Digital Communication

Ch5 Carrier and Symbol Synchronization

5-1. Signal Parameter Estimate.

- · Propagation delay t between Tx and Rx
- $S(t) = Re[S_1(t)e^{j2\pi\hbar t}]$
- $\rightarrow \gamma(t)$
- = $\operatorname{Re}\left[\operatorname{Se}\left(\frac{t-t}{t-t}\right) e^{\int_{-\infty}^{2\pi h} (t-t)} + \operatorname{Ne}(t) e^{\int_{-\infty}^{2\pi h} t}\right]$
- = $\operatorname{Re}\left\{\left[\operatorname{Se}\left(t-\underline{\tau}\right)e^{j\phi} + \operatorname{Ne}(t)\right]e^{j2\pi\hbar t}\right\}$
 - 7. Rx和Tx 振盪不同步 2. 時間 delay.

- · φ=-2πfcτ
- t and \$\phi\$ are different random variable.
- · φ~ U(-π,π)
- $\Rightarrow r(t) = S_{\ell}(t; \phi, \tau) + n(t)$
- ⇒ Let O = (中, T) ←要估計的參數

uniform

· MAP estimate

$$\hat{\theta} = \underset{\theta}{\text{argmax}} f(\theta \mid r, s)$$

- = $\underset{\theta}{\operatorname{argmax}} f(\theta) \cdot f(r|\theta,s)$
- · Assume O are uniform distribution.
- = $argmax f(\theta) \cdot f(r|\theta,s) \rightarrow ML estimate of <math>\theta$

· ML estimate

$$f(\gamma \mid \Theta, S) = \left(\frac{1}{\sqrt{\pi N_o}}\right)^N \exp\left(\frac{-\sum_{i=1}^N |\gamma_i - S(\theta)|^2}{N_o}\right)$$

- $\sum_{i=1}^{N} |r_i S(\theta)|^2 = \int |r(t) S(t;\theta)|^2 dt$ 數位
- $=) \hat{\theta} = \underset{\theta}{\text{arg max}} \cdot \exp\left(-\frac{1}{N_0} \int |Y(t) S(t;\theta)|^2 dt\right)$
 - where $\Lambda(\Theta) = \exp\left(-\frac{1}{N_0}\int |Y(t) S(t;\theta)|^2 dt\right)$

5.2-1 ML carrier Phase Estimation

$$\dot{X}$$
 Assume T=0 and Estimate ϕ , so Θ = θ

$$\rightarrow \wedge(\theta)$$

$$= \exp\left(-\frac{1}{N_0}\int |r(t) - S(t;\theta)|^2 dt\right)$$

$$= \exp\left(-\frac{1}{N_0}\int Y^2(t) - 2Y(t)S(t;\theta) + S^2(t;\theta)\right) dt$$

$$= \exp\left(\frac{2}{N_0} \int r(t) S(t;\theta) dt\right)$$

•
$$S(t;\theta) = \text{Re}\left[S_{\ell}(t)e^{j\theta}e^{j2\pi\hbar t}\right]$$

$$\| S(t;\theta) \|^2 = \frac{1}{2} \| S_{\ell}(t) e^{j\theta} \|^2 = \frac{1}{2} \| S_{\ell}(t) \|^2$$

•
$$\hat{\phi} = \underset{\theta}{\text{arg ma}} \times \exp\left(-\frac{2}{N_0} \int r(t) s(t;\theta) dt\right)$$

Dicision Directed Loops for Estimate &

· For general modulation.

$$\begin{cases} S_{\ell}(t) = \sum In g(t-nT) ; then \\ S(t; \underline{\theta}) = Re \left[S_{\ell}(t) e^{j\underline{\theta}} e^{j2\pi kt} \right] \end{cases}$$

·
$$\wedge_{\mathsf{L}}(\theta)$$

$$= \frac{2}{N_0} \int \gamma(t) \cdot S(t;\theta) dt$$

=
$$Re\left[\frac{2}{N_o}\int Y(t) \cdot \sum \ln g(t-nT) e^{j\theta} e^{j2\pi kt} dt\right]$$

$$= \operatorname{Re}\left[\frac{2}{N_0} \sum \operatorname{In} y_n e^{j\theta}\right]$$

$$\frac{\partial \Lambda_{L}(\theta)}{\partial \theta} = 0 \implies \hat{\phi} = -\tan^{-1} \left[\frac{\operatorname{Im} \left\{ \sum \operatorname{In} y_{n} \right\}}{\operatorname{Re} \left\{ \sum \operatorname{In} y_{n} \right\}} \right] \not$$

5.3 Symbol time estimation

$$\dot{\mathbb{X}}$$
 Assume $\phi = 0$ and Estimate T , so $\theta = T$

$$= \exp\left(-\frac{1}{N_0}\left[|\gamma(t) - S(t;\tau)|^2 dt\right]$$

$$= \exp\left(-\frac{1}{N_0}\int Y^2(t) - 2Y(t)S(t;t) + S(t,t)\right]dt$$

$$= \exp\left(\frac{2}{N_b} \int r(t) S(t;T) dt\right)$$

$$\begin{cases} F(t) = S(t;T) + Z(t) \\ S(t;T) = Re \left\{ \sum_{i=1}^{\infty} \sum_{j=1}^{\infty} \sum_{i=1}^{\infty} \sum_{j=1}^{\infty} \sum_{j=1}^{\infty}$$

$$= \frac{2}{N_0} \sum_{n} I_n \int_{\Gamma(t)} g(t-nT-\tau) \cos(2\pi h(t-\tau)) dt$$

$$= \frac{2}{N_0} \sum I_n y_n(\tau)$$

5.4 Joint Estimation of Carrier Phase and Symbol Timing.

$$\bigwedge(\phi, T) = e \times p \left(-\frac{1}{N_0} \int |Y(t) - S(t; \phi, T)|^2 dt \right)$$

. Assume $S_{\ell}(t) = \sum I_{n} g(t-nT) + j \sum Q_{n} p(t-nT)$

• PAM : In real and Qn = 0

· QAM and PSK: In complex and Qn=0

• DQPSK : $p(t) = q(t - \frac{1}{2})$

$$\begin{cases}
\frac{\partial \Lambda_{L}(\phi, \tau)}{\partial \phi} = 0 \\
\frac{\partial \Lambda_{L}(\phi, \tau)}{\partial \tau} = 0
\end{cases}$$