

ENGR-E 511; ENGR-E 399

# “Machine Learning for Signal Processing”

Module 02: Lecture 01:

## Time-Frequency Transforms

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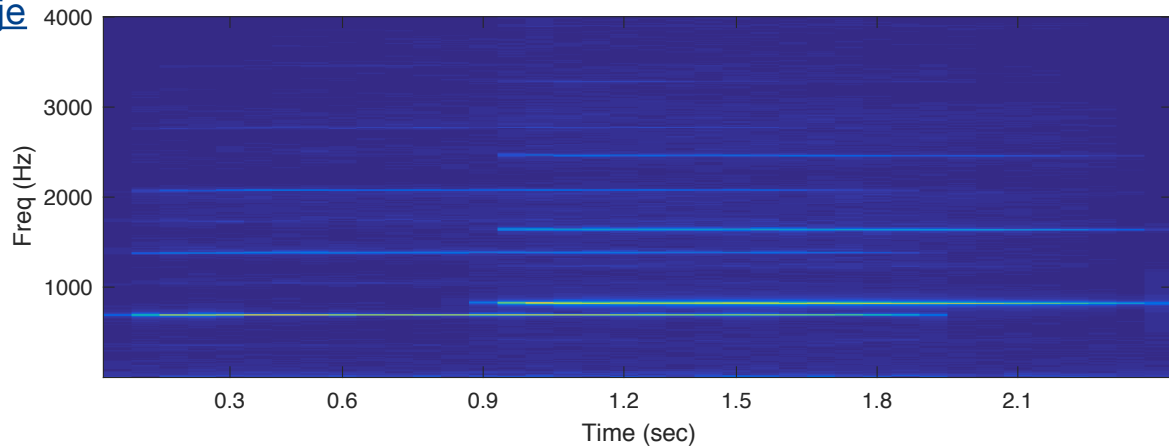
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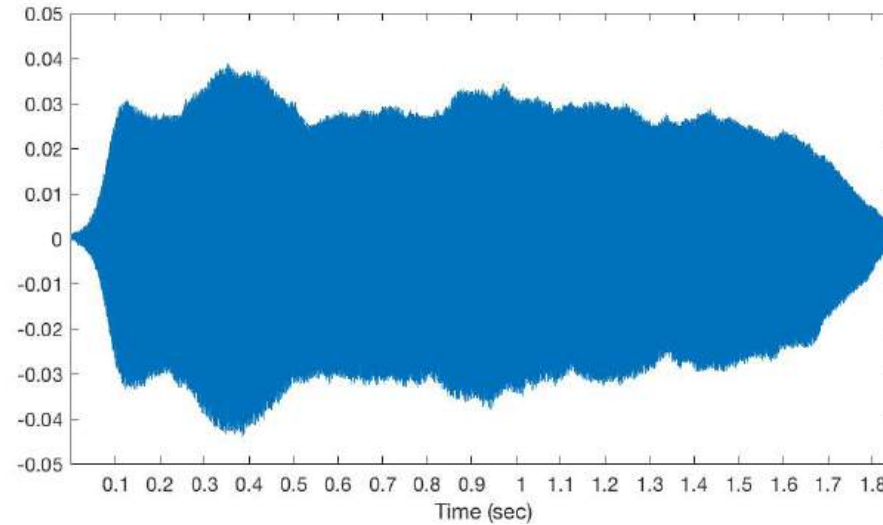
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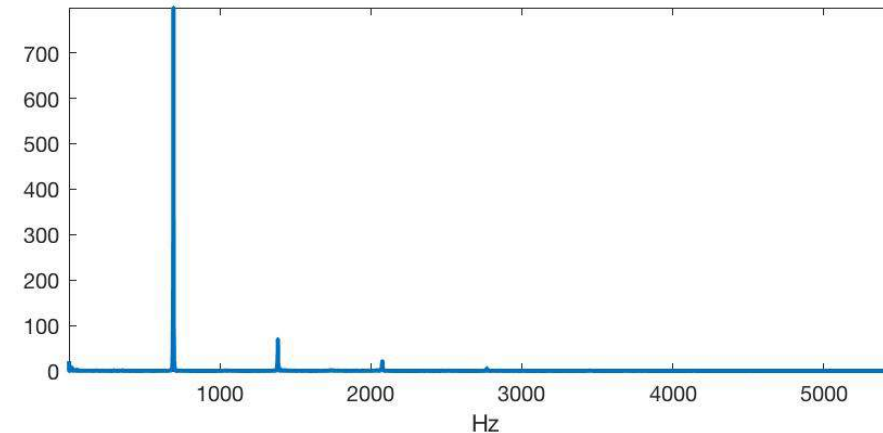
# Frequency Representation of Signals

## - Compactness

- How does it sound?



- It actually sounds like this: 
- A better way to represent the signal?
  - How about this?

