

Fig. 3: Structure of PLFRAME.

SHORT FRAME (16200 bits)								
PLFRAME	SOF	PLSC	Slot-i	#Slots	Pilot Symbols	Pilot Blocks	#Pilots	Without Pilots
QPSK	26	64	90	90	36	5	180	8190
8PSK	26	64	90	60	36	3	108	5490
16APSK	26	64	90	45	36	2	72	4140
32APSK	26	64	90	36	36	2	72	3330

TABLE I: Short frame details.

NORMAL FRAME (64800 bits)								
PLFRAME	SOF	PLSC	Slot-i	#Slots	Pilot Symbols	Pilot Blocks	#Pilots	Without Pilots
QPSK	26	64	90	360	36	22	792	32490
8PSK	26	64	90	240	36	14	504	21690
16APSK	26	64	90	180	36	11	396	16290
32APSK	26	64	90	144	36	8	288	13050

TABLE II: Long frame details.

$$I_{2i} = -Q_{2i} = -\frac{1}{\sqrt{2}}(1 - 2y_{2i-1}) \quad 1, 2, \dots, 45 \quad (2.2)$$

The PLHEADER, represented by binary stream  $y_1, \dots, y_{90}$  is modulated into 90  $\frac{\pi}{2}$ -BPSK symbols. Eqs.(2.1) and (2.2) shows the generation of  $\frac{\pi}{2}$ -BPSK mapping.

- PLHEADER consists of two fields,
  - Starting of Frame(SOF) 26 symbols
  - Physical Layer Signalling Code (PLSC) , 64 symbols

#### A. Generation of SOF

SOF consistutes a fixed sequence 18D2E82<sub>HEX</sub> in binary format which is from right to left.

#### B. Generation of PLSC

- Generation of PLSC involves definining 7 symbols and multiplying first 6 symbols with the defined  $G$  matrix in [1]. First 5 symbols called as MODCOD filed and next 2 symbols as TYPE field.

Mode	MOD COD	Mode	MOD COD	Mode	MOD COD	Mode	MOD COD
QPSK 1/4	1 <sub>D</sub>	QPSK 5/6	9 <sub>D</sub>	8PSK 9/10	17 <sub>D</sub>	32APSK 4/5	25 <sub>D</sub>
QPSK 1/3	2 <sub>D</sub>	QPSK 8/9	10 <sub>D</sub>	16APSK 2/3	18 <sub>D</sub>	32APSK 5/6	26 <sub>D</sub>
QPSK 2/5	3 <sub>D</sub>	QPSK 9/10	11 <sub>D</sub>	16APSK 3/4	19 <sub>D</sub>	32APSK 8/9	27 <sub>D</sub>
QPSK 1/2	4 <sub>D</sub>	8PSK 3/5	12 <sub>D</sub>	16APSK 4/5	20 <sub>D</sub>	32APSK 9/10	28 <sub>D</sub>
QPSK 3/5	5 <sub>D</sub>	8PSK 2/3	13 <sub>D</sub>	16APSK 5/6	21 <sub>D</sub>	Reserved	29 <sub>D</sub>
QPSK 2/3	6 <sub>D</sub>	8PSK 3/4	14 <sub>D</sub>	16APSK 8/9	22 <sub>D</sub>	Reserved	30 <sub>D</sub>
QPSK 3/4	7 <sub>D</sub>	8PSK 5/6	15 <sub>D</sub>	16APSK 9/10	23 <sub>D</sub>	Reserved	31 <sub>D</sub>
QPSK 4/5	8 <sub>D</sub>	8PSK 8/9	16 <sub>D</sub>	32APSK 3/4	24 <sub>D</sub>	DUMMY PLFRAME	0 <sub>D</sub>

Fig. 4: MODCOD coding for various mapping schemes

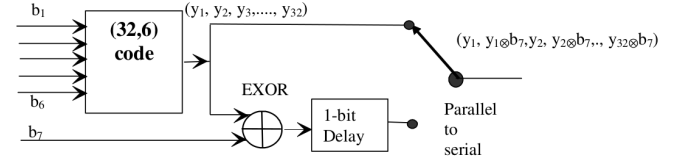


Fig. 5: Physical Layer Signalling generation

- First 5 symbols represents MODCOD which specifies Frame's mapping scheme and code rate.
- Next 2 symbols represents TYPE filed which specifies Frame length and presence and absence of pilot filed. (0 = normal: 64 800 bits; 1 = short: 16 200 bits) ; (0 = no pilots, 1 = pilots)

Fig. 4 shows the MODCOD coding for various mapping schemes. Similarly, Fig. 5 shows the generation of 64 bits.

After the generation of PLS code, we will again scramble the PLS Code with the fixed SCR sequence which is defined in the [1].

$$PLSC = PLSC \oplus SCR \quad (2.3)$$

#### C. Generation of Pilots

Pilot block consists of  $P = 36$  symbols. Each pilot is composed of un-modulated complex symbol. Where,  $I = Q = \frac{1}{\sqrt{2}}$  The first pilot block inserted

PARAMETERS OF THE normal AND short PLFRAME

normal frame: $\eta_{LDPC} = 64800$ bits					short frame: $\eta_{LDPC} = 16200$ bits			
$\eta_{MOD}$	$S$	$\alpha_{PIL}$	$K$	$\eta$ (%)	$S$	$\alpha_{PIL}$	$K$	$\eta$ (%)
QPSK: 2	360	22	33282	97.35	90	5	8370	96.77
8PSK: 3	240	14	22194	97.32	60	3	5598	96.46
16APSK: 4	180	11	16686	97.09	45	2	4212	96.15
32APSK: 5	144	8	13338	97.17	36	2	3402	95.24

Fig. 6: paramters of plframe

16 slots after the PLHEADER and next is inserted after the 32 slots and so on.

$$K = \begin{cases} 90 \times (S + 1) & \text{with out pilots} \\ 90 \times (S + 1) + 36 \times \alpha_{PIL} & \text{with pilots} \end{cases} \quad (2.4)$$

Where ,  $\alpha_{PIL} = \left\lfloor \frac{(S-1)}{16} \right\rfloor$ .

Eq.(2.4) specifies the total length  $K$  of the PL-FRAME. Smiliarly, Fig. 6 Shows the Parameters of PLFRAME.

### 3. PULSE SHAPING

$$Y_k(m) = H_k(m) * X_k + V_k(m) \quad m = 1, \dots, M; k = 1, \dots, N \quad (3.1)$$

Where,  $H_k$  represents the pulse shape,  $V_k(m) \sim \mathcal{N}(0, \sigma^2)$ .

At the Receiver we will,

$$Y_k(m) * H_k^*(M - m) = H_k^*(M - m) * H_k(m) * X_k + V_k(m) \quad (3.2)$$

$H(f)$  will be choosen from the [1] which is converted to time domain form to get  $H_k(m)$

$$H(f) = \begin{cases} 1 & |f| < f_N(1 - \alpha) \\ \left\{ \frac{1}{2} + \frac{1}{2} \sin \frac{\pi}{2} \left[ \frac{f_N - |f|}{\alpha} \right] \right\}^{\frac{1}{2}} & |f| = f_N(1 - \alpha) \\ 0 & |f| > f_N(1 - \alpha) \end{cases} \quad (3.3)$$

### REFERENCES

- [1] A. Morello and V. Mignone, "DVB-S2X: The New Extensions to the Second Generation DVB Satellite Standard DVB-S2," *Int. J. Satell. Commun. Netw.*, vol. 34, no. 3, pp. 323–325, May 2016. [Online]. Available: <https://doi.org/10.1002/sat.1167>