



INSTITUTO POLITÉCNICO NACIONAL

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*Escuela Superior de Cómputo
(ESCOM)*

TEORÍA DE COMUNICACIONES Y SEÑALES.

GUÍA DEL 2^{do} PARCIAL

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3CV6

02 – JULIO – 2020

Sección I. Convolución Continua.

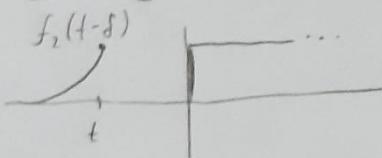
Problema 1. Calcular las siguientes integrales de convolución:

a) $u(t) * e^{-t} u(t) \Rightarrow f_1$

Solución

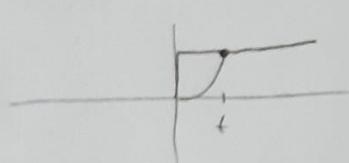
$$z(t) = \int_{-\infty}^{\infty} f_1(s) \cdot f_2(t-s) ds$$

Caso 0



$$\int_{-\infty}^{\infty} f_1(s) \cdot f_2(t-s) ds = 0$$

Caso 1



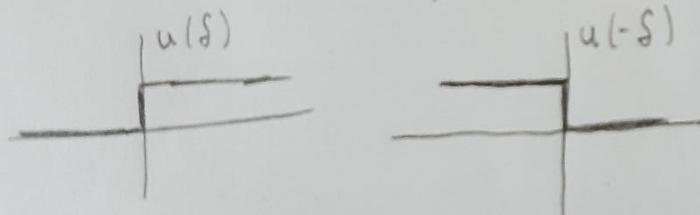
$$\int_{0}^{t} e^{s-t} ds = \frac{1}{2t} e^{2t} - \frac{1}{t} \cdot e^t = \frac{e^t}{2t} [e^t - t]$$

$\boxed{t \geq 0}$

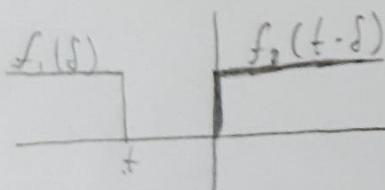
b) $u(t) * u(t)$

Solución

$$f_1(s) = u(s) \Rightarrow f_2(-s) = u(-s)$$

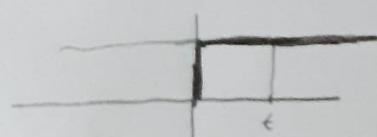


Caso 0: $t < 0$



$$z(t) = \int_{-\infty}^{\infty} f_1(s) \cdot f_2(t-s) ds = 0$$

Caso 1: $t \geq 0$



$$z(t) = \int_{-\infty}^{\infty} f_1(s) \cdot f_2(t-s) ds$$

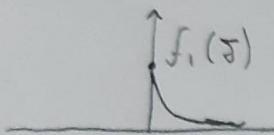
$$z(t) = \int_0^t 1 ds = t \quad \text{if } t \neq 0$$

$$c) e^{t u(t)} * e^{-3t} u(t)$$

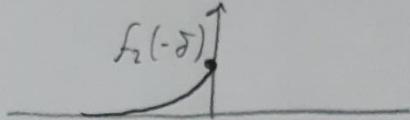
Solución

$$f_1(t) = e^{-t} \cdot u(t) \quad f_2(t) = e^{-3t} u(t) \quad \Rightarrow \quad \int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta$$

$$F_1(\sigma) = e^{-\sigma} \cdot u(\sigma)$$

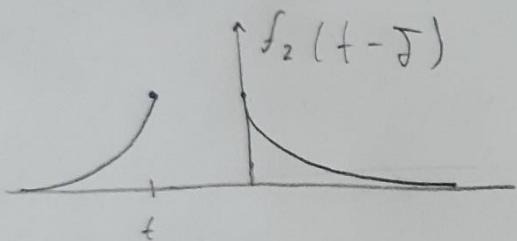


$$f_2(-\sigma) = e^{3\sigma} \cdot u(-\sigma)$$



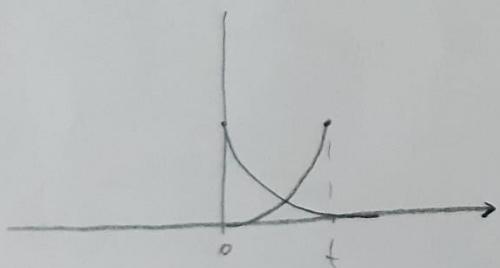
$$f_2(t-\tau) = e^{3(t-\tau)} \cdot u(t-\tau)$$

Caso 0:



$$\int_{-\infty}^{\infty} f_1(t) \cdot f_2(t - \tau) d\tau = 0$$

Caso 1:



$$\int_0^t f_1(t) \cdot f_2(t-\tau) d\tau = \int_0^t e^{3(t-\tau)} \cdot e^{-2} d\tau$$

$$-\int_0^t e^{3t-4s} ds = \frac{1}{3t-4t} \cdot e^{-\frac{3t-4t}{3t}} = e^{3t}$$

$$= \frac{1}{-t} e^{-t} - \frac{1}{3t} \cdot e^{3t} \Rightarrow -\frac{1}{3t} [3e^{-t} + e^{3t}]$$

~~$t \geq 0$~~

$$j) u(t) * f_u(t)$$

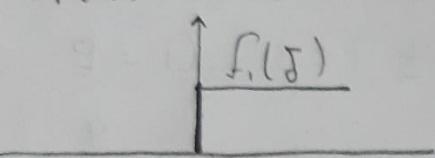
solución

$$f_1(t) = u(t)$$

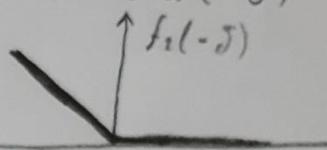
$$f_2(t) = t u(t)$$

$$z(t) = \int_{-\infty}^{\infty} f_1(\tau) \cdot f_2(t - \tau) d\tau$$

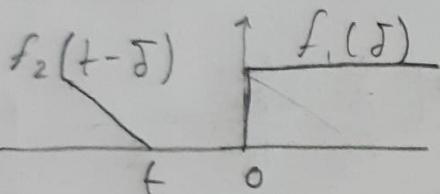
$$f_1(\tau) = u(\tau)$$



$$f_2(-\tau) = -\tau u(-\tau)$$

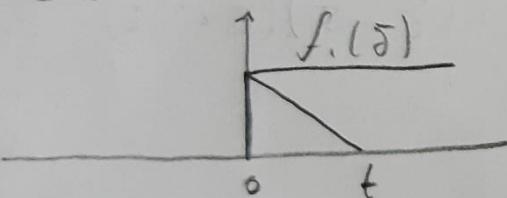


Caso 0:



$$\int_{-\infty}^{\infty} f_1(\tau) \cdot f_2(t - \tau) d\tau = 0$$

Caso 1:



$$\int_0^t f_1(\tau) \cdot f_2(t - \tau) d\tau$$

$$= \int_0^t u(\tau) \cdot (t - \tau) \cdot u(t - \tau) d\tau = \int_0^t (t - \tau) d\tau$$

$$= \left(t\tau - \frac{\tau^2}{2} \right) \Big|_0^t = t^2 - \frac{t^2}{2} \Rightarrow \underline{\frac{t^2}{2}} \quad t \geq 0$$

$$e) e^t u(t) * f u(t)$$

$$f_1(t) = e^{-t} u(t)$$

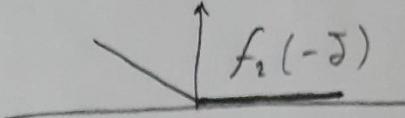
$$f_2(t) = t \cdot u(t)$$

$$f_1(\delta) = e^{-\delta} u(\delta)$$

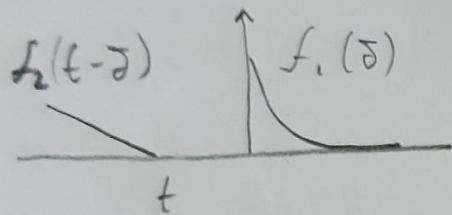


$$z = \int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta$$

$$f_2(-\delta) = -\delta u(-\delta)$$



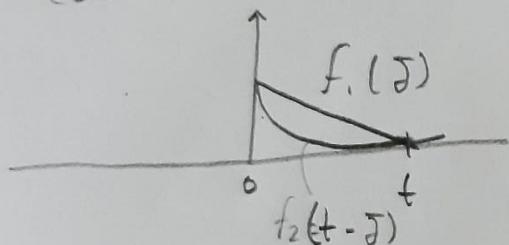
Caso 0:



$$\int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta = 0$$

$t < 0$

Caso 1:



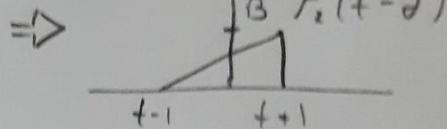
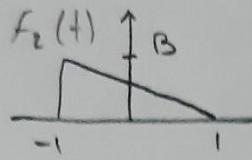
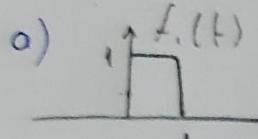
$$\begin{aligned} &= \int_0^t f_1(\delta) \cdot f_2(t-\delta) d\delta \\ &= \int_0^t e^{-\delta} u(\delta) \cdot (t-\delta) \cdot u(t-\delta) d\delta \end{aligned}$$

$$= - \int_0^t e^{-\delta} (\delta - t) d\delta = \left[\frac{1}{2} (\delta - t)^2 \cdot e^{-\delta} - \frac{1}{(-1)^2} e^{-\delta} \right]_0^t$$

$$= \left[-(\delta - t) \cdot e^{-\delta} - e^{-\delta} \right]_0^t = -e^{-t} - (t - 1) \cdot$$

$$= -e^{-t} - t + 1 \quad t \geq 0$$

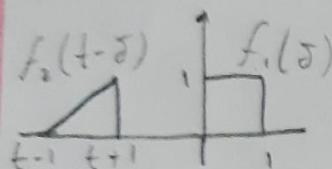
Problema 2: Calcular $f_1(t) * f_2(t)$ para cada par de señales de la figura.



$$f_1(\delta) = \begin{cases} \frac{B}{2}(\delta-1) \\ -1 \leq \delta \leq 1 \end{cases}$$

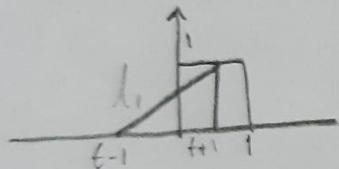
$$f_2(\delta) = \begin{cases} \frac{B}{2}(-\delta-1) \\ -1 \leq \delta \leq 1 \end{cases} \Rightarrow f_2(t-\delta) = \begin{cases} \frac{B}{2}[\delta+1-t] \\ t+1 \leq \delta \leq -1 \end{cases}$$

