

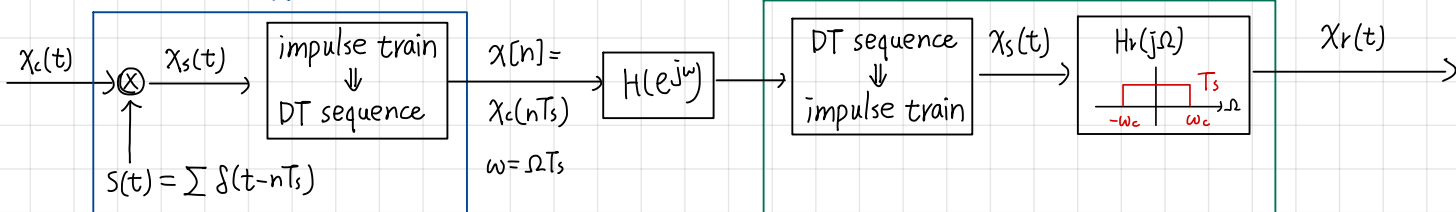
Digital Signal Processing

Ch4 Sampling

system

C/D converter

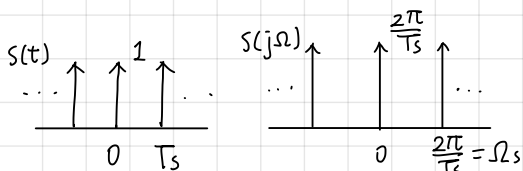
D/C converter



C/D converter

• ideal sampling

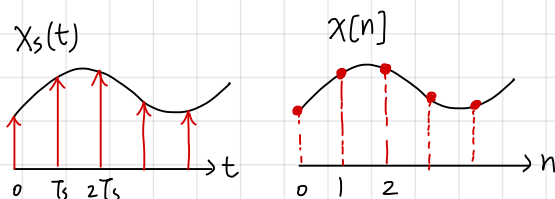
$$s(t) = \sum \delta(t - nT_s) \leftrightarrow S(j\Omega) = \frac{2\pi}{T_s} \sum \delta(\Omega - k\frac{2\pi}{T_s})$$



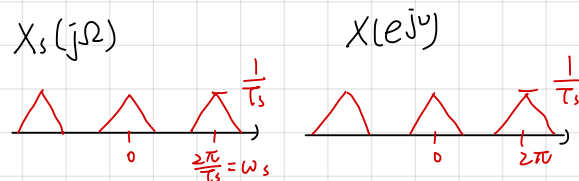
$$x_s(t) = \sum x_c(nT_s) \delta(t - nT_s) \leftrightarrow$$

$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) = \frac{1}{T_s} \sum X_c(j(\Omega - k\frac{2\pi}{T_s}))$$

橫軸尺度改變



$$X(e^{j\omega}) = X_s(j\omega/T_s)$$



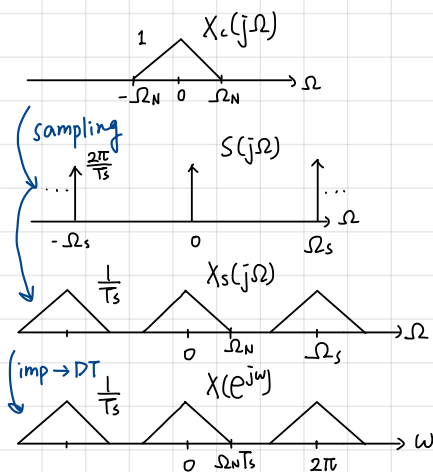
• impulse train \rightarrow DT sequence

$$\begin{cases} X_s(j\Omega) = \sum x_c(nT_s) \mathcal{F}\{\delta(t - nT_s)\} = \sum x_c(nT_s) e^{-jn\Omega T_s} \\ X(e^{j\omega}) = \sum x[n] e^{-j\omega n} = \sum x_c(nT_s) e^{-j\omega n} \end{cases}$$

