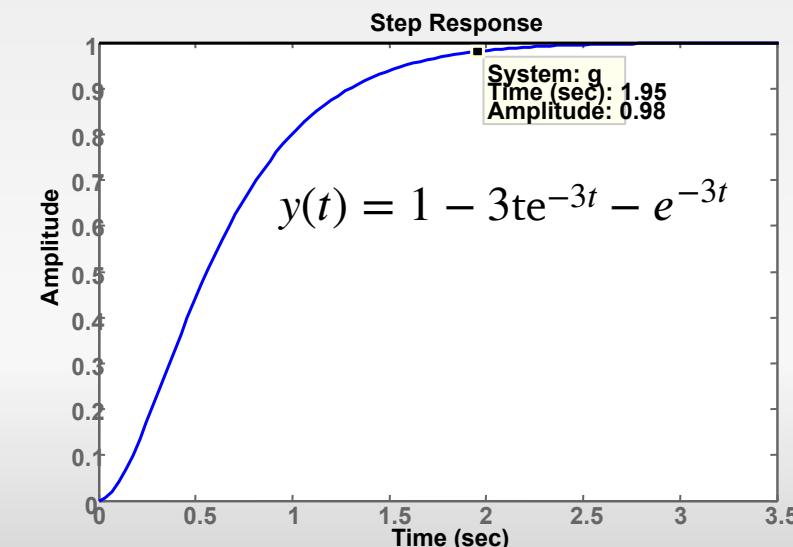
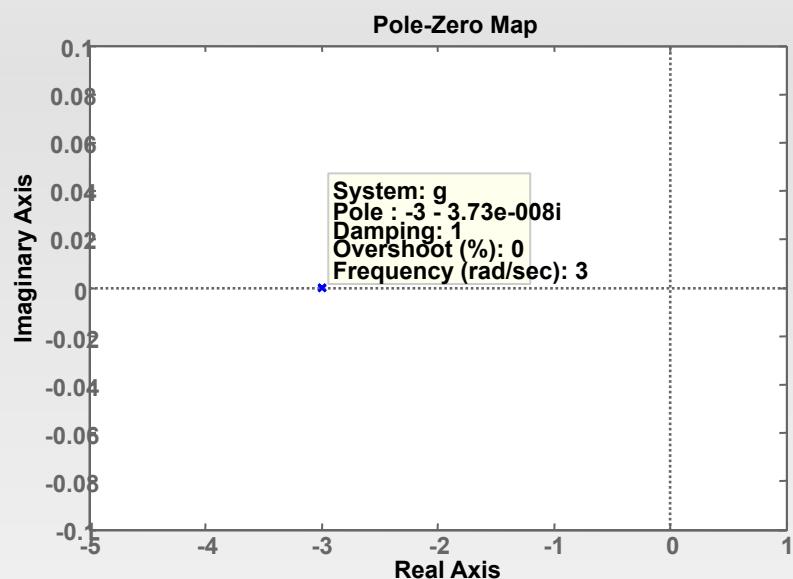
$$y(t) = K_0 + K_1 e^{-at} + K_2 t e^{-at}$$

$$y(t) = 1 - 3t e^{-3t} - e^{-3t}$$

Amortecimiento crítico \Rightarrow
 $(\zeta = 1)$

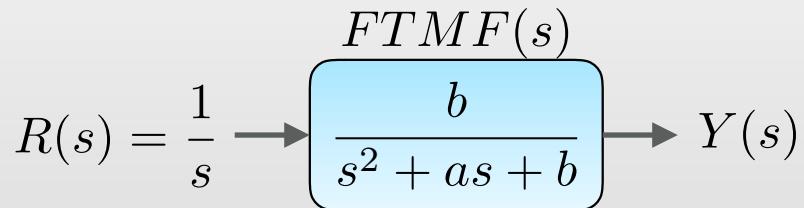
Critically damped \Rightarrow

Amortecimento crítico $\Rightarrow (\zeta = 1)$



Respuestas transitorias

Sistemas de 2a-orden:



3 casos:

- 1) Raízes reais:
 - a) Distintas;
 - b) Igualis

Matlab:

```
>> num=9;
>> den=conv([1 0],[1 6
9]);
>>
[r,p,k]=residue(num,den)
r =
-1
-3
1
p =
-3
-3
0
k =
[]
>> ftmf=tf(9,[1 6 9]);
>> pzmap(ftmf)
>> step(ftmf)
```

$$G(s) = \frac{9}{s^2 + 6s + 9}$$

2 pólos reais em $-p_1$ e $-p_2$

$$p_{1,2} = -3$$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s+p_1)(s+p_2)} = k \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n^2 = 9 \rightarrow \omega_n = 3$$

$$\zeta = \frac{6}{6} = 1$$

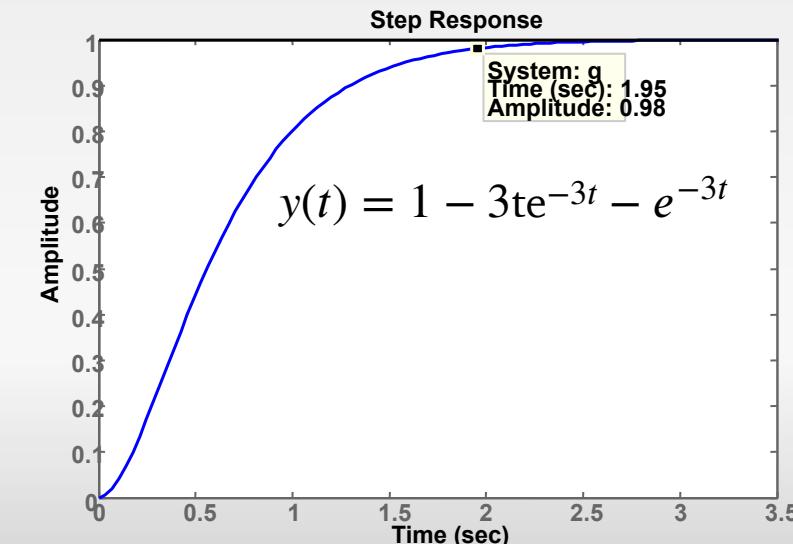
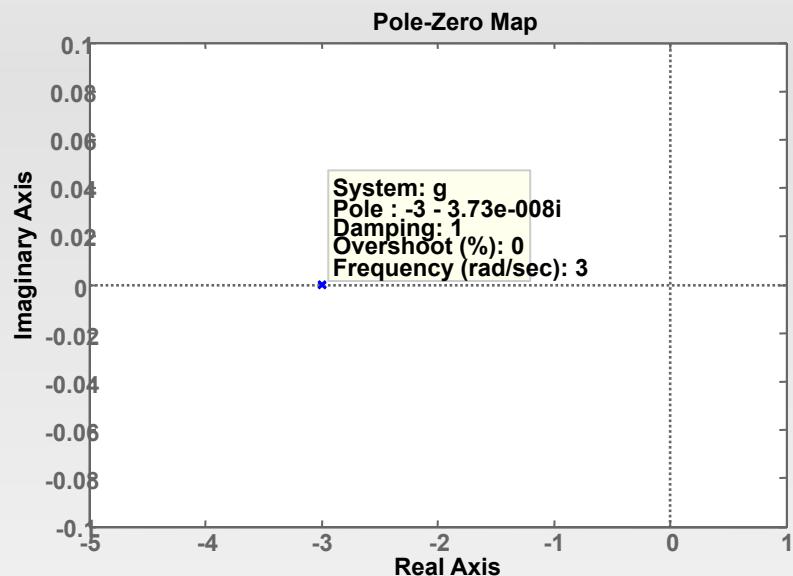
$$y(t) = K_0 + K_1 e^{-at} + K_2 t e^{-at}$$

$$y(t) = 1 - 3te^{-3t} - e^{-3t}$$

Amortecimento crítico \rightarrow
 $(\zeta = 1)$

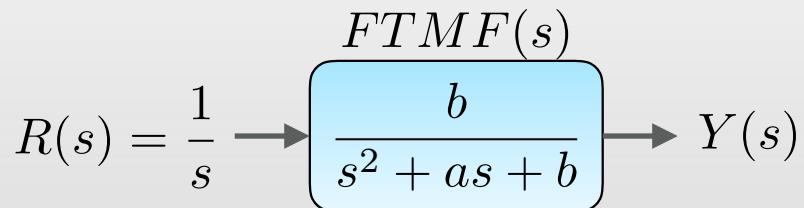
Critically damped \rightarrow

Amortecimento crítico $\rightarrow (\zeta = 1)$



Respuestas transitorias

Sistemas de 2a-orden:



3 casos:

- 1) Raízes reais:
 - a) Distintas;
 - b) Iguais
- 2) Raízes complexas:
 - a) Distante de $j\omega$;
 - b) Sobre $j\omega$.

$$G(s) = \frac{9}{s^2 + 2s + 9}$$

2 pólos complexos em $-p_1$ e $-p_2$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = k \frac{w_n^2}{s^2 + 2\zeta w_n s + w_n^2}$$

$$p_{1,2} = \sigma \pm j w_d \quad w_d = w_n \sin(\alpha) = w_n \sqrt{1 - \zeta^2}$$

$$p_{1,2} = -1 \pm j \sqrt{8}$$

$$w_n^2 = 9 \rightarrow w_n = 3$$

$$\zeta = \frac{2}{6} = 0,3333$$

$$y(t) = 1 - e^{-t} [\cos(\sqrt{8}t) + \frac{\sqrt{8}}{8} \sin(\sqrt{8}t)]$$

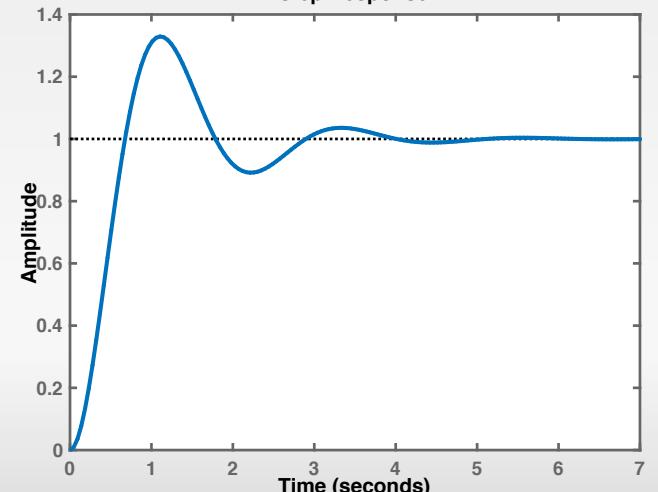
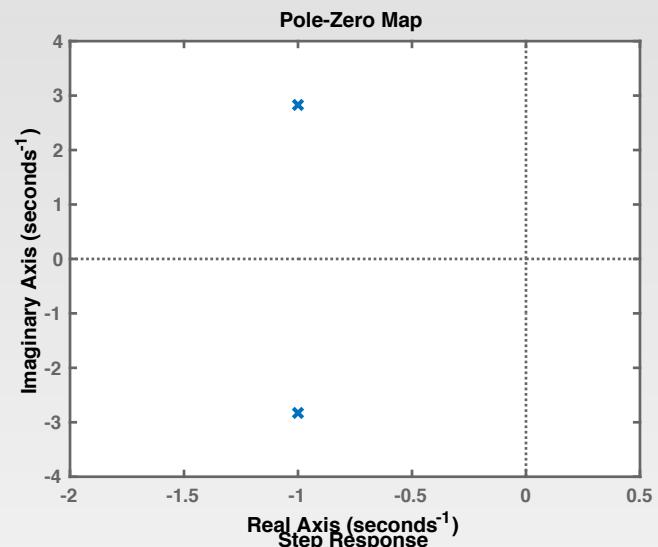
$$y(t) = 1 - 1,06 e^{-t} \cos(\sqrt{8}t - 19,47^\circ)$$

$$y(t) = 1 - A e^{-\sigma \% t} \cos(w_d t - \phi)$$

Sistema subamortecido \Rightarrow
 $(0 < \zeta < 1)$

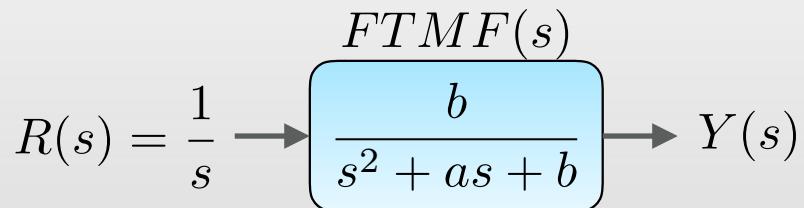
underdamped response \Rightarrow

Sistema subamortecido $\Rightarrow (0 < \zeta < 1)$



Respuestas transitorias

Sistemas de 2a-orden:



3 casos:

- 1) Raízes reais:
 - a) Distintas;
 - b) Iguais
- 2) Raízes complexas:
 - a) Distante de $j\omega$;
 - b) Sobre $j\omega$.

$$G(s) = \frac{9}{s^2 + 9}$$

2 pólos complexos em $\pm j\omega_n$:

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = k \frac{w_n^2}{s^2 + 2\zeta w_n s + w_n^2}$$
$$p_{1,2} = \pm j\sqrt{3}$$

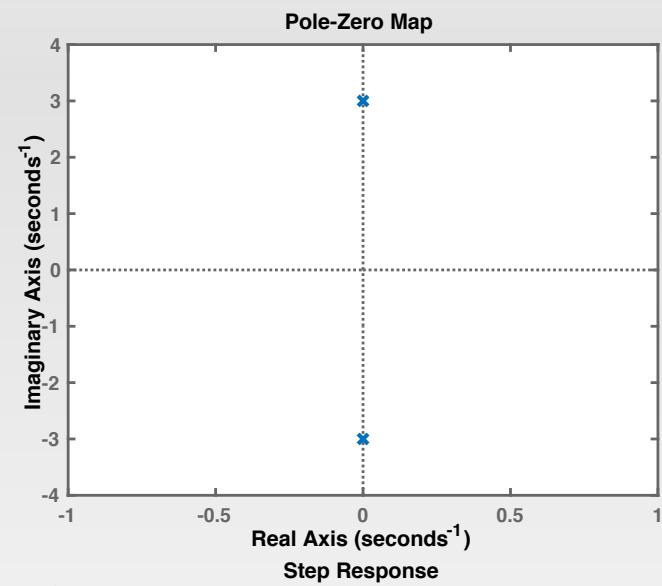
$$w_n^2 = 9 \rightarrow w_n = 3$$

$$\zeta = 0$$

$$y(t) = 1 - \cos(3t)$$

$$y(t) = A \cos(w_n t - \phi)$$

Sistema oscilatório $\rightarrow (\zeta = 0)$



Sistema oscilatório \rightarrow

$$(\zeta = 0)$$

undamped response \rightarrow

Respuestas transitorias

Sistemas de 2a-orden — Resumo:

- Super amortecido: $\zeta > 1$

$$\frac{9}{s^2 + 9s + 9}$$

```
>> pole(g1)
-7.8541
-1.1459
```

- Subamortecido: $0 < \zeta < 1$

$$\frac{9}{s^2 + 2s + 9}$$

```
>> pole(g2)
-1.0000 + 2.8284i
-1.0000 - 2.8284i
```

$$y(t) = 1 - \exp(-\zeta\omega_n t) \cdot \left[\cos(\omega_d t) + \frac{\zeta}{\sqrt{1-\zeta^2}} \sin(\omega_d t) \right]$$

pólos em: $s = \sigma \pm j\omega_d$ $\begin{cases} \sigma = \omega_n \zeta; \\ \omega_d = \omega_n \sqrt{1 - \zeta^2}; \end{cases}$

- Oscilatório: $\zeta = 0$

$$\frac{9}{s^2 + 9}$$

```
>> pole(g3)
0.0000 + 3.0000i
0.0000 - 3.0000i
```

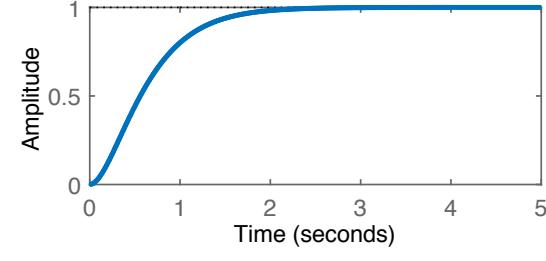
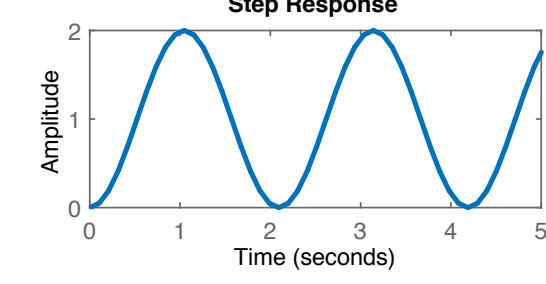
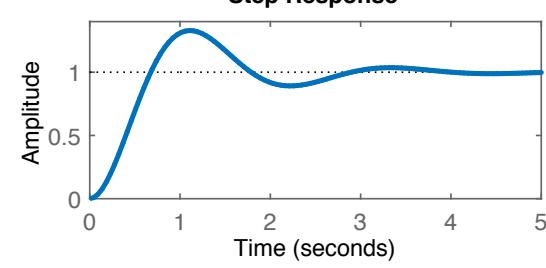
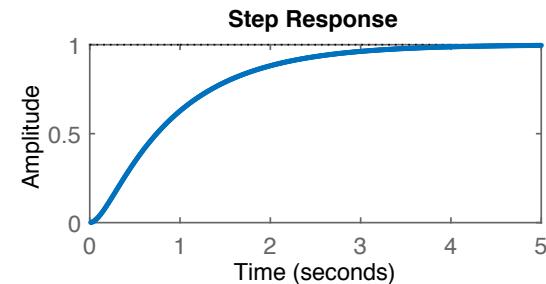
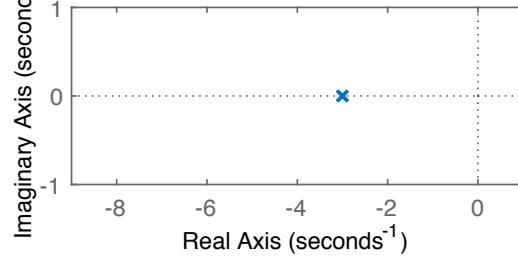
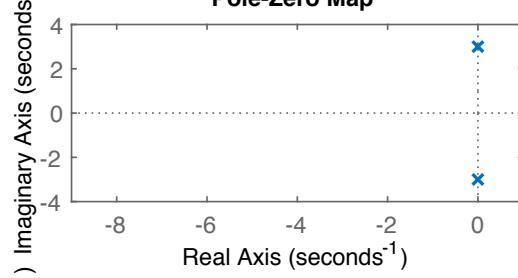
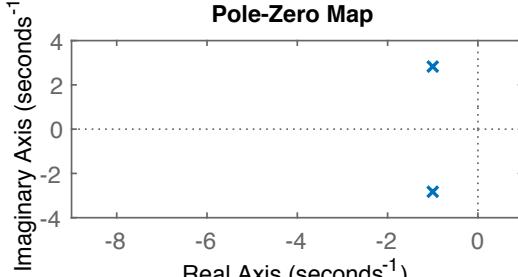
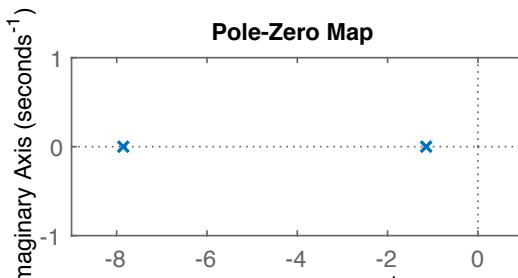
- Criticamente amortecido: $\zeta = 1$

$$\frac{9}{s^2 + 6s + 9}$$

```
>> pole(g4)
-3.0000 + 0.0000i
-3.0000 - 0.0000i
```

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{b}{s^2 + as + b}$$

$$R(s) = \frac{1}{s} \rightarrow \frac{b}{s^2 + as + b} \rightarrow Y(s)$$



Respuestas transitorias

Sistemas de 2a-orden — Resumo:

- Super amortecido: $\zeta > 1$

$$\frac{9}{s^2 + 9s + 9}$$

```
>> pole(g1)
-7.8541
-1.1459
```

- Subamortecido: $0 < \zeta < 1$

$$\frac{9}{s^2 + 2s + 9}$$

```
>> pole(g2)
-1.0000 + 2.8284i
-1.0000 - 2.8284i
```

$$y(t) = 1 - \exp(-\zeta\omega_n t) \cdot \left[\cos(\omega_d t) + \frac{\zeta}{\sqrt{1-\zeta^2}} \sin(\omega_d t) \right]$$

pólos em: $s = \sigma \pm j\omega_d$ $\begin{cases} \sigma = \omega_n \zeta; \\ \omega_d = \omega_n \sqrt{1 - \zeta^2}; \end{cases}$

- Oscilatório: $\zeta = 0$

$$\frac{9}{s^2 + 9}$$

```
>> pole(g3)
0.0000 + 3.0000i
0.0000 - 3.0000i
```

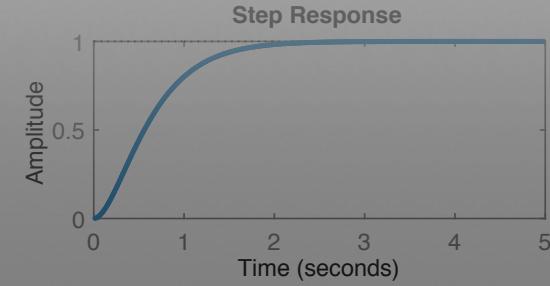
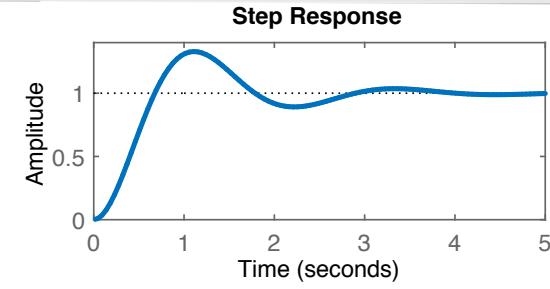
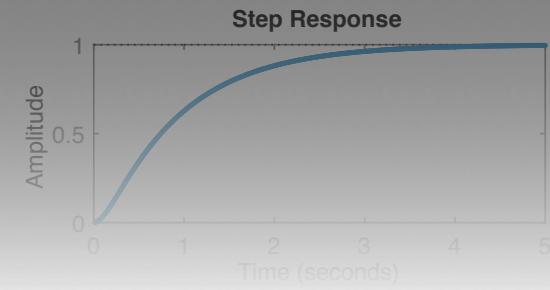
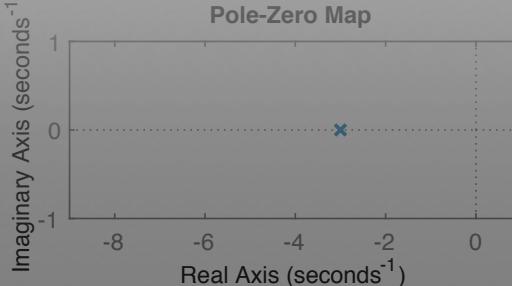
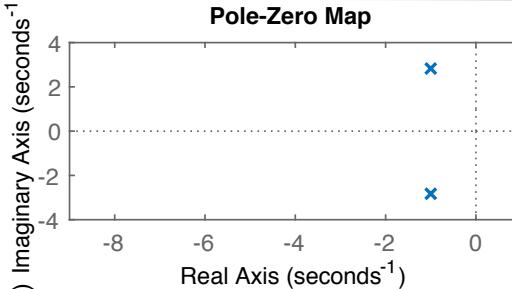
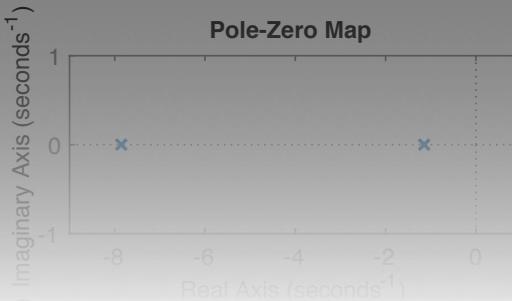
- Criticamente amortecido: $\zeta = 1$

