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EE 3900 - Assignment 1

VIBHAVASU

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Abstract—This manual provides a simple introduction to digital signal processing.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3-scipy python3-numpy python3matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://raw.
githubusercontent
.com/gadepall/
EE1310/master/
filter/codes/
Sound_Noise.
wav

- 2.2 You will find a spectrogram at https://academo.org/demos/spectrum-analyzer . Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find?
 - **Solution:** There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

2.4 The python output of the script Problem 2.3 is the audio file in Sound With ReducedNoise.wav. Plav the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 3.2.

wget https://github. com/gadepall/ EE1310/raw/ master/filter/ codes/xnyn.py

3.3 Repeat the above exercise using a C code. **Solution:** : C Code

Fig. 3.2

wget https://github. com/ DarkWake9/ EE3900/blob/ main/ Assignment %201/e3-3.c

Solution: : Python Code

wget https://github. com/ DarkWake9/ EE3900/blob/ main/ Assignment %201/e3-3.py

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)

resulting in (4.2). Similarly, it can be shown that

$$Z\{x(n-k)\} = z^{-k}X(z)$$
 (4.6)

4.2 Obtain X(z) for x(n) defined in problem 3.1.

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.7}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.8)

$$\implies H(z) = \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (4.9)

4.4 a) Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.10)

b) and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.11)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.12}$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.13}$$

and from (4.11),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.14)

$$=\frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.15}$$

using the fomula for the sum of an infinite geometric progression.

(i) Solution:

$$\Delta(z) = \mathcal{Z}\{\delta[n]\} = \sum_{n=-\infty}^{\infty} \delta(n)z^{-n} = 1 \quad (4.16)$$

(ii) Solution:

$$U(z) = \mathcal{Z}\{\delta(n)\} = \sum_{n=-\infty}^{\infty} u[n]z^{-n}$$
 (4.17)

$$= 1 + z^{-1} + z^{-2} + \dots = \frac{1}{1 - z^{-1}}$$
 (4.18)

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.19}$$

Solution:

$$a^{n}u[n] = \begin{cases} a^{n} & n \ge 0 \\ 0 & n < 0 \end{cases}$$
 (4.20)

$$U'(z) = \mathcal{Z}\{a^{n}u[n]\} = \sum_{n=-\infty}^{\infty} a^{n}u[n]z^{-n}$$
 (4.21)

$$= 1 + az^{-1} + a^2z^{-2} + \cdots$$
 (4.22)

Given: |z| > |a|

$$U'(z) = \frac{1}{1 - az^{-1}} \tag{4.23}$$

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.24)

Plot $|H(e^{j\omega})|$. Is it periodic? If so, find the period. $H(e^{j\omega})$ is known as the Discret Time Fourier Transform (DTFT) of h(n).

Solution: The following code plots Fig. 4.6

wget https://raw. githubusercontent .com/gadepall/ EE1310/master/ filter/codes/dtft. py

Fig. 4.6: $|H(e^{J\omega})|$

Solution: From (4.9) we get

$$H(e^{j\omega}) = \frac{1 + e^{-2j\omega}}{1 + \frac{1}{2}e^{-j\omega}}$$
(4.25)

$$= \frac{e^{j\omega} + e^{-j\omega}}{e^{j\omega} + \frac{1}{2}} = \frac{2\cos(\omega)}{e^{j\omega} + \frac{1}{2}}$$
(4.26)

$$\implies |H(e^{j\omega})| \qquad (4.27)$$

(4.28)

$$= \frac{1}{2\cos\omega} \frac{|H(e^{j\omega})|}{\sqrt{\left(\cos\omega + \frac{1}{2}\right)^2 + \sin^2\omega}}$$

 $=\frac{2\cos\omega}{\sqrt{\cos^2\omega+\sin^2\omega+\cos\omega+\frac{1}{4}}}$ (4.29)

$$=\frac{2\cos\omega}{\sqrt{\frac{5}{4}+\cos\omega}}\tag{4.30}$$

For a periodic function of period T,

$$f(x) = f(x+T), T \neq 0$$
 (4.31)

