

Iterative Filtering Algorithms

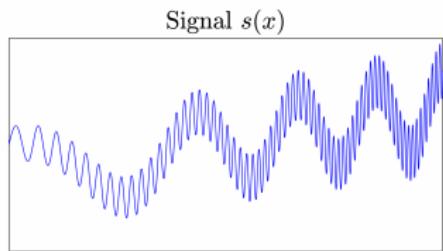
Giovanni Barbarino
Department of Mathematics
and Systems Analysis,
Aalto University

Antonio Cicone
Department of Information Engineering
Computer Science and Mathematics,
University of L'Aquila

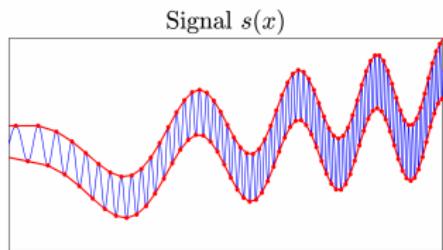
NoSAG21 Conference, L'Aquila

Model of Iterative Filtering Algorithms

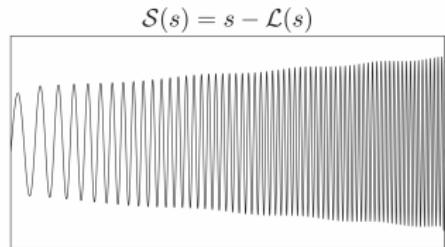
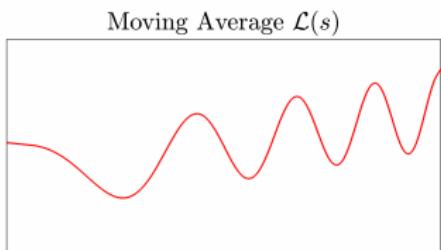
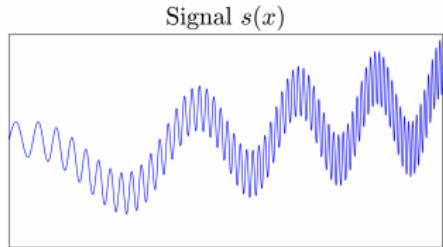
Empirical Mode Decomposition



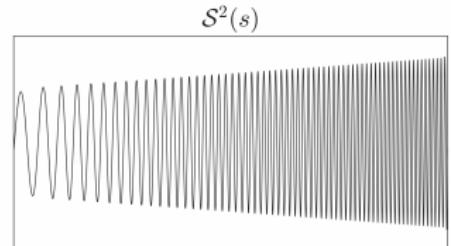
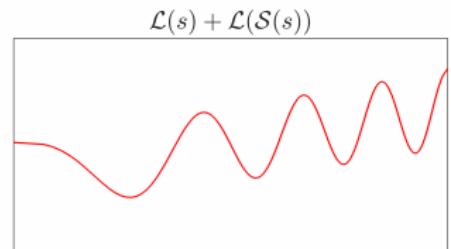
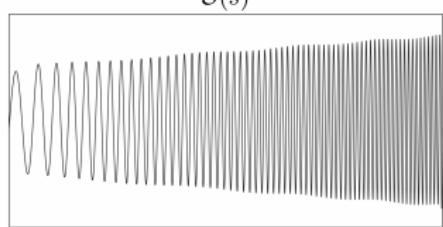
Empirical Mode Decomposition



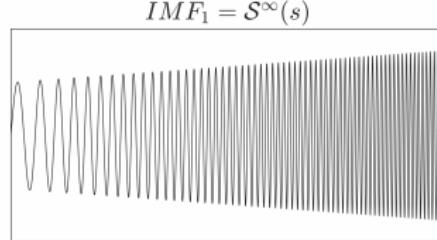
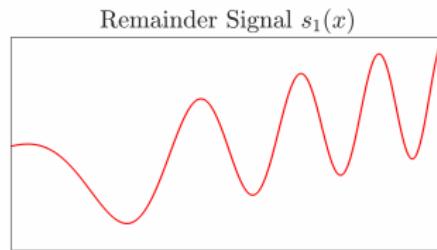
Empirical Mode Decomposition



Empirical Mode Decomposition

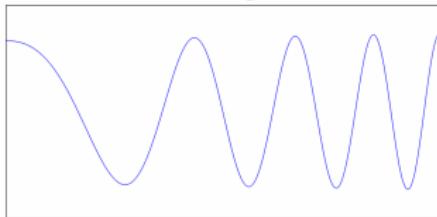


Empirical Mode Decomposition

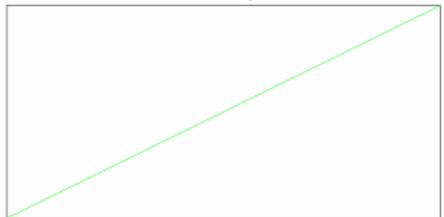


Empirical Mode Decomposition

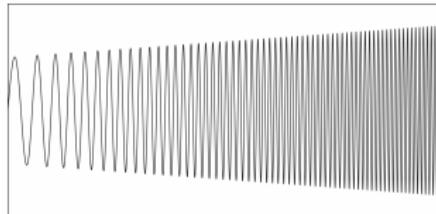
IMF_2



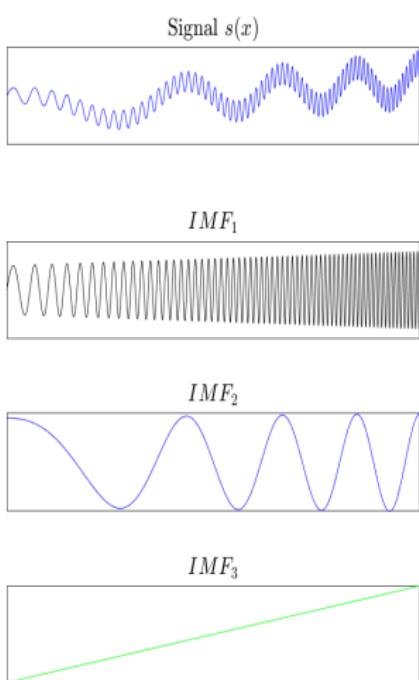
IMF_3



IMF_1



Empirical Mode Decomposition



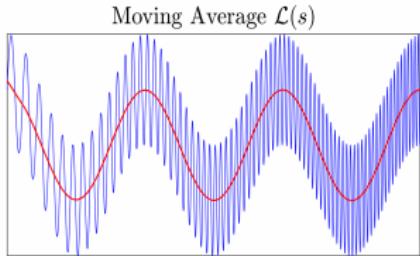
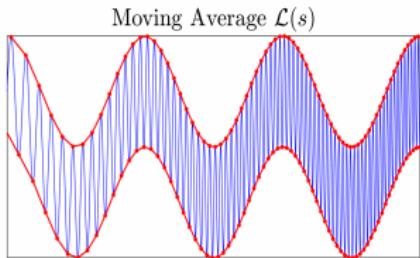
Decomposition of non-stationary signals into Intrinsic Mode Functions (IMF)

- Iterative Method
- Based on the computation of the moving average of the signal
- Splits the signal into simple oscillatory components

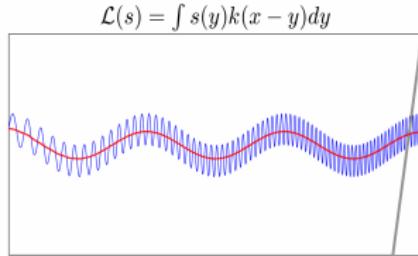
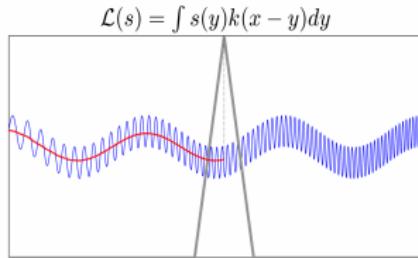
Numerous variants (EEMD, NA-MEMD, FMEMD, etc.) have been proposed in the years to deal with instability and mode splitting/mixing, and to prove its convergence

Empirical Mode Decomposition

The effect of the moving average is to flatten the highest frequency component



A way to emulate the effect is to use a filter on the signal



Iterative Filtering

Choose the filter k :