$$\frac{9}{s^2 + 6s + 9}$$

```
>> pole(g4)
-3.0000 + 0.0000i
-3.0000 - 0.0000i
```

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{b}{s^2 + as + b}$$

$$R(s) = \frac{1}{s} \rightarrow \frac{b}{s^2 + as + b} \rightarrow Y(s)$$



Sistemas de 2^a ordem — Detalhes

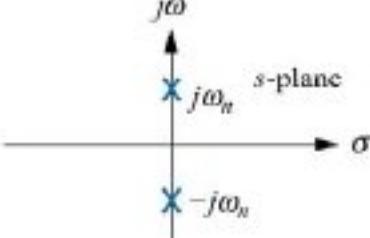
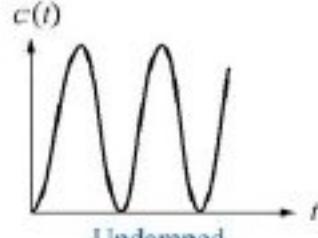
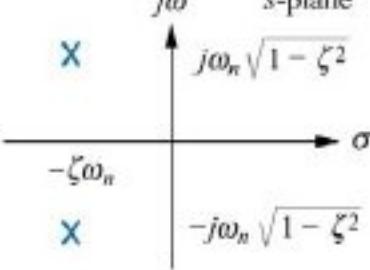
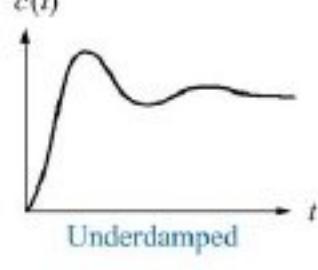
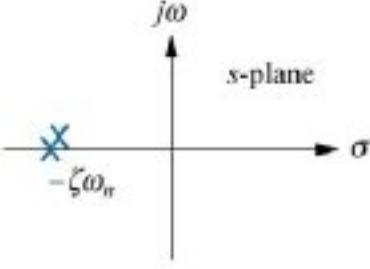
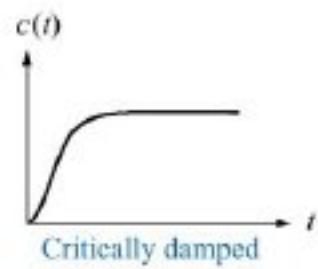
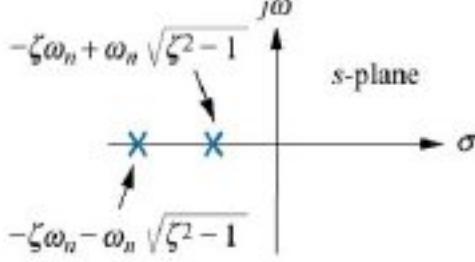
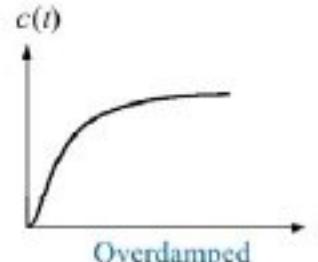
$$FTMF(s) = \frac{b}{s^2 + as + b} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

- Frequência natural de oscilação, ω_n : freqüência de oscilação do sistema sem amortecimento \Rightarrow polos puramente imaginários: $a = 0$, polos sobre $j\omega$ em: $\pm j\sqrt{b}$; $\omega_n = \sqrt{b}$; $b = \omega_n^2$.
- Coeficiente de amortecimento, ζ : os polos complexos contêm parte real, σ igual a $-a/2$. A magnitude deste coeficiente modula o decaimento exponencial:

$$\zeta = \frac{|\sigma|}{\omega_n} = \frac{a/2}{\omega_n}; \quad a = 2\zeta\omega_n$$

Sistemas de 2^a ordem — Detalhes

$$FTMF(s) = \frac{b}{s^2 + as + b} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

ζ	Poles	Step response
0		 Undamped
$0 < \zeta < 1$		 Underdamped
$\zeta = 1$		 Critically damped
$\zeta > 1$		 Overdamped

Sistema subamortecido

- Resposta ao degrau:

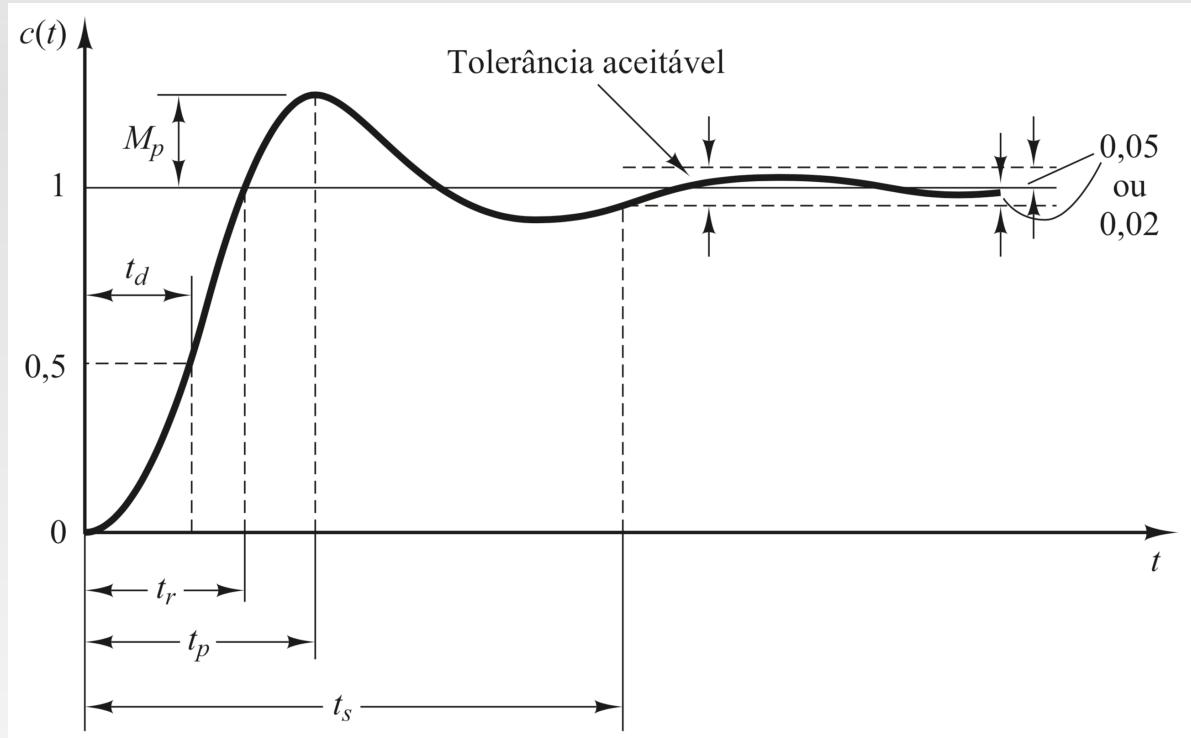
$$C(s) = R(s)Y(s) = \frac{K}{s(s+p_1)(s+p_2)} = k \frac{w_n^2}{s(s^2 + 2\zeta w_n + w_n^2)} = \frac{K_1}{s} + \frac{K_2 s + K_3}{s^2 + 2\zeta w_n^2 + w_n^2}$$

- Assumindo: $0 < \zeta < 1$ (sistema subamortecido):

$$C(s) = \frac{1}{s} - \frac{(s + \zeta w_n) + \frac{\zeta}{\sqrt{1-\zeta^2}} w_n \sqrt{1-\zeta^2}}{(s + \zeta w_n)^2 + w_n^2(1-\zeta^2)}$$

$$\begin{aligned} c(t) &= 1 - e^{-\zeta w_n t} (\cos w_n \sqrt{1-\zeta^2} t + \frac{\zeta}{\sqrt{1-\zeta^2}} \sin w_n \sqrt{1-\zeta^2} t) \\ &= 1 - \frac{1}{\sqrt{1-\zeta^2}} e^{-\zeta w_n t} \cos(w_n \sqrt{1-\zeta^2} t - \phi) \quad \text{where } \phi = \tan^{-1}(\zeta / \sqrt{1-\zeta^2}) \end{aligned}$$

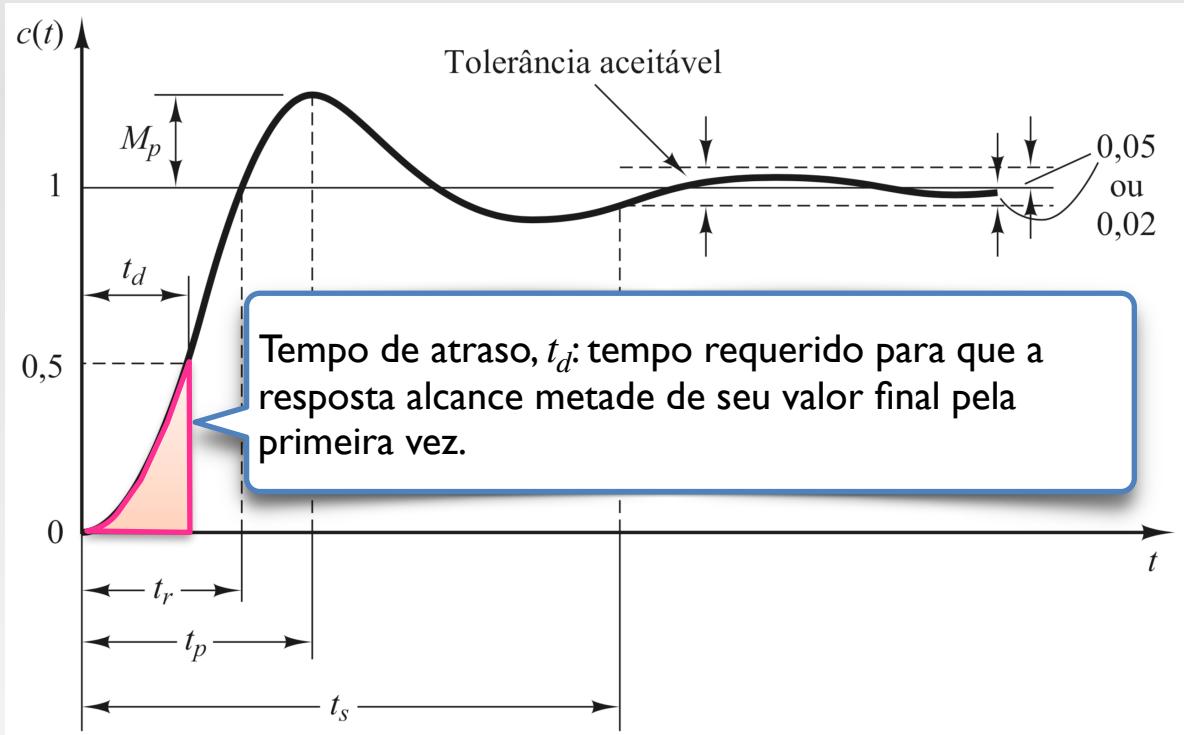
Respostas de Sistemas de 2^a ordem...



Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , % OS (*overshoot*);
4. Tempo de acomodação, t_s , (*settling time*);
5. Tempo de atraso, t_d .

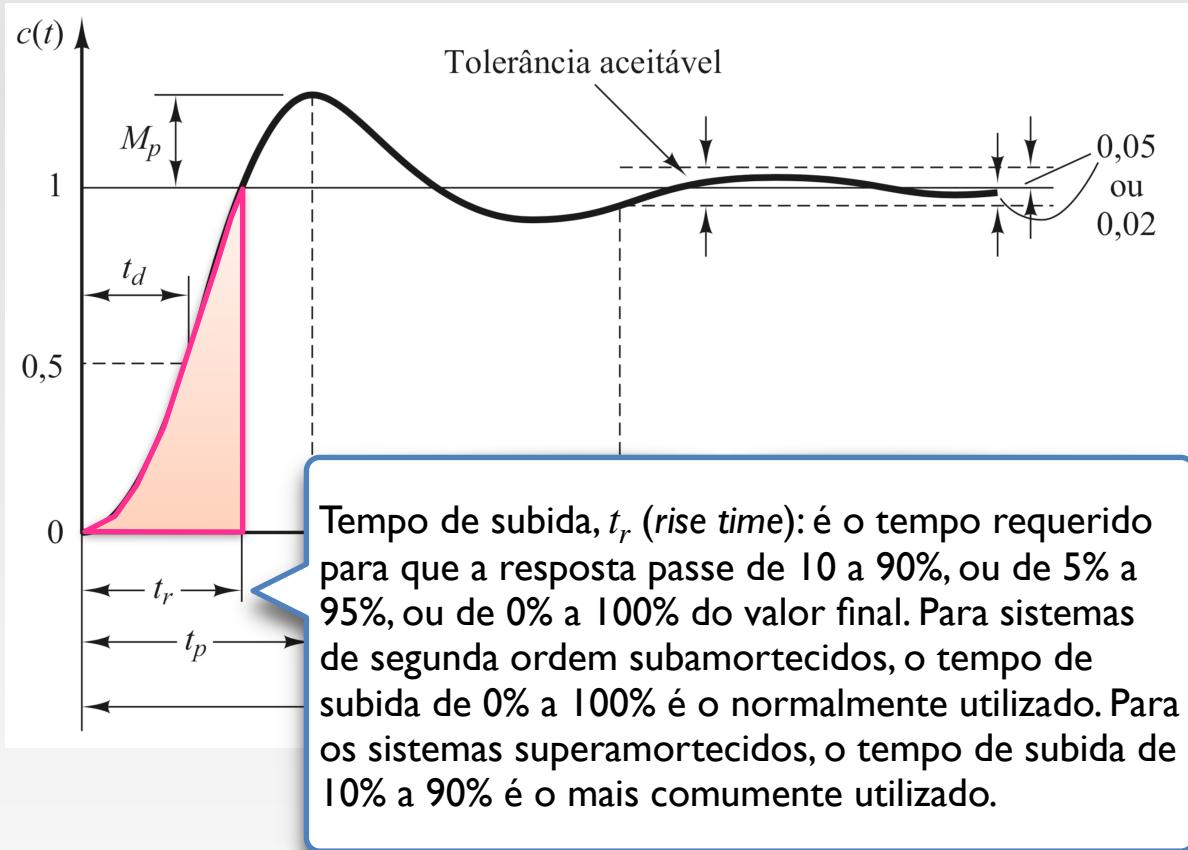
Respostas de Sistemas de 2^a ordem...



Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , % OS (overshoot);
4. Tempo de acomodação, t_s , (settling time);
5. Tempo de atraso, t_d .

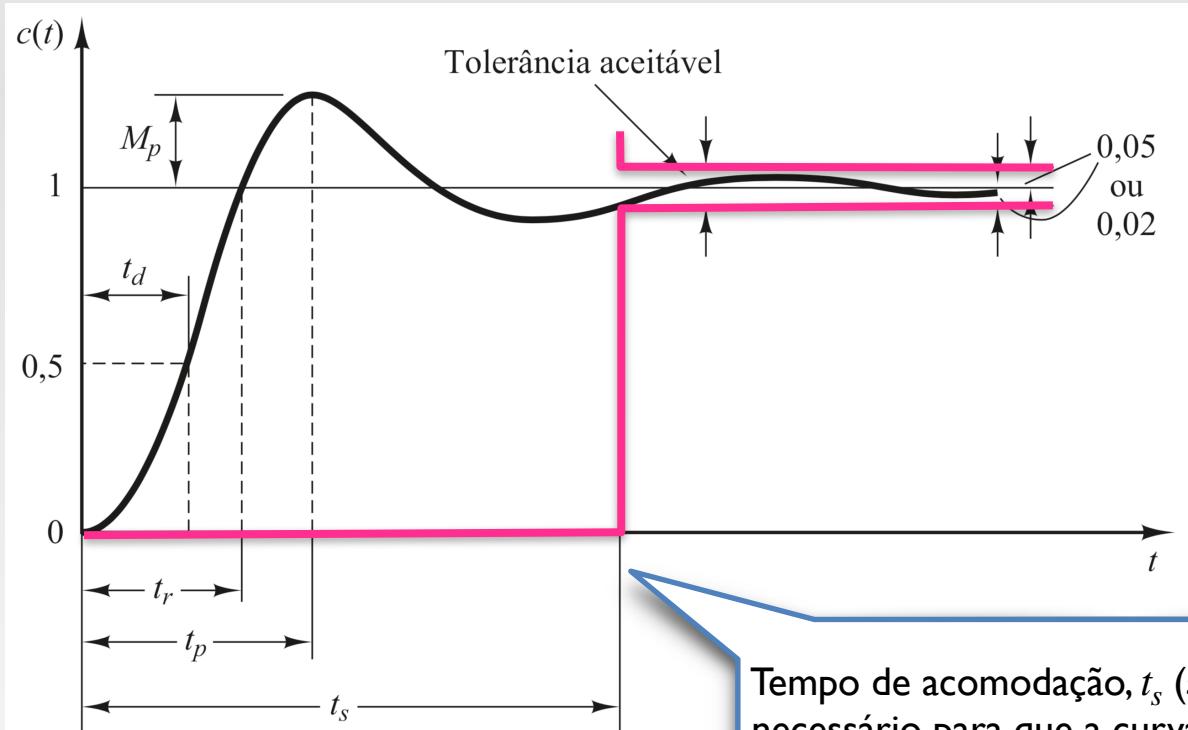
Respostas de Sistemas de 2^a ordem...



Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , % OS (*overshoot*);
4. Tempo de acomodação, t_s , (*settling time*);
5. Tempo de atraso, t_d .

Respostas de Sistemas de 2^a ordem...

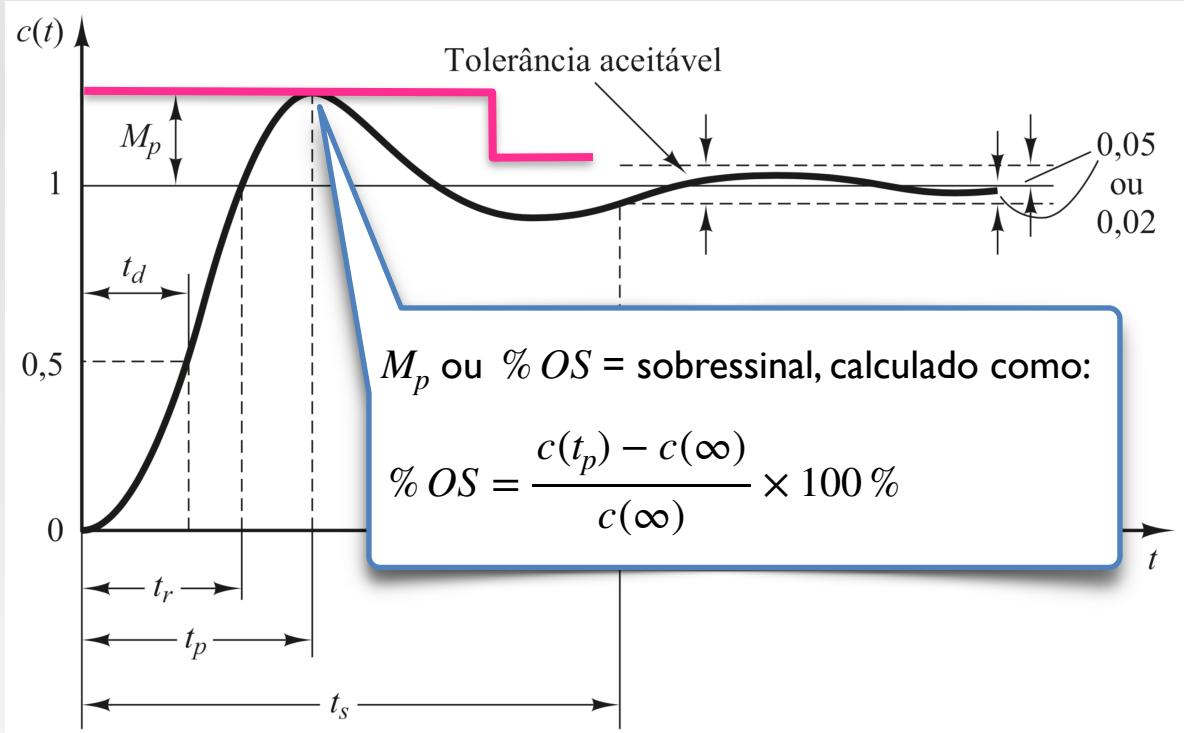


Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , % OS (overshoot);
4. Tempo de acomodação, t_s , (settling time);
5. Tempo de atraso, t_d .

Tempo de acomodação, t_s (settling time): é o tempo necessário para que a curva de resposta alcance valores em uma faixa (geralmente de 2% ou 5%) em torno do valor final, aí permanecendo indefinidamente. O tempo de acomodação está relacionado à maior constante de tempo do sistema de controle.