Caso 0:



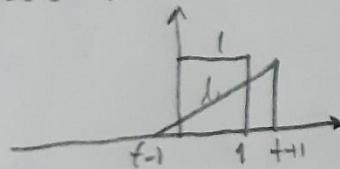
$$\int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta = 0$$

Caso 1:



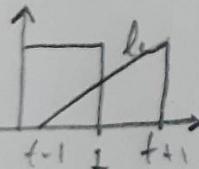
$$\int_0^{t+1} 1 \cdot l_1 d\delta \quad \begin{matrix} t+1 > 0 \\ t+1 \leq 1 \end{matrix} \Rightarrow -1 < t \leq 0$$

Caso 2:



$$\int_0^{t+1} 1 \cdot l_1 d\delta \quad \begin{matrix} t+1 > 1 \\ t-1 \leq 0 \end{matrix} \Rightarrow 0 < t \leq 1$$

Caso 3



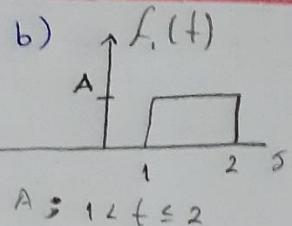
$$\int_{t-1}^t 1 \cdot l_1 d\delta \Rightarrow \begin{matrix} t+1 > 1 \\ t-1 \leq 1 \end{matrix} \quad \underline{0 < t \leq 2}$$

Entonces:

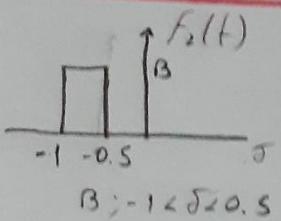
$$\underbrace{-\frac{B}{2}(\delta-1)d\delta}_{-1 < t \leq 0} +$$

$$\underbrace{\int_0^t -\frac{B}{2}(\delta-1)d\delta}_{0 < t \leq 1} +$$

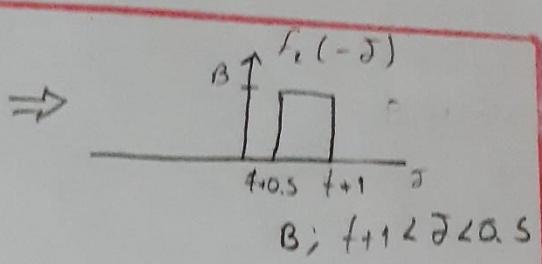
$$\underbrace{\int_{t-1}^t -\frac{B}{2}(\delta-1)d\delta}_{0 < t \leq 2}$$



$$A; 1 < t \leq 2$$

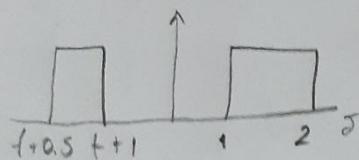


$$B; -1 < t \leq 0.5$$



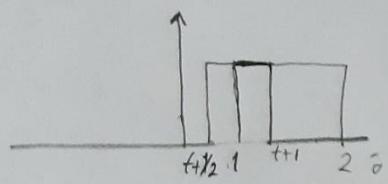
$$B; t+1/2 \leq t \leq 0.5$$

Caso 0:



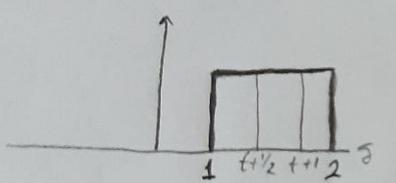
$$\int_{-\infty}^{\infty} f_1(\tau) \cdot f_2(t-\tau) d\tau = 0$$

Caso 1:



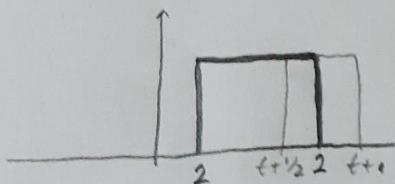
$$\int_1^{t+1} A \cdot B d\tau \Rightarrow t+1 > 1 \\ t+1/2 \leq 1 \Rightarrow 0 < t \leq 1$$

Caso 2:



$$\int_1^{t+1/2} A \cdot B d\tau \quad t+0.5 > 1 \\ t+1 \leq 2 \Rightarrow 1/2 < t \leq 1$$

Caso 3:

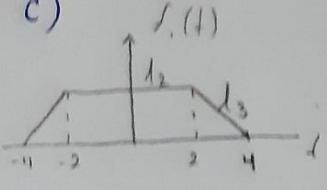


$$\int_{t+1/2}^2 A \cdot B d\tau \quad t+1 > 2 \\ t+1/2 \leq 2 \Rightarrow 1 < t \leq 1.5$$

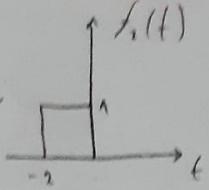
Entonces

$$f_1(t) \cdot f_2(t) = \underbrace{\int_0^{1/2} A \cdot B d\tau}_{0 < t < 1/2} + \underbrace{\int_{1/2}^1 A \cdot B d\tau}_{1/2 < t \leq 1} + \underbrace{\int_{1/2}^2 A \cdot B d\tau}_{1 < t \leq 1.5}$$

c) $f_1(t)$

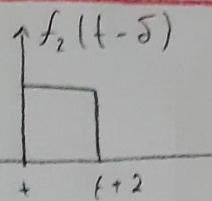


$f_1(t)$

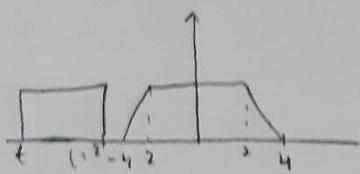


\Rightarrow

$f_2(t-2)$

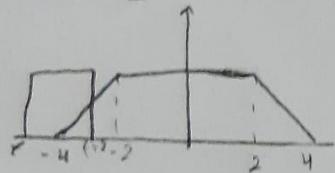


Caso 0:



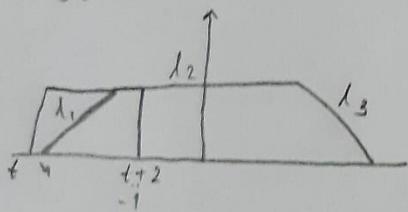
$$\int_{-\infty}^{\infty} f_1(t) f_2(t-2) dt = 0$$

Caso 1



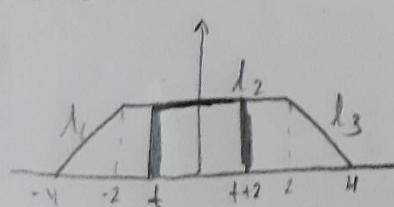
$$\int_{-4}^{t+2} 1 \cdot 1 dt \text{ en } \begin{cases} t+2 > -4 \\ t+2 \leq -2 \end{cases} \Rightarrow -6 < t \leq -4 \cancel{x}$$

Caso 2:



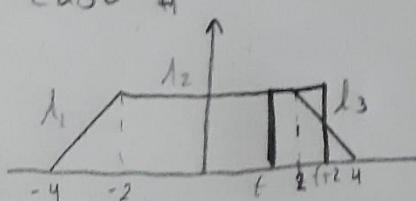
$$\int_{-4}^{-1} 1 \cdot 1 dt + \int_{-1}^{t+2} 1 dt \text{ en } \begin{cases} t+2 > -2 \\ t \leq -2 \end{cases} \Rightarrow -4 < t \leq -2 \cancel{x}$$

Caso 3:



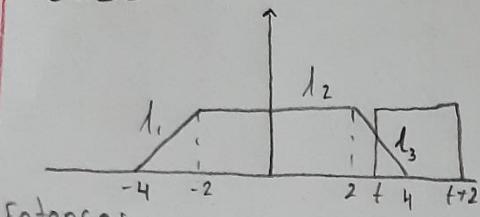
$$\int_t^{t+2} 1 \cdot 1 dt \text{ en } \begin{cases} t > -2 \\ t+2 \leq 2 \end{cases} \Rightarrow -2 < t \leq 0 \cancel{x}$$

Caso 4



$$\int_{-4}^{-2} 1 \cdot 1 dt + \int_2^{t+2} 1 dt \text{ en } \begin{cases} t+2 > 2 \\ t \leq 2 \end{cases} \Rightarrow 0 < t \leq 2 \cancel{x}$$

Caso 5:



$$\int_{t}^{4} 1 \lambda_3 d\delta$$

$$t+2 > 4 \\ t \leq 4 \Rightarrow 2 < t \leq 4$$

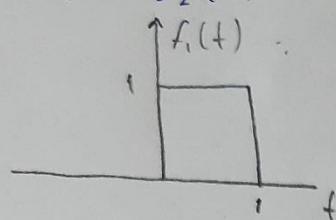
Entonces:

$$\int_{-4}^{t+2} 1 \lambda_1 d\delta + \underbrace{\int_{-4}^{-1} (1) \lambda_1 d\delta}_{-6 < t \leq -4} + \underbrace{\int_{-1}^{t+2} 1 \lambda_2 d\delta}_{-4 < t \leq 2} + \underbrace{\int_{t}^{t+2} (1) \lambda_2 d\delta}_{-2 < t \leq 0} +$$

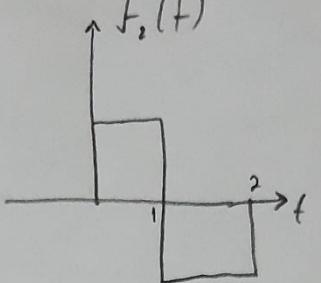
$$+ \underbrace{\int_{t}^{2} 1 \lambda_2 d\delta}_{0 < t \leq 2} + \underbrace{\int_{2}^{t+2} 1 \lambda_3 d\delta}_{2 < t \leq 4}$$

Problema 3: Evalúe las funciones de convolución para las señales mostradas

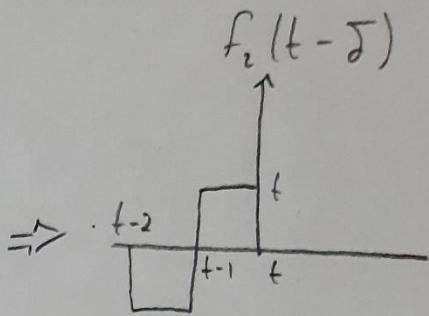
a) $f_1(t) * f_2(t)$



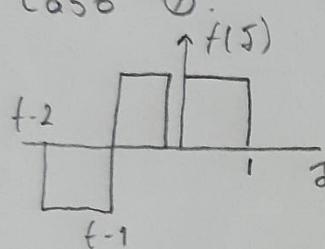
$$f_1(t)$$



$$f_2(t-\delta)$$

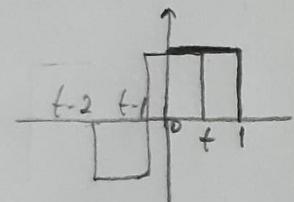


Caso 0:



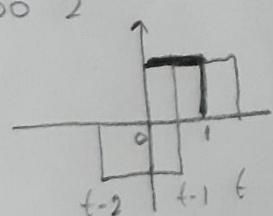
$$\int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta = 0$$

Caso 1:



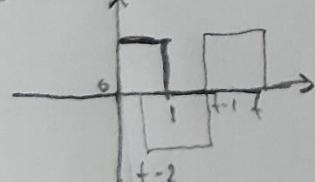
$$\int_{-2}^t (-1)(1) d\delta \quad \begin{cases} t > 0 \\ t \leq 1 \end{cases} \Rightarrow 0 < t \leq 1$$

Caso 2



$$\int_0^{t-1} (1)(-1) d\delta + \int_{t-1}^1 (1)(1) d\delta \quad \begin{cases} t > 1 \\ t-1 \leq 1 \end{cases} \Rightarrow 1 < t \leq 2$$

Caso 3:



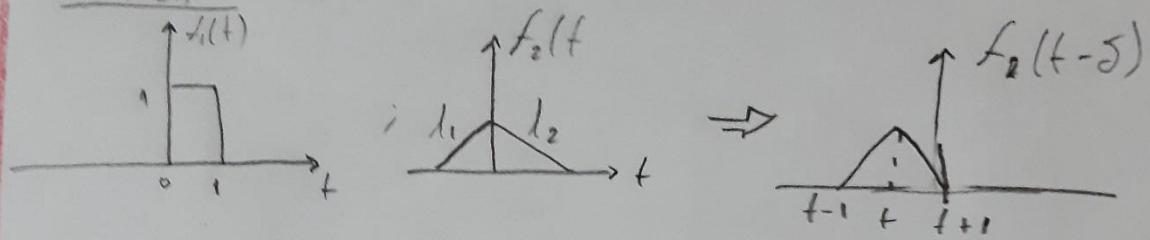
$$\int_{-2}^{t-1} (1)(-1) d\delta \quad \begin{cases} t > 1 \\ -2 \leq 1 \end{cases} \Rightarrow 2 < t \leq 3$$

Entonces:

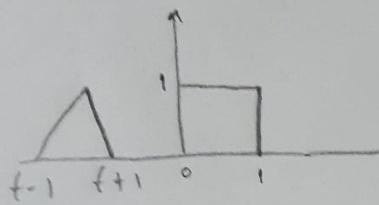
$$f_1(t) * f_2(t) = \underbrace{\int_0^t (-1)(1) d\delta}_{0 < t \leq 1} + \underbrace{\int_0^{t-1} (1)(-1) d\delta + \int_{t-1}^1 (1)(1) d\delta}_{1 < t \leq 2} + \underbrace{\int_{-2}^{t-1} (1)(-1) d\delta}_{2 < t \leq 3}$$

b) $f_1(t) * f_3(t)$

Solución

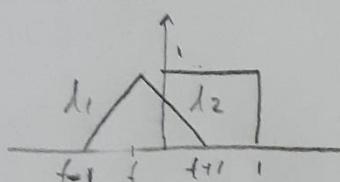


Caso 0:



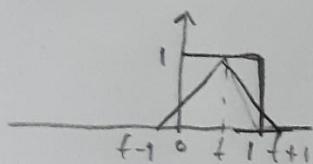
$$\int_{-\infty}^{\infty} f_1(\delta) \cdot f_2(t-\delta) d\delta = 0$$

Caso 1.



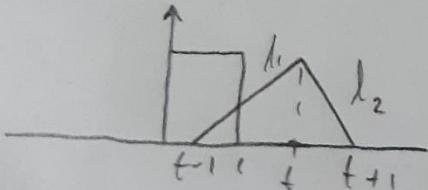
$$\int_0^{t+1} f_1(\delta) f_2(\delta) d\delta \quad \begin{cases} t+1 > 0 \\ t \leq 0 \end{cases} \Rightarrow -1 < t \leq 0$$

Caso 2



$$\int_0^t f_1(\delta) f_2(\delta) d\delta + \int_t^1 f_1(\delta) f_2(\delta) d\delta \quad \begin{cases} t > 0 \\ t \leq 1 \end{cases} \Rightarrow 0 < t \leq 1$$

Caso 3

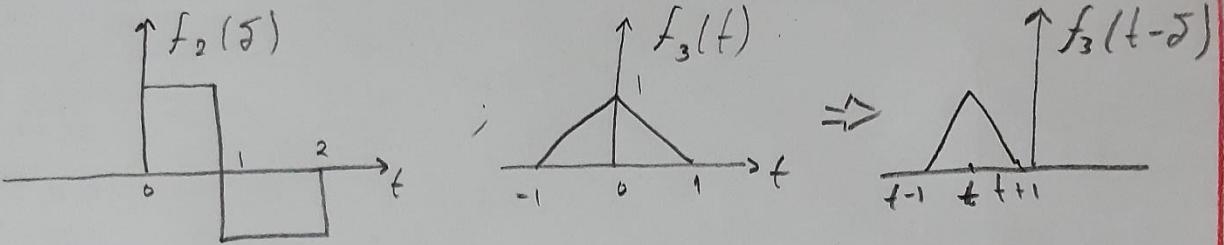


$$\int_{t-1}^t f_1(\delta) f_2(\delta) d\delta \quad \begin{cases} t > 1 \\ t \leq 1 \end{cases} \Rightarrow 1 < t \leq 2$$

Entonces:

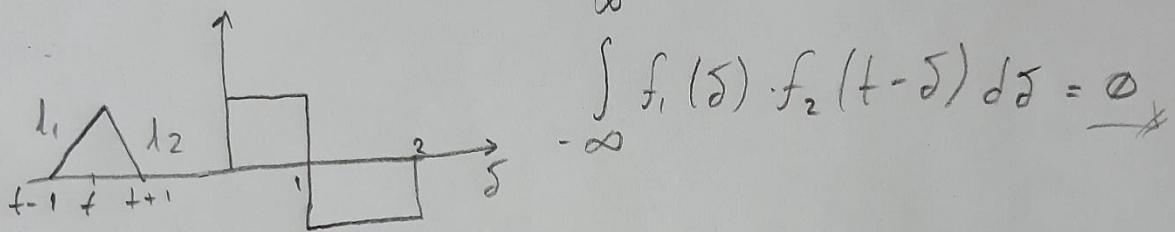
$$f_1(t) * f_3(t) = \underbrace{\int_0^{t+1} f_2(\delta) d\delta}_{-1 < t \leq 0} + \underbrace{\int_0^t f_2(\delta) d\delta}_{0 < t \leq 1} + \underbrace{\int_{t-1}^1 f_2(\delta) d\delta}_{1 < t \leq 2}$$

c) $f_2(t) * f_3(t)$

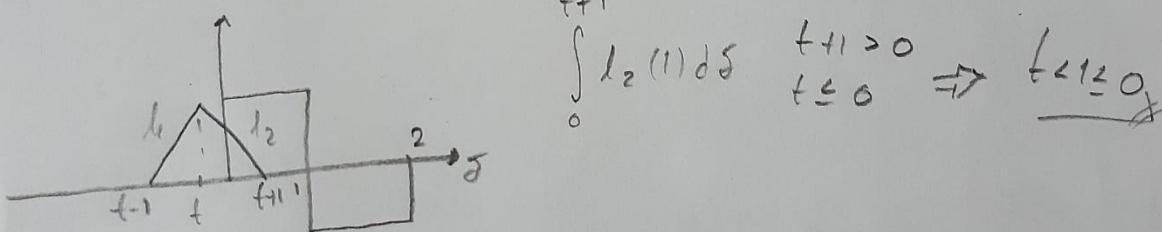


solución

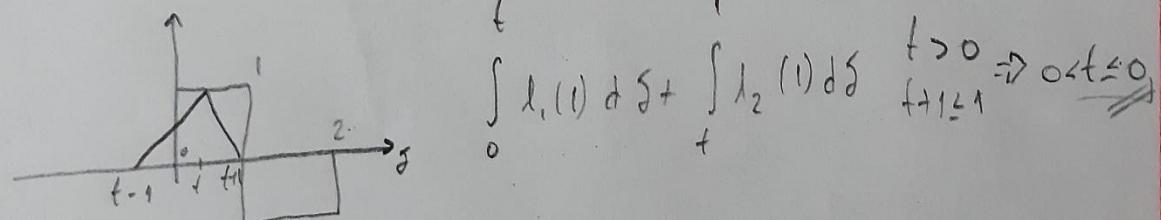
Caso 0:



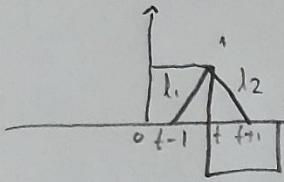
Caso 1:



Caso 2:

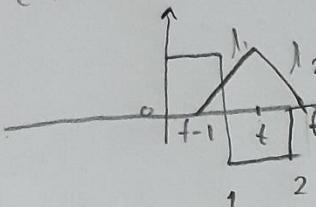


Caso 3:



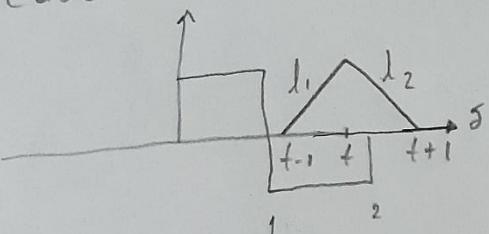
$$\int_{t-1}^1 l_1(1) d\delta + \int_1^{t+1} l_2(-1) d\delta; \quad t+1 > 1 \Rightarrow 0 < t \leq 1$$

Caso 4:



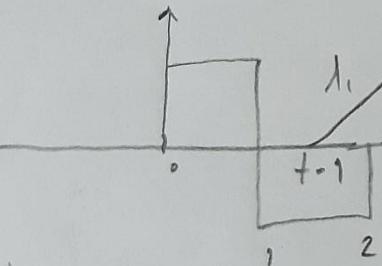
$$\int_{t-1}^1 l_1(1) d\delta + \int_1^t l_2(-1) d\delta; \quad t+1 > 1 \Rightarrow 1 < t \leq 2$$

Caso 5:



$$\int_1^t l_1(-1) d\delta + \int_t^{t+1} l_2(-1) d\delta; \quad t+1 > 1 \Rightarrow 2 < t \leq 2$$

Caso 6:

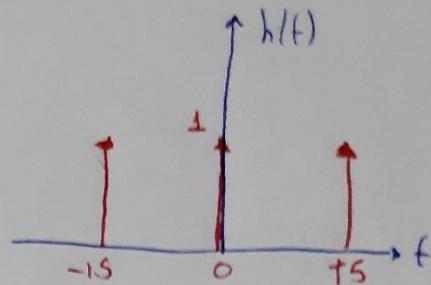
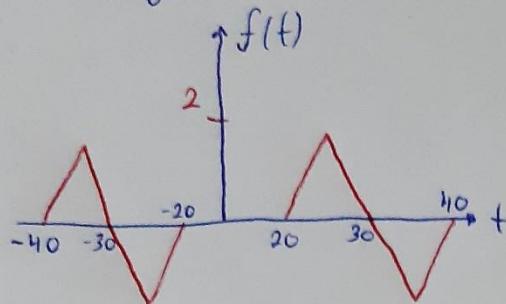


$$\int_{t-1}^2 l_1(-1) d\delta; \quad t+1 \leq 2 \Rightarrow 2 < t \leq 3$$

Entonces:

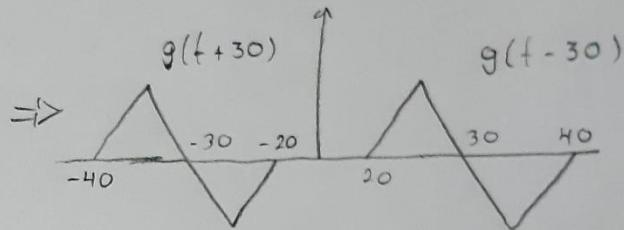
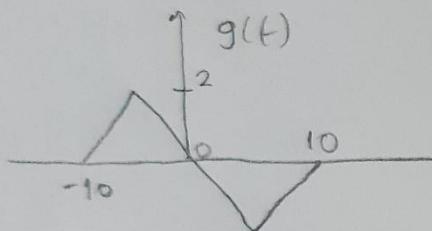
$$f_2(t) * f_3(t) = \underbrace{\int_{-1}^0 l_2(1) d\delta}_{-1 < t < 0} + \underbrace{\int_{0}^{t-1} l_1(1) d\delta}_{0 < t \leq 1} + \underbrace{\int_1^{t+1} l_2(-1) d\delta}_{1 < t \leq 2} + \underbrace{\int_{t+1}^2 l_1(-1) d\delta}_{2 < t \leq 3}$$

Problema 41: Obtener y dibujar $f_1(t) * h(t)$ para la siguiente figura:



Solución

Se representa a $f(t)$



$$f(t) * g(t) = f(t) * [\delta(t+15) + \delta(t-15)] =$$

$$= [g(t+30) + g(t-30)] * [\delta(t+15) + \delta(t-15)]$$

$$= g(t+30) * \delta(t+15) + g(t+30) * \delta(t-15) + g(t-30) * \delta(t+15) + g(t-30) * \delta(t-15)$$

$$= g(t+30+15) + g(t+30-15) + g(t-30+15) + g(t-30-15)$$

Entonces:

∴

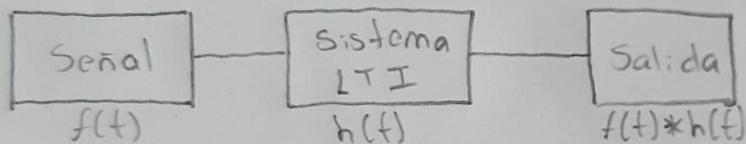
$$f_1(t) * f_2(t) = g(t+45) + g(t+15) + g(t-15) + g(t-45)$$

~~✓~~

Sección 2.

1. Defina Procesamiento Digital de señales, procesador digital de señales y dibuje el diagrama a bloques de un sistema de procesamiento digital de señales (adquisición de datos)

- Procesamiento digital es una manipulación matemática de una señal de información para modificarla y mejorarla. Caracterizado por su representación en tiempo discreto o frecuencia discreta.
- Procesador digital: Es un sistema basado en un procesador que posee un conjunto de instrucciones para optimizaciones que requieran operaciones a muy alta velocidad.



2: Enuncie el teorema del muestreo.

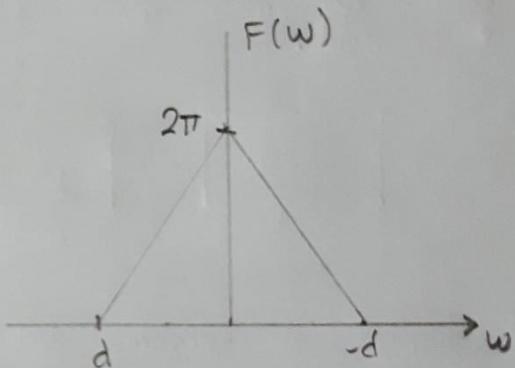
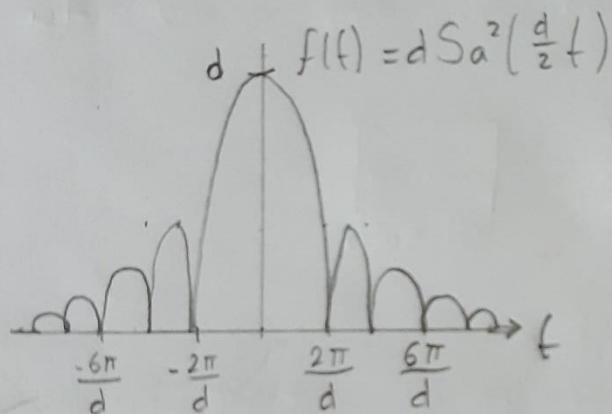
Toda señal limitada en banda que no contiene frecuencias mayores a H_z y $\frac{r_{at}}{s}$ está completamente representada por el conjunto de sus muestras formadas a intervalos uniformes no mayores de $1/2 f_m$ (s)

Es decir $T \leq 1/2 f_m$ ó $f_s \geq 2 f_m$

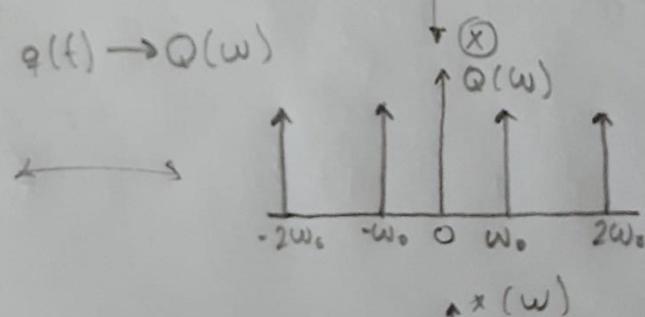
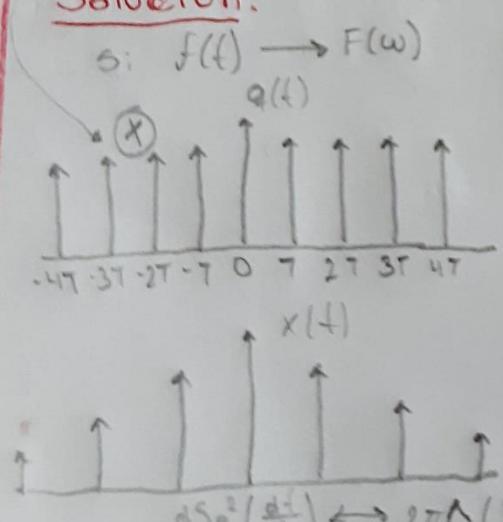
3: ¿A que se refiere el efecto Alias?

Surge cuando no se cumple el teorema de muestreo y provoca un deformamiento en la señal, perdiendo que se quiera volver a la señal original, es decir, se traslapan los señales y nos da una señal distinta a la original.

4. Considere la siguiente función en el tiempo $f(t)$ y su transformada $F(w)$. Desarrolle gráfica y matemáticamente el muestreo ideal de $f(t)$.



Solución.



$$\sum_{n=-\infty}^{\infty} f(nT) \delta(f-nT) \leftrightarrow \frac{1}{T} \sum_{n=-\infty}^{\infty} F(w-nw_0) //$$

$$\sum_{n=-\infty}^{\infty} \delta(f-nT) \leftrightarrow \sum_{n=-\infty}^{\infty} \delta(w-nw_0)$$

5. Determine la rapidez de las siguientes señales y el intervalo de Nyquist

a) $\text{Sa}(100t)$

Solución

$$\text{De tablas. } A\text{Cd}(t) \leftrightarrow A\text{dSa}\left(\frac{\omega d}{2}\right)$$

$$Cd(t) \leftrightarrow d\text{Sa}\left(\frac{\omega d}{2}\right)$$

$$d\text{Sa}\left(\frac{\omega d}{2}\right) \leftrightarrow 2\pi Cd(\omega)$$

$$s: d = 200$$

$$200\text{Sa}(100t) \leftrightarrow 2\pi C_{200}(\omega)$$

La componente frecuencial $\omega_m = 200 \frac{\text{rad}}{\text{s}}$

$$As: f_m = \frac{200}{2\pi} = 100/\pi \text{ Hz}$$

$$2f_m = \frac{200}{\pi} \text{ Hz} \quad T \leq \frac{1}{2f_m} = \frac{\pi}{200} \text{ s} \cancel{x}$$

$$T_w = \frac{1}{2f_m} = \frac{\pi}{200} \text{ s} \cancel{x}$$

b) $\text{Sa}^2(100t)$

Solución

$$A\left(\frac{t}{a}\right) \leftrightarrow d\text{Sa}^2\left(\frac{\omega d}{2}\right)$$

$$d\text{Sa}^2\left(\frac{t}{a}\right) \leftrightarrow 2\pi A\left(\frac{\omega}{a}\right)$$

$$s: d = 200$$

$$200\text{Sa}^2(100t) \leftrightarrow 2\pi A\left(\frac{\omega}{200}\right)$$

$$\text{Sa}^2(100t) \leftrightarrow \frac{\pi}{200} A\left(\frac{\omega}{200}\right)$$

Componente Frecuencial:

$$\omega_m = 200 \frac{\text{rad}}{\text{s}}$$

$$f_m = \frac{200}{2\pi} = \frac{100}{\pi} \text{ Hz} \cancel{x}$$

$$2f_m = \frac{200}{\pi} \text{ Hz}$$

$$T \leq \frac{1}{2f_m} = \frac{\pi}{200} \text{ seg} \cancel{x}$$

$$T_w = \frac{1}{2f_m} = \frac{\pi}{200} \text{ s} \cancel{x}$$

c) $\text{Sa}(100t) + \text{Sa}(50t)$

Solución

Para $\text{Sa}(100t) \leftrightarrow ?$

$$Cd(t) \leftrightarrow d\text{Sa}\left(\frac{\omega d}{2}\right)$$

$$d\text{Sa}\left(\frac{\omega d}{2}\right) \leftrightarrow 2\pi Cd(\omega)$$

$$s: d = 200$$

$$200\text{Sa}(100t) \leftrightarrow 2\pi C_{200}(\omega)$$

$$\text{Sa}(100t) \leftrightarrow \frac{\pi}{200} C_{200}(\omega)$$

Entonces:

$$\text{Sa}(100t) + \text{Sa}(50t) \leftrightarrow$$

Para: $\text{Sa}(50t)$

$$Cd(t) \leftrightarrow d\text{Sa}\left(\frac{\omega d}{2}\right)$$

$$d\text{Sa}\left(\frac{\omega d}{2}\right) \leftrightarrow 2\pi Cd(\omega)$$

$$s: \omega = 100$$

$$100\text{Sa}(50t) \leftrightarrow 2\pi C_{100}(\omega)$$

$$\text{Sa}(50t) \leftrightarrow \frac{\pi}{50} C_{100}(\omega)$$

$$\frac{\pi}{200} C_{200}(\omega) + \frac{\pi}{50} C_{100}(\omega) \cancel{x}$$

La componente frecuencial más grande es:

$$\omega_m = 200 \frac{\text{rad}}{\text{s}}$$

Así:

$$f_m = \frac{\omega_m}{2\pi} = \frac{100}{\pi} \text{ Hz}$$

$$2f_m = \frac{200}{\pi} \text{ Hz}$$

$$T = \frac{1}{2f_m} = \frac{\pi}{200} \text{ s}$$

$$T_N = \frac{1}{2f_m} = \frac{\pi}{200} \text{ seg} \cancel{\text{Hz}}$$

- 6.- Se sabe que una señal de valor real $x(t)$ ha sido determinada solo por sus muestras cuando la frecuencia de muestreo es $\omega_s = 10,000\pi t$, para que valores de ω se garantiza que $F(\omega)$ sea cero.

