



# **Model Predictive Control**

7. Reference Tracking and Disturbance Rejection

Jun.-Prof. Dr.-Ing. Daniel Görges
Juniorprofessur für Elektromobilität
Technische Universität Kaiserslautern



## **Reference Tracking based on Target Calculation**

Discrete-Time Linear Time-Invariant (LTI) System

$$x(k+1) = Ax(k) + Bu(k)$$
 state equation (7.1)

$$y(k) = Cx(k)$$
 measured output equation (7.2)

$$\mathbf{y}_r(k) = \mathbf{C}_r \mathbf{x}(k)$$
 controlled output equation (7.3)

Symbols

$$\boldsymbol{x}(k) \in \mathbb{R}^n$$
 state vector  $\boldsymbol{u}(k) \in \mathbb{U} \subseteq \mathbb{R}^m$  input vector

$$y(k) \in \mathbb{Y} \subseteq \mathbb{R}^p$$
 measured output vector  $y_r(k) \in \mathbb{R}^{p_r}$  controlled output vector

$$\pmb{A} \in \mathbb{R}^{n \times n}$$
 system matrix  $\pmb{B} \in \mathbb{R}^{n \times m}$  input matrix

$$\pmb{\mathcal{C}} \in \mathbb{R}^{p \times n}$$
 measured output matrix  $\pmb{\mathcal{C}_r} \in \mathbb{R}^{p_r \times n}$  controlled output matrix

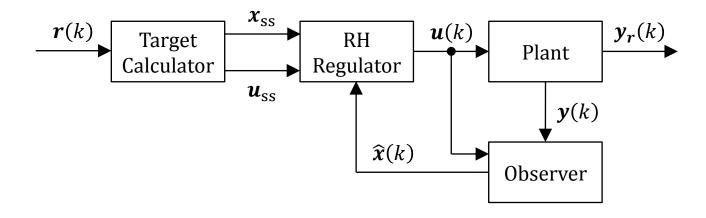
#### Remarks

- The measured output y(k) is used for the observer
- The controlled output  $oldsymbol{y_r}(k)$  is considered for reference tracking



### **Reference Tracking based on Target Calculation**

#### Structure



### Objective

- Control the discrete-time LTI system such that  $y_r(k) \to r(k)$  if  $r(k) \to \text{const.}$  as  $k \to \infty$ 

### Approach

- The target calculator computes the target state  $oldsymbol{x}_{ ext{SS}}$  and the target input  $oldsymbol{u}_{ ext{SS}}$
- The RH regulator controls the discrete-time LTI system to the target pair  $(x_{ss}, u_{ss})$



### **Reference Tracking based on Target Calculation**

#### Approach

- Consider that the discrete-time LTI system (7.1)/(7.3) is in the steady target state  $x_{ss}$ , i.e.

$$x_{SS} = Ax_{SS} + Bu_{SS} \Leftrightarrow (I_{n \times n} - A)x_{SS} - Bu_{SS} = 0_{n \times 1}$$
  
 $y_{rSS} = C_r x_{SS} = r \Leftrightarrow C_r x_{SS} = r$ 

Rewrite the equations in matrix form, i.e.

$$\begin{pmatrix} \mathbf{I}_{n \times n} - \mathbf{A} & -\mathbf{B} \\ \mathbf{C}_{r} & \mathbf{0}_{p_{r} \times m} \end{pmatrix} \begin{pmatrix} \mathbf{x}_{SS} \\ \mathbf{u}_{SS} \end{pmatrix} = \begin{pmatrix} \mathbf{0}_{n \times 1} \\ \mathbf{r} \end{pmatrix}$$
 (7.4)

- The steady target pair  $(x_{ss}, u_{ss})$  can be then calculated from (7.4) provided that a solution exists
- Introduce the state deviation and input deviation

$$\widetilde{\mathbf{x}}(k) = \widehat{\mathbf{x}}(k) - \mathbf{x}_{SS}, \quad \widetilde{\mathbf{u}}(k) = \mathbf{u}(k) - \mathbf{u}_{SS}$$

This leads to the discrete-time LTI state equation

$$\widetilde{\boldsymbol{x}}(k+1) = \widehat{\boldsymbol{x}}(k+1) - \boldsymbol{x}_{SS} = A\widehat{\boldsymbol{x}}(k) + \boldsymbol{B}\boldsymbol{u}(k) - (\boldsymbol{A}\boldsymbol{x}_{SS} + \boldsymbol{B}\boldsymbol{u}_{SS}) = A\widetilde{\boldsymbol{x}}(k) + \boldsymbol{B}\widetilde{\boldsymbol{u}}(k)$$
(7.5)



## **Reference Tracking based on Target Calculation**

### Approach

the discrete-time quadratic cost function

$$\widetilde{V}_{N}\left(\widetilde{\boldsymbol{x}}(k),\widetilde{\boldsymbol{U}}(k)\right) = \widetilde{\boldsymbol{x}}^{T}(k+N)\boldsymbol{P}\widetilde{\boldsymbol{x}}(k+N) + \sum_{i=0}^{N-1} \widetilde{\boldsymbol{x}}^{T}(k+i)\boldsymbol{Q}\widetilde{\boldsymbol{x}}(k+i) + \widetilde{\boldsymbol{u}}^{T}(k+i)\boldsymbol{R}\widetilde{\boldsymbol{u}}(k+i) \quad (7.6)$$

the state and input constraints

$$C(\widetilde{\boldsymbol{x}}(k+i)+\boldsymbol{x}_{SS}) \in \mathbb{Y}, i = 1,2,...,N$$

$$\widetilde{\boldsymbol{u}}(k+i)+\boldsymbol{u}_{SS} \in \mathbb{U}, \quad i = 0,1,...,N-1$$
(7.7)

- Rewrite Problem 4.1/5.1 w.r.t. the state equation (7.5), cost function (7.6) and constraints (7.7), i.e.

$$\min_{\widetilde{\boldsymbol{U}}(k)} \widetilde{V}_N(\widetilde{\boldsymbol{x}}(k), \widetilde{\boldsymbol{U}}(k))$$

subject to 
$$\begin{cases} \widetilde{\boldsymbol{x}}(k+i+1) = \boldsymbol{A}\widetilde{\boldsymbol{x}}(k+i) + \boldsymbol{B}\widetilde{\boldsymbol{u}}(k+i), i = 0,1,...,N-1 \\ \boldsymbol{C}(\widetilde{\boldsymbol{x}}(k+i) + \boldsymbol{x}_{ss}) \in \mathbb{Y}, & i = 1,2,...,N \\ \widetilde{\boldsymbol{u}}(k+i) + \boldsymbol{u}_{ss} \in \mathbb{U}, & i = 0,1,...,N-1 \end{cases}$$
(7.8)