• When $\omega = \Omega T_s$: $X(e^{j\omega}) = X_s(j\Omega) \big|_{\omega = \Omega T_s}$

$$\begin{cases} X_s(j\Omega) = \frac{1}{T_s} \sum X_c(j(\Omega - k\frac{2\pi}{T_s})) \\ X(e^{j\omega}) = \frac{1}{T_s} \sum X_c(j(\frac{\omega - 2\pi k}{T_s})) \end{cases} \rightarrow x[n] = x_c(nT_s)$$

以 2π 為週期



Nyquist-Shannon Sampling theorem

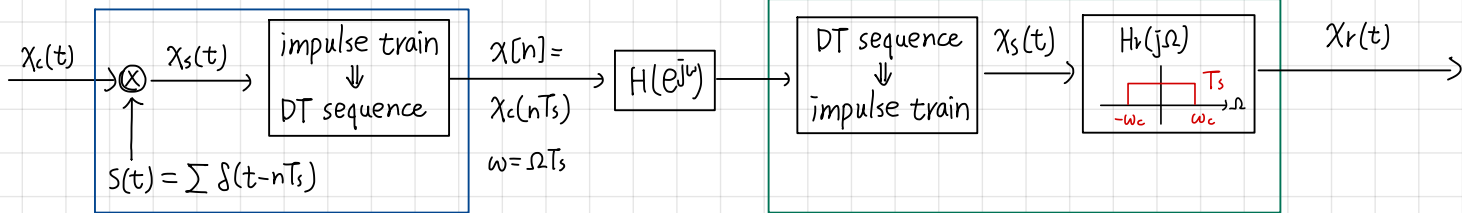
When $\Omega_s \geq 2\Omega_N$

$x_c(t)$ is uniquely determined by its sample $x[n] = x_c(nT_s)$

system

C/D converter

D/C converter



D/C converter

- DT sequence \rightarrow impulse train

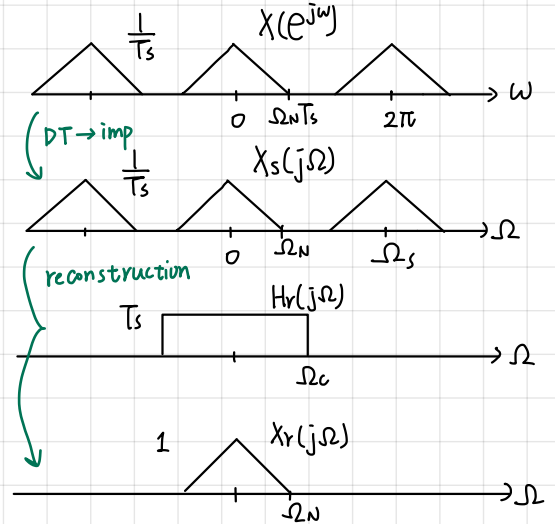
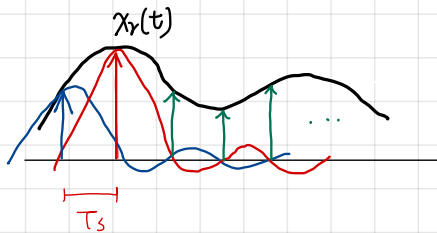
$$\begin{cases} x_s(t) = \sum x[n] \delta(t - nT_s) \\ x_s(j\Omega) = \sum x[n] e^{-j\Omega nT_s} = X(e^{j\omega})|_{\omega = \Omega T_s} \end{cases}$$

