

23 - 51 - Q4

$$(a) (i) \quad k? \quad \rightarrow z_{1,2} = 0.9 \pm j0.1$$

$$\text{Solution: } k = [0 \ 1] W_c^{-1} \mathcal{A}_c(A)$$

$$W_c^{-1} = [B \ AB]^{-1} = \begin{bmatrix} 1 & 4 \\ 0 & 2 \end{bmatrix}^{-1} = \frac{1}{2} \begin{bmatrix} 2 & -4 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 1 & -2 \\ 0 & \frac{1}{2} \end{bmatrix}$$

$$AB = \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 4 \\ 2 \end{bmatrix}$$

$$\mathcal{A}_c(z) = [z - (0.9 - j0.1)][z - (0.9 + j0.1)]$$

$$= z^2 - (0.9 + j0.1)z - (0.9 - j0.1)z + (0.9 - j0.1)(0.9 + j0.1)$$

$$= z^2 - [0.9 + j0.1 + 0.9 - j0.1]z + 0.82$$

$$= z^2 - 1.8z + 0.82$$

$$\mathcal{A}_c(A) = \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix} \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix} - 1.8 \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix} + 0.82 \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

$$= \begin{bmatrix} 7.62 & -2.2 \\ 4.4 & -1.18 \end{bmatrix}$$

$$k = [0 \ 1] \begin{bmatrix} 1 & -2 \\ 0 & \frac{1}{2} \end{bmatrix} \begin{bmatrix} 7.62 & -2.2 \\ 4.4 & -1.18 \end{bmatrix}$$

$$= [0 \ 0.5] \begin{bmatrix} 7.62 & -2.2 \\ 4.4 & -1.18 \end{bmatrix}$$

$$= [2.2 \ -0.59]$$

(ii) Q: deadbeat

Solution $L_0 = \alpha_0(A) [w_0]^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$

$$\alpha_0(z) = z^2$$

$$\alpha_0(A) = A^2 = \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix}^2 = \begin{bmatrix} 14 & -4 \\ 8 & -2 \end{bmatrix}$$

$$w_0 = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 10 & -2 \end{bmatrix}$$

$$CA = \begin{bmatrix} 2 & 1 \end{bmatrix} \begin{bmatrix} 4 & -1 \\ 2 & 0 \end{bmatrix}$$

$$w_0^{-1} = \begin{bmatrix} \frac{1}{7} & \frac{1}{14} \\ \frac{5}{7} & -\frac{1}{7} \end{bmatrix} = \begin{bmatrix} 0.1428 & 0.0714 \\ 0.7142 & -0.1428 \end{bmatrix} = \begin{bmatrix} 10 & -2 \end{bmatrix}$$

$$L_0 = \begin{bmatrix} 14 & -4 \\ 8 & -2 \end{bmatrix} \begin{bmatrix} \frac{1}{7} & \frac{1}{14} \\ \frac{5}{7} & -\frac{1}{7} \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \begin{bmatrix} \frac{11}{7} \\ \frac{6}{7} \end{bmatrix} = \begin{bmatrix} 1.5714 \\ 0.8571 \end{bmatrix}$$

(b) Q: 帶輸入的系統 - lec 4 - Example 4.8

from (a)(i) $K = \begin{bmatrix} 2.2 & -0.59 \end{bmatrix}$

$$\frac{Y(z)}{R(z)} = K_r C [zI - A + BK]^{-1} B$$

where $[zI - A + BK]$

$$= \begin{bmatrix} z-4 & 1 \\ -2 & z \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} 2.2 & -0.59 \end{bmatrix}$$

$$= \begin{bmatrix} z-4 & 1 \\ -2 & z \end{bmatrix} + \begin{bmatrix} 2.2 & -0.59 \\ 0 & 0 \end{bmatrix}$$

$$= \begin{bmatrix} z-1.8 & 0.41 \\ -2 & z \end{bmatrix}$$

where $\begin{bmatrix} z-1.8 & 0.41 \\ -2 & z \end{bmatrix}^{-1}$

$$= \frac{1}{z(z-1.8)+0.82} \begin{bmatrix} z & -0.41 \\ 2 & z-1.8 \end{bmatrix}$$