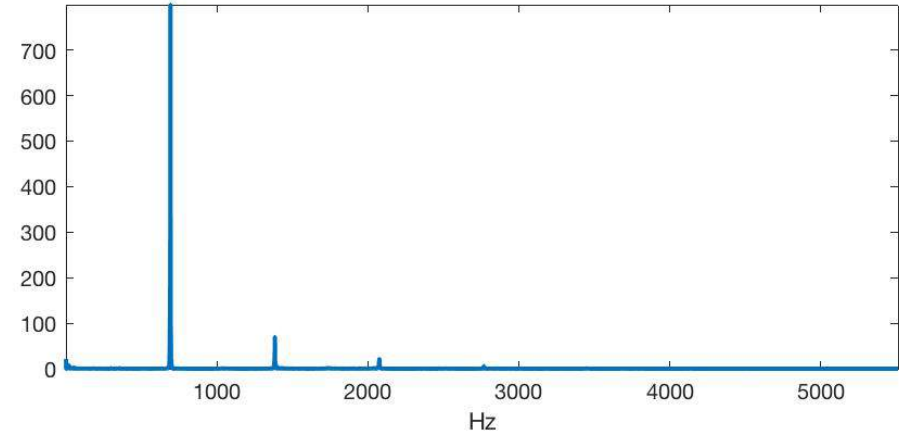
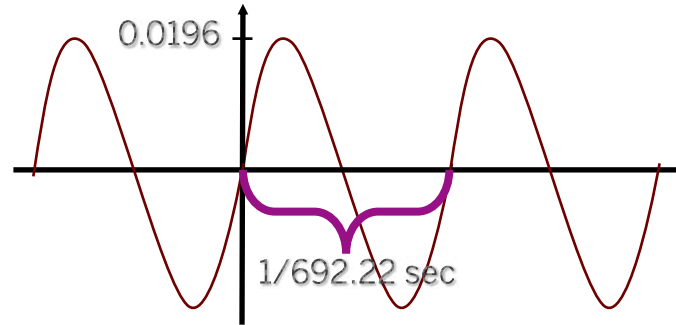


# Frequency Representation of Signals

## - Compactness

- I prefer the frequency representation for this signal
  - Because it says about the signal in a compact way
  - i.e. it's a mixture of four sine waves, whose frequencies and amplitudes are defined by the peaks in the spectrum

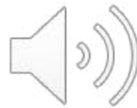
$$\begin{aligned} &0.0196 \sin(2\pi 692.22t) \\ &+ 0.0017 \sin(2\pi 1384.45t) \\ &+ 0.0005 \sin(2\pi 2077.21t) \\ &+ 0.0001 \sin(2\pi 2768.9t) \\ &+ \dots \end{aligned}$$



- This (manual) frequency analysis misses some other components, but is quite good



Reconstruction  
from 4 sinusoids

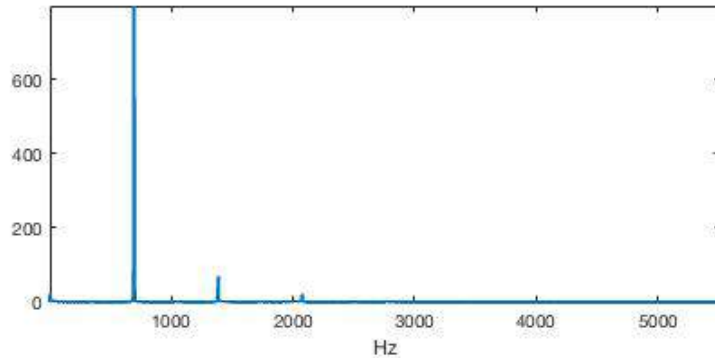
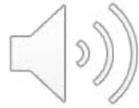


Original

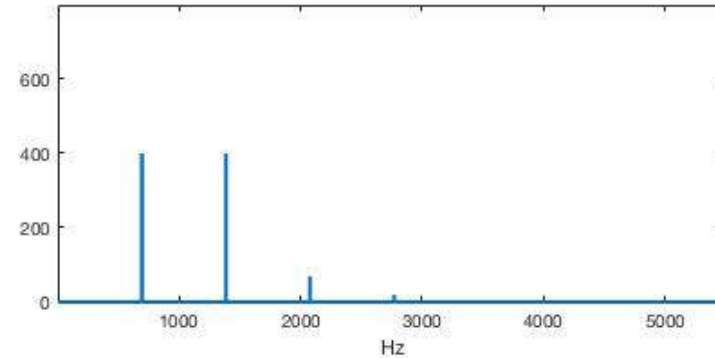
# Frequency Representation of Signals

## - Amplitudes of the basis functions

- Same sinusoids but with different amplitudes
- Height of the spectral peaks reflects amplitudes



$$\begin{aligned} &0.0196 \sin(2\pi 692.22t) \\ &+ 0.0017 \sin(2\pi 1384.45t) \\ &+ 0.0005 \sin(2\pi 2077.21t) \\ &+ 0.0001 \sin(2\pi 2768.9t) \end{aligned}$$