- Signal reconstruction

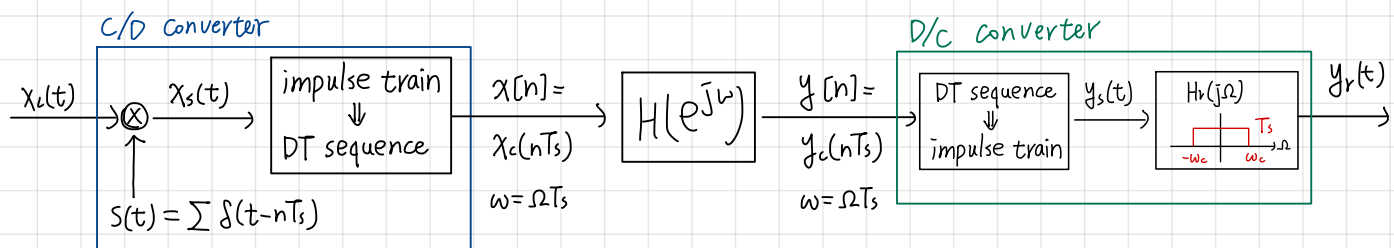
$$\begin{cases} H_r(j\Omega) = T_s \Pi\left(\frac{\Omega}{2\omega_c}\right) \\ h_r(t) = \frac{T_s \omega_c}{\pi} \text{sinc}\left(\frac{\omega_c t}{\pi}\right) \end{cases} \quad \begin{matrix} \omega_s = 2\omega_c \\ \omega_c = \frac{\pi}{T_s} \end{matrix} \rightarrow \text{sinc}\left(\frac{t}{T_s}\right)$$

$$\begin{aligned} X_r(j\Omega) &= X(e^{j\Omega T_s}) H_r(j\Omega) \\ &= T_s X(e^{j\Omega T_s}) ; |\Omega| < \Omega_c \end{aligned}$$

$$x_r(t) = \sum x_c(nT_s) \text{sinc}\left(\frac{t - nT_s}{T_s}\right)$$



Conclusion



$$\begin{aligned} \bullet x_s(t) &= \sum x_c(nT_s) \delta(t - nT_s) \longleftrightarrow X_s(j\Omega) = \frac{1}{T_s} \sum X_c(j(\Omega - k\Omega_s)) \\ \bullet x_s[n] &= x_c(nT_s) \xleftrightarrow{\omega = \Omega T_s} X(e^{j\omega}) = X_s(j\Omega) \end{aligned}$$

$$\begin{aligned} \bullet y_s(t) &= \sum y[n] \delta(t - nT_s) \longleftrightarrow Y_s(j\Omega) = \sum y[n] e^{-j\Omega nT_s} = Y(e^{j\omega}) \\ \bullet y_r(t) &= y_s(t) * h_r(t) \longleftrightarrow Y_r(j\Omega) = Y_s(j\Omega) H_r(j\Omega) \end{aligned}$$

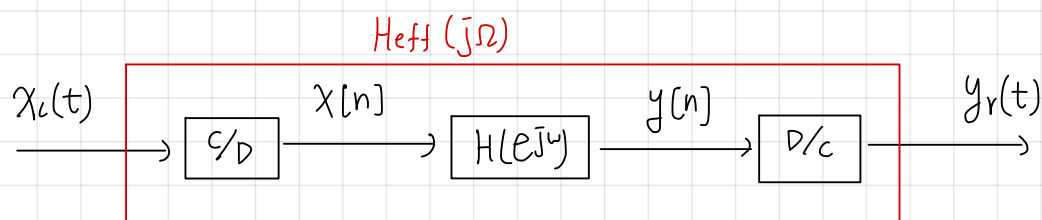
$$\bullet X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) = \frac{1}{T_s} \sum X_c(j(\Omega - k\Omega_s))$$

$$\bullet X(e^{j\omega}) = X_s(j\frac{\omega}{T_s}) = \frac{1}{T_s} \sum X_c(j(\frac{\omega}{T_s} - \frac{2\pi k}{T_s}))$$

$$\bullet Y(e^{j\omega}) = X(e^{j\omega}) H(e^{j\omega}) \quad \Omega = \omega T_s$$

