23 - S(-QG

(a) (i)
$$k? \rightarrow Z_{12} = 0.9 \pm j0.1$$

Solwtim: $k = [01]$ Wc $[-]$

(ii) Design of Deadbeat Observer Gain L_0 :

A deadbeat observer requires the observer error dynamics to have eigenvalues at zero.

Step 1: Compute $A-L_0 C$

Let
$$L_0 = egin{bmatrix} l_1 \ l_2 \end{bmatrix}$$
 and $C = egin{bmatrix} 2 & 1 \end{bmatrix}$.

Compute L_0C :

$$L_0C=egin{bmatrix} 2l_1 & l_1\ 2l_2 & l_2 \end{bmatrix}.$$

Then:

$$A-L_0C = egin{bmatrix} 4-2l_1 & -1-l_1 \ 2-2l_2 & -l_2 \end{bmatrix}.$$

Step 2: Set Characteristic Polynomial to $z^2=0\,$

Compute the determinant of $zI - (A - L_0C)$:

$$\det(zI - (A - L_0C)) = z^2 + z(l_2 + 2l_1 - 4) + (2 + 2l_1 - 6l_2) = 0.$$

Set coefficients to zero:

1.
$$l_2 + 2l_1 - 4 = 0$$

2.
$$2 + 2l_1 - 6l_2 = 0$$

Step 3: Solve for \emph{l}_1 and \emph{l}_2

From equation 1:

$$l_2 = 4 - 2l_1$$

Substitute into equation 2:

$$2 + 2l_1 - 6(4 - 2l_1) = 0$$

 $2 + 2l_1 - 24 + 12l_1 = 0$
 $14l_1 - 22 = 0$
 $l_1 = \frac{22}{14} = \frac{11}{7}$

Then:

$$l_2=4-2\left(\frac{11}{7}\right)=\frac{6}{7}$$

Answer:

$$L_0 = \left\lceil rac{11}{7}
ight
ceil$$

(b) Design of Control Law with Reference Input:

We need to find k_r such that y(k) reaches unity for a unit-step reference r(k).

Step 1: Compute $(I-A_{ m cl})^{-1}B$

Using $A_{
m cl}=A-BK$ and K from part (a):

$$A_{
m cl} = egin{bmatrix} 1.8 & -0.41 \ 2 & 0 \end{bmatrix}$$

Compute $I-A_{
m cl}$:

$$I-A_{
m cl}=egin{bmatrix} -0.8 & 0.41 \ -2 & 1 \end{bmatrix}$$

Compute the inverse:

$$(I - A_{\rm cl})^{-1} = rac{1}{0.02} egin{bmatrix} 1 & -0.41 \ 2 & -0.8 \end{bmatrix} = 50 egin{bmatrix} 1 & -0.41 \ 2 & -0.8 \end{bmatrix}$$

Step 2: Compute the DC Gain

$$ext{DC Gain} = C(I-A_{ ext{cl}})^{-1}B = egin{bmatrix} 2 & 1 \end{bmatrix} \cdot 50 egin{bmatrix} 1 \ 2 \end{bmatrix} = 200$$

Step 3: Compute k_r

Set the DC gain to unity:

$$1 = 200 \cdot k_r \quad \Rightarrow \quad k_r = \frac{1}{200} = 0.005$$

Answer:

$$k_r = 0.005$$

(c) (i) Determination of Optimal Control Law:

Given the system:

$$x(k+1) = 2x(k) + 3u(k)$$

and performance index:

$$J=rac{1}{2}\sum_{k=0}^{\infty}\left(x^2(k)+5u^2(k)
ight)$$

Step 1: Solve the Discrete Riccati Equation

Let S satisfy:

$$S=A^2S+Q-rac{(ABS)^2}{r+B^2S}$$

Substitute A=2, B=3, Q=1, r=5:

$$S = 4S + 1 - \frac{(6S)^2}{9S + 5}$$

Simplify:

$$-3S = 1 - \frac{36S^2}{9S + 5}$$

Multiply both sides by 9S + 5:

$$-3S(9S+5) = 1(9S+5) - 36S^2$$

Simplify:

$$-27S^2 - 15S = 9S + 5 - 36S^2$$

 $9S^2 - 24S - 5 = 0$

Solve the quadratic equation:

$$S = \frac{24 \pm \sqrt{24^2 + 180}}{18} = \frac{24 \pm 27.4955}{18}$$

Select the positive root:

$$S = rac{24 + 27.4955}{18} pprox 2.8619$$

Step 2: Compute Optimal Gain K

$$K = rac{BSA}{B^2S + r} = rac{3 imes 2.8619 imes 2}{9 imes 2.8619 + 5} pprox 0.5583$$

Answer:

$$u^*(k) = -0.5583 x(k)$$

(ii) Final Value of x(k) with Optimal Control:

The closed-loop system is:

$$x(k+1) = (2-3K)x(k) = 0.3251x(k)$$

Since |0.3251| < 1, x(k) converges to zero for any non-zero initial x(0). Therefore, the final value of x(k) is zero.

Explanation:

The optimal control law stabilizes the system by placing the eigenvalue inside the unit circle. Consequently, the state x(k) decays exponentially to zero, minimizing the performance index J.

Answer:

The final value of x(k) is zero for any non-zero x(0), as the optimal control law ensures the closed-loop system is stable and x(k) converges to zero.