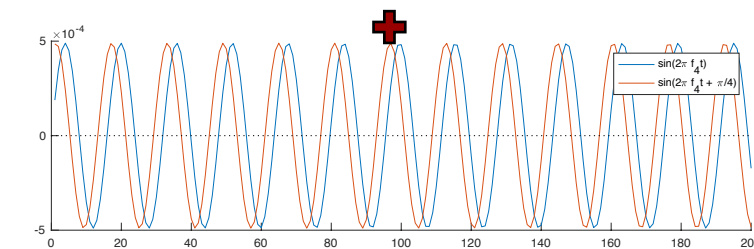
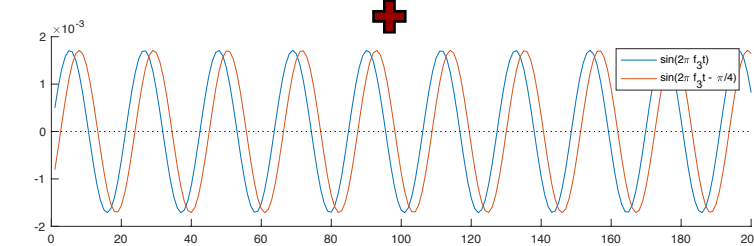
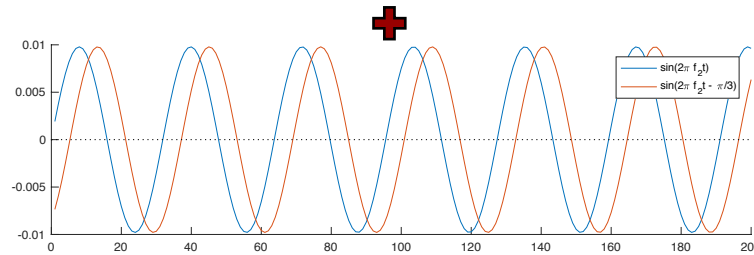
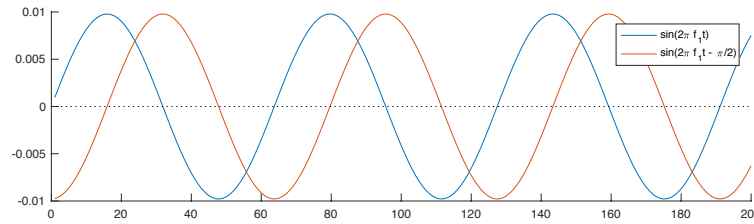


$$\begin{aligned} &0.0098 \sin(2\pi 692.22t) \\ &+ 0.0098 \sin(2\pi 1384.45t) \\ &+ 0.0017 \sin(2\pi 2077.21t) \\ &+ 0.0005 \sin(2\pi 2768.9t) \end{aligned}$$

- Any guess as to how they would sound?
  - They sound different!

# Frequency Representation of Signals

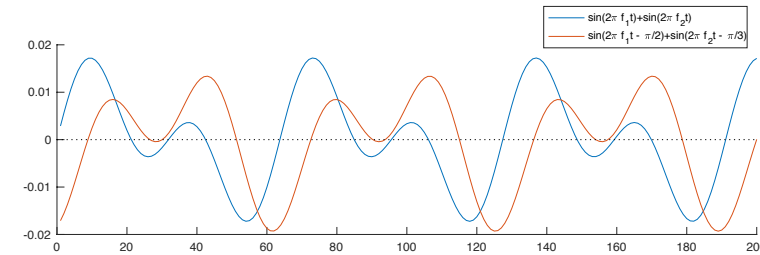
- Phase of the basis



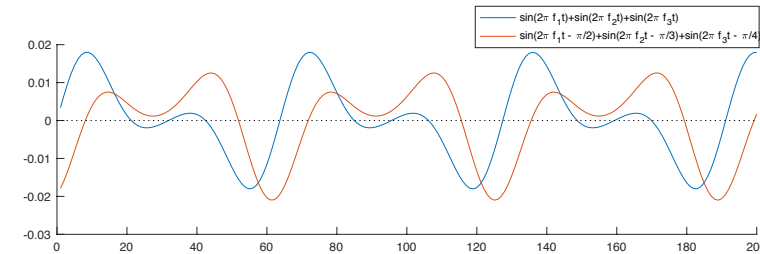
○ But they sound the same!