## **Reference Tracking based on Target Calculation**

#### • Remark on the Optimization Problem

- Problem (7.8) can be formulated as a quadratic program using the methods from Chapter 4 and 5
- Problem (7.8) relates to a regulation problem, i.e.  $\widetilde{x}(k) \to 0$ ,  $\widetilde{u}(k) \to 0$  as  $k \to \infty$
- The reference tracking problem is, however, simultaneously addressed since

$$\widetilde{\mathbf{x}}(k) \to \mathbf{0}, \widetilde{\mathbf{u}}(k) \to \mathbf{0} \Rightarrow \widehat{\mathbf{x}}(k) \to \mathbf{x}_{SS}, \mathbf{u}(k) \to \mathbf{u}_{SS} \Rightarrow \mathbf{y}_r(k) \to \mathbf{r}$$

### Remark on the Target Calculator

- Generally it is not possible to control the state x(k) to an arbitrary target state  $x_{ss}$
- E.g. it is not possible to maintain a constant position and a constant velocity of a car simultaneously
- A sufficient condition for the existence of a solution of (7.4) for any reference input r is that

$$\begin{pmatrix} \textbf{\textit{I}}_{n\times n} - \textbf{\textit{A}} & -\textbf{\textit{B}} \\ \textbf{\textit{C}}_{r} & \textbf{\textit{0}}_{p_{r}\times m} \end{pmatrix} \in \mathbb{R}^{(n+p_{r})\times (n+m)} \text{ has full rank } n+p_{r}$$

- This implies that  $C_r$  must have full rank and number of controlled outputs  $p_r \leq$  number of inputs m
- The solution may not be unique



### **Reference Tracking based on Target Calculation**

- Remark on the Constrained Case
  - For the constrained case target pair  $(x_{ss}, u_{ss})$  must fulfill output constraint  $\mathbb{Y}$  and input constraint  $\mathbb{U}$
  - For this purpose the target calculator based on (7.4) must be modified to

$$\min_{\boldsymbol{x}_{SS}, \boldsymbol{u}_{SS}} \frac{1}{2} \left( (\boldsymbol{C}_{r} \boldsymbol{x}_{SS} - \boldsymbol{r})^{T} \boldsymbol{Q}_{SS} (\boldsymbol{C}_{r} \boldsymbol{x}_{SS} - \boldsymbol{r}) + (\boldsymbol{u}_{SS} - \boldsymbol{u}_{SS}^{unc})^{T} \boldsymbol{R}_{SS} (\boldsymbol{u}_{SS} - \boldsymbol{u}_{SS}^{unc}) \right)$$
subject to
$$\begin{cases}
 \begin{pmatrix} \boldsymbol{I}_{n \times n} - \boldsymbol{A} & -\boldsymbol{B} \\ \boldsymbol{C}_{r} & \boldsymbol{0}_{p_{r} \times m} \end{pmatrix} \begin{pmatrix} \boldsymbol{x}_{SS} \\ \boldsymbol{u}_{SS} \end{pmatrix} = \begin{pmatrix} \boldsymbol{0}_{n \times 1} \\ \boldsymbol{r} \end{pmatrix}$$

$$\boldsymbol{C} \boldsymbol{x}_{SS} \in \mathbb{Y}$$

$$\boldsymbol{u}_{SS} \in \mathbb{U}$$

$$(7.9)$$

where  $\boldsymbol{u}_{ss}^{unc}$  is the target input resulting from (7.4) for the unconstrained case and  $\boldsymbol{Q}_{ss} = \boldsymbol{Q}_{ss}^T \geq \mathbf{0}$  and  $\boldsymbol{R}_{ss} = \boldsymbol{R}_{ss}^T > \mathbf{0}$  are weighting matrices

- Problem (7.9) can be formulated as a quadratic program
- Feasibility of Problem (7.9) is discussed in [RM09, Section 1.5.1]



### **Reference Tracking based on Target Calculation**

### Remark on Receding Horizon Control

- 1. Estimate the current state  $\widehat{x}(k)$
- 2. Solve Problem (7.9) for the given reference input r(k) to determine the target pair  $(x_{ss}, u_{ss})$
- 3. Solve Problem (7.8) for  $\widetilde{x}(k) = \widehat{x}(k) x_{ss}$  to determine the optimal input sequence  $\widetilde{U}^*(k)$
- 4. Compute first element of the optimal input sequence  $\tilde{\boldsymbol{u}}^*(k) = (\boldsymbol{I}_{m \times m} \quad \boldsymbol{0}_{m \times m} \quad \cdots \quad \boldsymbol{0}_{m \times m}) \tilde{\boldsymbol{U}}^*(k)$
- 5. Implement the optimal input  $m{u}^*(k) = \widetilde{m{u}}^*(k) + m{u}_{\mathrm{SS}}$
- 6. Increment the time instant k := k + 1 and go to 1.

#### Further Remarks

- The extension for state constraints and controlled output constraints is straightforward
- The structure on Slide 7-3 is essentially equivalent to the state-command structure on Slide 2-49ff
- Disturbances and uncertainties in A, B and  $C_r$  lead to a steady-state error or offset
- More details and references are given in [RM09, Section 1.5.1], [BBM15, Section 13.6], and [MR93]