Solución

Sea: $\omega_s = 10,000\pi t$; \Rightarrow Dividiendo entre 2π : $\omega_s = \frac{10000}{2\pi} \pi t = 5000t$ \therefore para valores de $5000t$

- 7.- Aquella frecuencia que de acuerdo con el teorema de muestreo, debe ser excedida por la frecuencia de muestreo se llama razón de Nyquist. Determine la razón de Nyquist correspondiente a cada una de las siguientes señales:

a) $x(t) = 10 \operatorname{sen} \omega t + 5 \operatorname{sen} 2\omega t$

Solución

$$10 \operatorname{sen} \omega t \longleftrightarrow ?$$

Por tablas

$$10 \operatorname{sen}(\omega t) \longleftrightarrow 10\pi i [\delta(\omega - \omega) - \delta(\omega + \omega)]$$

$$5 \operatorname{sen}(\omega t) \longleftrightarrow ?$$

Por tablas:

$$5 \operatorname{sen}(\omega t) \longleftrightarrow 5\pi [\delta(\omega - \omega) - \delta(\omega + \omega)]$$

La componente frecuencial es $\omega_m = \omega$

Así $f_m = \frac{\omega}{2\pi}$ $2f_m = \frac{\omega}{\pi} \text{ Hz}$

$$T_N = \frac{1}{2f_m} = \frac{\pi}{\omega} \text{ seg} \cancel{\text{Hz}}$$

$$f_N = 2f_m = \frac{\omega}{\pi} \text{ Hz}$$

$$b) x(t) = 1 + \cos(2000\pi t) + \sin(4000\pi t)$$

Solución

$$\begin{array}{l} 1 \leftrightarrow ? \\ \text{Por tablas } \delta(t) \leftrightarrow 1 \\ 1 \leftrightarrow 2\pi\delta(\omega) \end{array}$$

$$\left. \begin{array}{l} \cos(2000\pi t) \leftrightarrow ? \\ \text{Por tablas} \\ \cos 2000\pi t \leftrightarrow \pi[\delta(\omega+2000\pi) + \delta(\omega-2000\pi)] \end{array} \right\}$$

$$\begin{array}{l} \sin(4000\pi t) \leftrightarrow ? \\ \text{Por tablas} \\ \sin(4000\pi t) \leftrightarrow \pi i[\delta(\omega+4000\pi) - \delta(\omega-4000\pi)] \end{array}$$

$$x(t) \leftrightarrow 2\pi\delta(\omega) + \pi i[\delta(\omega+4000\pi) - \delta(\omega-4000\pi)] + \pi[\delta(\omega+2000\pi) + \delta(\omega-2000\pi)]$$

La componente frecuencial más grande es $\omega_m = 4000\pi$

$$f_m = \frac{4000\pi}{2\pi} = \underline{2000 \text{ Hz}}$$

$$2f_m = 4000 \underline{\text{Hz}}$$

$$T_N = \frac{1}{2f_m} = \frac{1}{4000} \text{ Hz} = 4 \times 10^{-3} \text{ Hz} = \underline{2.5 \text{ ms}}$$

$$f_N = 2f_m = 4000 \underline{\text{Hz}}$$

8) Una señal continua $x(t)$ se obtiene a la salida de un filtro pasa bajo ideal con frecuencia de corte $\omega_c = 1000\pi$. Si el muestreo con tren impulsos se realiza sobre $x(t)$: ¿Cuál de los siguientes períodos de muestreo garantiza que $x(t)$ se puede recuperar a partir de sus versiones muestreadas usando un filtro paso bajas adecuado?

- a) $T = 0.5 \times 10^{-3}$
- b) $T = 2 \times 10^{-3}$
- c) $T = 10^{-4}$

Solución

$$\text{Si } \omega_c = 1000\pi \Rightarrow f_m = \frac{1000\pi}{2\pi} = \underline{500 \text{ Hz}}$$

$$2f_m = 1000 \text{ Hz}$$

$$T \leq \frac{1}{1000 \text{ Hz}} = \underline{1 \text{ ms}}$$

a) $0.5 \text{ ms} \leq 1 \text{ ms} \Rightarrow \text{Sí cumple.}$

b) $4 \text{ ms} \geq 1 \text{ ms} \Rightarrow \text{No cumple.}$

c) $0.1 \text{ ms} \leq 1 \text{ ms} \Rightarrow \text{Sí cumple.}$

9. Dibuje la función de transferencia $H(\omega)$ de un filtro pasa bajas con ganancia 5 y frecuencia de corte $f=100\text{Hz}$

Solución

$$f_m = 100\text{Hz}$$

$$\omega_m = 200\pi \frac{\text{rad}}{\text{s}}$$

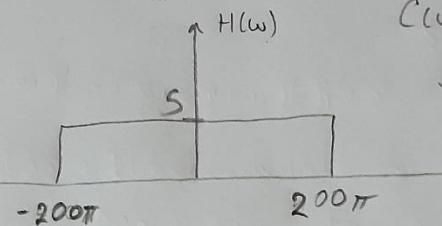
$$H(\omega) = T C_2 \omega_m (\omega) = 5 C_{400\pi}$$

$$H(\omega) = 5$$

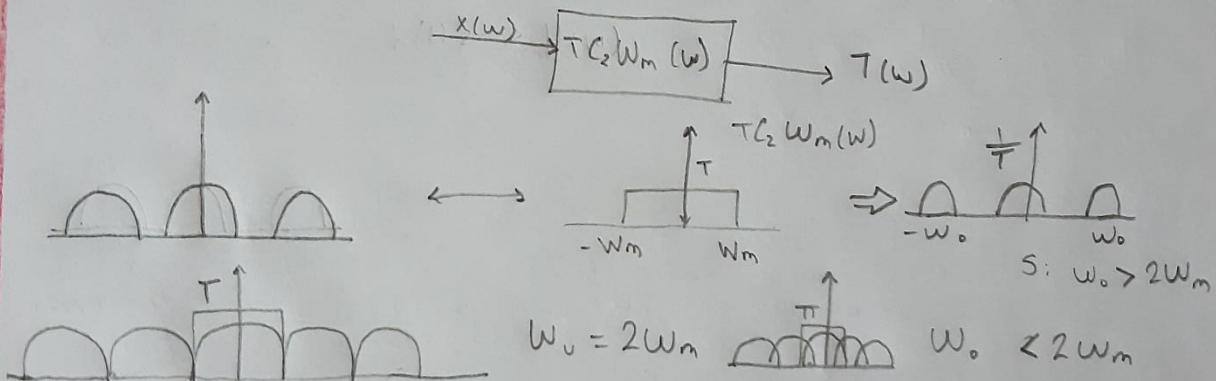
$$C(\omega) = 5_{400} C(\omega)$$

$$f = 100\text{Hz}$$

$$\Rightarrow \omega = 2\pi(100\text{Hz})$$

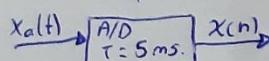


10) Dibuje la función característica $h(t)$ del filtro anterior.



11) Considere el sistema mostrado en la figura siguiente. La señal de entrada al sistema es: $x_a(t) = 3\cos 100\pi t + 2\sin 250\pi t$

Determine la versión discreta de $x_a(t)$ ¿Es posible recuperar la señal original a partir de $x(n)$ usando un filtro pasa bajas adecuado?



Solución

$$S: f_m = \frac{100\pi}{2\pi} = 50 \text{ Hz} \Rightarrow 2f_m = 100\text{Hz}$$

$$T_N = \frac{1}{2f_m} = \frac{1}{200} \text{ sec} \Rightarrow f_N = 2f_m = 100\text{Hz}$$

$\therefore S:$ se puede recuperar usando un filtro pasa bajas

12. Analice las siguientes secuencias (esto es, su secuencia digital); e indique si son o no periódicas. En caso de ser periódicas, halle su periodo.

a) $10 \sin\left(\frac{3}{2}\pi n\right)$

Solución

$$10 \sin\left(\frac{3}{2}\pi n\right) = 10 \cos\left(\frac{3}{2}\pi n + \frac{\pi}{2}\right)$$

$$2\pi F_N = \frac{3}{2}\pi N$$

$$F = \frac{3}{4} \Rightarrow N = 4 \cancel{\text{X}}$$

b) $5 \cos\left(\frac{4}{9}\pi n\right)$

Solución

$$5 \cos\left(\frac{4}{9}\pi n\right) \Rightarrow A \cos(2\pi F_n + d)$$

$$2\pi F_n = \frac{4}{9}\pi n$$

$$F = \frac{4}{18} \Rightarrow N = 18 \cancel{\text{X}}$$

c) $2 \cos\left(\frac{4}{9}n\right)$

Solución

$$2 \cos\left(\frac{4}{9}n\right) \Rightarrow A \cos(2\pi F_n + d)$$

$$2\pi F = \frac{4}{9}n$$

$$F = \frac{4}{18\pi} : \text{No es periódica} \cancel{\text{X}}$$

d) $x(n) = \cos \frac{2\pi n}{3} + e^{\pi n}$

Solución

$x_a(n) = \cos \frac{2\pi n}{3}$; es periódica $T=3$

$x_b(n) = e^{\pi n}$; no es periódica

$\therefore x(n)$ no es periódica.