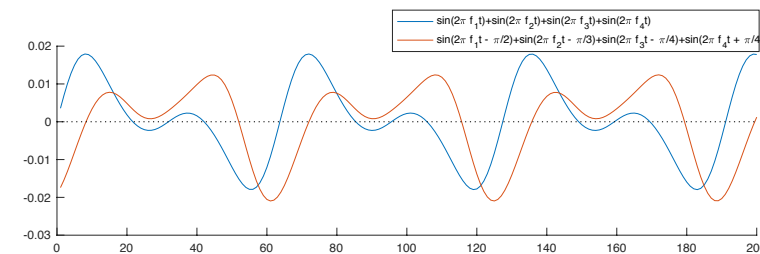
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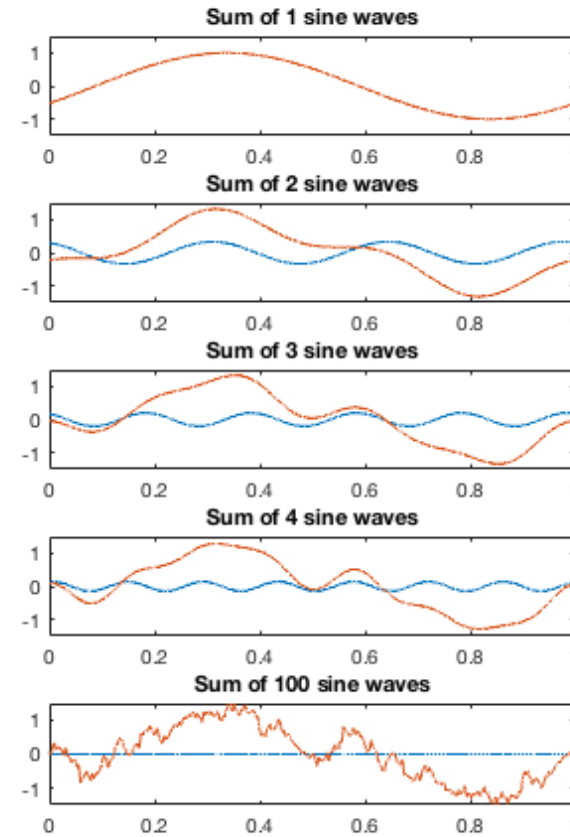
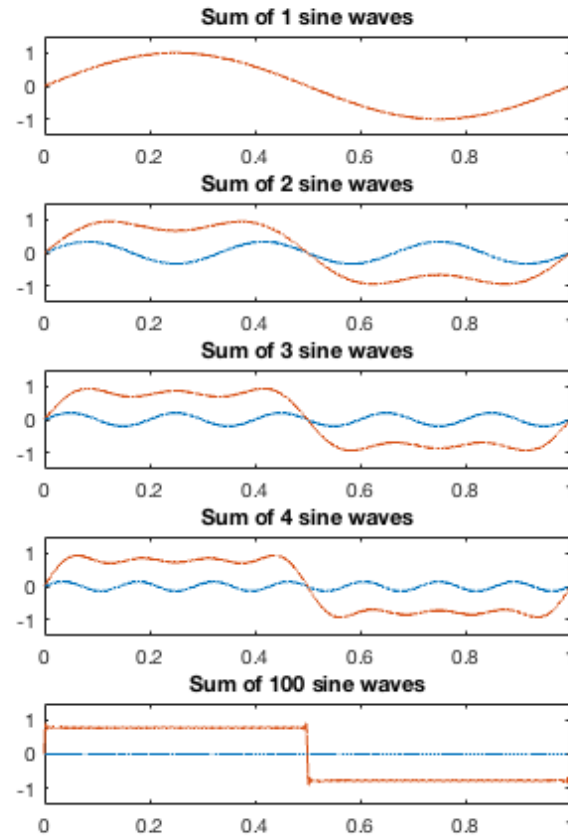


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# Frequency Representation of Signals

- Phase of the basis functions

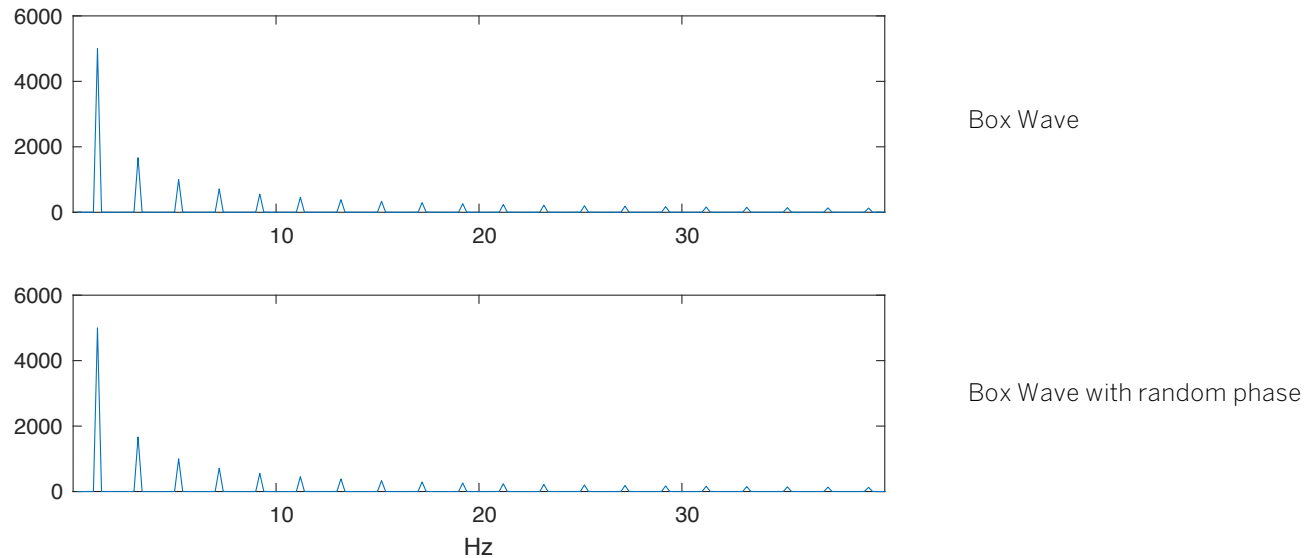


○ Now they sound different!