Checking if π is a period,

$$\frac{2\cos(\omega+\pi)}{\sqrt{\frac{5}{4}+\cos(\omega+\pi)}}$$
 (4.32)

$$=\frac{-2\cos\omega}{\sqrt{\frac{5}{4}-\cos\omega}} \quad (4.33)$$

$$\implies H(e^{J(\omega+\pi)}) \neq H(e^{J\omega}) \quad (4.34)$$

Checking if 2π is a period

$$\frac{2\cos(\omega + 2\pi)}{\sqrt{\frac{5}{4} + \cos(\omega + 2\pi)}} = \frac{2\cos\omega}{\sqrt{\frac{5}{4} + \cos\omega}}$$
 (4.35)

$$\implies H\left(e^{J(\omega+2\pi)}\right) = H\left(e^{J\omega}\right) \quad (4.36)$$

 \therefore Period of $H(e^{j\omega})$ is 2π

4.7 Express h(n) in terms of $H(e^{j\omega})$

Solution:

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h(n')e^{-j\omega n'} \qquad (4.37)$$

$$\implies \int_{\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.38)$$

$$=\sum_{n'=-\infty}^{\infty}\int_{-\pi}^{\pi}h(n')e^{-j\omega n'}e^{j\omega n}d\omega \qquad (4.39)$$

$$= \sum_{n'=-\infty}^{\infty} h(n') 2\pi \delta(n'-n) = 2\pi h(n)$$
 (4.40)

$$\therefore h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega} d\omega \qquad (4.41)$$

5 IMPULSE RESPONSE

5.1 Using long division, find

$$h(n), \quad n < 5 \tag{5.1}$$

for H(z) in (4.9). **Solution:** from (4.9)

$$H(z) = (1 + z^{-2}) \left(1 + \frac{1}{2} z^{-1} \right)^{-1}$$
 (5.2)

For ROC:
$$\left| \frac{1}{2} z^{-1} \right| < 1$$
 (5.3)

$$\implies |z| > \frac{1}{2}$$
 is the ROC for this case (5.4)

(5.12)

$$\frac{1 - \frac{1}{2}z^{-1} + \frac{5}{4}z^{-2} - \frac{5}{8}z^{-3} + \frac{5}{16}z^{-4}}{1 + z^{-2}} \dots \dots \\
1 + \frac{1}{2}z^{-1}) \qquad 1 + z^{-2} \qquad \qquad \frac{1 + \frac{1}{2}z^{-1}}{-\frac{z^{-1}}{2} + z^{-2}} \\
-\frac{z^{-1}}{2} - \frac{1}{4}z^{-2} \\
-\frac{\frac{5}{4}z^{-2}}{-\frac{5}{4}z^{-2}} \\
-\frac{5}{8}z^{-3} \\
-\frac{5}{8}z^{-3} - \frac{5}{16}z^{-4} \\
-\frac{\frac{5}{16}z^{-4} + \frac{5}{32}z^{-5}}{16z^{-4} + \frac{5}{32}z^{-5}} \\
\vdots$$

$$\implies H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} = 1 - \frac{1}{2}z^{-1} + \sum_{n=2}^{\infty} \frac{5}{4}z^{-n}$$

(5.5)

We know that
$$H(z) = \sum_{n=-\infty}^{\infty} h(n)z^{-n}$$
 (5.6)

comparing coefficients: in the ROC $|z| > \frac{1}{2}$

$$h(n) = \begin{cases} 0 & \text{if } n < 0 \\ 1, & \text{if } n = 0 \\ -\frac{1}{2}, & \text{if } n = 1 \\ 5\left(-\frac{1}{2}\right)^{n}, & \text{if } n \ge 2 \end{cases}$$

$$h(0) = 1, \quad h(1) = -\frac{1}{2}, \quad h(2) = \frac{5}{4}$$

$$(5.8)$$

(5.9)

 $h(3) = -\frac{5}{8}, \quad h(4) = \frac{5}{16}$

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z)$$
 (5.10)

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.9),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$

$$\therefore h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$

using (4.19) and (4.6).

5.3 Sketch h(n). Is it bounded? Justify theoretically.

Solution: The following code plots Fig. 5.3.