e) $y(n) = 2 + \operatorname{Re}[e^{i\frac{n\pi}{3}}] + \cos \frac{3\pi n}{2}$

Solución

$$y_1(n) = 2 = \{ \dots, 2, 2, 2, 2, 2, 2, 2, 2, 2, 2, 2, 2, \dots \}$$

$$y_2(n) = \operatorname{Re}\left\{ e^{i\frac{n\pi}{3}} \right\} = \cos \frac{n\pi}{3} + i \sin \frac{n\pi}{3}$$

$$\cos \frac{n\pi}{3} = \{ \dots, 1, \frac{1}{2}, \frac{1}{2}, -1, \frac{1}{2}, \frac{1}{2}, 1, \frac{1}{2}, \frac{1}{2}, -1, \frac{1}{2}, \frac{1}{2}, \dots \}$$

$$y_3(n) = \cos\left(\frac{3n\pi}{2}\right) = \{ \dots, -1, 0, 1, 0, -1, 0, 1, 0, -1, 0, 1, 0, -1, 0, \dots \}$$

$$\therefore N = \cancel{12} / \cancel{3}$$

$$N = \cancel{3}$$

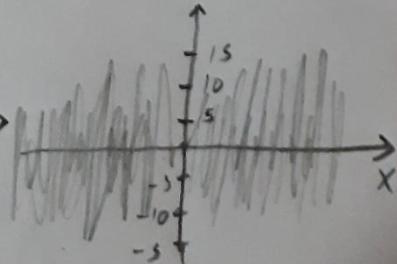
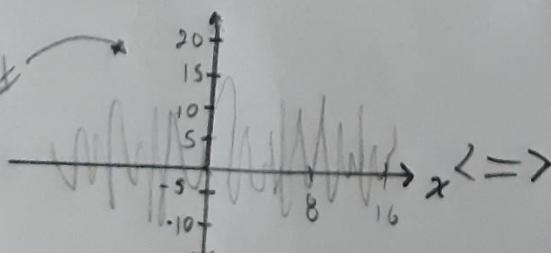
13: Gráfique la siguiente señal $y(n)$ que es una suma sinusoides, indique su periodo. ¿Cuál es el periodo de la suma de 2 sinusoides de periodo N_1 y N_2 ?

$$y(n) = \underbrace{10 \sin\left(\frac{3}{2}\pi n\right)}_{N_1} + \underbrace{5 \cos\left(\frac{4}{3}\pi n\right)}_{N_2}$$

Solución

$$N_1 = 4; N_2 = 18$$

$$\begin{array}{r|rr} 1 & 18 \\ 2 & 9 \\ 1 & 9 \\ 1 & 3 \\ 1 & 3 \end{array} \geq 4 \quad \begin{array}{r|rr} 2 & 36 \\ 2 & 18 \\ 1 & 9 \\ 1 & 3 \\ 1 & 3 \end{array} \geq 9$$



Sección 3: Operaciones Básicas entre secuencias.

Problema 1. Considere las secuencias siguientes y realice con ellas las operaciones indicadas.

$$x[n] = \{0, 1, 2, 3, 4, 5, 6, 7, 8\}$$

$$y[n] = \{2, 4, 8, 16, 32\}$$

$$z[n] = \sum_{k=-3}^3 x(n-k) =$$

$$g[n] = \left\{ \frac{1}{3}, \frac{1}{6}, \frac{1}{9}, \frac{1}{12}, \frac{1}{15}, \frac{1}{18}, \dots \right\}$$

$$k[n] = 2u(n-2)$$

Encuentre:

a) $g[-n]$

Solución

$$g[-n] = \{\dots, \frac{1}{18}, \frac{1}{15}, \frac{1}{12}, \frac{1}{9}, \frac{1}{6}, \frac{1}{3}\}$$

b) $z[n] + y[n]$

Solución

$$z[n] + y[n] = \{1, 1, 1, 1, 2, 3, 4, 4, 5, 6, 7, 8\}$$

c) $3g[n] - 6z[n]$

Solución

$$3g[n] = \{1, \frac{1}{2}, \frac{1}{3}, \frac{1}{4}, \frac{1}{5}, \frac{1}{6}\}$$

$$-6z[n] = \{-6, -6, -6, -6, -6, -6\}$$

$$\therefore = \left\{ -5, -\frac{11}{2}, -\frac{17}{3}, -\frac{23}{4}, -\frac{29}{5}, -\frac{35}{6}, \dots \right\}$$

d) $y[n-6]$

Solución

$$y[n-6] = \{0, 0, 2, 4, 8, 16, 32\}$$

$$\text{c) } \frac{1}{2}y[n+3]$$

Solución

$$y[n+3] = \{2, 4, 8, 16, 32, 0, 0, 0\}$$

$$\frac{1}{2} y[n+3] = \{1, 2, 4, 8, \underline{16}, 000\}$$

$$f) x[n-2]$$

Solución

$$x[n-2] = \{0, 0, 0, 1, 2, 3, 4, 5, 6, 7, 8\}$$

g) $x[3n-3]$

Solución

$$x[n-3] = \{ \textcircled{0}, 0, 0, \textcircled{0}, 1, 2, \textcircled{3}, 4, 5, \textcircled{6}, 7, 8 \}$$

$$x[3n-3] = \{ \bar{0}, 0, 3, 6 \}$$

$$h) x \left[\frac{n}{2} + 5 \right]$$

Solución

$$x[n+5] = \{0, 1, 2, 3, 4, \overline{5}, 6, 7, 8\}$$

$$x[\frac{n}{2}+5] = \{0, 0, 1, 1, 2, 2, 3, 3, 4, 4, 5, 5, 6, 6, 7, 7, 8, 8\}$$

$$i) y \left[\frac{n-3}{3} \right]$$

Solución

$$y[n-3] = \{2, 4, 8, 16, 32\}$$

$$y \left\lceil \frac{n-3}{3} \right\rceil = \{2, 2, 2, 4, 4, 4, 8, 8, 8, 16, 16, 16, 32, 32, 32\}$$

$$j) K \left[\frac{4n-3}{10} \right]$$

Solución

$$k[n] = \{0, 0, 2, 2, 2, \dots\}$$

$$K[0,3] = \{0,0,0,0,0,2,2,2\}$$

$$K) \in [-\frac{7}{4} + 10]$$

Solución

$$K[n] = \{ \bar{0}, 0, 2, 2, 2, 2, 2, \}.$$

$$K[\neg n] = \{ \bar{0}, 0, 2, \bar{2}, 2, 2, \dots \}$$

$$\mathbb{K}[\frac{1}{x}] = \{0, 2\}$$

$$K \left[-\frac{9}{4} + 10 \right] = \{ 0, 2, 0, 2, 0, 2, 0, 2, 0, 2, \underline{\underline{0, 2, 3}} \}$$

$$1) x\left[\frac{3n-3}{3}\right] - g\left[-\frac{8n-7}{3}\right]$$

Solución

$$X[3n] = \{\overline{0}, 1, 2, \cancel{3}, 4, 5, \cancel{6}, 7, 8\} = \{\overline{0}, 3, 6\}$$

$$x[3n-3] = \{0, 0, 0, 0, 3, 6\}$$

$$x[\frac{3n-3}{3}] = \{0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 3, 3, 3, 3, 3, 6, 6, 6, 6\}$$

$$g\left[-\frac{Bn-2}{3}\right] = \left\{ \frac{1}{15}, \frac{1}{15}, \frac{1}{15}, \frac{1}{12}, \frac{1}{12}, \frac{1}{12}, \frac{1}{9}, \frac{1}{9}, \frac{1}{9}, \frac{1}{3}, \frac{1}{3}, \frac{1}{3} \right\}$$

$$x \left[\frac{3n-3}{3} \right] - 9 \left[-\frac{8n-7}{3} \right] = \left\{ -\dots, -\frac{1}{18}, -\frac{1}{9}, \frac{2}{3}, 2, 3, 4, 5, 6, 7, 8 \right\}$$

$$m) u[n] \cdot g[n] + z[n]$$

Solución

$$u[n] = \{ \bar{1}, 1, \bar{1}, 1, 1, 1, \dots \}$$

$$g[n] = \{ \frac{1}{3}, \frac{1}{6}, \frac{1}{9}, \frac{1}{12}, \frac{1}{15}, \frac{1}{18}, \dots \}$$

$$u(n) \cdot g(n) = \{ \frac{\bar{1}}{12}, \frac{1}{15}, \frac{1}{18}, \frac{1}{21}, \frac{1}{24}, \dots \}$$

$$+ z(n) = \{ 1, 1, 1, \bar{1}, 1, 1, 1 \}$$

$$u(n) \cdot g(n) + z(n) = \{ 1, 1, 1, \frac{13}{12}, \frac{16}{15}, \frac{19}{18}, \frac{22}{21}, \frac{25}{24}, \dots \}$$

$$n) x[\frac{n}{2}] * y[\frac{n}{3}]$$

Solución

$$x(n) = \{ 0, 1, 2, 3, 4, 5, 6, \bar{7}, 8 \} \quad y(n) = \{ 2, 4, 8, 16, \bar{32} \}$$

$$x(\frac{n}{2}) = \{ 0, 0, 1, 1, 2, 2, 3, 3, 4, 4, 5, 5, 6, 6, \bar{7}, 7, 8, 8 \}$$

$$y(\frac{n}{3}) = \{ 2, 2, 2, 4, 4, 4, 8, 8, 8, 16, 16, \bar{32}, 32, 32 \}$$

$$x(\frac{n}{2}) * y(\frac{n}{3}) = \{ 0, 0, 2, 4, 8, 16, 22, 32, 48, 66, 90, \bar{124}, 164, 214, 286, 308, 442, 559, 640, 709, 788, 832, 896, \bar{936}, 960, 960, 970, 896, 832, 766, 256 \}$$

$$\tilde{n}) g(\frac{3n-1}{2}) * y(\frac{2n}{2} + 2)$$

Solución

$$g(3n) = \{ \cancel{\frac{1}{3}}, \frac{1}{6}, \cancel{\frac{1}{9}}, \cancel{\frac{1}{12}}, \frac{1}{15}, \frac{1}{18}, \dots \} = \{ \frac{1}{3}, \frac{1}{12} \}$$

$$g(3n-1) = \{ \frac{1}{3}, \frac{1}{12} \}$$

$$g(\frac{3n-1}{2}) = \{ \frac{1}{3}, \frac{1}{3}, \frac{1}{12}, \frac{1}{12} \}$$

$$y(2n) = \{ 2, 4, \cancel{8}, 16, \cancel{32} \} = \{ 2, 8, \bar{32} \}$$

$$y(\frac{2n}{2}) = \{ 2, 2, 8, 8, \bar{32}, 32 \}$$

$$y(\frac{2n}{2} + 2) = \{ 2, 2, 8, 8, 32, 32, \bar{0} \}$$

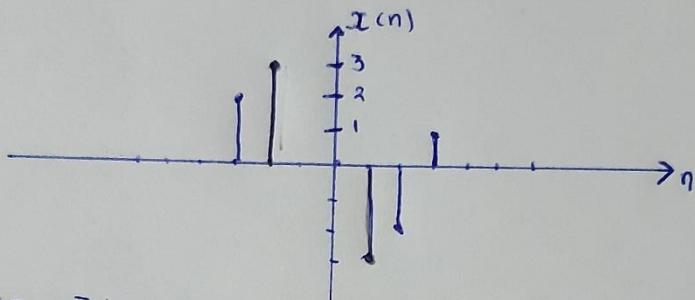
$$g(\frac{3n-1}{2}) * y(\frac{2n}{2} + 2) = \{ \frac{1}{3}, \frac{8}{9}, \frac{86}{45}, \frac{59}{45}, \frac{118}{45}, \frac{76}{45}, \frac{136}{45}, \frac{16}{9}, 0, 0 \}$$

Problema 2. Encuentre la gráfica de la secuencia de convolución de dos secuencias definidas como:
 $x(n) = \{1, 2, 3, 4, 3, 2, 1\}$ y $y(n) = \{-2, -1, 0, 1, 2\}$. Use cualquier método.