# Frequency Representation of Signals

## - Magnitude spectra

- So, the “magnitude” spectrum representation is somehow misleading



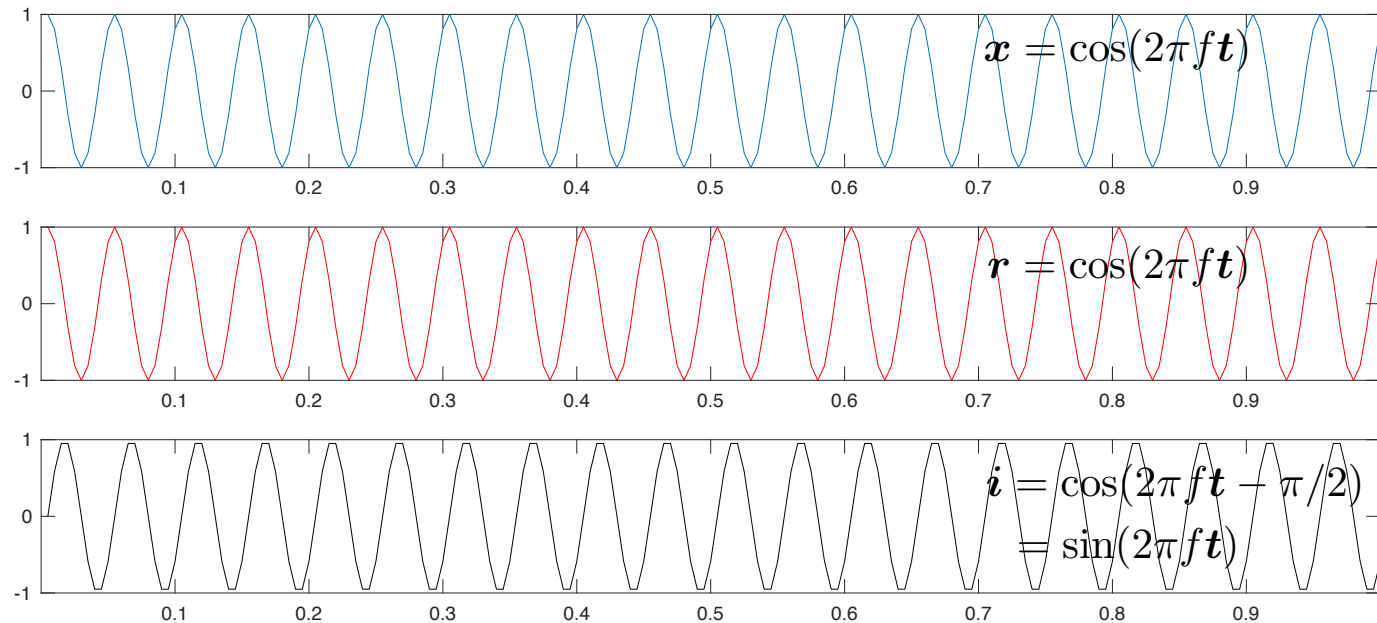
- The magnitudes of the box wave and its random-phase version are the same
- Representing a signal with a sum of sinusoids
  - Needs period, amplitude, and phase of each of the sinusoids
- How do we find this information?

# Discrete Fourier Transform

## - Basis vectors with different phase

- We are projecting a signal  $x$  onto basis vectors  $r$  and  $i$

$$x = \cos(2\pi ft) \quad f = 20\text{Hz}, \quad t = [0, 0.0001, 0.0002, \dots, 1]$$