- Unit-norm, even, nonnegative and compact supported
- $k = \omega * \omega$
- Smooth

The IF method iteratively apply the filter through convolution

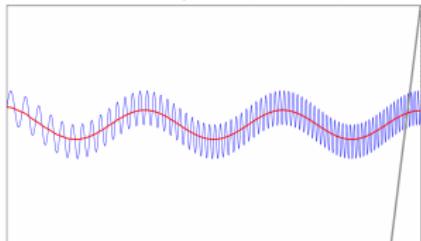
$$\mathcal{S}(f) := f(x) - \int f(y)k(x-y)dy$$

$$IMF = IMF \cup \{\mathcal{S}^\infty(s)\}$$

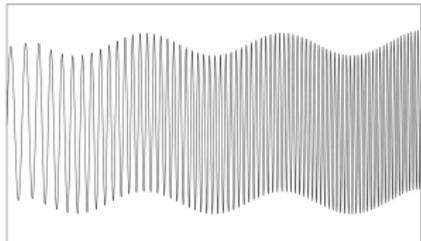
$$s = s - \mathcal{S}^\infty(s)$$

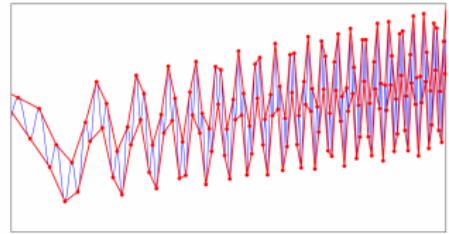
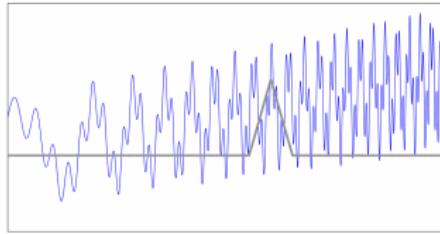
$\mathcal{S}^\infty(s)$ always converges and the method is fast (cyclic matrix, FFT), but it is not as flexible as EMD...

$$\mathcal{L}(s) = \int s(y)k(x-y)dy$$

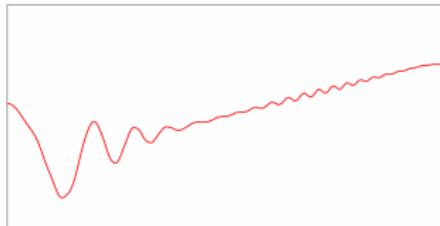


$$\mathcal{S}(s) = s(x) - \int s(y)k(x-y)dy$$

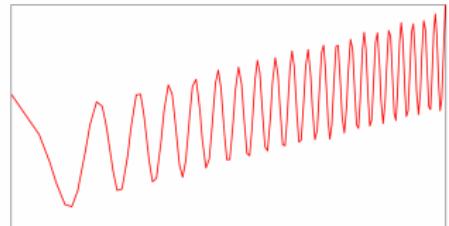




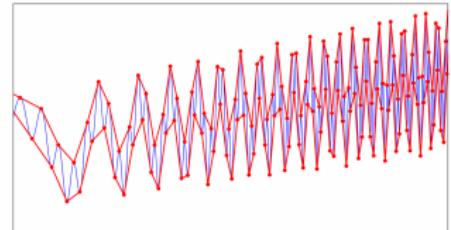
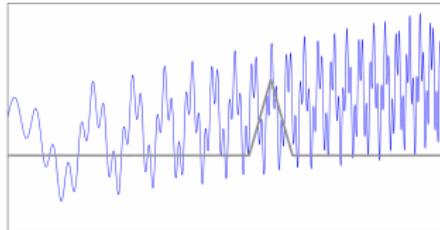
$$\mathcal{L}(s) = \int s(y)k(x-y)dy$$



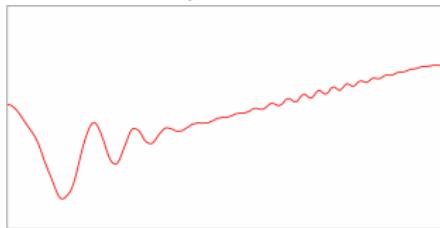
$$\text{EMD } \mathcal{L}(s)$$



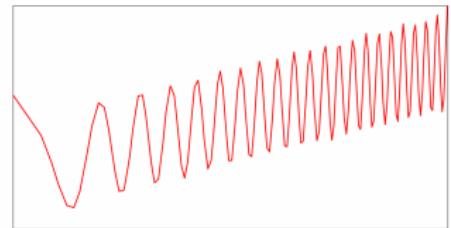
Let's take a look at the instantaneous frequencies



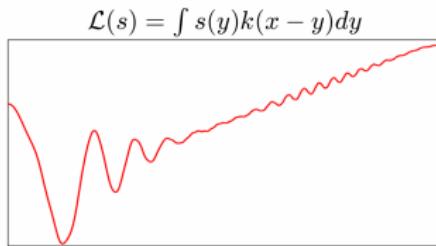
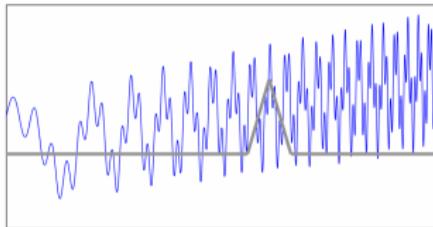
$$\mathcal{L}(s) = \int s(y)k(x-y)dy$$



$$\text{EMD } \mathcal{L}(s)$$



Let's take a look at the instantaneous frequencies



Instantaneous Frequencies

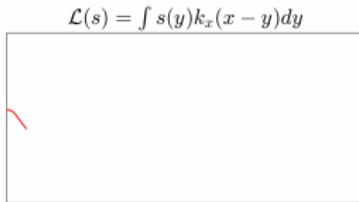
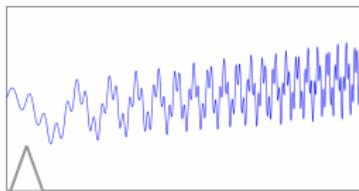


$$\hat{\mathcal{L}}(s) = \hat{s}(y) \cdot \hat{k}(y)$$

IF does not work with non-disjoint bands of frequencies

Adaptive Local Iterative Filtering

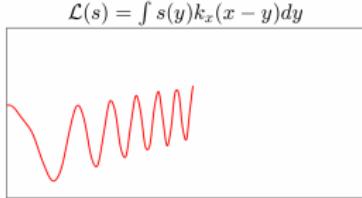
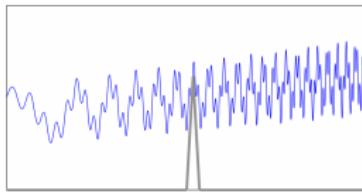
$$k_x(y) := k(\ell(x)^{-1}y)\ell(x)^{-1}$$



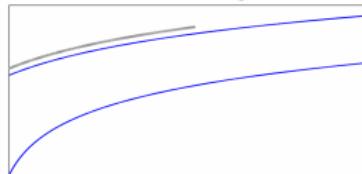
Instantaneous Frequencies



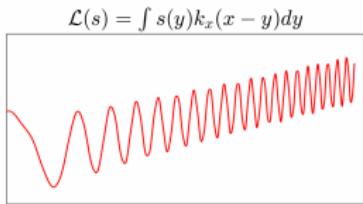
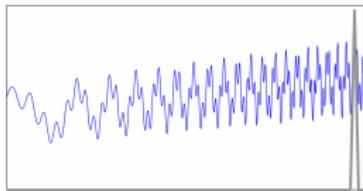
$$k_x(y) := k(\ell(x)^{-1}y)\ell(x)^{-1}$$