## Reference Tracking based on the Delta Input Formulation

Discrete-Time Linear Time-Invariant (LTI) System

$$x(k+1) = Ax(k) + Bu(k)$$

state equation 
$$(7.10)$$

$$y(k) = Cx(k)$$

Symbols

$$x(k) \in \mathbb{R}^n$$
 state vector

$$\boldsymbol{u}(k) \in \mathbb{U} \subseteq \mathbb{R}^m$$
 input vector

$$\mathbf{y}(k) \in \mathbb{Y} \subseteq \mathbb{R}^p$$
 output vector

$$A \in \mathbb{R}^{n \times n}$$
 system matrix

$$\mathbf{B} \in \mathbb{R}^{n \times m}$$
 input matrix

 $C \in \mathbb{R}^{p \times n}$  output matrix

### Objective

- Control the discrete-time LTI system such that  $y(k) \rightarrow r(k)$ 



## Reference Tracking based on the Delta Input Formulation

### Approach

Introduce the input increment

$$\Delta \boldsymbol{u}(k) = \boldsymbol{u}(k) - \boldsymbol{u}(k-1)$$

Introduce the discrete-time quadratic cost function

$$V_{N}(\mathbf{y}(k), \Delta \mathbf{U}(k), \mathbf{r}(k)) = (\mathbf{y}(k+N) - \mathbf{r}(k))^{T} \mathbf{P}(\mathbf{y}(k+N) - \mathbf{r}(k))$$

$$+ \sum_{i=0}^{N-1} (\mathbf{y}(k+i) - \mathbf{r}(k))^{T} \mathbf{Q}(\mathbf{y}(k+i) - \mathbf{r}(k)) + \Delta \mathbf{u}^{T}(k+i) \mathbf{R} \Delta \mathbf{u}(k+i)$$
(7.12)

with the reference input r(k) and the weighting matrices  $P = P^T > 0$ ,  $Q = Q^T > 0$ ,  $R = R^T > 0$ 

- Note that r(k) is held over the whole prediction horizon
- Note that y(k+i) r(k) is the control error



### Reference Tracking based on the Delta Input Formulation

#### Approach

- Reformulate Problem 4.1/5.1 w.r.t. the system (7.10)/(7.11), cost function (7.12) and  $\Delta u(k)$ , i.e.

$$\min_{\Delta U(k)} V_N(y(k), \Delta U(k), r(k))$$

$$\begin{cases} x(k+i+1) = Ax(k+i) + Bu(k+i), i = 0,1, ..., N-1 \\ y(k+i) = Cx(k+i), & i = 0,1, ..., N \end{cases}$$

subject to 
$$\begin{cases} y(k+i) = 0x(k+i), & i = 0,1,...,N \\ y(k+i) \in \mathbb{Y}, & i = 1,2,...,N \\ u(k+i) \in \mathbb{U}, & i = 0,1,...,N-1 \\ u(k+i) = u(k+i-1) + \Delta u(k+i), & i = 0,1,...,N-1 \end{cases}$$

- Problem (7.13) can be formulated as a quadratic program using the methods from Chapter 4 and 5
- Problem (7.13) relates to a regulation problem, i.e.  $y(k) r(k) \rightarrow 0$ ,  $\Delta u(k) \rightarrow 0$  in steady state
- Note that the reference tracking problem has been transformed into a regulation problem using y(k+i)-r(k) instead of x(k+i) and  $\Delta u(k)$  instead of u(k) (generally  $u(k) \nrightarrow 0$  in steady state)

(7.13)



### Reference Tracking based on the Delta Input Formulation

#### Approach

- The prediction model (4.4), the cost function (4.5), and the constraint model (5.1)
   must be reformulated to obtain a quadratic program
- For this purpose the discrete-time LTI system (7.10)/(7.11) is augmented w.r.t.  $\Delta u(k)$ , i.e.

$$\underbrace{\begin{pmatrix} \boldsymbol{x}(k+1) \\ \boldsymbol{u}(k) \end{pmatrix}}_{\boldsymbol{x}(k+1)} = \underbrace{\begin{pmatrix} \boldsymbol{A} & \boldsymbol{B} \\ \boldsymbol{0} & \boldsymbol{I} \end{pmatrix}}_{\boldsymbol{A}} \underbrace{\begin{pmatrix} \boldsymbol{x}(k) \\ \boldsymbol{u}(k-1) \end{pmatrix}}_{\boldsymbol{x}(k)} + \underbrace{\begin{pmatrix} \boldsymbol{B} \\ \boldsymbol{I} \end{pmatrix}}_{\boldsymbol{A}} \Delta \boldsymbol{u}(k) \tag{7.14}$$

$$y(k) = (C \quad 0) \underbrace{\begin{pmatrix} x(k) \\ u(k-1) \end{pmatrix}}_{\mathbf{y}(k)}$$

$$y(k) = \widetilde{C} \quad \widetilde{x}(k)$$
(7.15)

- The prediction model (4.4) can then be reformulated w.r.t. the augmented system (7.14)/(7.15) as

$$\widetilde{X}(k) = \widetilde{\Phi}\widetilde{x}(k) + \widetilde{\Gamma}\Delta U(k) \tag{7.16}$$

where  $\widetilde{X}(k)$ ,  $\widetilde{\Phi}$ ,  $\widetilde{\Gamma}$ ,  $\Delta U(k) = (\Delta u^T(k) \ \Delta u^T(k+1) \ \cdots \ \Delta u^T(k+N-1))^T$  are defined as in (4.4)





### Reference Tracking based on the Delta Input Formulation

#### Approach

- The cost function (4.5) can then be reformulated w.r.t. the augmented system (7.14)/(7.15), the prediction model (7.16), and the cost function (7.12) as



## Reference Tracking based on the Delta Input Formulation

### Approach

- The constraints in standard form (cf. Slide 5-10) can be reformulated w.r.t. Problem (7.13) as

$$My(k+i) + E\Delta u(k+i) \le b, i = 0,1,...,N-1$$
  
 $My(k+N) \le b$ 

- The constraint model (5.1) can then be reformulated w.r.t. the augmented system (7.14)/(7.15) and the prediction model (7.16) as

$$\mathcal{D}\widetilde{C}\widetilde{x}(k) + \mathcal{M}\widetilde{C}\left(\widetilde{\Phi}\widetilde{x}(k) + \widetilde{\Gamma}\Delta U(k)\right) + \mathcal{E}\Delta U(k) \leq \mathcal{E} \qquad \Leftrightarrow \\ \left(\mathcal{D}\widetilde{C} + \mathcal{M}\widetilde{C}\widetilde{\Phi}\right)\widetilde{x}(k) + \left(\mathcal{M}\widetilde{C}\widetilde{\Gamma} + \mathcal{E}\right)\Delta U(k) \qquad \leq \mathcal{E} \qquad \Leftrightarrow \\ \left(\mathcal{M}\widetilde{C}\widetilde{\Gamma} + \mathcal{E}\right)\Delta U(k) \qquad \leq \mathcal{E} + \left(-\mathcal{D}\widetilde{C} - \mathcal{M}\widetilde{C}\widetilde{\Phi}\right)\widetilde{x}(k) \Leftrightarrow \\ \widetilde{\mathcal{A}} \qquad \Delta U(k) \qquad \leq \mathcal{E} + \widetilde{\mathcal{W}} \qquad \widetilde{x}(k)$$

where  $\mathcal{D}$ ,  $\mathcal{M}$ ,  $\mathcal{E}$  and  $\mathcal{E}$  are defined as in (5.1)