$$x \cdot r = 100$$

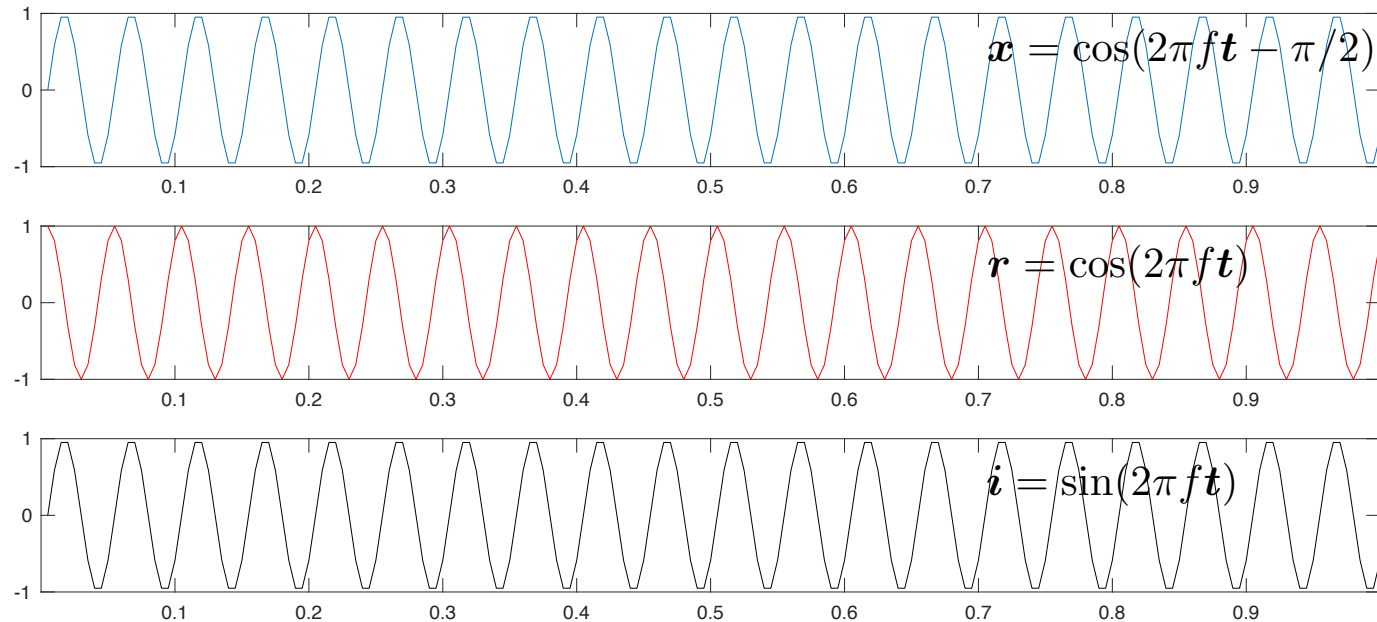
$$x \cdot i = 0$$



# Discrete Fourier Transform

- Basis vectors with different phase

$$f = 20\text{Hz}, \mathbf{t} = [0, 0.0001, 0.0002, \dots, 1]$$



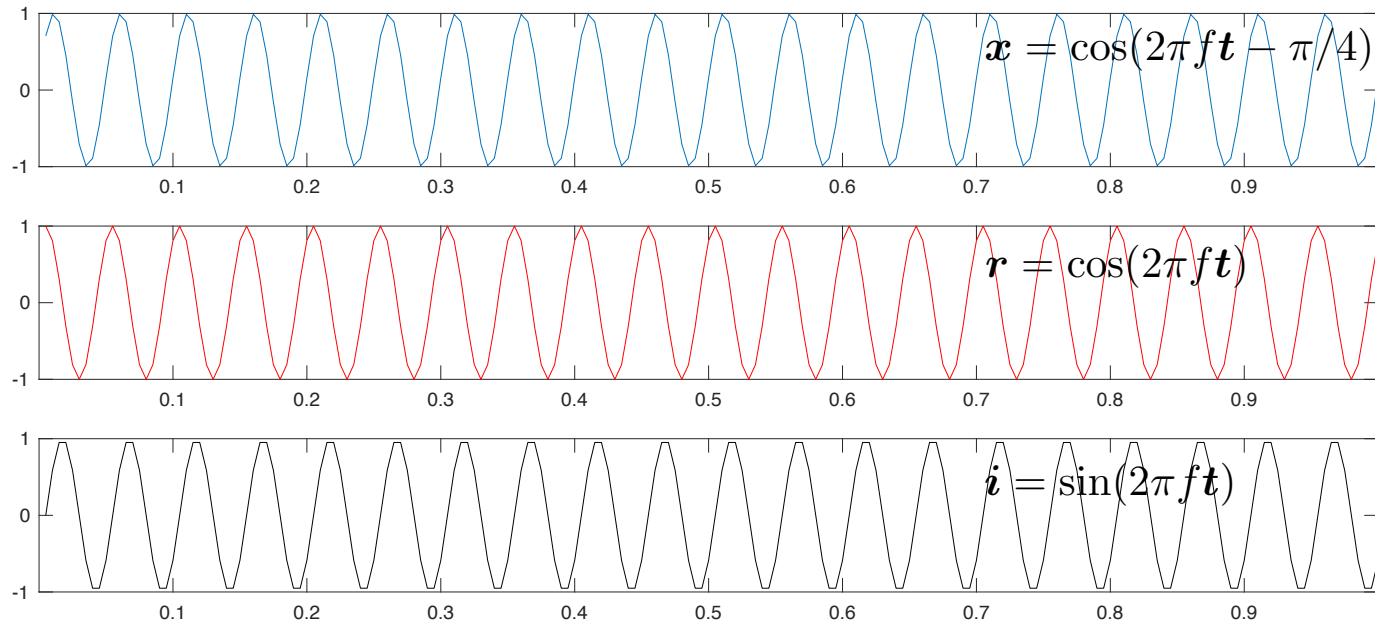
$$\mathbf{x} \cdot \mathbf{r} = 0$$

$$\mathbf{x} \cdot \mathbf{i} = 100$$

# Discrete Fourier Transform

- Basis vectors with different phase

$$f = 20\text{Hz}, \mathbf{t} = [0, 0.0001, 0.0002, \dots, 1]$$



$$\mathbf{x} \cdot \mathbf{r} = 70.71$$

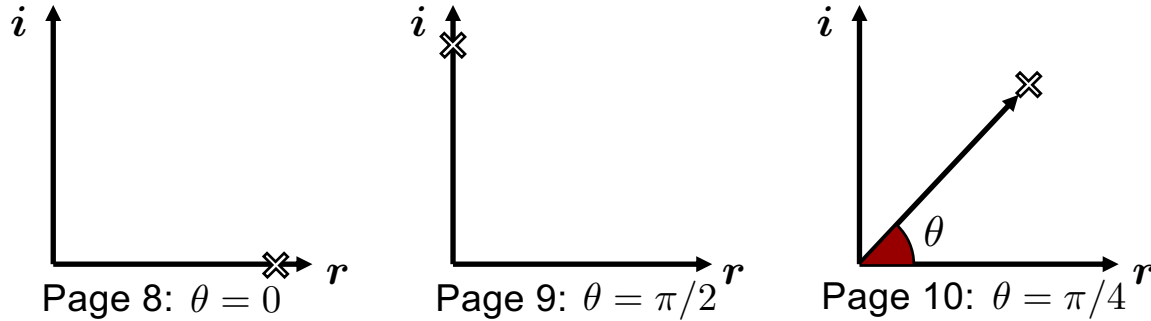
$$\mathbf{x} \cdot \mathbf{i} = 70.71$$

$$\sqrt{70.71^2 + 70.71^2} = 100$$

# Discrete Fourier Transform

## - Basis vectors with different phase

- We defined two orthogonal basis vectors  $\mathbf{r}$  and  $\mathbf{i}$ 
  - Which are the same cosine functions with a phase shift by  $\pi/2$



- Two coefficients:  $X_r(f) = \mathbf{x} \cdot \mathbf{r} = \sum_n x[n] \cos(2\pi f \frac{n}{N})$

$$X_i(f) = \mathbf{x} \cdot \mathbf{i} = \sum_n x[n] \sin(2\pi f \frac{n}{N})$$

- Or, using Euler's rule:

$$\begin{aligned} X(f) &= \mathbf{x} \cdot \mathbf{r} - j \mathbf{x} \cdot \mathbf{i} = \sum_n x[n] \left\{ \cos\left(2\pi f \frac{n}{N}\right) - j \sin\left(2\pi f \frac{n}{N}\right) \right\} \\ &= \sum_n x[n] \exp\left\{ -j\left(2\pi f \frac{n}{N}\right) \right\} \end{aligned}$$