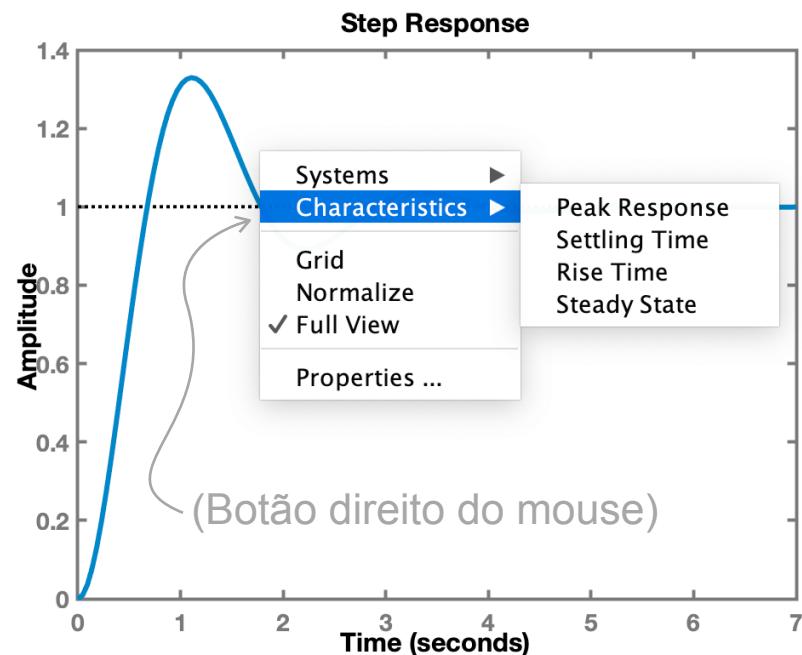
Respostas de Sistemas de 2^a ordem...



Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , $\% OS$ (*overshoot*);
4. Tempo de acomodação, t_s , (*settling time*);
5. Tempo de atraso, t_d .

Respostas de Sistemas de 2^a ordem...



Matlab:

```
>> G=tf(9, [1 2 9]);  
>> zpk(G)
```

$$\frac{9}{s^2 + 2s + 9}$$

Continuous-time zero/pole/gain model.

```
>> step(G)
```

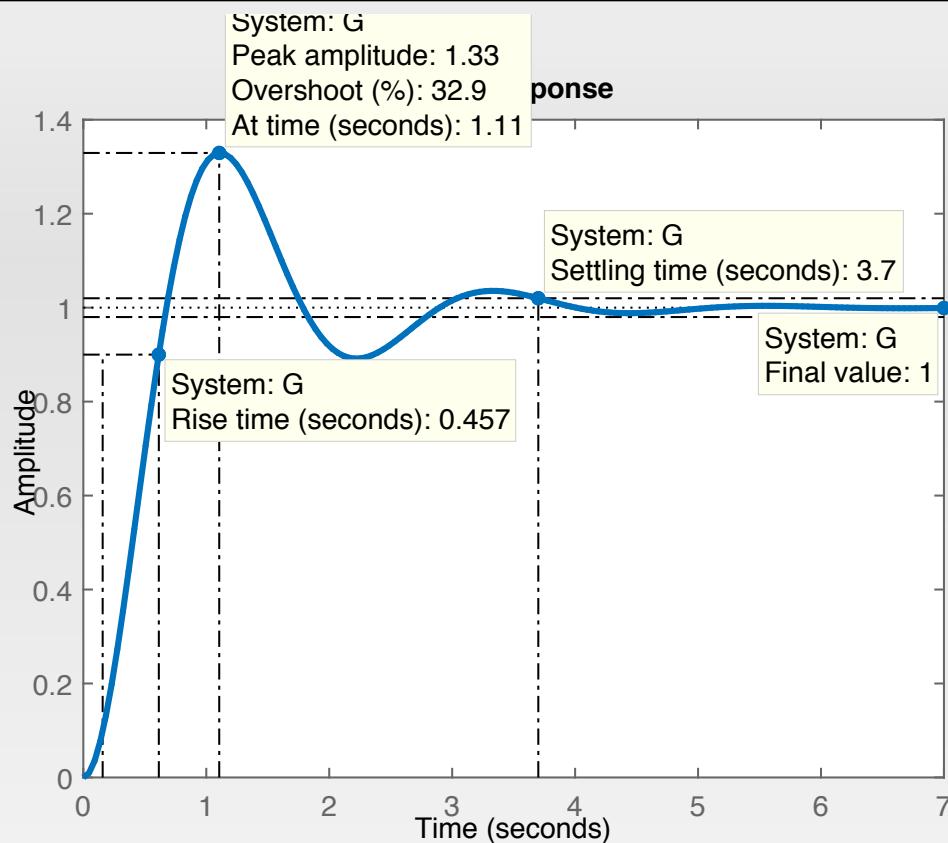
Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p , % OS (*overshoot*);
4. Tempo de acomodação, t_s , (*settling time*);
5. Tempo de atraso, t_d .

```
>> stepinfo(G)  
ans =  
struct with fields:  
  
    RiseTime: 0.4568  
SettlingTime: 3.7005  
SettlingMin: 0.8916  
SettlingMax: 1.3293  
Overshoot: 32.9277  
Undershoot: 0  
Peak: 1.3293  
PeakTime: 1.1052
```

>>

Respostas de Sistemas de 2^a ordem...



Matlab:

```
>> G=tf(9, [1 2 9]);  
>> zpk(G)
```

$$\frac{9}{s^2 + 2s + 9}$$

Continuous-time zero/pole/gain model.

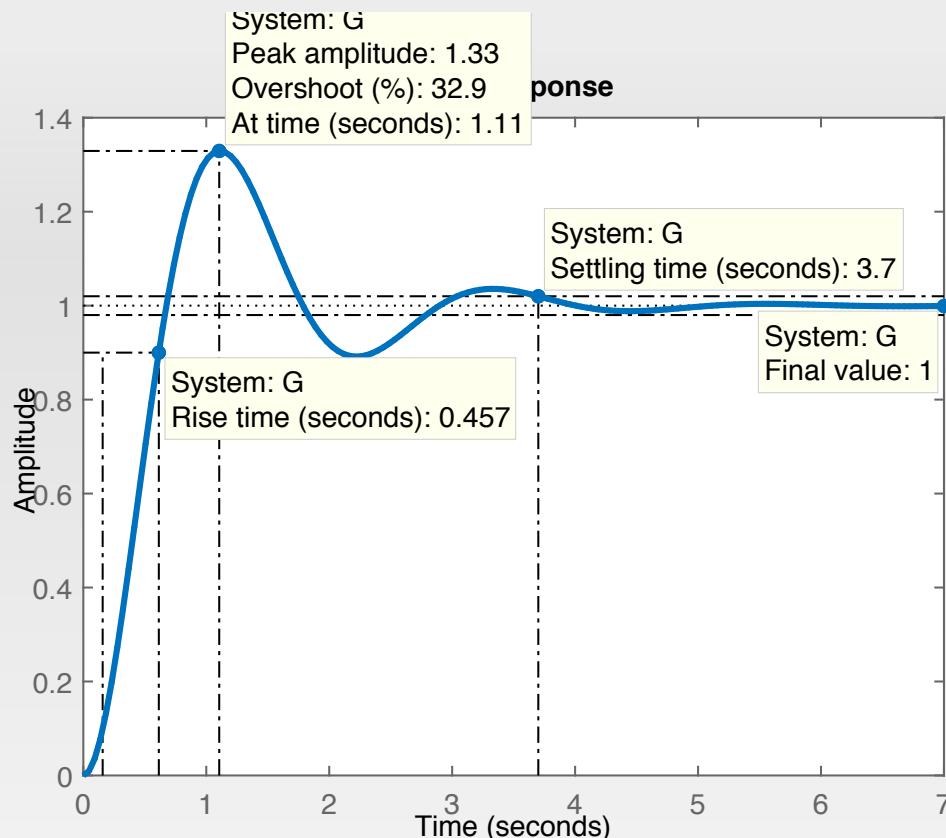
```
>> step(G)
```

Parâmetros temporais:

1. Tempo de subida, t_r ;
2. Tempo de pico, t_p ;
3. Máximo sobressinal, M_p ,
4. Tempo de acomodação, t_s , (*settling time*);
5. Tempo de atraso, t_d .

```
>> stepinfo(G)  
ans =  
struct with fields:  
  
    RiseTime: 0.4568  
    SettlingTime: 3.7005  
    SettlingMin: 0.8916  
    SettlingMax: 1.3293  
    Overshoot: 32.9277  
    Undershoot: 0  
    Peak: 1.3293  
    PeakTime: 1.1052
```

Respostas de Sistemas de 2^a ordem subamortecidos ($0 < \zeta < 1$)



Tempo de pico, t_p (*peak time*): se obtém resolvendo $\frac{dy}{dt} = 0$:

$$0 = \frac{dy}{dt} = L^{-1}\{\dot{c}(t)\} = L^{-1}\left\{ s \cdot \frac{1}{s^2 + 2\zeta w_n s + w_n^2} \right\}$$

$$\dot{c}(t) = \frac{w_n}{\sqrt{1 - \zeta^2}} e^{-\zeta w_n t} \sin(w_n \sqrt{1 - \zeta^2} t)$$

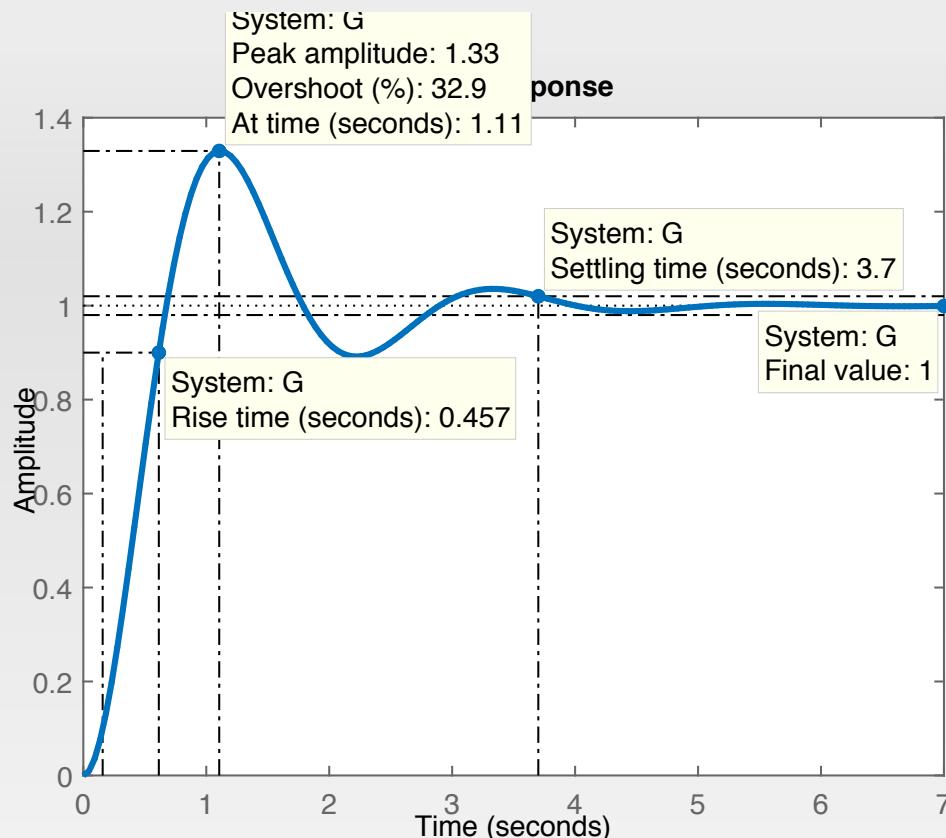
$$w_n \sqrt{1 - \zeta^2} t = n\pi$$

$n = 0$: primeiro ponto da curva de resposta (inclinação zero);