Fig. 5.3: h(n) as the inverse of H(z)

5.4 Convergent? Justify using the ratio test. **Solution:**

$$h(n) = \begin{cases} 0 & \text{if} & n < 0 \\ 1, & \text{if} & n = 0 \\ -\frac{1}{2}, & \text{if} & n = 1 \\ 5\left(-\frac{1}{2}\right)^{n}, & \text{if} & n \ge 2 \end{cases}$$
 (5.13)

Using ratio test:

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \left| \frac{5\left(-\frac{1}{2}\right)^{n+1}}{5\left(-\frac{1}{2}\right)^n} \right| = \frac{1}{2} < \infty \quad (5.14)$$

 $\implies h(n)$ is convergent

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.15}$$

Is the system defined by (3.2) stable for the impulse response in (5.10)

Solution:

$$\sum_{n=-\infty}^{\infty} h(n) = 0 + 1 - \frac{1}{2} + 5 \sum_{n=2}^{\infty} \left(-\frac{1}{2} \right)^n$$
 (5.16)
$$= \frac{1}{2} + 5 \left(1 - \frac{1}{2} - \left(\frac{1}{1 + \frac{1}{2}} \right) \right)$$
 (5.17)
$$= \frac{1}{2} + \frac{5}{6} = \frac{8}{6} = 1.333 < \infty$$
 (5.18)

 $\therefore h(n)$ is Stable

5.6 Verify the above result using a python code. **Solution:** The Following code computes and proves the aboves result

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.19)$$

This is the definition of h(n).

Solution: The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

Fig. 5.7: h(n) from the definition

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{n = -\infty}^{\infty} x(k)h(n - k)$$
 (5.20)

Comment. The operation in (5.20) is known as *convolution*.

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

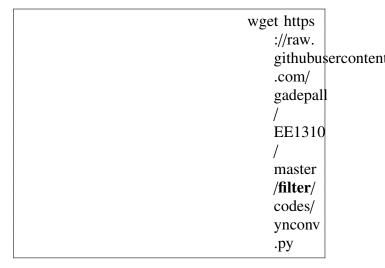


Fig. 5.8: y(n) from the definition of convolution

://raw.
githubusercontent
.com/
gadepall
/
EE1310 5.9 Express the above convolution using a Teoplitz
/
matrix.
master
Solution: From (3.1) $x(n) = \{1, 2, 3, 4, 2, 1\}$ /filter/
From (5.20) y(n) = x(n) * h(n)

wget https

$$\begin{pmatrix} 1\\ 1.5\\ 3.25\\ 4.375\\ 2.8125\\ 3.59375\\ 0.203125\\ 0.8984375\\ -0.44921875\\ 0.224609375\\ -0.112304688\\ 0.0561523438\\ -0.0280761719\\ 0.0140380859\\ -7.01904297\times10^{-3}\\ 3.50952148\times10^{-3}\\ -1.75476074\times10^{-3}\\ 8.77380371\times10^{-4}\\ -4.38690186\times10^{-4}\\ 0 \end{pmatrix}$$
 (5.23)

5.10 Show that

$$y(n) = \sum_{n=-\infty}^{\infty} x(n-k)h(k)$$
 (5.24)

Solution:

$$X(z) = \mathbb{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (5.25)

$$H(z) = \mathcal{Z}{h(m)} = \sum_{m=-\infty}^{\infty} h(m)z^{-m}$$
 (5.26)

$$Y(z) = \mathcal{Z}\{h(m)\} = \sum_{k=-\infty}^{\infty} y(m)z^{-k}$$
 (5.27)

$$X(z)H(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \sum_{m=-\infty}^{\infty} h(m)z^{-m} \quad (5.28)$$

$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} x[n]h[m]z^{-(n+m)}$$
 (5.29)

Let m = k - n

$$= \sum_{k=-\infty}^{\infty} \left(\sum_{n=-\infty}^{\infty} x[n]h[n-k] \right) z^{-k}$$
 (5.30)

$$= \sum_{k=-\infty}^{\infty} y[n]z^{-k} = Y(z)$$
 (5.31)

$$\implies Y(z) = X(z) \cdot H(z)0$$
 (5.32)

now put n + m = k $n = -\infty$

$$\Rightarrow Y(z) = \sum_{k=-\infty}^{\infty} x(m-k) \sum_{m=-\infty}^{\infty} h(m) z^{-k} \quad (5.33)$$

$$= \sum_{k=-\infty}^{\infty} \left(\sum_{m=-\infty}^{\infty} x[m-k]h[k] \right) z^{-k} \quad (5.34)$$

but
$$Y(z) = \sum_{k=-\infty}^{\infty} y(m)z^{-k}$$
 (5.35)