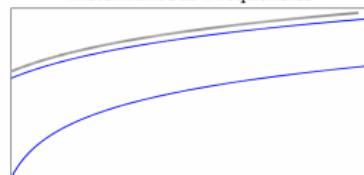
Instantaneous Frequencies



$$k_x(y) := k(\ell(x)^{-1}y)\ell(x)^{-1}$$



Instantaneous Frequencies



Adaptive Local Iterative Filtering

Given the signal $s(x)$, fix the filter

$$k_x(y) := k(\ell(x)^{-1}y)\ell(x)^{-1}$$

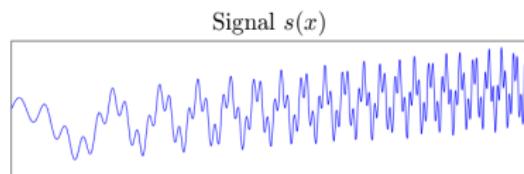
and apply iteratively the filter through convolution

$$\mathcal{S}(f) := f(x) - \int f(y)k_x(x-y)dy$$

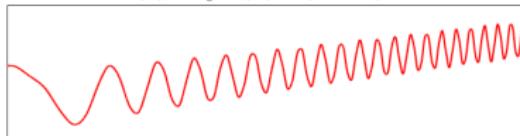
$$IMF = IMF \cup \{\mathcal{S}^\infty(s)\}$$

$$s = s - \mathcal{S}^\infty(s)$$

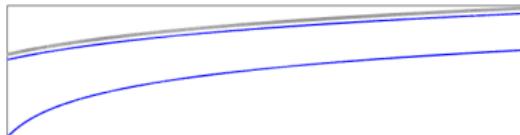
ALIF is now as flexible as EMD, and empirically converges, but..



$$\mathcal{L}(s) = \int s(y)k_x(x-y)dy$$



Instantaneous Frequencies



- No structure, not fast as IF ($O(n^2)$ against $O(n)$)
- Has no clean formal analysis
- $\mathcal{S}^\infty(s)$ is not always convergent (in the discrete setting)

Discrete Setting

$$\mathbf{s} = [s(h) \ s(2h) \ \dots \ s(1-h) \ s(1)] \quad h = 1/N$$

$$s(x) - \int_0^1 s(y) k_x(x-y) dy|_{x=ah} \sim \mathbf{s}_a - \frac{1}{N} \sum_{b=1}^N k\left(\frac{(a-b)h}{\ell(ah)}\right) \frac{1}{\ell(ah)} \mathbf{s}_b$$

$$\mathcal{S}(\mathbf{s}) := \mathbf{s} - Ms = (I - M)\mathbf{s}$$

- $\mathcal{S}^\infty(\mathbf{s})$ converges when

$$|\lambda_i(I - M)| < 1 \vee \lambda_i(I - M) = 1$$

- Converges to the kernel of M

The kernel is the same in αM where $\alpha \in \mathbb{R}$, so the real condition is

$$\Im(\lambda_i(M)) > 0 \vee \lambda_i(M) = 0$$

Setting a stopping condition in the iteration makes $\mathcal{S}^\infty(\mathbf{s})$ a near-kernel vector

For big enough N and if $\ell(x)$ is continuous, positive and

$$k(x) = \omega(x) * \omega(x),$$

then the spectrum of M respects the condition for almost every eigenvalue

There are artificial examples where M has negative eigenvalues, so the convergence is not always assured

Discrete Setting

$$\mathbf{s} = [s(h) \ s(2h) \ \dots \ s(1-h) \ s(1)] \quad h = 1/N$$

$$s(x) - \int_0^1 s(y) k_x(x-y) dy|_{x=ah} \sim \mathbf{s}_a - \frac{1}{N} \sum_{b=1}^N k\left(\frac{(a-b)h}{\ell(ah)}\right) \frac{1}{\ell(ah)} \mathbf{s}_b$$

$$\mathcal{S}(\mathbf{s}) := \mathbf{s} - M\mathbf{s} = (I - M)\mathbf{s}$$

- $\mathcal{S}^\infty(\mathbf{s})$ converges when

$$|\lambda_i(I - M)| < 1 \vee \lambda_i(I - M) = 1$$

- Converges to the kernel of M

The kernel is the same in αM where $\alpha \in \mathbb{R}$, so the real condition is

$$\Im(\lambda_i(M)) > 0 \vee \lambda_i(M) = 0$$

Setting a stopping condition in the iteration makes $\mathcal{S}^\infty(\mathbf{s})$ a near-kernel vector

For big enough N and if $\ell(x)$ is continuous, positive and

$$k(x) = \omega(x) * \omega(x),$$

then the spectrum of M respects the condition for almost every eigenvalue

There are artificial examples where M has negative eigenvalues, so the convergence is not always assured

Discrete Setting

$$\mathbf{s} = [s(h) \ s(2h) \ \dots \ s(1-h) \ s(1)] \quad h = 1/N$$

$$s(x) - \int_0^1 s(y) k_x(x-y) dy|_{x=ah} \sim \mathbf{s}_a - \frac{1}{N} \sum_{b=1}^N k\left(\frac{(a-b)h}{\ell(ah)}\right) \frac{1}{\ell(ah)} \mathbf{s}_b$$

$$\mathcal{S}(\mathbf{s}) := \mathbf{s} - M\mathbf{s} = (I - M)\mathbf{s}$$

- $\mathcal{S}^\infty(\mathbf{s})$ converges when

$$|\lambda_i(I - M)| < 1 \vee \lambda_i(I - M) = 1$$

- Converges to the kernel of M

The kernel is the same in αM where $\alpha \in \mathbb{R}$, so the real condition is

$$\Im(\lambda_i(M)) > 0 \vee \lambda_i(M) = 0$$

Setting a stopping condition in the iteration makes $\mathcal{S}^\infty(\mathbf{s})$ a near-kernel vector

For big enough N and if $\ell(x)$ is continuous, positive and

$$k(x) = \omega(x) \star \omega(x),$$

then the spectrum of M respects the condition for almost every eigenvalue

There are artificial examples where M has negative eigenvalues, so the convergence is not always assured

Stable ALIF

Given the ALIF matrix M , let

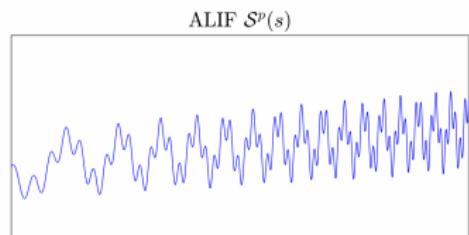
$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

$N = 3000$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



Stable ALIF

Given the ALIF matrix M , let

$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

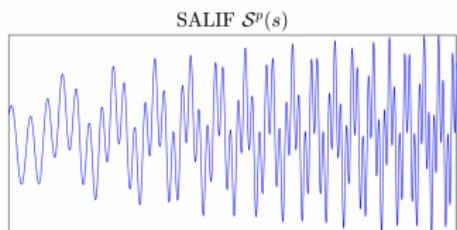
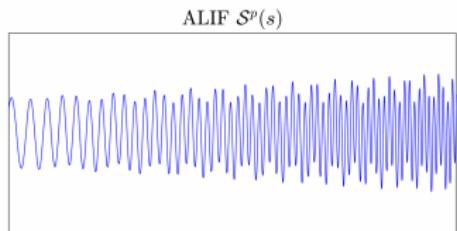
$$T = 1/20$$

$$N = 3000$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



Stable ALIF

Given the ALIF matrix M , let

$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

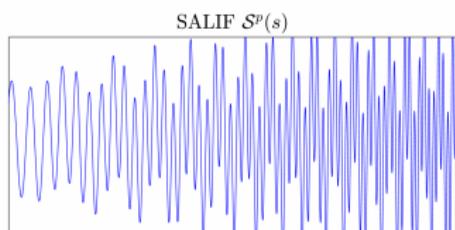
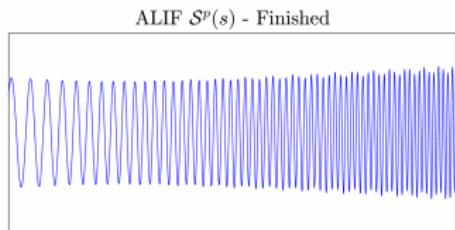
$$T = 2/20$$