$n = 1$: primeiro pico, equivale a $t = t_p$:

$$t_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}$$

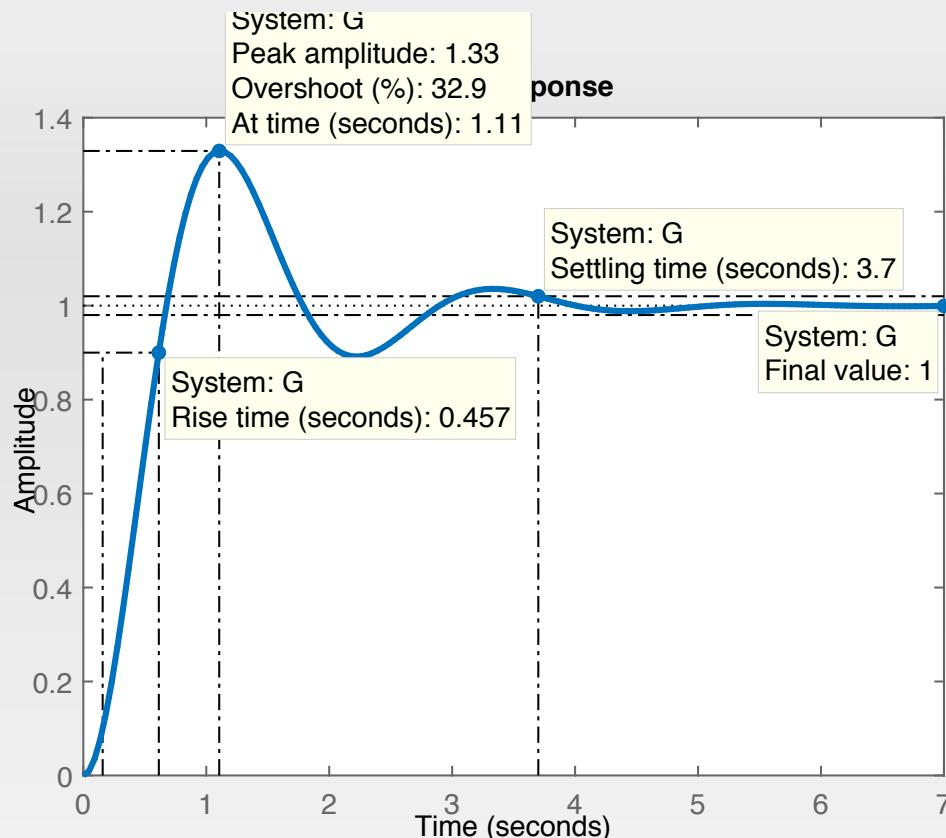
Respostas de Sistemas de 2^a ordem subamortecidos ($0 < \zeta < 1$)



$$\% OS = \exp\left(\frac{-\zeta\pi}{\sqrt{1-\zeta^2}}\right) \times 100\%$$

$$\zeta = \frac{-\ln(\%OS/100)}{\sqrt{\pi^2 + \ln^2(\%OS/100)}}$$

Respostas de Sistemas de 2^a ordem subamortecidos ($0 < \zeta < 1$)



Tempo de assentamento (settling time), t_s
($\pm 2\%$ de tolerância):

A partir de:

$$c(t) = 1 - e^{-\zeta w_n t} \left(\cos(w_n \sqrt{1-\zeta^2} t) + \frac{\zeta}{\sqrt{1-\zeta^2}} \sin(w_n \sqrt{1-\zeta^2} t) \right)$$

Isolamos:

$$e^{-\zeta w_n t} \frac{1}{\sqrt{1-\zeta^2}} = 0,02 \quad (4.40)$$

Sabe-se ainda que em $t = t_p$:

$$\cos(w_n \sqrt{1-\zeta^2} t_p - \phi) = 1$$

Resolvendo (4.40) para $t = t_p$:

$$t_s = \frac{-\ln(0,02\sqrt{1-\zeta^2})}{\zeta \omega_n}$$

O numerador desta eq. varia entre 3,91 à 4,74 enquanto $0 < \zeta < 1$, \Rightarrow
então uma aproximação seria:

$$t_s = \frac{4}{\zeta \omega_n}$$

Respostas de Sistemas de 2^a ordem subamortecidos ($0 < \zeta < 1$) — Resumo

Equações do sistema em MF:

$$FTMF(s) = \frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{b}{s^2 + as + b}$$

$$\frac{Y(s)}{R(s)} = \frac{K}{(s + p_1)(s + p_2)} = K \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right)$$

pólos em: $s = \sigma \pm j\omega_d$ ou: $s = -\zeta\omega_n \pm j\omega_n\sqrt{1 - \zeta^2}$

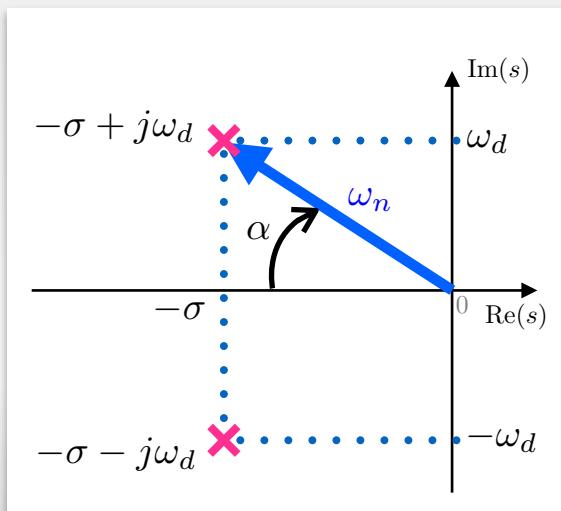
$$\sigma = \omega_n \cos(\alpha) = \omega_n \zeta;$$

$$\omega_d = \omega_n \sin(\alpha) = \omega_n \sqrt{1 - \zeta^2};$$

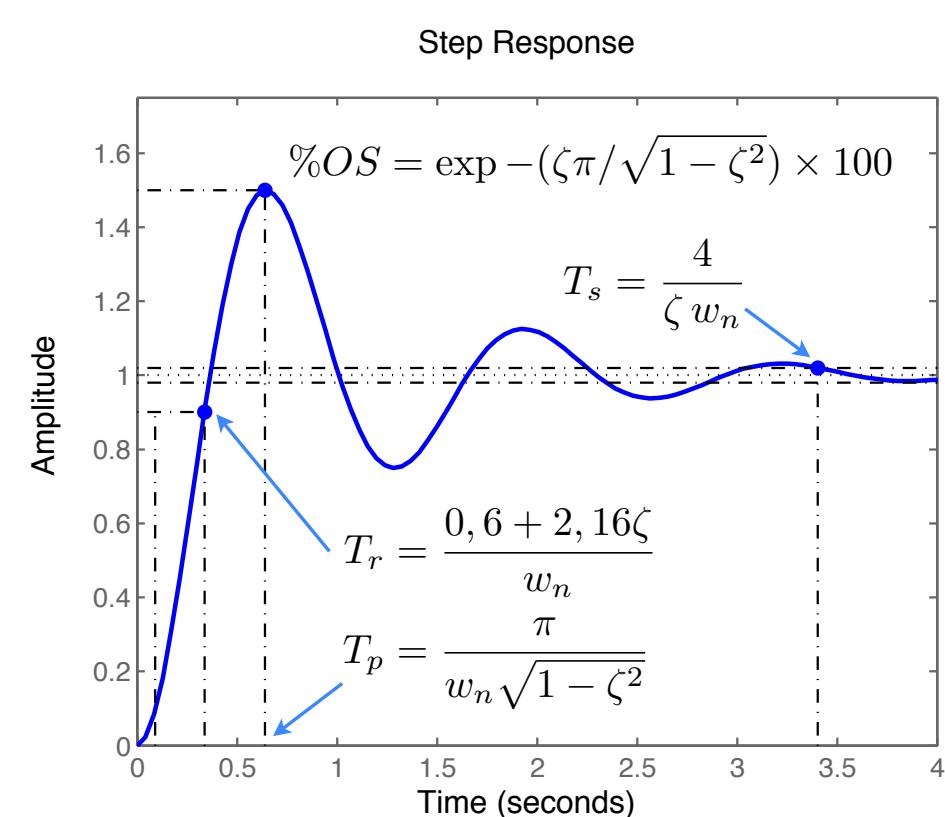
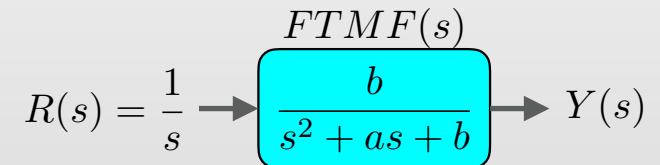
$$\zeta = \cos(\alpha);$$