### Reference Tracking based on the Delta Input Formulation

#### Approach

Problem 4.1 is then solved by the optimal state feedback control law

$$\frac{\partial}{\partial \Delta U(k)} V_N(\widetilde{\mathbf{x}}(k), \Delta U(k), \mathbf{R}(k)) = \widetilde{\mathbf{H}} \Delta U(k) + \widetilde{\mathbf{F}} \widetilde{\mathbf{x}}(k) + \widetilde{\mathbf{F}}_{\mathbf{R}} \mathbf{R}(k) = \mathbf{0} \Leftrightarrow \Delta U^*(k) = -\widetilde{\mathbf{H}}^{-1} \widetilde{\mathbf{F}} \widetilde{\mathbf{x}}(k) - \widetilde{\mathbf{H}}^{-1} \widetilde{\mathbf{F}}_{\mathbf{R}} \mathbf{R}(k)$$

Problem 5.1 can then be formulated as the quadratic program

$$\min_{\Delta \boldsymbol{U}(k)} \frac{1}{2} \Delta \boldsymbol{U}^T(k) \, \widetilde{\boldsymbol{H}} \Delta \boldsymbol{U}(k) + \Delta \boldsymbol{U}^T(k) \left( \widetilde{\boldsymbol{F}} \widetilde{\boldsymbol{x}}(k) + \widetilde{\boldsymbol{F}}_{\boldsymbol{R}} \boldsymbol{R}(k) \right) + f\left( \widetilde{\boldsymbol{x}}(k), \boldsymbol{R}(k) \right) \quad \text{Term is independent of } \Delta \boldsymbol{U}(k) \\ \text{Subject to } \widetilde{\boldsymbol{\mathcal{A}}} \Delta \boldsymbol{U}(k) \leq \boldsymbol{\mathcal{E}} + \widetilde{\boldsymbol{\mathcal{W}}} \widetilde{\boldsymbol{x}}(k) \quad \text{The reference input sequence } \boldsymbol{R}(k) \text{ occurs here!}$$

The receding horizon controller in the unconstrained case is given by

$$\mathbf{u}^{*}(k) = (\mathbf{I}_{m \times m} \ \mathbf{0}_{m \times m} \cdots \ \mathbf{0}_{m \times m}) \Delta \mathbf{U}^{*}(k) + \mathbf{u}^{*}(k-1)$$

$$= \underbrace{-(\mathbf{I}_{m \times m} \ \mathbf{0}_{m \times m} \cdots \ \mathbf{0}_{m \times m}) \widetilde{\mathbf{H}}^{-1} \widetilde{\mathbf{F}} \widetilde{\mathbf{x}}(k)}_{\mathbf{K}} \underbrace{-(\mathbf{I}_{m \times m} \ \mathbf{0}_{m \times m} \cdots \ \mathbf{0}_{m \times m}) \widetilde{\mathbf{H}}^{-1} \widetilde{\mathbf{F}}_{\mathbf{R}} \mathbf{R}(k) + \mathbf{u}^{*}(k-1)$$

$$= \underbrace{\widetilde{\mathbf{K}}_{RHC}} \widetilde{\mathbf{x}}(k) + \underbrace{\widetilde{\mathbf{K}}_{RRHC}} \mathbf{R}(k) + \mathbf{u}^{*}(k-1)$$

The receding horizon controller is an affine TI state feedback controller in the unconstrained case



### Reference Tracking based on the Delta Input Formulation

### Approach

- The feedback matrices  $\widetilde{K}_{RHC}$  and  $\widetilde{K}_{RRHC}$  can be calculated offline in the unconstrained case
- The closed-loop system is globally asymptotically stable iff  $\rho(\widetilde{A} + \widetilde{B}\widetilde{K}_{RHC}) < 1$  (cf. Theorem 2.3)
- The receding horizon controller in the constrained case results as

$$\boldsymbol{u}^*(k) = (\boldsymbol{I}_{m \times m} \quad \boldsymbol{0}_{m \times m} \quad \cdots \quad \boldsymbol{0}_{m \times m}) \Delta \boldsymbol{U}^*(k) + \boldsymbol{u}^*(k-1)$$

The receding horizon controller is a nonlinear state feedback controller in the constrained case

#### Remarks

- The extension for state constraints is straightforward
- Constraints on  $m{u}(k+i-1)$  and thus on  $m{u}(k+i)$  can be written as  $(m{0}_{m imes n} \quad m{I}_{m imes m}) \widetilde{m{x}}(k+i) \in \mathbb{U}$
- Disturbances and uncertainties in A, B and C do not lead to a steady-state error or offset
- More details and references are given in [MacO2] and [BBM15, Section 13.6]



## **Preview Control for Reference Tracking**

#### Motivation

- Sometimes the reference input r(k+i) is known over the prediction horizon
- E.g. in motion control problems the reference sequence is usually precomputed
- The reference input r(k+i) can then be included in Problem (7.13) and previewed in this way
- The receding horizon controller can then work proactively

#### Approach

- Problem (7.13) with reference r(k+i) can be solved in "batch" way using quadratic programming
- Assume that the reference input r(k+i) is known for i=0,1,...,N
- The reference input r(k) can then be replaced by the reference input r(k+i) in cost function (7.12)
- The reference input sequence on Slide 7-13ff is then  $\mathbf{R}(k) = (\mathbf{r}^T(k+1) \ \mathbf{r}^T(k+2) \ \cdots \ \mathbf{r}^T(k+N))^T$
- The formulation as a quadratic program then follows analogously to Slide 7-12ff



## Reference Tracking based on Target Calculation with Disturbance Estimation

• Discrete-Time Linear Time-Invariant (LTI) System

$$x(k+1) = Ax(k) + Bu(k) + B_w w(k)$$
 state equation (7.17)

$$y(k) = Cx(k) + C_w w(k)$$
 measured output equation (7.18)

$$\mathbf{y}_{r}(k) = \mathbf{C}_{r}\mathbf{x}(k) + \mathbf{C}_{rw}\mathbf{w}(k)$$
 controlled output equation (7.19)