$$N = 3000$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



Stable ALIF

Given the ALIF matrix M , let

$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

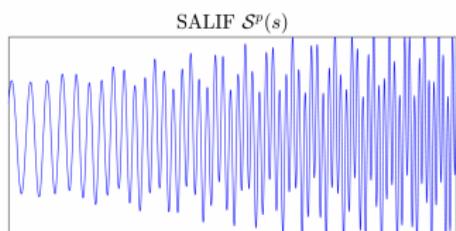
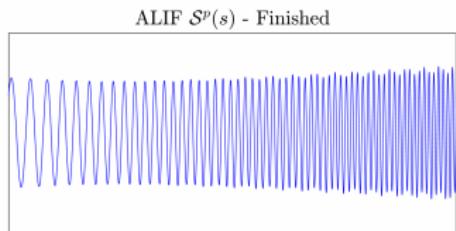
$$T = 3/20$$

$$N = 3000$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



Stable ALIF

Given the ALIF matrix M , let

$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

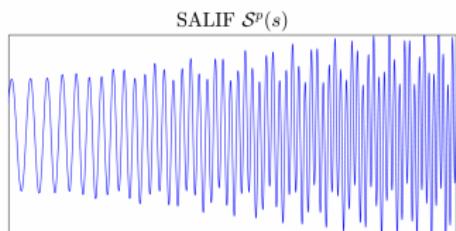
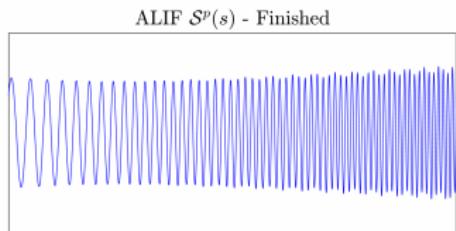
$$T = 4/20$$

$$N = 3000$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



Stable ALIF

Given the ALIF matrix M , let

$$\mathcal{S}(s) := s - M^T M s = (I - M^T M)s$$

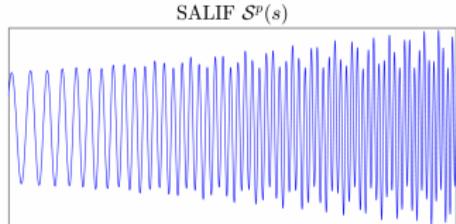
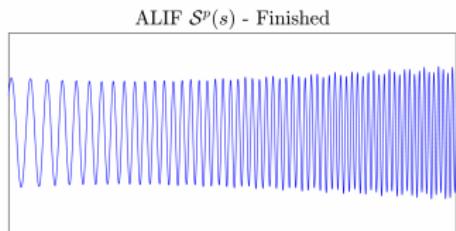
$$T = 5/20$$

$$N = 3000$$

- $M^T M$ Has the same kernel of M
- $\lambda_i(M^T M) \geq 0$

As a consequence, $\mathcal{S}^\infty(s)$ always converges, but the method is way slower

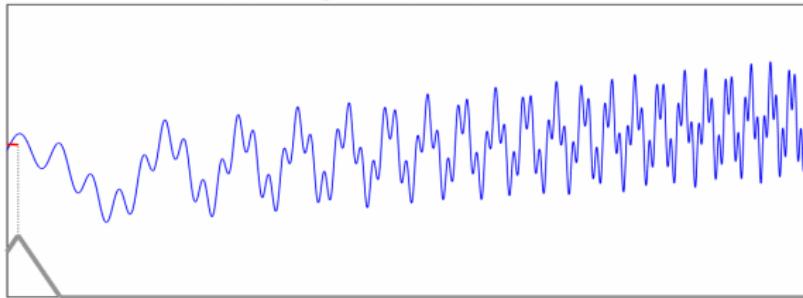
- The cost per iteration is doubled
- There are more eigenvalues close to zero, so it takes more iterations to extract the exact component