Solución

$$x(n)*y(n) = \{-2, -5, -8, -10, -6, 0, 6, 10, 8, 5, 2\}$$

Problema 3. A partir de la secuencia mostrada en la figura.



Encuentre: $x\left(\frac{3n+3}{4}\right)$ y $g(n) = x(n) * \{x[n] - (u[n+1] - u[n-3])\}$

Solución

$$\begin{aligned} x(n) &= \sum_{n=-\infty}^{\infty} x(n) e^{j \frac{\pi}{2} n} \Rightarrow x(n) = \sum_{n=-\infty}^{\infty} x(n) e^{-jn} = x(-2)e^{j2n} + x(-1)e^{jn} \\ &\quad + x(0)e^0 + x(1)e^{j\pi} + x(2)e^{j2\pi} \\ &= 2(e^{j2\pi} + e^{-2j\pi}) + e^{jn} + e^{-jn} \\ &= 4\cos 2\pi + 2\cos \pi \end{aligned}$$

Problema 4. Las extensiones periódicas de dos secuencias, $x_1(n)$ y $x_2(n)$, se definen como:

$$x_1(n) = \{1, 2, 0, -1, 1\}, \quad x_2(n) = \{1, 3, -1, -2\}. \text{ Halle la secuencia de convolución.}$$

Solución

$$\begin{array}{r} 1 \ 2 \ 0 \ -1 \ 1 \\ 1 \ 3 \ -1 \ -2 \\ \hline 1 \ 2 \ 0 \ -1 \ 1 \\ 3 \ 6 \ 0 \ -3 \ 3 \\ -1 \ -2 \ 0 \ 1 \ -1 \\ -2 \ -4 \ 0 \ 2 \ -2 \\ \hline 1 \ 5 \ 5 \ -5 \ -6 \ 4 \ 1 \ -2 \end{array}$$

$$\begin{array}{r} 1 \ 5 \ 5 \ 6 \ -6 \\ 4 \ 1 \cdot 2 \ 0 \ 0 \\ \hline 5 \ 6 \ 3 \ 6 \ -6 \end{array}$$

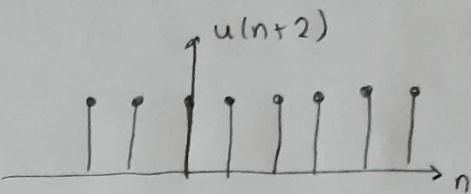
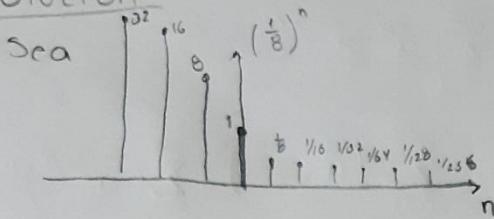
$$\therefore x_1(n)*x_2(n) = \{5, 6, \bar{3}, 6, -6\}$$

Sección 4: Transformado de Fourier y Transformada Z.

Problema 1: Halle la transformada de Fourier de:

$$y(n) = \left(\frac{1}{8}\right)^n u(n+2) + 2\delta(-n)$$

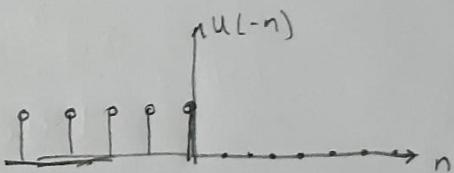
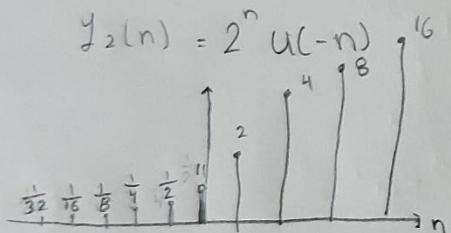
Solución



$$y_1(n) = \begin{cases} 0 & n < -2 \\ \left(\frac{1}{8}\right)^n & n \geq -2 \end{cases}$$

$$Y_1(n) = \sum_{n=-2}^{\infty} \left(\frac{1}{8}\right)^n e^{-jn\pi} = \sum_{n=2}^{\infty} \left(\frac{1}{8} e^{j\pi}\right)^{n+2} < 0$$

$$\left| \frac{1}{8} e^{j\pi} \right| = \frac{1}{8} < 1 \Rightarrow Y_1(n) = \frac{\left(\frac{1}{8} e^{j\pi}\right)^{-2}}{1 - \frac{1}{8} e^{-jn\pi}} = \frac{8^2 e^{-j2\pi}}{8 - e^{-jn\pi}} = \frac{8e^{j2\pi}}{8 - e^{-jn\pi}}$$



$$y_2(n) = \begin{cases} 0 & n > 0 \\ 2^n & n \leq 0 \end{cases}$$

$$Y_2(n) = \sum_{n=-\infty}^0 2^n e^{-jn\pi} = \sum_{n=-\infty}^0 (2e^{-j\pi})^n = \sum_{n=0}^{\infty} (2e^{-j\pi})^{-n} = \sum_{n=0}^{\infty} \left(\frac{1}{2} e^{jn\pi}\right)^n$$

$$\left|\frac{1}{2} e^{jn\pi}\right| = \frac{1}{2} < 1 \Rightarrow Y_2(n) = \frac{1}{1 - \frac{1}{2} e^{-jn\pi}} = \frac{2}{2 - e^{-jn\pi}}$$

$$\therefore Y(n) = Y_1(n) + Y_2(n) = \frac{8^3 e^{2jn\pi}}{8 - e^{-jn\pi}} + \frac{2}{2 - e^{-jn\pi}} = \frac{(8^3 e^{2jn\pi})(2 - e^{jn\pi}) + 2(8 - e^{jn\pi})}{(8 - e^{-jn\pi})(2 - e^{-jn\pi})}$$

$$= \frac{(2 \cdot 8^3) e^{2jn\pi} - 8 e^{3jn\pi} + 16 - 2 e^{jn\pi}}{17 - 8 e^{jn\pi} - 2 e^{-jn\pi}}$$

Problema 2: Halle la transformada de Fourier de:

$$y(n) = \begin{cases} a^n & n \geq 0 \\ \bar{a}^n & n < 0 \end{cases}$$

Solución

$$y(n) = \sum_{n=-\infty}^{\infty} \Rightarrow y(n) = \sum_{n=-\infty}^{-1} \bar{a}^n e^{-in\omega} + \sum_{n=0}^{\infty} a^n e^{-in\omega}$$

$$y(n) = \sum_{n=1}^{\infty} a^n e^{in\omega} + \sum_{n=0}^{\infty} (a e^{-i\omega})^n = \sum_{n=1}^{\infty} (a e^{i\omega})^n + \sum_{n=0}^{\infty} (a e^{-i\omega})^n$$

Por series geométricas...

$$y(n) = \frac{a e^{i\omega}}{1 - a e^{i\omega}} + \frac{1}{1 - a e^{-i\omega}} \quad \text{para } |a| < 1$$

$$y(n) = \frac{a e^{i\omega} (1 - a e^{i\omega}) + 1 - a e^{i\omega}}{(1 - a e^{i\omega})(1 - a e^{-i\omega})}$$

$$y(n) = \frac{a e^{i\omega} - a e^{i\omega + i\omega} + 1 - a e^{i\omega}}{1 - a e^{i\omega} - a e^{i\omega} + a^2 e^{i\omega}} =$$

$$\cancel{y(n) = \frac{1 - a^2}{1 + a^2 - 2a \cos \omega}}$$

Calcule la transformada discreta de Fourier (DFT)

$$de x(n). \quad x(n) = \{0, \frac{1}{2}, 1, 2, -3, -4\}$$

Solución

$$x(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi}{N} kn}$$

$$N=6 \quad x(k) = \sum_{n=0}^{6} x(n) e^{-j \frac{2\pi}{6} kn}$$

con $k=0$

$$x(0) = \sum_{n=0}^{6} x(n) e^0$$

$$x(0) = x(0) + x(1) + x(2) + x(3) + x(4) + x(5)$$

$$x(0) = 0 + \frac{1}{2} + 1 + 2 - 3 - 4 = -\frac{7}{2}$$

$$k=1 \quad x(1) = \sum_{n=0}^{5} x(n) e^{-j \frac{2\pi}{6} n}$$

$$x(1) = x(0) e^0 + x(1) e^{-j \frac{2\pi}{6}} + x(2) e^{-j \frac{4\pi}{6}} + x(3) e^{-j \frac{6\pi}{6}} + x(4) e^{-j \frac{8\pi}{6}} + x(5) e^{-j \frac{10\pi}{6}}$$

$$x(1) = 0 - \frac{1}{2}i - 1 - i - 3i - 4 = -5 - \frac{9}{2}i$$

$k=2$

$$x(2) = \sum_{n=0}^{5} x(n) e^{-j \frac{4\pi}{6} n}$$

$$x(2) = x(0) e^0 + x(1) e^{-j \frac{4\pi}{6}} + x(2) e^{-j \frac{8\pi}{6}} + x(3) e^{-j \frac{12\pi}{6}} + x(4) e^{-j \frac{16\pi}{6}} + x(5) e^{-j \frac{20\pi}{6}}$$
$$= -\frac{3}{2} + \frac{3}{2}i$$

$$x(3) = -1 - 4i$$

$$x(4) = x(0)e^0 + x(1)e^{-i\frac{8\pi}{5}} + x(2)e^{-i\frac{16\pi}{5}} + x(3)e^{-i\frac{24\pi}{5}} + x(4)e^{-i\frac{32\pi}{5}} + \\ x(5)e^{-i\frac{40\pi}{5}}$$

$$x(4) = \frac{1}{2} - 1 - i - 2 - 2i + 3i - 4 = \frac{15}{2}$$

$$x(5) = \sum_{n=0}^5 x(n) e^{-i\frac{2\pi}{5}n}$$

$$x(5) = x(0)e^0 + x(1)e^{-i\frac{2\pi}{5}} + x(2)e^{-i\frac{4\pi}{5}} + x(3)e^{-i\frac{8\pi}{5}} + x(4)e^{-i\frac{12\pi}{5}} + \\ x(5)e^{-i\frac{16\pi}{5}}$$

$$x(5) = \frac{1}{2} + 1 + 2 - 3 - 4 = -\frac{7}{2}$$

$$x(k) = \left\{ \dots, -\frac{7}{2}, -5, -\frac{9}{2}i, -\frac{3}{2} + \frac{3}{2}i, -1 - 4i, \frac{15}{2}, \dots \right\}$$