Symbols

$$x(k) \in \mathbb{R}^n$$
 state vector  $u(k) \in \mathbb{U} \subseteq \mathbb{R}^m$  input vector

$$\mathbf{w}(k) \in \mathbb{R}^{m_{\mathbf{w}}}$$
 disturbance vector

$$y(k) \in \mathbb{Y} \subseteq \mathbb{R}^p$$
 measured output vector

$$A \in \mathbb{R}^{n \times n}$$
 system matrix

$$\boldsymbol{B}_{\boldsymbol{w}} \in \mathbb{R}^{n \times m_{\boldsymbol{w}}}$$
 disturbance input matrix

$$\mathbf{C} \in \mathbb{R}^{p \times n}$$
 measured output matrix

$$\boldsymbol{C}_{\boldsymbol{w}} \in \mathbb{R}^{p \times m_{\boldsymbol{w}}}$$

$$y_r(k) \in \mathbb{R}^{p_r}$$
 controlled output vector

$$\mathbf{B} \in \mathbb{R}^{n \times m}$$
 input matrix

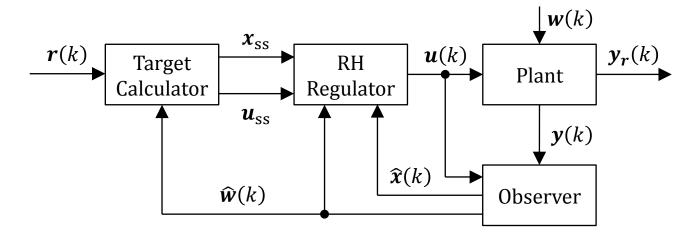
$$C_r \in \mathbb{R}^{p_r \times n}$$
 controlled output matrix

$$C_{rw} \in \mathbb{R}^{p_r \times m_w}$$



### Reference Tracking based on Target Calculation with Disturbance Estimation

#### Structure



### Objective

- Control the discrete-time LTI system such that  $y_r(k) \to r(k)$  if r(k),  $w(k) \to \text{const.}$  as  $k \to \infty$ 

### Approach

- The observer estimates the current state  $\widehat{x}(k)$  and the disturbance  $\widehat{w}(k)$
- The target calculator and RH regulator are essentially utilized as outlined on Slide 7-3



### Reference Tracking based on Target Calculation with Disturbance Estimation

#### Approach

Introduce the disturbance model

$$\mathbf{w}(k+1) = \mathbf{w}(k)$$

- Augment the discrete-time LTI system (7.17)/(7.18) by the disturbance model, i.e.

$$\begin{pmatrix} \mathbf{x}(k+1) \\ \mathbf{w}(k+1) \end{pmatrix} = \begin{pmatrix} \mathbf{A} & \mathbf{B}_{\mathbf{w}} \\ \mathbf{0} & \mathbf{I} \end{pmatrix} \begin{pmatrix} \mathbf{x}(k) \\ \mathbf{w}(k) \end{pmatrix} + \begin{pmatrix} \mathbf{B} \\ \mathbf{0} \end{pmatrix} \mathbf{u}(k)$$
 (7.20)

$$\mathbf{y}(k) = (\mathbf{C} \quad \mathbf{C}_{\mathbf{w}}) \begin{pmatrix} \mathbf{x}(k) \\ \mathbf{w}(k) \end{pmatrix} \tag{7.21}$$

- Design an observer for the augmented discrete-time LTI system (7.20)/(7.21) based on the methods developed in Chapter 2



### Reference Tracking based on Target Calculation with Disturbance Estimation

#### Approach

- Consider that the discrete-time LTI system (7.17)/(7.19) is in the steady target state  $x_{ss}$ , i.e.

$$x_{SS} = Ax_{SS} + Bu_{SS} + B_{w}\widehat{w} \iff (I_{n \times n} - A)x_{SS} - Bu_{SS} = B_{w}\widehat{w}$$

$$y_{rSS} = C_{r}x_{SS} + C_{rw}\widehat{w} = r \iff C_{r}x_{SS} = r - C_{rw}\widehat{w}$$

- Note that the estimated disturbance  $\hat{w}$  is used since the steady-state disturbance  $w_{ss}$  is unknown
- Rewrite the equations in matrix form, i.e.

$$\begin{pmatrix} \mathbf{I}_{n \times n} - \mathbf{A} & -\mathbf{B} \\ \mathbf{C}_{r} & \mathbf{0}_{p_{r} \times m} \end{pmatrix} \begin{pmatrix} \mathbf{x}_{SS} \\ \mathbf{u}_{SS} \end{pmatrix} = \begin{pmatrix} \mathbf{B}_{w} \widehat{\mathbf{w}} \\ \mathbf{r} - \mathbf{C}_{rw} \widehat{\mathbf{w}} \end{pmatrix}$$
(7.22)

- The steady target pair  $(x_{ss}, u_{ss})$  can be then calculated from (7.22) provided that a solution exists
- Note that the estimated disturbance  $\hat{w}$  is regarded in the target pair  $(x_{\rm ss}, u_{\rm ss})$
- Introduce the state deviation and input deviation

$$\widetilde{\boldsymbol{x}}(k) = \widehat{\boldsymbol{x}}(k) - \boldsymbol{x}_{SS}, \quad \widetilde{\boldsymbol{u}}(k) = \boldsymbol{u}(k) - \boldsymbol{u}_{SS}$$