$$\sin(\alpha) = \sqrt{1 - \zeta^2};$$

$$0 < \zeta < 1$$



Prof. Fernando



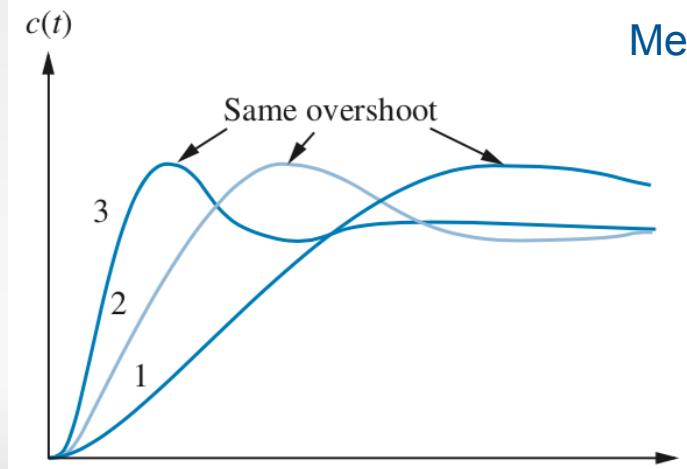
Linhas guia no plano-s

$$\%OS = \exp\left(-\zeta\pi/\sqrt{1-\zeta^2}\right) \times 100$$

$$\zeta = \cos(\theta)$$

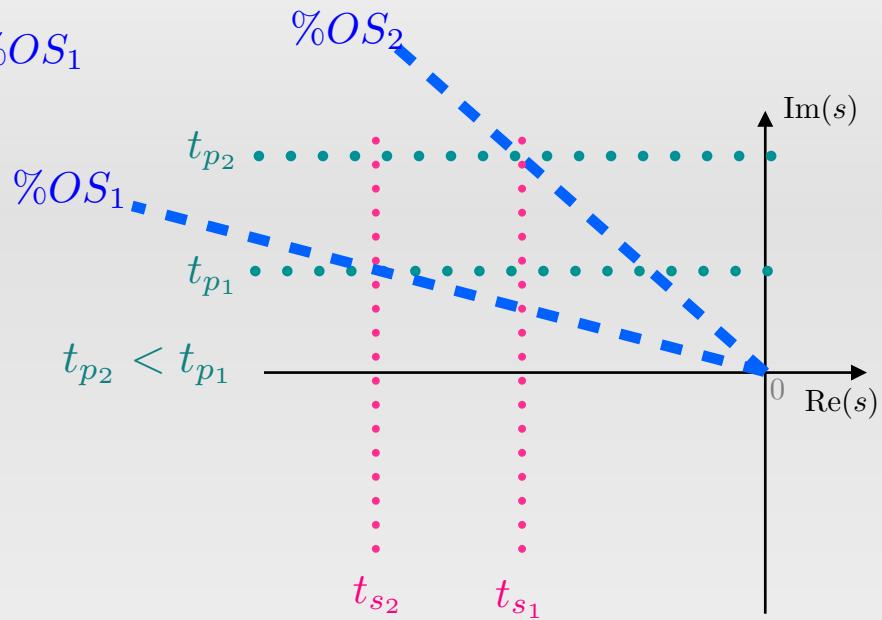
↓

Mesmas linhas radiais:
= mesmo ζ

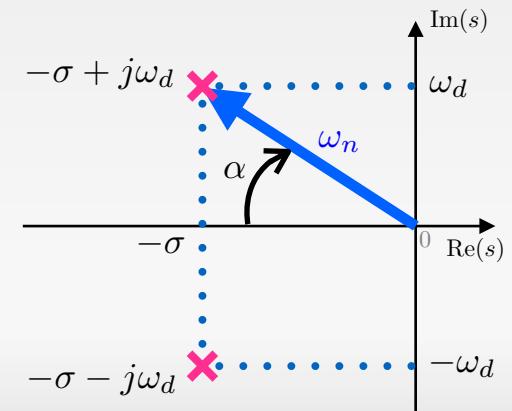
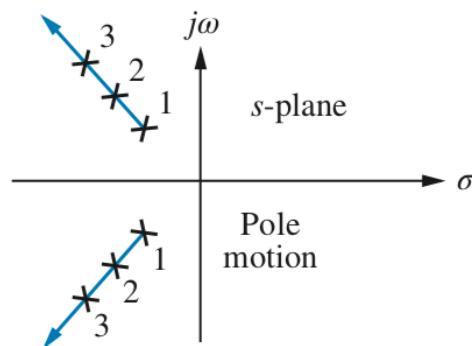


Mesmo ζ :

$$\%OS_2 > \%OS_1$$



$$t_{s_2} < t_{s_1}$$

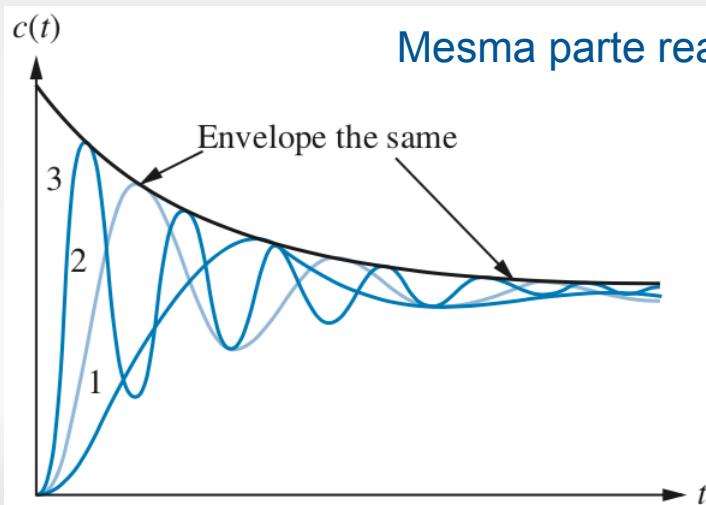


Linhas guia no plano-s

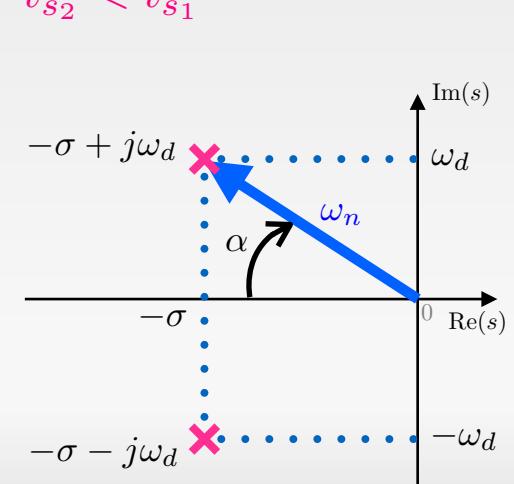
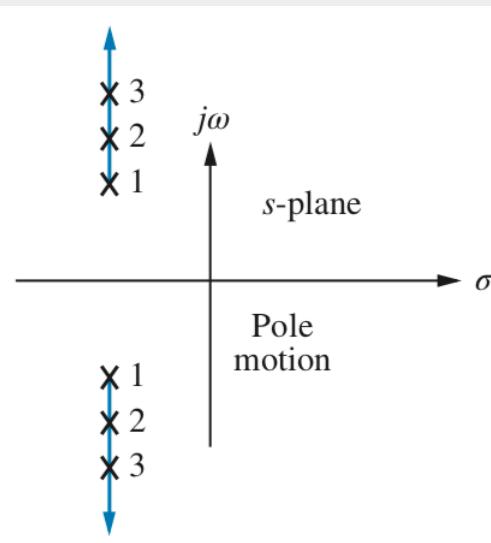
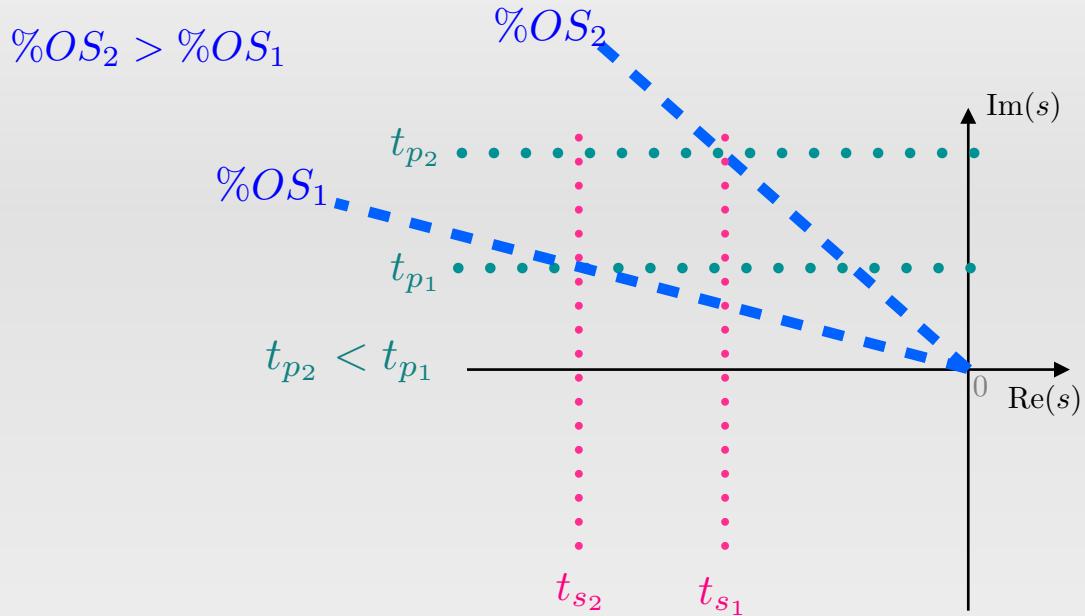
$$t_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}} = \frac{\pi}{\omega_d}$$

$$\begin{matrix} t_p \\ \Downarrow \end{matrix}$$

Inversamente proporcional à parte imaginária do pólo.



Mesma parte real ($= \sigma$):



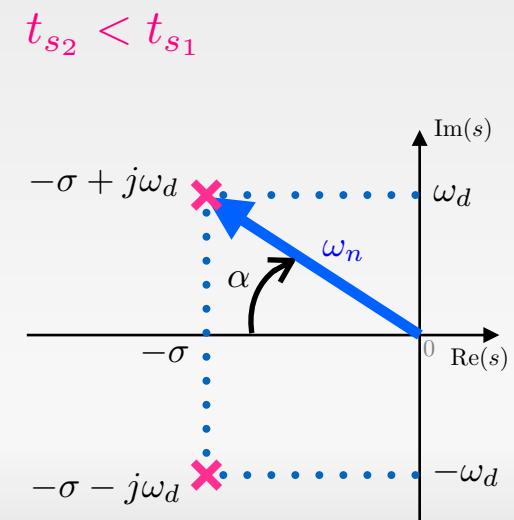
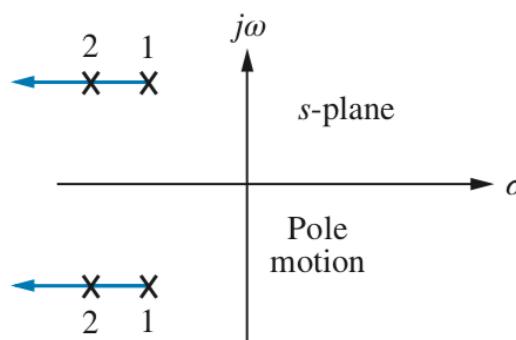
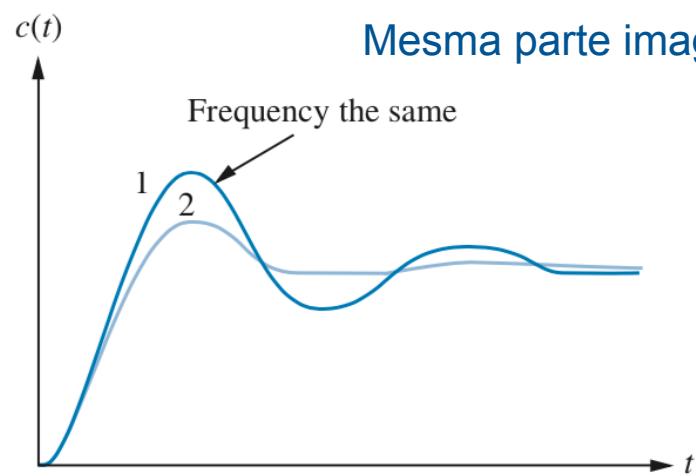
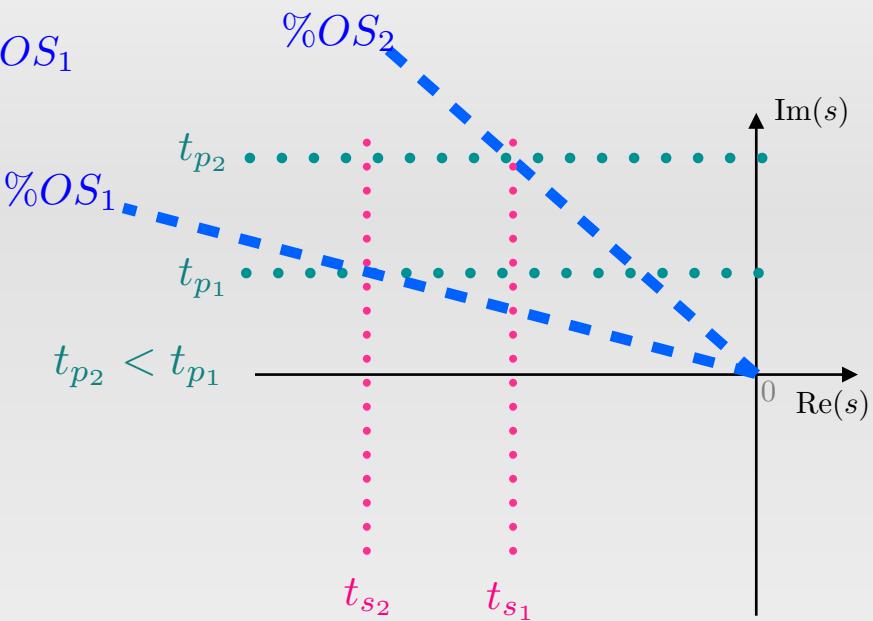
Linhas guia no plano-s