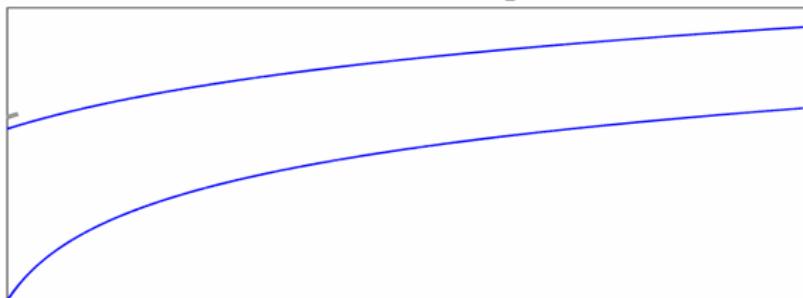
Resampled Iterative Filtering

ALIF

$$\mathcal{L}(s) = \int s(y)k_x(x - y)dy$$

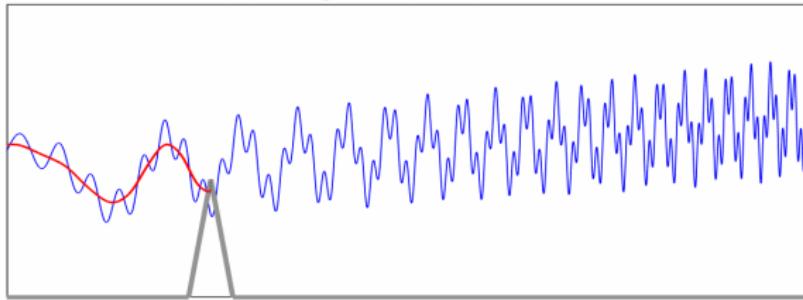


Instantaneous Frequencies

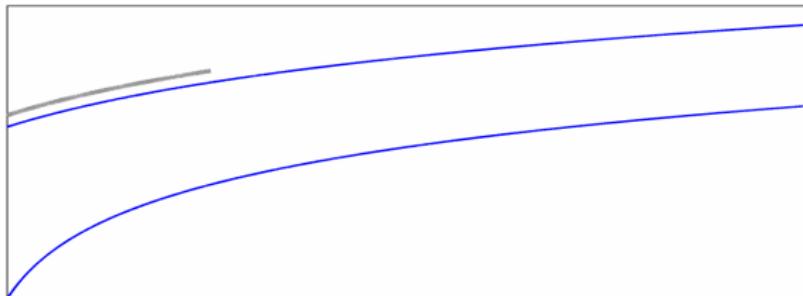


ALIF

$$\mathcal{L}(s) = \int s(y)k_x(x - y)dy$$

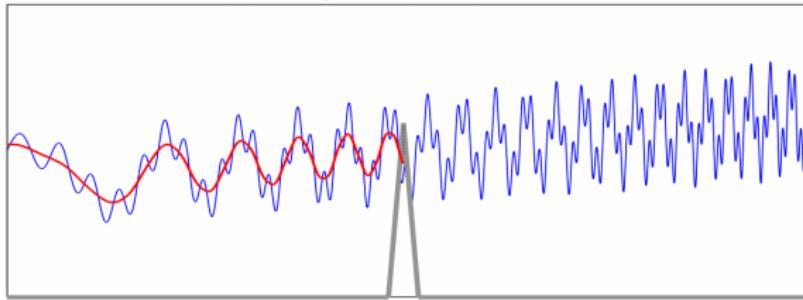


Instantaneous Frequencies

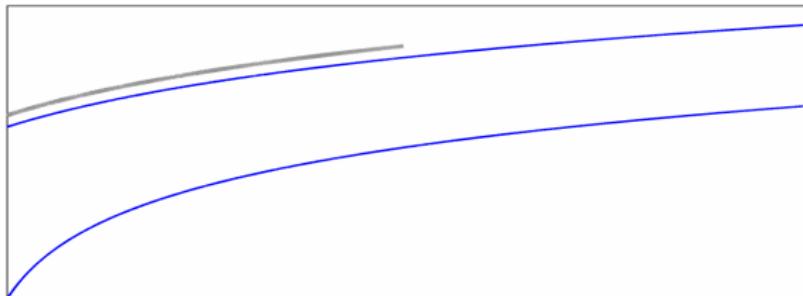


ALIF

$$\mathcal{L}(s) = \int s(y)k_x(x - y)dy$$

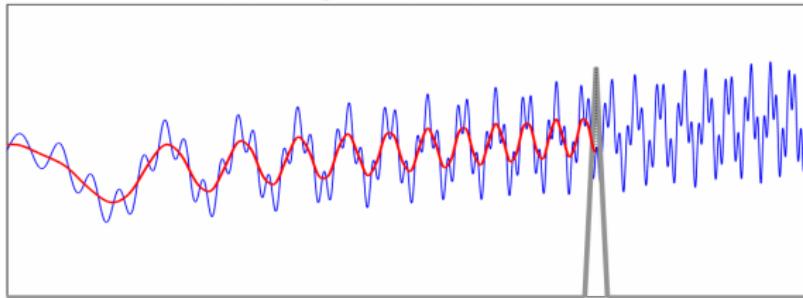


Instantaneous Frequencies

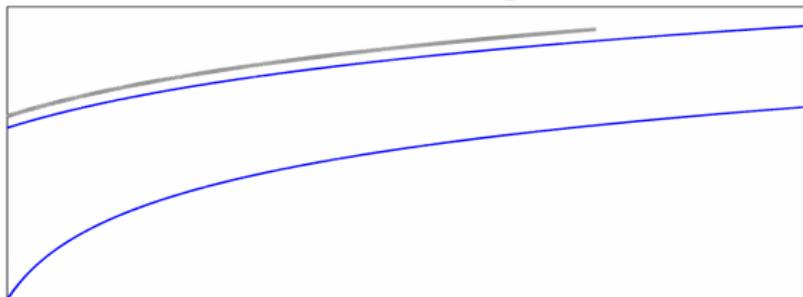


ALIF

$$\mathcal{L}(s) = \int s(y)k_x(x - y)dy$$

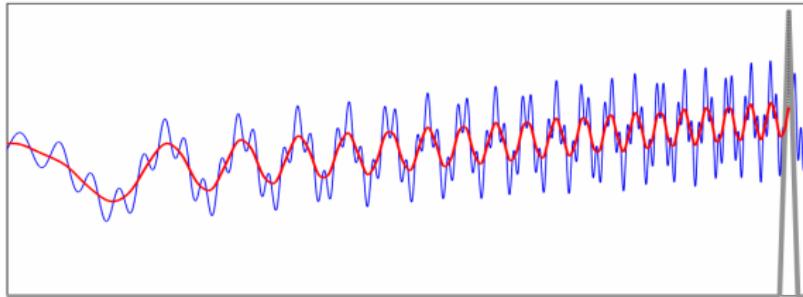


Instantaneous Frequencies

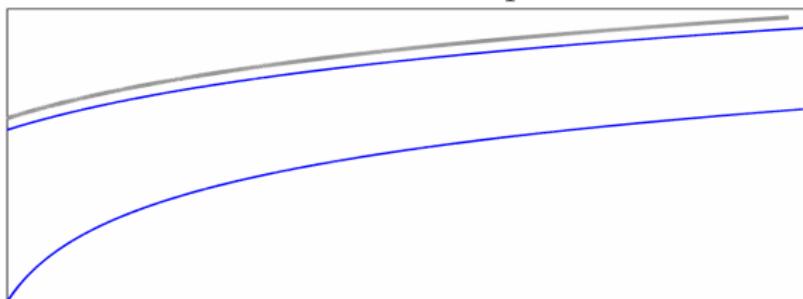


ALIF

$$\mathcal{L}(s) = \int s(y)k_x(x - y)dy$$

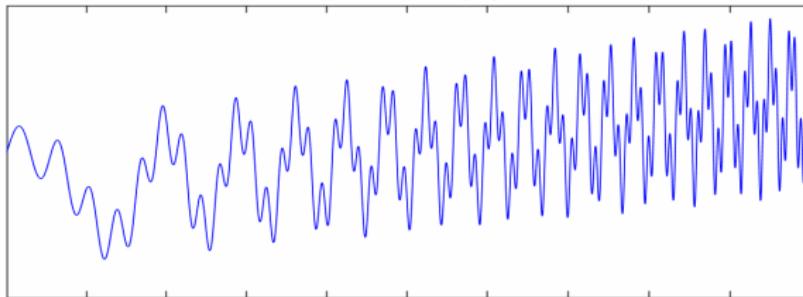


Instantaneous Frequencies

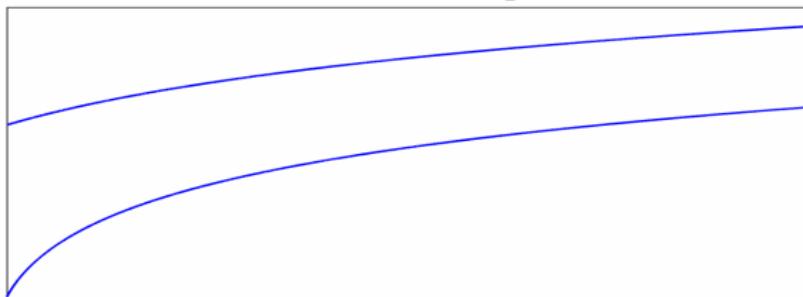


Resampling

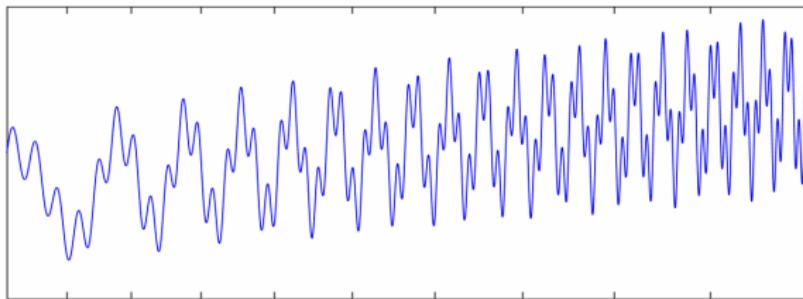
Signal $s(x)$



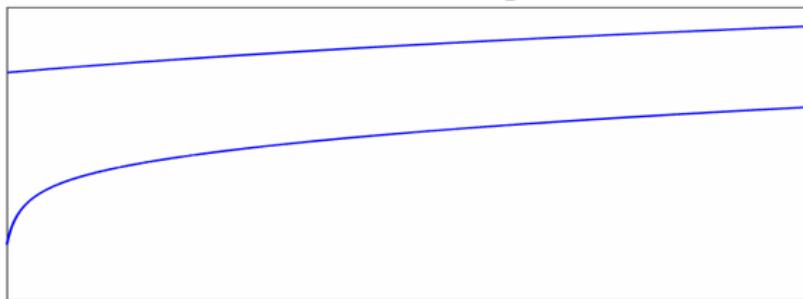
Instantaneous Frequencies



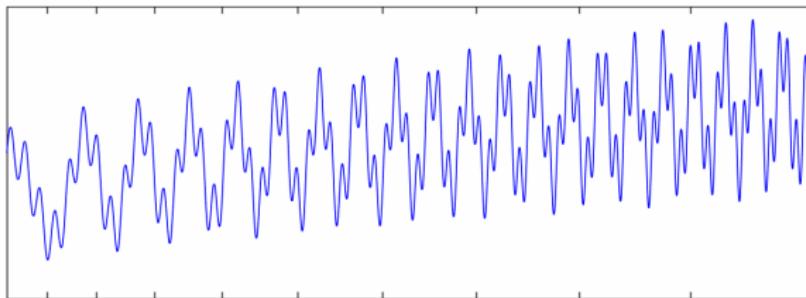
Resampling



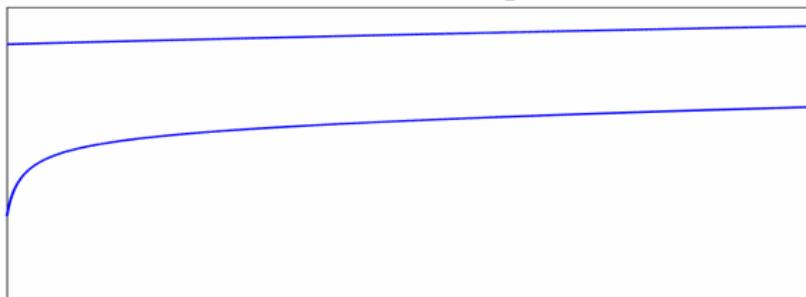
Instantaneous Frequencies



Resampling

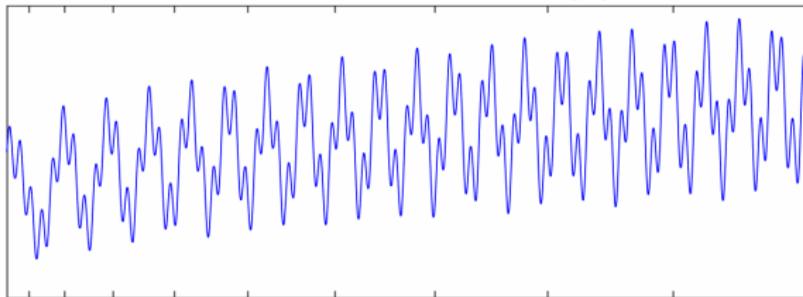


Instantaneous Frequencies



Resampling

Resampled Signal $s_r(x)$

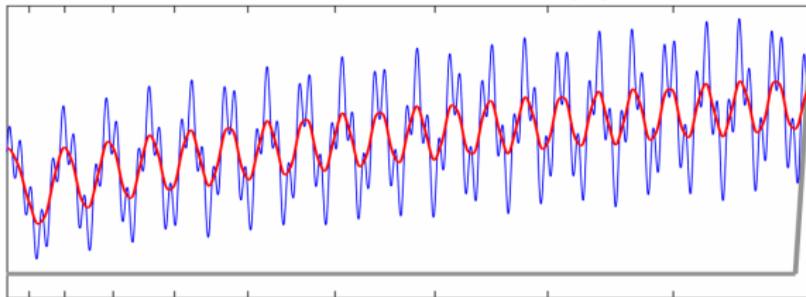


Instantaneous Frequencies



Resampling

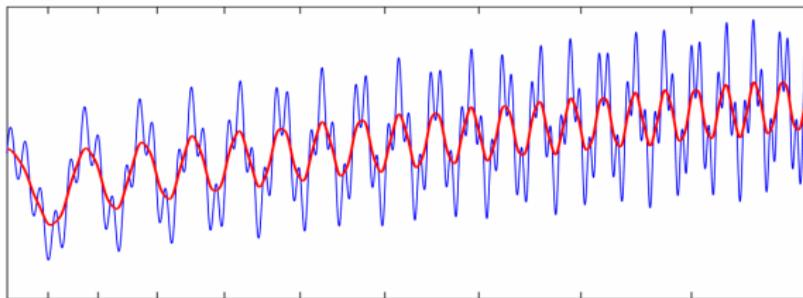
Resampled Signal $s_r(x)$



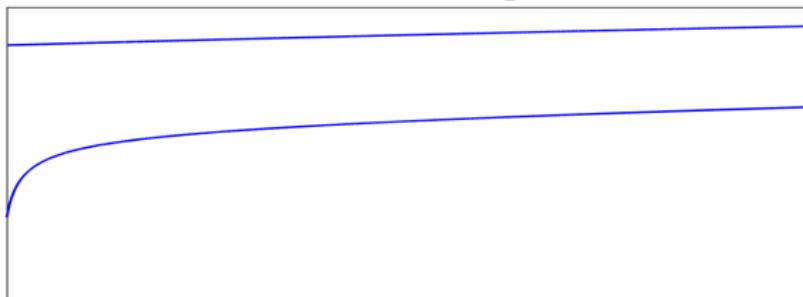