### Reference Tracking based on Target Calculation with Disturbance Estimation

#### Approach

This leads to the discrete-time LTI state equation

$$\widetilde{\boldsymbol{x}}(k+1) = \widehat{\boldsymbol{x}}(k+1) - \boldsymbol{x}_{ss} = A\widehat{\boldsymbol{x}}(k) + \boldsymbol{B}\boldsymbol{u}(k) + \boldsymbol{B}\boldsymbol{w}\widehat{\boldsymbol{w}}(k) - (\boldsymbol{A}\boldsymbol{x}_{ss} + \boldsymbol{B}\boldsymbol{u}_{ss} + \boldsymbol{B}\boldsymbol{w}\boldsymbol{w}_{ss}) = A\widetilde{\boldsymbol{x}}(k) + \boldsymbol{B}\widetilde{\boldsymbol{u}}(k) \quad (7.23)$$

the discrete-time quadratic cost function

$$\widetilde{V}_{N}\left(\widetilde{\boldsymbol{x}}(k),\widetilde{\boldsymbol{U}}(k)\right) = \widetilde{\boldsymbol{x}}^{T}(k+N)\boldsymbol{P}\widetilde{\boldsymbol{x}}(k+N) + \sum_{i=0}^{N-1}\widetilde{\boldsymbol{x}}^{T}(k+i)\boldsymbol{Q}\widetilde{\boldsymbol{x}}(k+i) + \widetilde{\boldsymbol{u}}^{T}(k+i)\boldsymbol{R}\widetilde{\boldsymbol{u}}(k+i) \quad (7.24)$$

the state and input constraints

$$C(\widetilde{\boldsymbol{x}}(k+i) + \boldsymbol{x}_{ss}) + \boldsymbol{C}_{\boldsymbol{w}}\widehat{\boldsymbol{w}}(k) \in \mathbb{Y}, i = 1, 2, ..., N$$

$$\widetilde{\boldsymbol{u}}(k+i) + \boldsymbol{u}_{ss} \in \mathbb{U}, \qquad i = 0, 1, ..., N - 1$$

$$(7.25)$$

- Note that  $\widehat{w}(k) = w_{ss}$  is assumed in (7.23) as detailed in [PR03] and [RM09, Section 1.5.1]
- Note that the deviations are resubstituted in (7.25) since the constraint sets Y and U are considered



### Reference Tracking based on Target Calculation with Disturbance Estimation

#### Approach

- Rewrite Problem 4.1/5.1 w.r.t. the state equation (7.23), cost function (7.24) and constr. (7.25), i.e.  $\min_{\widetilde{\boldsymbol{U}}(k)} \tilde{V}_N(\widetilde{\boldsymbol{x}}(k), \widetilde{\boldsymbol{U}}(k))$ 

subject to 
$$\begin{cases} \widetilde{\boldsymbol{x}}(k+i+1) = \boldsymbol{A}\widetilde{\boldsymbol{x}}(k+i) + \boldsymbol{B}\widetilde{\boldsymbol{u}}(k+i), i = 0,1,...,N-1 \\ \boldsymbol{C}(\widetilde{\boldsymbol{x}}(k+i) + \boldsymbol{x}_{ss}) + \boldsymbol{C}_{\boldsymbol{w}}\widehat{\boldsymbol{w}}(k) \in \mathbb{Y}, & i = 1,2,...,N \\ \widetilde{\boldsymbol{u}}(k+i) + \boldsymbol{u}_{ss} \in \mathbb{U}, & i = 0,1,...,N-1 \end{cases}$$
(7.26)

#### · Remark on the Detectability

- The augmented discrete-time LTI system (7.20)/(7.21) is detectable iff
  - (1) (C, A) is detectable

(2) 
$$\begin{pmatrix} I_{n \times n} - A & -B_w \\ C & C_w \end{pmatrix} \in \mathbb{R}^{(n+p) \times (n+m_w)}$$
 has full rank  $n+m_w$ 

- This implies that number of disturbances  $m_w \leq$  number of measured outputs p is required for detectability of the disturbance. Note that  $B_w$  and  $C_w$  can always be chosen such that (2) is fulfilled.



### Reference Tracking based on Target Calculation with Disturbance Estimation

- Remark on the Constrained Case
  - For the constrained case target pair  $(x_{ss}, u_{ss})$  must fulfill output constraint  $\mathbb{Y}$  and input constraint  $\mathbb{U}$
  - For this purpose the target calculator based on (7.22) must be modified to

$$\min_{\boldsymbol{x}_{SS}, \boldsymbol{u}_{SS}} \frac{1}{2} \left( (\boldsymbol{C}_{r} \boldsymbol{x}_{SS} + \boldsymbol{C}_{rw} \widehat{\boldsymbol{w}} - \boldsymbol{r})^{T} \boldsymbol{Q}_{SS} (\boldsymbol{C}_{r} \boldsymbol{x}_{SS} + \boldsymbol{C}_{rw} \widehat{\boldsymbol{w}} - \boldsymbol{r}) + (\boldsymbol{u}_{SS} - \boldsymbol{u}_{SS}^{unc})^{T} \boldsymbol{R}_{SS} (\boldsymbol{u}_{SS} - \boldsymbol{u}_{SS}^{unc}) \right)$$
subject to
$$\begin{cases}
 \begin{pmatrix} \boldsymbol{I}_{n \times n} - \boldsymbol{A} & -\boldsymbol{B} \\ \boldsymbol{C}_{r} & \boldsymbol{0}_{p_{r} \times m} \end{pmatrix} \begin{pmatrix} \boldsymbol{x}_{SS} \\ \boldsymbol{u}_{SS} \end{pmatrix} = \begin{pmatrix} \boldsymbol{B}_{w} \widehat{\boldsymbol{w}} \\ \boldsymbol{r} - \boldsymbol{C}_{rw} \widehat{\boldsymbol{w}} \end{pmatrix} \\
 \begin{pmatrix} \boldsymbol{C}_{r} & \boldsymbol{v}_{SS} + \boldsymbol{C}_{w} \widehat{\boldsymbol{w}} \in \mathbb{Y} \\ \boldsymbol{u}_{SS} \in \mathbb{U} \end{pmatrix}$$
(7.27)

where  $u_{ss}^{unc}$  is the target input resulting from (7.22) for the unconstrained case and  $Q_{ss} = Q_{ss}^T \ge 0$  and  $R_{ss} = R_{ss}^T > 0$  are weighting matrices

Problem (7.27) can be formulated as a quadratic program



### Reference Tracking based on Target Calculation with Disturbance Estimation

### • Remark on Receding Horizon Control

- 1. Estimate the current state  $\widehat{x}(k)$  and the disturbance  $\widehat{w}(k)$
- 2. Solve Problem (7.27) for the given r(k) and estimated  $\hat{w}(k)$  to determine the target pair  $(x_{ss}, u_{ss})$
- 3. Solve Problem (7.26) for  $\tilde{x}(k) = \hat{x}(k) x_{ss}$  to determine the optimal input sequence  $\tilde{U}^*(k)$
- 4. Compute first element of the optimal input sequence  $\tilde{\boldsymbol{u}}^*(k) = (\boldsymbol{I}_{m \times m} \quad \boldsymbol{0}_{m \times m} \quad \cdots \quad \boldsymbol{0}_{m \times m}) \tilde{\boldsymbol{U}}^*(k)$
- 5. Implement the optimal input  ${m u}^*(k) = \widetilde{m u}^*(k) + {m u}_{\rm ss}$
- 6. Increment the time instant k = k + 1 and go to 1.