- So, why are we doing this?
- The projection preserves information
  - It's onto two basis vectors that are with the same frequency with my input signal
  - Basis vectors are orthogonal (shifted by  $\pi/2$ )
  - Coefficients preserve the amplitudes (100)
  - Coefficients preserve the phase of the signal  $\theta = \pi/4$
- This procedure introduces a new representation  
 $\mathbf{x}$  versus  $(\mathbf{X}_r(f), \mathbf{X}_i(f))$

# Discrete Fourier Transform

## - Definition

- We just analyzed a sinusoid with a fixed frequency
  - By using a pair of cos and sin functions with a matching frequency
- What if  $x$  is not a pure sinusoid, e.g. a weighted sum of multiple sinusoids?
  - We repeat this procedure for various pre-defined pairs of cos/sin basis by varying  $f$

$$X[f] = \sum_{n=0}^{N-1} x[n] \exp \left\{ -j \left( 2\pi f \frac{n}{N} \right) \right\}$$

- It's called **discrete Fourier transform**
- The magnitude of the coefficient tells the amount of the contribution of the given frequency
- Its phase encodes the phase shift of the sinusoid

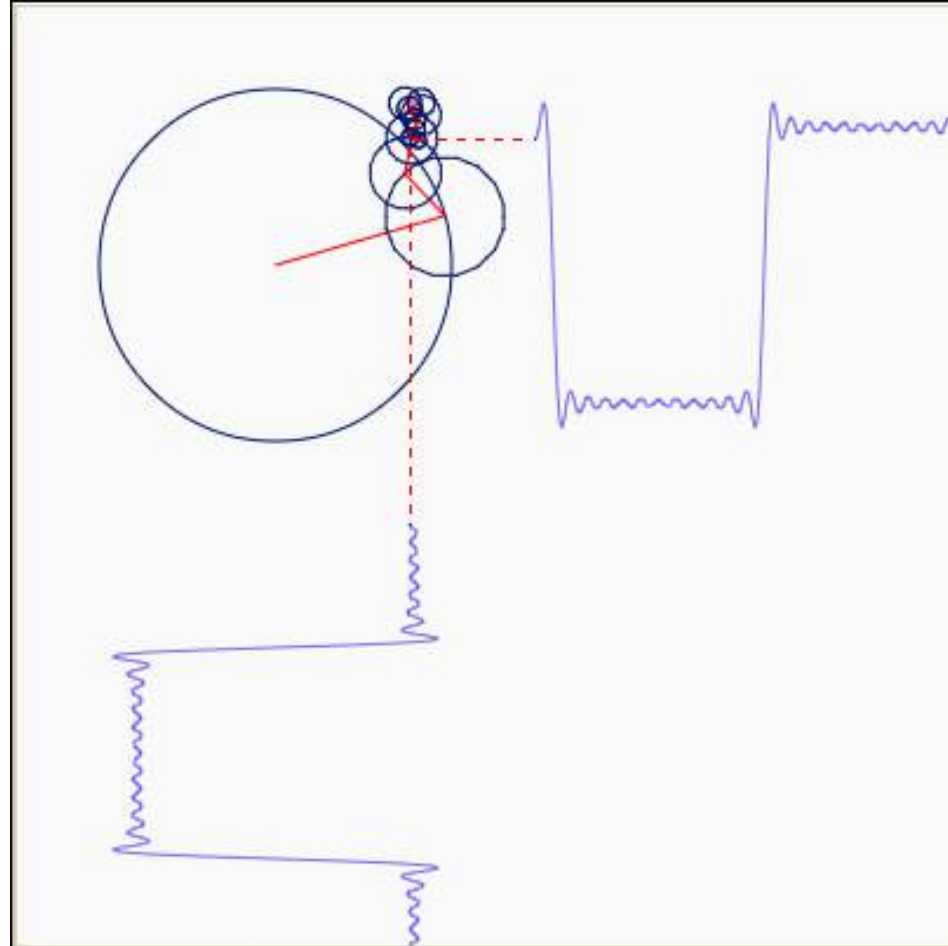
$$\angle(X[f]) = \arctan \left( \frac{\sum_n x[n] \sin(2\pi f \frac{n}{N})}{\sum_n x[n] \cos(2\pi f \frac{n}{N})} \right)$$