$$= \frac{1}{z^2-1.8z+0.82} \begin{bmatrix} z & -0.41 \\ 2 & z-1.8 \end{bmatrix}$$

$$\frac{Y(z)}{R(z)} = k_r C [zI - A + BK]^{-1} B$$

$$= k_r [2 \ 1] \frac{1}{z^2-1.8z+0.82} \begin{bmatrix} z & -0.41 \\ 2 & z-1.8 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$= k_r \frac{1}{z^2-1.8z+0.82} [2 \ 1] \begin{bmatrix} z \\ 2 \end{bmatrix}$$

$$= k_r \frac{2z+2}{z^2-1.8z+0.82}$$

$$Y(z) = k_r \frac{z^2 + 2}{z^2 - 1.8z + 0.82} R(z)$$

$$\lim_{k \rightarrow \infty} y(k) = \lim_{z \rightarrow 1} (1 - z^{-1}) Y(z)$$

$$= \lim_{z \rightarrow 1} (1 - z^{-1}) k_r \frac{z^2 + 2}{z^2 - 1.8z + 0.82} \frac{1}{1 - z^{-1}}$$

$$= k_r \frac{4}{1 - 1.8 + 0.82}$$

$$= 200 k_r$$

$$\text{So } 200 k_r = 1 \quad k_r = 0.005$$

c) (i) Q: optimal control law $\min J$?

Solution

$$x(k+1) = 2x(k) + 3u(k)$$

$$J = \frac{1}{2} \sum_{k=0}^{\infty} (x^2(k) + 5u^2(k))$$

$$S = A^T S A + Q - A^T S B (r + B^T S B)^{-1} B^T S A$$

$$A = 2 \quad B = 3 \quad Q = 1 \quad r = 5$$

$$\text{So } S = 4S + 1 - 6S(5 + 9S)^{-1} 3S \quad 2$$

$$S = 4s + 1 - \frac{36s^2}{9s + 5}$$

$$-3s = 1 - \frac{36s^2}{9s + 5}$$

$$-3s(9s + 5) = 9s + 5 - 36s^2$$

$$36s^2 - 27s^2 - 15s - 9s - 5 = 0$$

$$9s^2 - 24s - 5 = 0$$

$$S_{1,2} = \frac{4 \pm \sqrt{21}}{3}$$

$$s_1 = 2.8609 \quad s_2 = -0.1942$$

So select the positive root

$$S = 2.8609$$

$$K = (B^T S B + r)^{-1} B^T S A$$

$$A = 2 \quad B = 3 \quad Q = 1 \quad r = 5$$

$$K = (3 \times 5 \times 3 + 5)^{-1} \times 3 \times 5 \times 2$$

$$= \frac{6s}{9s + 5} = 0.5583$$

$$\text{So } u^*(k) = -0.5583 x(k)$$

$$\begin{aligned} \text{(ii)} \quad x(k+1) &= 2x(k) - 3 \times 0.5583 x(k) \\ &= 0.3251 x(k) \end{aligned}$$

since $|0.3251| < 1$, $x(k)$ converges to zero
for $x(0) \neq 0$

So, the final value of $x(k) = 0$

(ii) Design of Deadbeat Observer Gain L_0 :

A deadbeat observer requires the observer error dynamics to have eigenvalues at zero.

Step 1: Compute $A - L_0C$

Let $L_0 = \begin{bmatrix} l_1 \\ l_2 \end{bmatrix}$ and $C = \begin{bmatrix} 2 & 1 \end{bmatrix}$.

Compute L_0C :

$$L_0C = \begin{bmatrix} 2l_1 & l_1 \\ 2l_2 & l_2 \end{bmatrix}.$$

Then:

$$A - L_0C = \begin{bmatrix} 4 - 2l_1 & -1 - l_1 \\ 2 - 2l_2 & -l_2 \end{bmatrix}.$$

Step 2: Set Characteristic Polynomial to $z^2 = 0$

Compute the determinant of $zI - (A - L_0C)$:

$$\det(zI - (A - L_0C)) = z^2 + z(l_2 + 2l_1 - 4) + (2 + 2l_1 - 6l_2) = 0.$$

Set coefficients to zero:

1. $l_2 + 2l_1 - 4 = 0$
2. $2 + 2l_1 - 6l_2 = 0$

Step 3: Solve for l_1 and l_2

From equation 1:

$$l_2 = 4 - 2l_1$$

Substitute into equation 2:

$$\begin{aligned} 2 + 2l_1 - 6(4 - 2l_1) &= 0 \\ 2 + 2l_1 - 24 + 12l_1 &= 0 \\ 14l_1 - 22 &= 0 \\ l_1 &= \frac{22}{14} = \frac{11}{7} \end{aligned}$$

Then:

$$l_2 = 4 - 2\left(\frac{11}{7}\right) = \frac{6}{7}$$

Answer:

$$L_0 = \begin{bmatrix} \frac{11}{7} \\ \frac{6}{7} \end{bmatrix}$$

(b) Design of Control Law with Reference Input:

We need to find k_r such that $y(k)$ reaches unity for a unit-step reference $r(k)$.

Step 1: Compute $(I - A_{cl})^{-1}B$

Using $A_{cl} = A - BK$ and K from part (a):

$$A_{cl} = \begin{bmatrix} 1.8 & -0.41 \\ 2 & 0 \end{bmatrix}$$

Compute $I - A_{cl}$:

$$I - A_{cl} = \begin{bmatrix} -0.8 & 0.41 \\ -2 & 1 \end{bmatrix}$$

Compute the inverse:

$$(I - A_{cl})^{-1} = \frac{1}{0.02} \begin{bmatrix} 1 & -0.41 \\ 2 & -0.8 \end{bmatrix} = 50 \begin{bmatrix} 1 & -0.41 \\ 2 & -0.8 \end{bmatrix}$$

Step 2: Compute the DC Gain

$$\text{DC Gain} = C(I - A_{cl})^{-1}B = [2 \quad 1] \cdot 50 \begin{bmatrix} 1 \\ 2 \end{bmatrix} = 200$$

Step 3: Compute k_r

Set the DC gain to unity:

$$1 = 200 \cdot k_r \quad \Rightarrow \quad k_r = \frac{1}{200} = 0.005$$

Answer:

$$k_r = 0.005$$

(c) (i) Determination of Optimal Control Law:

Given the system:

$$x(k+1) = 2x(k) + 3u(k)$$

and performance index:

$$J = \frac{1}{2} \sum_{k=0}^{\infty} (x^2(k) + 5u^2(k))$$

Step 1: Solve the Discrete Riccati Equation

Let S satisfy:

$$S = A^2 S + Q - \frac{(ABS)^2}{r + B^2 S}$$

Substitute $A = 2$, $B = 3$, $Q = 1$, $r = 5$:

$$S = 4S + 1 - \frac{(6S)^2}{9S + 5}$$

Simplify:

$$-3S = 1 - \frac{36S^2}{9S + 5}$$

Multiply both sides by $9S + 5$:

$$-3S(9S + 5) = 1(9S + 5) - 36S^2$$

Simplify:

$$\begin{aligned} -27S^2 - 15S &= 9S + 5 - 36S^2 \\ 9S^2 - 24S - 5 &= 0 \end{aligned}$$

Solve the quadratic equation:

$$S = \frac{24 \pm \sqrt{24^2 + 180}}{18} = \frac{24 \pm 27.4955}{18}$$

Select the positive root:

$$S = \frac{24 + 27.4955}{18} \approx 2.8619$$

Step 2: Compute Optimal Gain K

$$K = \frac{BSA}{B^2S + r} = \frac{3 \times 2.8619 \times 2}{9 \times 2.8619 + 5} \approx 0.5583$$

Answer:

$$u^*(k) = -0.5583 x(k)$$

(ii) Final Value of $x(k)$ with Optimal Control:

The closed-loop system is:

$$x(k+1) = (2 - 3K)x(k) = 0.3251 x(k)$$

Since $|0.3251| < 1$, $x(k)$ converges to zero for any non-zero initial $x(0)$. Therefore, the final value of $x(k)$ is zero.

Explanation:

The optimal control law stabilizes the system by placing the eigenvalue inside the unit circle. Consequently, the state $x(k)$ decays exponentially to zero, minimizing the performance index J .

Answer:

The final value of $x(k)$ is zero for any non-zero $x(0)$, as the optimal control law ensures the closed-loop system is stable and $x(k)$ converges to zero.