#### Further Remarks

- The remarks on the optimization problem and on the target calculator given on Slide 7-6 analogously apply to reference tracking with disturbance estimation
- Note that reference tracking with disturbance estimation is also denoted as offset-free control
- More details and references are given in [RM09, Section 1.5.1], [BBM15, Section 13.6], and [PR03]



### **Preview Control for Disturbance Rejection**

#### Motivation

- Consider the discrete-time LTI state equation (7.17) including the disturbance w(k)
- Sometimes the disturbance w(k+i) can be predicted or measured over the prediction horizon
- E.g. the renewable generation in a power system can be predicted with good accuracy
- E.g. the road displacement before a car can be measured with a camera sensor
- The disturbance w(k + i) can then be included in Problem 4.1/5.1 and previewed in this way
- The receding horizon controller can then work proactively

### Approach

- Problem 4.1/5.1 with disturbance w(k+i) can be solved in "batch" way using quadratic programming
- The prediction model (4.4), the cost function (4.5), and the constraint model (5.1) must be reformulated w.r.t. the disturbance w(k+i) to this end
- Assume that the disturbance w(k+i) can be predicted or measured for i=0,1,...,N-1



## **Preview Control for Disturbance Rejection**

#### Approach

- The solution of the discrete-time LTI state equation (7.17) is then given by

$$x(k+1) = Ax(k) + Bu(k) + B_{w}w(k)$$

$$x(k+2) = Ax(k+1) + Bu(k+1) + B_{w}w(k+1)$$

$$= A^{2}x(k) + ABu(k) + AB_{w}w(k) + Bu(k+1) + B_{w}w(k+1)$$

$$\vdots$$

$$x(k+N) = A^{N}x(k) + A^{N-1}Bu(k) + A^{N-1}B_{w}w(k) + \dots + ABu(k+N-2) + AB_{w}w(k+N-2)$$

$$+ Bu(k+N-1) + B_{w}w(k+N-1)$$

The prediction model is then given by

$$\begin{pmatrix}
x(k+1) \\
x(k+2) \\
\vdots \\
x(k+N)
\end{pmatrix} = \begin{pmatrix}
A \\
A^{2} \\
\vdots \\
A^{N}
\end{pmatrix} x(k) + \begin{pmatrix}
B & 0 & \cdots & 0 \\
AB & B & \cdots & 0 \\
\vdots & \vdots & \ddots & \vdots \\
A^{N-1}B & A^{N-2}B & \cdots & B
\end{pmatrix} \begin{pmatrix}
u(k) \\
u(k+1) \\
\vdots \\
u(k+N-1)
\end{pmatrix} + \begin{pmatrix}
B_{w} & 0 & \cdots & 0 \\
AB_{w} & B_{w} & \cdots & 0 \\
\vdots & \vdots & \ddots & \vdots \\
A^{N-1}B_{w} & A^{N-2}B_{w} & \cdots & B_{w}
\end{pmatrix} \begin{pmatrix}
w(k) \\
w(k+1) \\
\vdots \\
w(k+N-1)
\end{pmatrix}$$

$$X(k) = \Phi x(k) + \Gamma \qquad U(k) + \Gamma_{w} \qquad W(k) (7.28)$$



### **Preview Control for Disturbance Rejection**

- Approach
  - Substituting the prediction model (7.28) into the cost function (4.5) leads to

$$\begin{split} V_{N}\big(x(k), \boldsymbol{U}(k), \boldsymbol{W}(k)\big) &= \boldsymbol{x}^{T}(k)\boldsymbol{Q}\boldsymbol{x}(k) + \boldsymbol{X}^{T}(k)\boldsymbol{\Omega}\boldsymbol{X}(k) + \boldsymbol{U}^{T}(k)\boldsymbol{\Psi}\boldsymbol{U}(k) \\ &= \boldsymbol{x}^{T}(k)\boldsymbol{Q}\boldsymbol{x}(k) + \left(\boldsymbol{\Phi}\boldsymbol{x}(k) + \boldsymbol{\Gamma}\boldsymbol{U}(k) + \boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k)\right)^{T}\boldsymbol{\Omega}\big(\boldsymbol{\Phi}\boldsymbol{x}(k) + \boldsymbol{\Gamma}\boldsymbol{U}(k) + \boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k)\big) + \boldsymbol{U}^{T}(k)\boldsymbol{\Psi}\boldsymbol{U}(k) \\ &= \boldsymbol{x}^{T}(k)\boldsymbol{Q}\boldsymbol{x}(k) + \boldsymbol{x}^{T}(k)\boldsymbol{\Phi}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) + \boldsymbol{x}^{T}(k)\boldsymbol{\Phi}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}\boldsymbol{U}(k) + \boldsymbol{x}^{T}(k)\boldsymbol{\Phi}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) \\ &+ \boldsymbol{U}^{T}(k)\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) + \boldsymbol{U}^{T}(k)\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}\boldsymbol{U}(k) + \boldsymbol{U}^{T}(k)\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) \\ &+ \boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) + \boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}\boldsymbol{U}(k) + \boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) + \boldsymbol{U}^{T}(k)\boldsymbol{\Psi}\boldsymbol{U}(k) \\ &= \boldsymbol{U}^{T}(k)(\boldsymbol{\Psi} + \boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma})\boldsymbol{U}(k) + 2\boldsymbol{U}^{T}(k)\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) + 2\boldsymbol{U}^{T}(k)\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) \\ &+ \boldsymbol{x}^{T}(k)(\boldsymbol{Q} + \boldsymbol{\Phi}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi})\boldsymbol{x}(k) + \boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) + 2\boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) \\ &= \frac{1}{2}\boldsymbol{U}^{T}(k)2(\boldsymbol{\Psi} + \boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma})\boldsymbol{U}(k) + \boldsymbol{U}^{T}(k)(2\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) + 2\boldsymbol{\Gamma}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k)) \\ &+ \boldsymbol{x}^{T}(k)(\boldsymbol{Q} + \boldsymbol{\Phi}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi})\boldsymbol{x}(k) + \boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Gamma}_{\boldsymbol{w}}\boldsymbol{W}(k) + 2\boldsymbol{W}^{T}(k)\boldsymbol{\Gamma}_{\boldsymbol{w}}^{T}\boldsymbol{\Omega}\boldsymbol{\Phi}\boldsymbol{x}(k) \\ &= \frac{1}{2}\boldsymbol{U}^{T}(k) \boldsymbol{H} \boldsymbol{W}(k) \boldsymbol{H} \boldsymbol{W}^{T}(k)\boldsymbol{H} \boldsymbol{W}^$$