$$t_s = \frac{4}{\zeta \omega_n} = \frac{4}{\sigma}$$

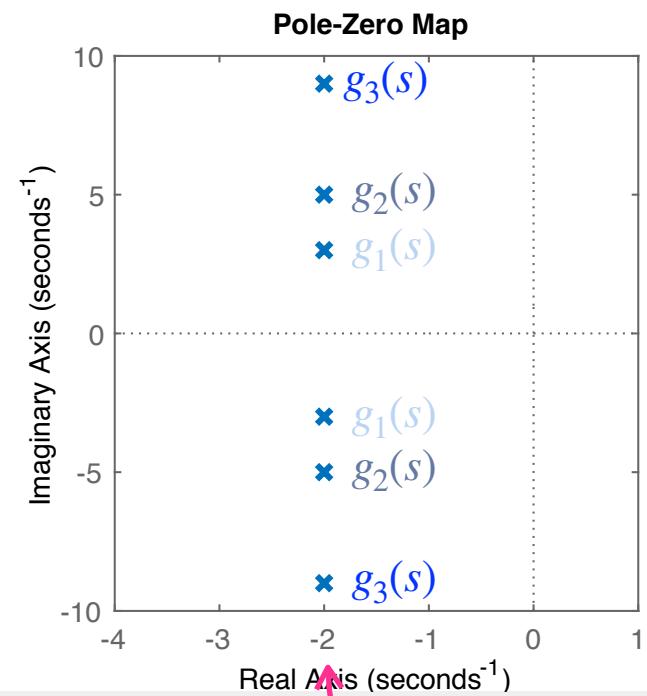
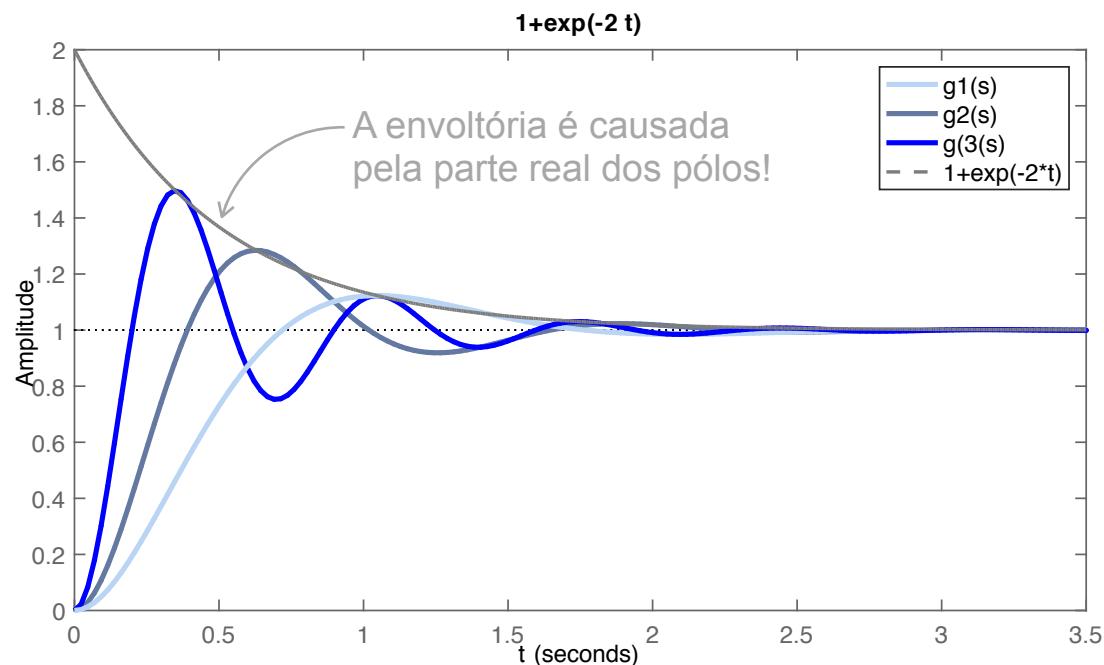
$$\downarrow t_s$$

Inversamente
proporcional à parte real
do pólo

$$\%OS_2 > \%OS_1$$



Sistemas subamortecidos (mesmo σ)



$$\begin{aligned} g_1(s) &= \frac{13}{(s^2+4s+13)} = \frac{13}{(s+2+j3)(s+2-j3)} \\ &= \frac{(3,6056)^2}{s^2+2(0,5547)(3,6056)s+(3,6056)^2} \end{aligned}$$

$$\begin{aligned} g_2(s) &= \frac{9}{(s^2+2s+9)} = \frac{13}{(s+2+j5)(s+2-j5)} \\ &= \frac{(5,3852)^2}{s^2+2(0,3714)(5,3852)s+(5,3852)^2} \end{aligned}$$

$$\begin{aligned} g_3(s) &= \frac{85}{(s^2+4s+85)} = \frac{85}{(s+2+j9)(s+2-j9)} \\ &= \frac{(9,2195)^2}{s^2+2(0,2169)(9,2195)s+(9,2195)^2} \end{aligned}$$

Mesma parte real

Resumo sistemas subamortecidos ($0 < \zeta < 1$)

$$FTMF(s) = \frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{b}{s^2 + as + b}$$

$$y(t) = 1 - \exp(-\zeta\omega_n t) \cdot \left[\cos(\omega_d t) + \frac{\zeta}{\sqrt{1-\zeta^2}} \sin(\omega_d t) \right]$$

pólos em: $s = \sigma \pm j\omega_d$ ou: $s = -\zeta\omega_n \pm j\omega_n\sqrt{1-\zeta^2}$

$$\%OS = \exp\left(-\zeta\pi/\sqrt{1-\zeta^2}\right) \times 100$$

$$\zeta = \frac{-\ln(\%OS/100)}{\sqrt{\pi^2 + \ln^2(\%OS/100)}}$$

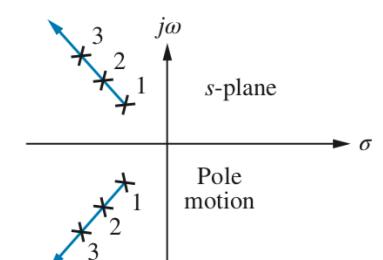
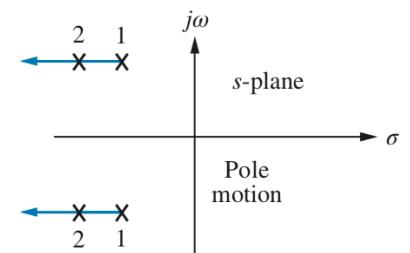
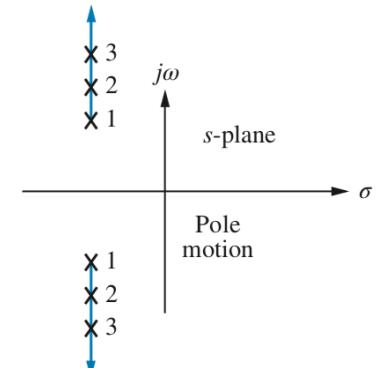
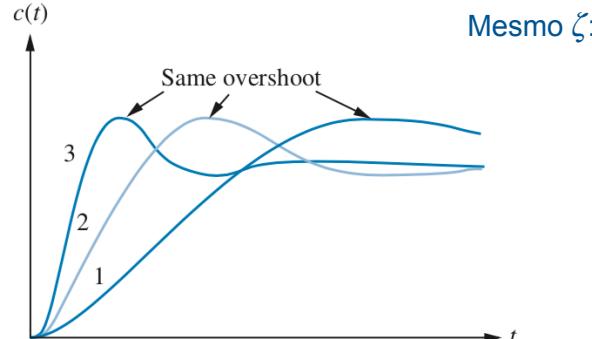
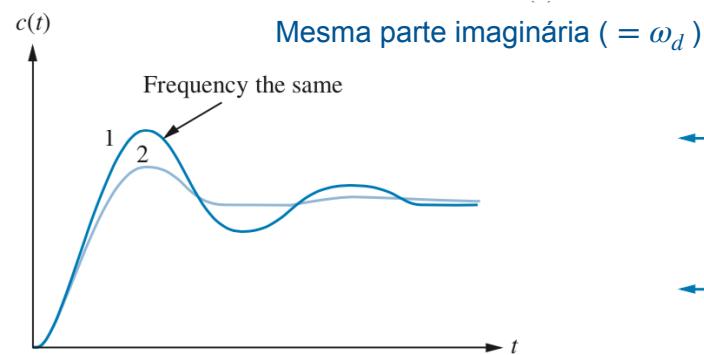
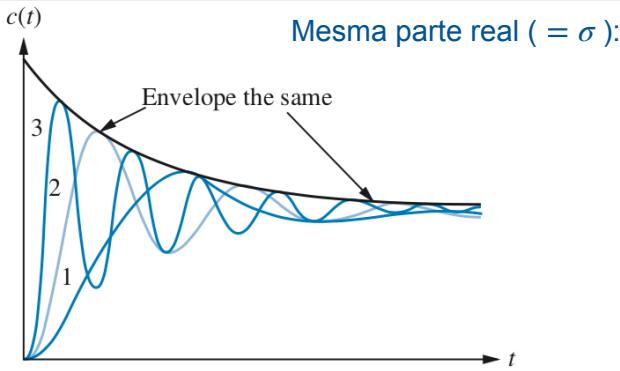
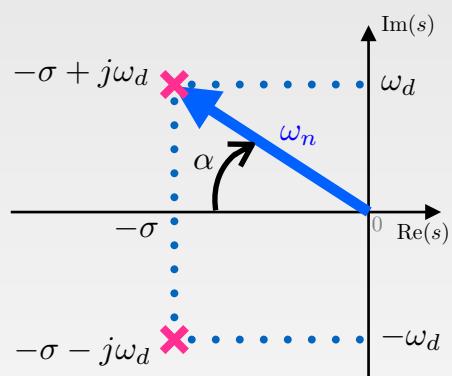
$$\zeta = \cos(\alpha)$$

$$t_p = \frac{\pi}{\omega_n\sqrt{1-\zeta^2}} = \frac{\pi}{\omega_d}$$

$$t_s = \frac{-\ln(0,02\sqrt{1-\zeta^2})}{\zeta\omega_n}$$

$$t_s = \frac{4}{\zeta\omega_n} = \frac{4}{\sigma} \quad \text{para: } 0 < \zeta < 0,9$$

$$t_r = \frac{0,6 + 2,16\zeta}{\omega_n}$$



Procedimento Para Projetos

- 1) Desenhe o diagrama do lugar das raízes para o sistema;
- 2) Assuma que é um sistema de 2^a ordem com ou sem zeros e encontre o ganho que cumpre com as especificações de resposta transitória;
- 3) Justifique a aproximação para um sistema de 2^a ordem encontrando a posição de todos os polos mais rápidos e considerando o fato de que estão muito mais afastados do eixo jw que os pares dominantes de 2a-orden!
- 4) Também verifique que os zeros de laço fechado são aproximadamente cancelados pelos polos de maior ordem (os dominantes).
- 5) Se esta hipótese não pode ser justificada, sua solução terá que ser simulada com o fim de garantir de que se cumpre com a especificação para a resposta transitória.