Instantaneous Frequencies



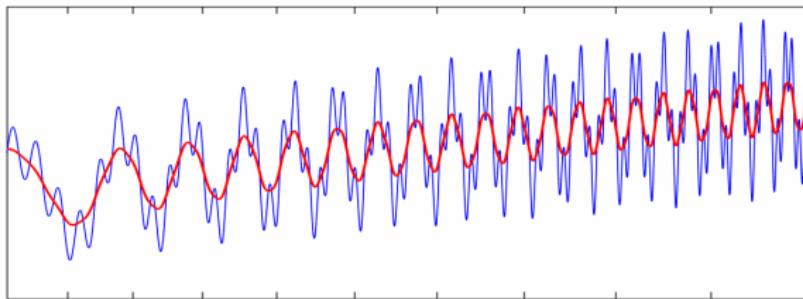
Resampling



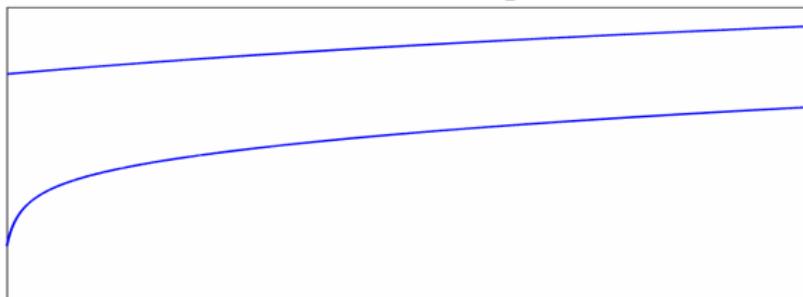
Instantaneous Frequencies



Resampling

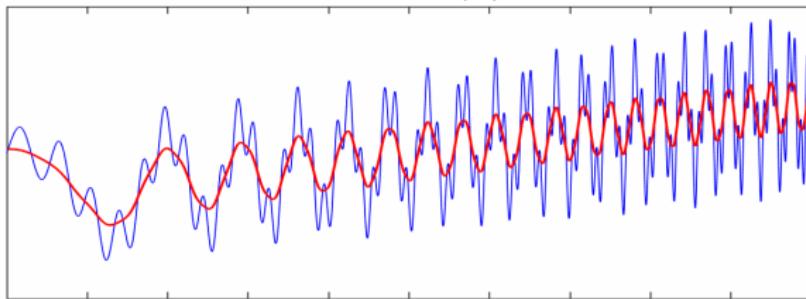


Instantaneous Frequencies

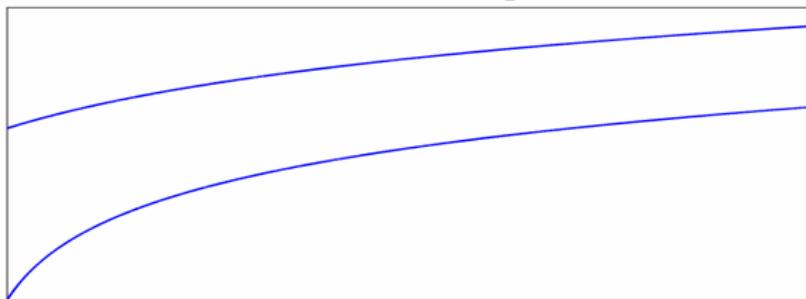


Resampling

Signal $s(x)$



Instantaneous Frequencies



Resampling Function $G(x)$

ALIF: $\mathcal{S}(s)(x) := s(x) - \int s(y)k\left(\frac{x-y}{\ell(x)}\right)\frac{1}{\ell(x)}dy$

$$t = (x - y)/\ell(x) \quad x = G(z)$$

$$\mathcal{S}(s)(G(z)) := s(G(z)) - \int s(G(z) - t\ell(G(z)))k(t)dt$$

$$G'(z) = \ell(G(z)) \quad G(z-t) \sim G(z) - tG'(z)$$

RIF: $\mathcal{S}(s)(G(y)) := s(G(y)) - \int s(G(z-t))k(z)dz$

ALIF is a "first order" RIF where

$$G^{-1}(z) = \int_0^z \frac{1}{\ell(x)}dx$$

ALIF \neq RIF except when $\ell(x) = \ell$ and both are IF

Resampling Function $G(x)$

ALIF: $\mathcal{S}(s)(x) := s(x) - \int s(y)k\left(\frac{x-y}{\ell(x)}\right)\frac{1}{\ell(x)}dy$

$$t = (x - y)/\ell(x) \quad x = G(z)$$

$$\mathcal{S}(s)(G(z)) := s(G(z)) - \int s(G(z) - t\ell(G(z)))k(t)dt$$

$$G'(z) = \ell(G(z)) \quad G(z-t) \sim G(z) - tG'(z)$$

RIF: $\mathcal{S}(s)(G(y)) := s(G(y)) - \int s(G(z-t))k(z)dz$

ALIF is a "first order" RIF where

$$G^{-1}(z) = \int_0^z \frac{1}{\ell(x)}dx$$

ALIF \neq RIF except when $\ell(x) = \ell$ and both are IF

Resampling Function $G(x)$

ALIF: $\mathcal{S}(s)(x) := s(x) - \int s(y)k\left(\frac{x-y}{\ell(x)}\right) \frac{1}{\ell(x)} dy$