### **Preview Control for Disturbance Rejection**

#### Approach

Substituting the prediction model (7.28) into the constraint model (5.1) leads to

$$\mathcal{D}(k)x(k) + \mathcal{M}(k) \left( \Phi x(k) + \Gamma U(k) + \Gamma_{w} W(k) \right) + \mathcal{E}(k)U(k) \leq \mathcal{B}(k) \iff (\mathcal{D}(k) + \mathcal{M}(k)\Phi)x(k) + \left( \mathcal{M}(k)\Gamma + \mathcal{E}(k) \right) U(k) + \mathcal{M}(k)\Gamma_{w} W(k) \leq \mathcal{B}(k) \iff (\mathcal{M}(k)\Gamma + \mathcal{E}(k))U(k) \leq \mathcal{B}(k) + \left( -\mathcal{D}(k) - \mathcal{M}(k)\Phi \right) x(k) + \left( -\mathcal{M}(k)\Gamma_{w} \right) W(k) \iff \mathcal{A}(k) \quad U(k) \leq \mathcal{B}(k) + \mathcal{W}(k) \quad x(k) + \mathcal{W}_{w}(k) \quad W(k)$$

Problem 4.1 is then solved by the optimal state feedback control law

$$\frac{\partial}{\partial U(k)}V_N(x(k),U(k),W(k)) = HU(k) + Fx(k) + F_wW(k) = 0 \Leftrightarrow U^*(k) = -H^{-1}Fx(k) - H^{-1}F_wW(k)$$

Problem 5.1 can then be formulated as the quadratic program Term is independent of  $\boldsymbol{U}(k)$  $\min_{\boldsymbol{U}(k)} \frac{1}{2} \boldsymbol{U}^{T}(k) \boldsymbol{H} \boldsymbol{U}(k) + \boldsymbol{U}^{T}(k) \left( \boldsymbol{F} \boldsymbol{x}(k) + \boldsymbol{F}_{\boldsymbol{w}} \boldsymbol{W}(k) \right) + f \left( \boldsymbol{x}(k), \boldsymbol{W}(k) \right)$ Term is therefore not relevant!

subject to 
$$\mathcal{A}(k)\mathbf{U}(k) \leq \mathcal{B}(k) + \mathcal{W}(k)\mathbf{X}(k) + \mathcal{W}_{\mathbf{W}}(k)\mathbf{W}(k)$$
 The disturbance sequence  $\mathbf{W}(k)$  occurs here!



### **Preview Control for Disturbance Rejection**

#### Approach

The receding horizon controller in the unconstrained case is given by

$$\mathbf{u}^{*}(k) = (\mathbf{I}_{m \times m} \quad \mathbf{0}_{m \times m} \quad \cdots \quad \mathbf{0}_{m \times m}) \mathbf{U}^{*}(k)$$

$$= \underbrace{-(\mathbf{I}_{m \times m} \quad \mathbf{0}_{m \times m} \quad \cdots \quad \mathbf{0}_{m \times m}) \mathbf{H}^{-1} \mathbf{F} \mathbf{x}(k)}_{\mathbf{K}_{RHC}} \qquad \underbrace{-(\mathbf{I}_{m \times m} \quad \mathbf{0}_{m \times m} \quad \cdots \quad \mathbf{0}_{m \times m}) \mathbf{H}^{-1} \mathbf{F}_{\mathbf{w}} \mathbf{W}(k)}_{\mathbf{W}(k)}$$

- The receding horizon controller is an affine TI state feedback controller in the unconstrained case
- The feedback matrices  $K_{RHC}$  and  $K_{wRHC}$  can be calculated offline in the unconstrained case
- The closed-loop system is globally asymptotically stable iff  $\rho(A + BK_{RHC}) < 1$  (cf. Theorem 2.3)
- The receding horizon controller in the constrained case results as

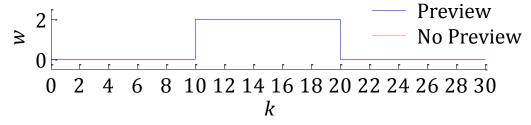
$$\boldsymbol{u}^*(k) = (\boldsymbol{I}_{m \times m} \quad \boldsymbol{0}_{m \times m} \quad \cdots \quad \boldsymbol{0}_{m \times m}) \boldsymbol{U}^*(k)$$

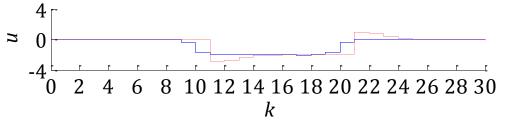
The receding horizon controller is an nonlinear state feedback controller in the constrained case

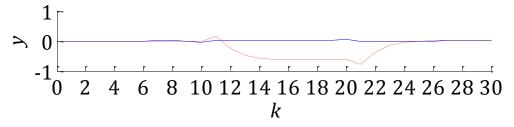


### **Preview Control for Disturbance Rejection**

#### • Illustrative Example







### **Example from Chapter 4**

$$\mathbf{x}(0) = (0 \quad 0)^T$$

$$y(k) = (-1 \quad 1)x(k)$$

No state and input constraints

Disturbance input matrix  $\boldsymbol{B}_{\boldsymbol{w}} = \boldsymbol{B}$ 

Input weight R = 0.01

Terminal weight  $m{P} = m{P}_{
m LQR}$ 

RHC (prediction horizon N = 5)

Preview RHC works proactively

Very good performance



## Literature

### Miscellaneous

[MR93] Kenneth R. Muske and James B. Rawlings. Model predictive control with linear models. *AIChE Journal*, 39(2):262–287, 1993.

[PR03] Gabriele Pannocchia and James B. Rawlings. Disturbance models for offset-free model predictive control. *AIChE Journal*, 49(2):426–437, 2003.