$$\bullet Y_s(j\omega) = Y(e^{j\omega T_s})$$

$$\bullet Y_r(j\omega) = Y_s(j\Omega) H_r(j\Omega)$$

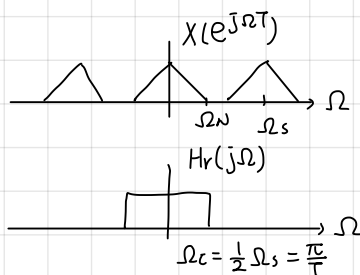


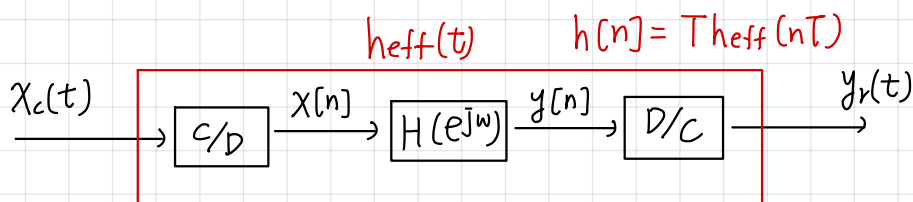
$$Y_r(j\Omega) = X_c(j\Omega) H_{eff}(j\Omega) \quad ; \quad H_{eff}(j\Omega) = \begin{cases} H(e^{j\Omega T_s}) & ; |\Omega| < \frac{\pi}{T_s} \\ 0 & ; \text{else} \end{cases}$$

<pt>

$$Y_r(j\Omega) = X(e^{j\Omega T_s}) H(e^{j\Omega T_s}) H_r(j\Omega)$$

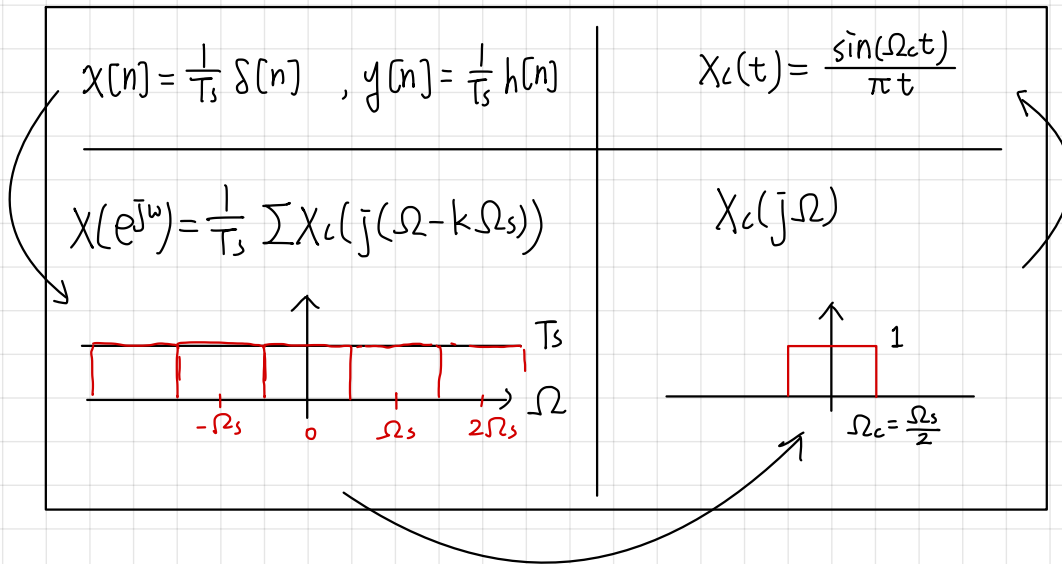
$$= H(e^{j\Omega T_s}) X_c(j\Omega)$$





Differentiator: $y_r(t) = \frac{d}{dt} x_c(t)$

- Find impulse response $h[n]$: let $x[n] = \frac{1}{T_s} \delta[n]$



$$\begin{aligned}
 x[n] &= x_c(nT) = \frac{1}{T} \delta[n] \\
 y[n] &= y_r(nT) = \frac{1}{T} h[n]
 \end{aligned}
 \Rightarrow h[n] = T y_r(nT)$$

$$\begin{aligned}
 h[n] &= T h_{\text{eff}}(nT) = T \frac{\sin(\Omega_c nT)}{\pi nT} = \frac{\sin(\omega_c n)}{\pi n} \quad \neq \\
 H(e^{j\omega}) &= \begin{cases} 1 & : |\omega| < \omega_c \\ 0 & : \text{else} \end{cases}
 \end{aligned}$$