$$t = (x - y)/\ell(x) \quad x = G(z)$$

$$\mathcal{S}(s)(G(z)) := s(G(z)) - \int s(G(z) - t\ell(G(z)))k(t) dt$$

$$G'(z) = \ell(G(z)) \quad G(z-t) \sim G(z) - tG'(z)$$

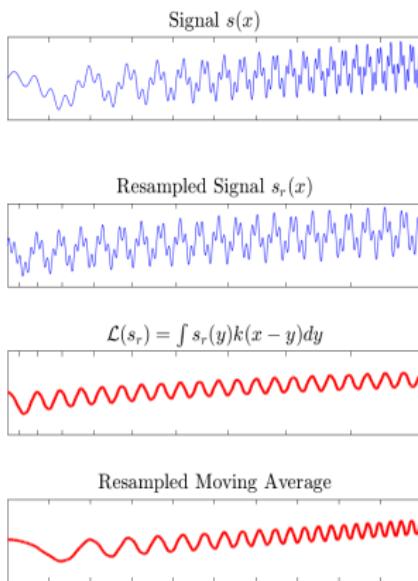
RIF: $\mathcal{S}(s)(G(y)) := s(G(y)) - \int s(G(z-t))k(z) dz$

ALIF is a "first order" RIF where

$$G^{-1}(z) = \int_0^z \frac{1}{\ell(x)} dx$$

ALIF \neq RIF except when $\ell(x) = \ell$ and both are IF

Resampled Iterative Filtering



Given the signal $s(x)$, compute the resampling

$$s_r(x) := s(G(x)) \quad G^{-1}(z) = \int_0^z \frac{1}{\ell(x)} dx$$

and apply iteratively the filter through convolution

$$\begin{aligned} \mathcal{S}(f) &:= f(x) - \int f(y)k(x - y)dy \\ IMF &= IMF \cup \{\mathcal{S}^\infty(s_r)(G^{-1}(x))\} \\ s &= s - \mathcal{S}^\infty(s_r)(G^{-1}(x)) \end{aligned}$$

At the cost of two interpolations per IMF, we have an algorithm that is

- As flexible as ALIF
- Fast as IF, the resampling is outside the iterations
- $\mathcal{S}^\infty(s_r)$ is always convergent

Numerical Experiments

Experiment 1

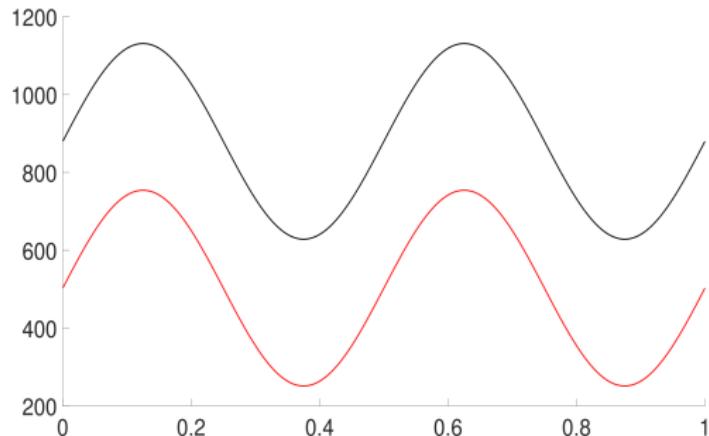
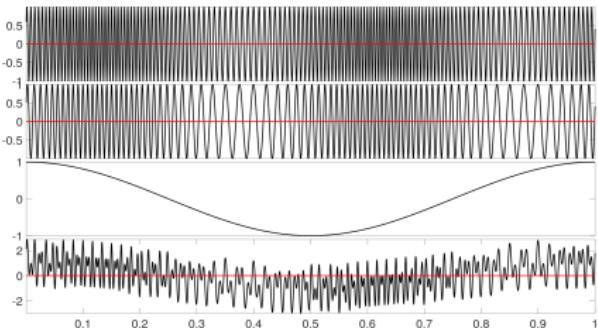
$$N = 8000$$

$$h_1(x) = \cos(20 \cos(4\pi t) - 160\pi t)$$

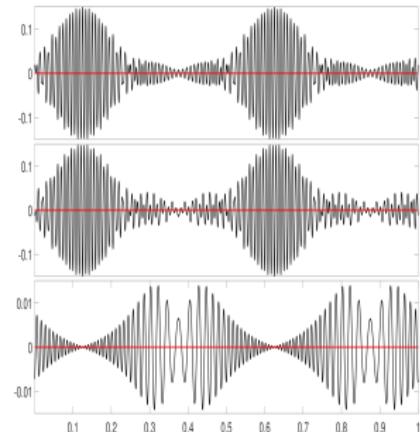
$$h_2(x) = \cos(20 \cos(4\pi t) - 280\pi t)$$