$$\Rightarrow y(m) = \sum_{m=-\infty}^{\infty} x[m-k]h(k) \quad (5.36)$$

$$\implies y(n) = \sum_{n=-\infty}^{\infty} x(n-k)h(k) \quad (5.37)$$

6 DFT and FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

6.2 Compute

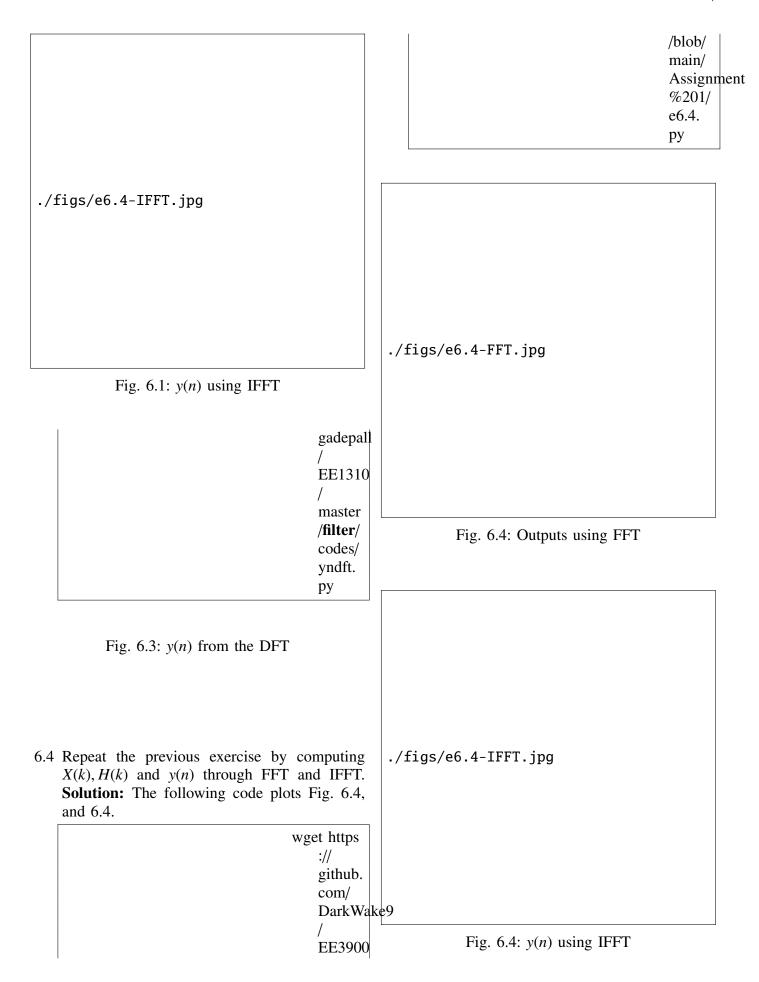
$$Y(k) = X(k)H(k) \tag{6.2}$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https
://raw.
githubusercontent
.com/



a) Note that the figure 6.4 is the same as y(n)in Fig. 3.2

7 FFT

7.1 Definitions

1. The DFT of x(n) is given by

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(7.1)

2. Let

$$W_N = e^{-j2\pi/N} \tag{7.2}$$

Then the N-point DFT matrix is defined as

$$\vec{F}_N = [W_N^{mn}] \tag{7.3}$$

where W_N^{mn} are the elements of \vec{F}_N .

3. Let

$$\vec{I}_4 = (\vec{e}_4^1 \quad \vec{e}_4^2 \quad \vec{e}_4^3 \quad \vec{e}_4^4) \tag{7.4}$$

be the 4×4 identity matrix. Then the 4 point DFT permutation matrix is defined as

$$\vec{P}_4 = (\vec{e}_4^1 \quad \vec{e}_4^3 \quad \vec{e}_4^2 \quad \vec{e}_4^4) \tag{7.5}$$

4. The 4 point *DFT diagonal matrix* is defined as

$$\vec{D}_4 = diag \left(W_N^0 \quad W_N^1 \quad W_N^2 \quad W_N^3 \right) \tag{7.6}$$

7.2 Problems

1. Show that

$$W_N^2 = W_{N/2} (7.7)$$

$$W_N = e^{-j2\pi/N} (7.8)$$

$$\implies W_{N/2} = e^{-j2\pi/(N/2)}$$
 (7.9)

$$\implies W_{N/2} = e^{-j2\pi/(N/2)}$$
 (7.9)
$$W_N^2 = e^{2(-j2\pi/N)} = e^{-j2\pi/(N/2)} = W_{N/2}$$
 (7.11)

(7.11)

2. Find \vec{P}_6 .

Solution:

$$\vec{P}_6 = (\vec{e}_6^1 \quad \vec{e}_6^3 \quad \vec{e}_6^5 \quad \vec{e}_6^2 \quad \vec{e}_6^4 \quad \vec{e}_6^4) \tag{7.12}$$