Problema 5: Use la forma matricial de la DFT y calcule:

$$g(n) = \{0, 5, 0, 2, 5\}$$

Solución

$$x(k) = \sum_{n=0}^{N-1} x(n) e^{-i\frac{2\pi}{N} kn} ; \quad X(k) = \sum_{n=0}^3 x(n) e^{-i\frac{2\pi}{3} kn}$$

$$X(k) = \sum_{n=0}^{N-1} x(n) W^k \quad X_k = W_N X_n$$

$$W_N = \begin{vmatrix} 1 & 1 & 1 & 1 \\ 1 & -i & -1 & i \\ 1 & -1 & 1 & -1 \\ 1 & i & -1 & -i \end{vmatrix} \begin{vmatrix} 5 \\ 0 \\ 2 \\ 5 \end{vmatrix} = \begin{vmatrix} 0.5 & 0.5 & 0.5 & 0.5 \\ 0 & 0 & 0 & 0 \\ 2 & -2 & 2 & -2 \\ 5 & 5i & -5 & -5i \end{vmatrix} \begin{vmatrix} 2 \\ 0 \\ 0 \\ 0 \end{vmatrix}$$

$$g(k) = \{2, 0, 0, 0\}$$

Halle la FFT para la siguiente secuencia:

Solución

$$x[n] = [0.5, 1, 0, 12]$$

$$N=4; \quad \log_2 N = \log_2 4 = 2 \quad \omega^2 = \left(e^{-j\frac{2\pi}{4}}\right) = \left(e^{-j\frac{\pi}{2}}\right) = -i$$

$$\begin{array}{ll} \frac{N}{2}=2; & \begin{array}{l} x(0) \\ \cdot \omega^0 \\ x(1) \\ \cdot \omega^1 \\ x(2) \\ \cdot \omega^2 \\ x(3) \\ \cdot \omega^3 \end{array} \end{array} \quad \left. \begin{array}{l} x'(0) = 0.5 + \omega^0(0) = 0.5 \\ x'(1) = 0.5 + \omega^1(0) = 0.5 \\ x'(2) = 1 + \omega^2(12) = 13 \\ x'(3) = 1 + \omega^3(12) = -1 \\ \\ X_0 = 0.5 + 13 = 13.5 \\ X_1 = 0.5 + \omega^1(-11) = 0.5 + (-i)(11) = 0.5 - 11i \\ X_2 = 0.5 + 13(-1) = -8 \\ X_3 = 0.5 - (\omega^3)(-11) = 0.5 - (-i)(-11) = 0.5 - 11i \end{array} \right\}$$

Encuentre la transformada Z y grafique la región de convergencia

$$h(n) = \begin{cases} \left(\frac{1}{6}\right)^n & n \geq 0 \\ 0 & n \leq 0 \end{cases}$$

Solución

$$Z\{h(n)\} = h(z) = \sum_{n=-\infty}^{\infty} h(n) z^{-n}$$

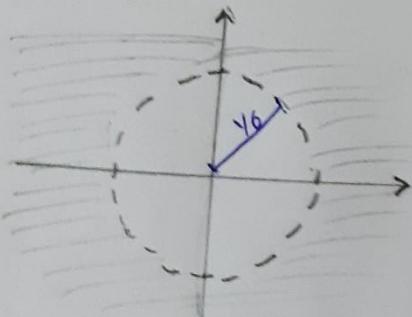
$$\begin{aligned} Z\{h(n)\} = h(z) &= \sum_{n=-\infty}^0 0 + \sum_{n=0}^{\infty} \left(\frac{1}{6}\right)^n z^{-n} \\ &= \sum_{n=0}^{\infty} \left(\frac{1}{6} z^{-1}\right)^n \end{aligned}$$

Por series geométricas

$$\sum_{k=m}^{\infty} \alpha^k = \frac{\alpha^m}{1-\alpha} \quad |\alpha| < 1$$

$$h(z) = \frac{\frac{1}{6z}}{1 - \frac{1}{6z}} ; \left|\frac{1}{6z}\right| < 1 \Rightarrow h(z) = \frac{6z}{(6z-1)(6z)} = \frac{1}{6z-1}$$

región de convergencia



Polos de $h(z)$
 $z = \frac{1}{6}$

Encuentre la transformada z y grafique la región de convergencia de $h(n)$

$$h(n) = \begin{cases} (10)^n & n \leq 0 \text{ anticausal} \\ (\frac{1}{10})^n & n \text{ par causal} \\ (\frac{1}{5})^n & n \text{ impar causal} \end{cases}$$

Solución

$$Z\{h(n)\} = h(z) = \sum_{n=-\infty}^{\infty} h(n) z^{-n}$$

$$h(z) = \sum_{n=-\infty}^{n=1} 10^n z^{-n} + \sum_{n=0}^{\infty} (\frac{1}{10})^n z^{-n} + \sum_{n=0}^{\infty} (\frac{1}{5})^n z^{-n}$$

"par" "impar"

$$h(z) = \sum_{n=-\infty}^{n=-1} (10z^{-1})^n + \sum_{n=0}^{\infty} (\frac{1}{10})^{2n} z^{-2n} + \sum_{n=0}^{\infty} (\frac{1}{5})^{2n+1} z^{-(2n+1)}$$

$$h(z) = \sum_{n=-\infty}^{n=-1} (\frac{10}{z})^n + \sum_{n=0}^{\infty} \left[(\frac{1}{10})^2 \frac{1}{z^2} \right]^n + \sum_{n=0}^{\infty} (\frac{1}{5})^{2n} \left(\frac{1}{z} \right) \cdot z^{2n} z^{-1}$$

$$h(z) = \sum_{n=1}^{\infty} \left(\frac{z}{10} \right)^n + \sum_{n=0}^{\infty} \left[\frac{1}{100z^2} \right]^n + \frac{1}{5z} \sum_{n=0}^{\infty} \left(-\frac{1}{25z^2} \right)^n$$

De series geométricas

$$\sum_{k=0}^{\infty} \alpha^k = \frac{1}{1-\alpha} \quad |\alpha| < 1 \quad ; \quad \sum_{k=m}^{\infty} \alpha^k = \frac{\alpha^m}{1-\alpha} \quad |\alpha| < 1$$

Entonces:

$$h(z) = \underbrace{\frac{z}{10}}_{|\frac{z}{10}| < 1} + \underbrace{\frac{1}{1 - \frac{1}{100z^2}}}_{|\frac{1}{100z^2}| < 1} + \frac{1}{5z} \cdot \underbrace{\frac{1}{1 - \frac{1}{25z^2}}}_{|\frac{1}{25z^2}| < 1}$$

$$h(z) = \frac{z}{\frac{10}{10-z}} + \frac{1}{\frac{100z^2-1}{100z^2}} + \frac{1}{5z} \cdot \frac{1}{\frac{25z^2-1}{25z^2}}$$

$$h(z) = \frac{z}{10-z} + \frac{100z^2}{100z^2-1} + \frac{1}{5z} \cdot \frac{25z^2}{25z^2-1}$$

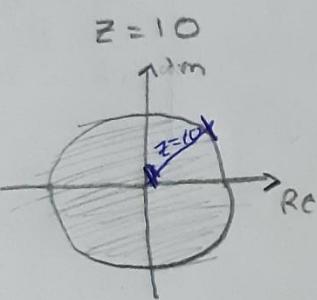
$$h(z) = \underbrace{\frac{z}{10-z}}_{|z| < 10} + \underbrace{\frac{z^2}{z^2 - \frac{1}{100}}}_{\frac{1}{100} < |z^2|} + \underbrace{\frac{1}{5z} \cdot \frac{z^2}{z^2 - \frac{1}{25}}}_{\frac{1}{25} < |z^2|}$$

Convergencias:

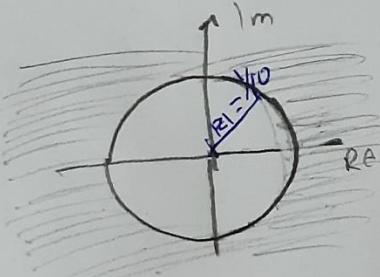
$$\frac{1}{100} < |z^2|$$

$$\frac{1}{25} < |z^2|$$

Poles

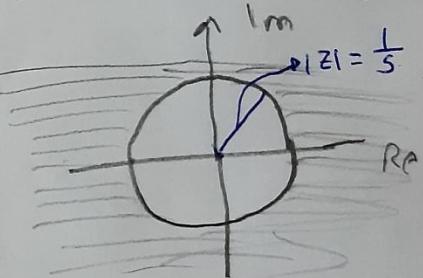


Polo
 $z^2 = \frac{1}{100} \Rightarrow z = \pm \frac{1}{10}$

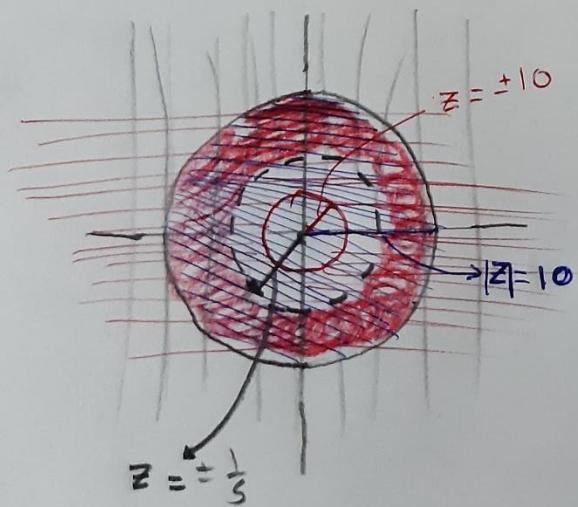


Tren' 2 polos

$z_1 = 0 ; |z_2| = \frac{1}{25} \Rightarrow z_2 = \pm \frac{1}{5}$



∴



$$\frac{1}{5} < |z| < 10$$