$$h_3(x) = \cos(2\pi t)$$

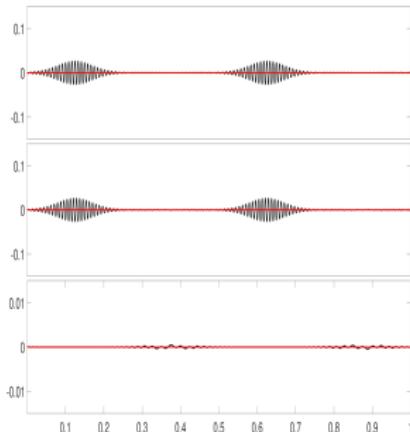
$$h(x) = h_1(x) + h_2(x) + h_3(x)$$



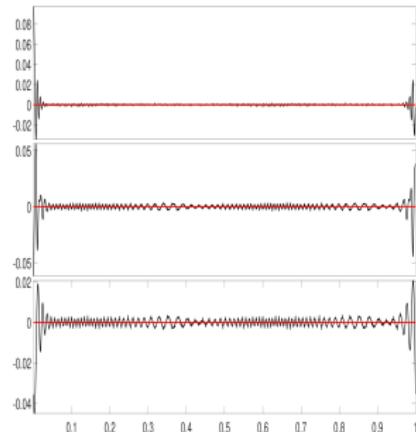
ALIF



SALIF

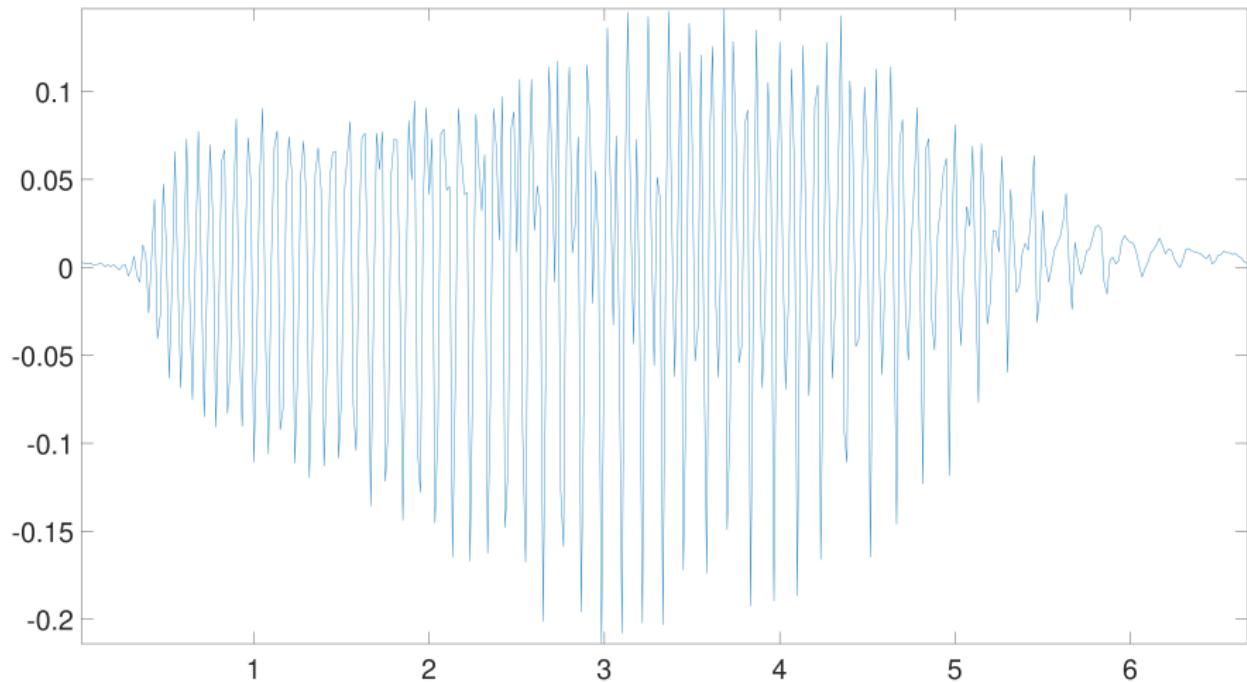


RIF

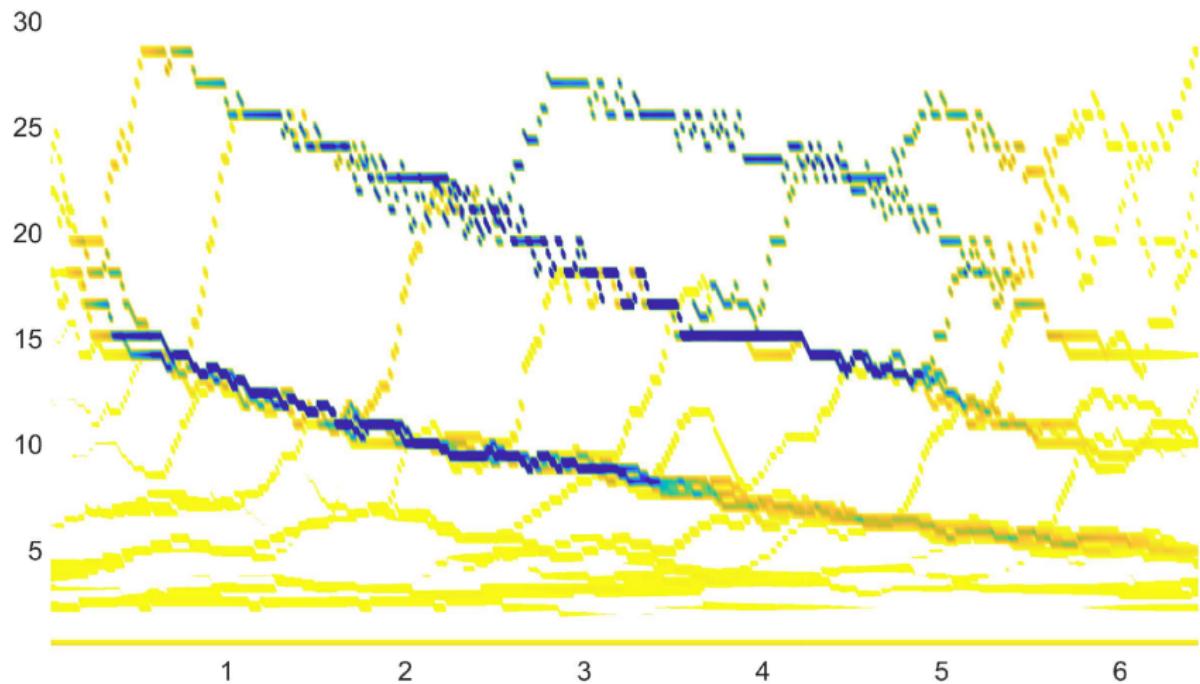


	Time	err1	err2	err3	Niter1	Niter2
ALIF	4.0860	0.070388	0.071158	0.008549	18	2
SALIF	19.7919	0.010054	0.010055	0.000161	353	5
RIF	1.4724	0.003426	0.003292	0.000908	81	11

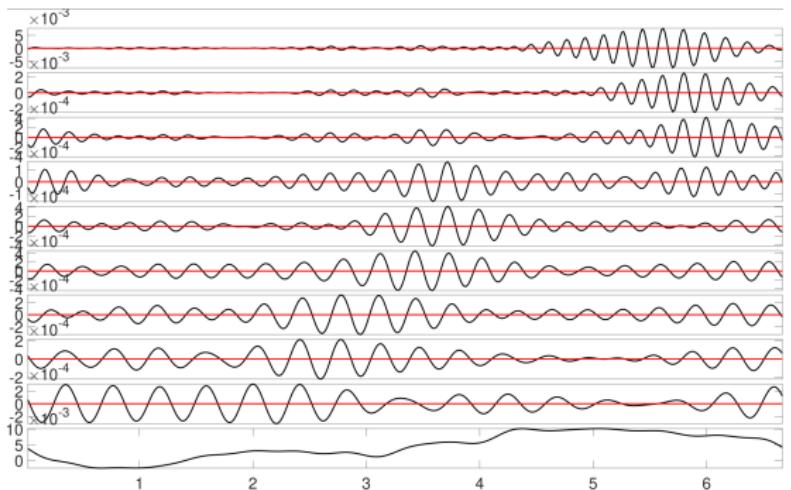
Experiment 2



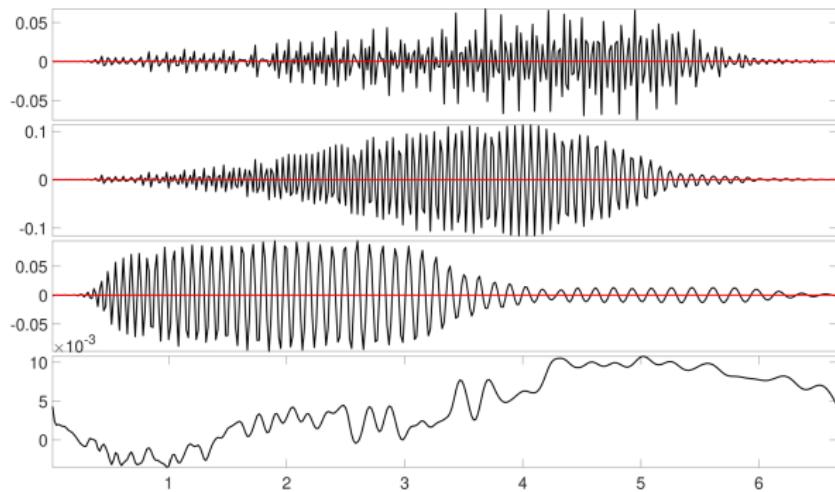
Experiment 2



IF



RIF



Future Work

- More involved analysis (borders, length regularity, etc.)
- Multidimensional and Multi-Signals methods
- Direct computation of re-sampling
- Comparison with Synchroqueezing
- RIF/ALIF as denoising methods

Thank You!

- ❑ J. Liu A. Cicone and H. Zhou. **Adaptive local iterative filtering for signal decomposition and instantaneous frequency analysis.** *Applied and Computational Harmonic Analysis*, 41(2), 2016.
- ❑ S. Serra-Capizzano A. Cicone, C. Garoni. **Spectral and convergence analysis of the discrete alif method.** *Linear Algebra and its Applications*, 580, 2019.
- ❑ A. Cicone G. Barbarino. **Conjectures on spectral properties of alif algorithm**, <https://arxiv.org/abs/2009.00582>. 2020.
- ❑ A. Cicone G. Barbarino. **Stabilization and variations to the alif algorithm: the fast resampled iterative filtering method.** (In preparation).
- ❑ Y. Wang L. Lin and H. Zhou. **Iterative filtering as an alternative algorithm for empirical mode decomposition.** *Advances in Adaptive Data Analysis*, 1(4), 2009.
- ❑ S. R. Long M. C. Wu H. H. Shih Q. Zheng N.-C. Yen C. C. Tung N. E. Huang, Z. Shen and H. H. Liu. **The empirical mode decomposition and the hilbert spectrum for nonlinear and non-stationary time series analysis.** *Proceedings of the Royal Society of London. Series A: mathematical, physical and engineering sciences*, 454(1971), 1998.