3. Find \vec{D}_3 .

Solution:

$$\vec{D}_3 = diag(W_3^0 \ W_3^1 \ W_3^2) \tag{7.13}$$

4. Show that

$$\vec{F}_6 = \begin{bmatrix} \vec{I}_3 & \vec{D}_3 \\ \vec{I}_3 & -\vec{D}_3 \end{bmatrix} \begin{bmatrix} \vec{F}_3 & 0 \\ 0 & \vec{F}_3 \end{bmatrix} \vec{P}_3 \tag{7.14}$$

5. Show that

$$\vec{F}_{N} = \begin{bmatrix} \vec{I}_{N/2} & \vec{D}_{N/2} \\ \vec{I}_{N/2} & -\vec{D}_{N/2} \end{bmatrix} \begin{bmatrix} \vec{F}_{N/2} & 0 \\ 0 & \vec{F}_{N/2} \end{bmatrix} \vec{P}_{N} \quad (7.15)$$

Solution:

$$\vec{F}_{N} = \begin{bmatrix} \vec{F}_{N/2} & \vec{D}_{N/2} \vec{F}_{N/2} \\ \vec{F}_{N/2} & -\vec{D}_{N/2} \vec{F}_{N/2} \end{bmatrix}$$
(7.16)

6. Find

$$\vec{P}_6 \vec{x} \tag{7.17}$$

Solution:

$$\vec{P}_6 = (\vec{e}_6^1 \quad \vec{e}_6^3 \quad \vec{e}_6^5 \quad \vec{e}_6^2 \quad \vec{e}_6^4 \quad \vec{e}_6^6) \qquad (7.18)$$

From(3.1):
$$\vec{x} = \{1, 2, 3, 4, 2, 1\}$$
 (7.19)

$$\vec{P}_6 \vec{x} = (1, 3, 2, 2, 4, 1) \tag{7.20}$$

7. Show that

$$\vec{X} = \vec{F}_N \vec{x} \tag{7.21}$$

where \vec{x}, \vec{X} are the vector representations of x(n), X(k) respectively.

8. Let

$$\begin{pmatrix} X_1(0) \\ X_1(1) \\ X_1(2) \end{pmatrix} = F_3 \begin{pmatrix} x(0) \\ x(2) \\ x(4) \end{pmatrix}$$
 (7.22)

$$\begin{pmatrix} X_2(0) \\ X_2(1) \\ X_2(2) \end{pmatrix} = F_3 \begin{pmatrix} x(1) \\ x(3) \\ x(5) \end{pmatrix}$$
(7.23)

Show that

$$\begin{pmatrix} X(0) \\ X(1) \\ X(2) \end{pmatrix} = \begin{pmatrix} X_1(0) \\ X_1(1) \\ X_1(2) \end{pmatrix} + \begin{pmatrix} W_6^0 & 0 & 0 \\ 0 & W_6^1 & 0 \\ 0 & 0 & W_6^2 \end{pmatrix} \begin{pmatrix} X_2(0) \\ X_2(1) \\ X_2(2) \end{pmatrix}$$
 (7.24)
$$\begin{pmatrix} X(3) \\ X(4) \\ X(5) \end{pmatrix} = \begin{pmatrix} X_1(0) \\ X_1(1) \\ X_1(2) \end{pmatrix} - \begin{pmatrix} W_6^0 & 0 & 0 \\ 0 & W_6^1 & 0 \\ 0 & 0 & W_6^2 \end{pmatrix} \begin{pmatrix} X_2(0) \\ X_2(1) \\ X_2(2) \end{pmatrix}$$
 (7.25)

9. For

$$\vec{x} = \begin{pmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \end{pmatrix} \tag{7.26}$$

compte the DFT using (7.21)

- 10. Repeat the above exercise using (7.25)
- 11. Write a C program to compute the 8-point FFT.

8 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

8.1 The command

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m) y(n-m) = \sum_{k=0}^{N} b(k) x(n-k) \quad (8.1)$$

where the input signal is x(n) and the output signal is y(n) with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

- 8.2 Repeat all the exercises in the previous sections for the above a and b.
- 8.3 What is the sampling frequency of the input signal?

Solution: Sampling frequency(fs)=44.1kHZ.

8.4 What is type, order and cutoff-frequency of the above butterworth filter

Solution: The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

8.5 Modifying the code with different input parameters and to get the best possible output.