- It has its inverse transform as well:

$$x[n] = \frac{1}{N} \sum_{f=0}^{N-1} X[f] \exp \left\{ j \left( 2\pi f \frac{n}{N} \right) \right\}$$

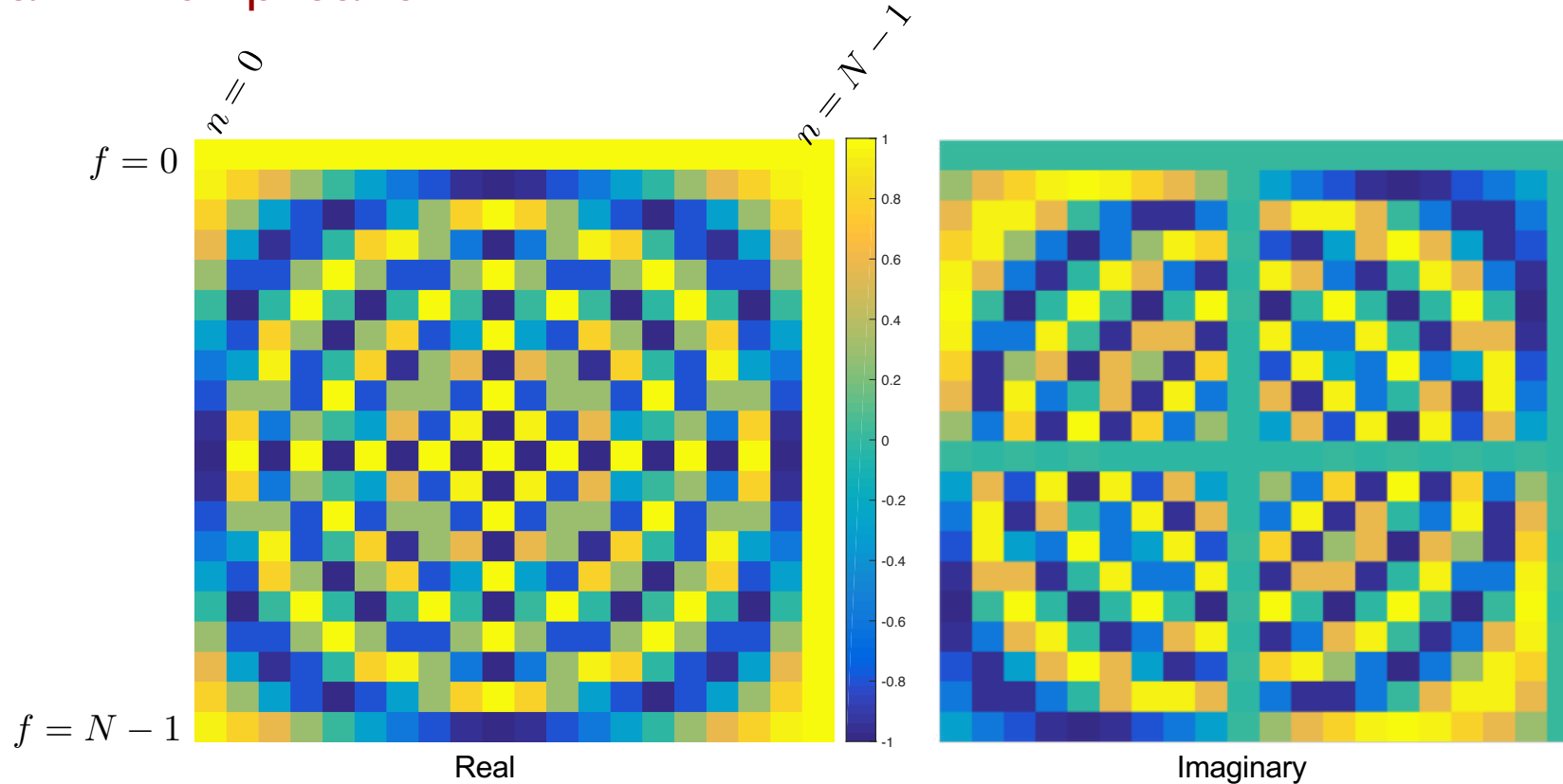
# Discrete Fourier Transform

- Box wave example



# Discrete Fourier Transform

- As a matrix multiplication

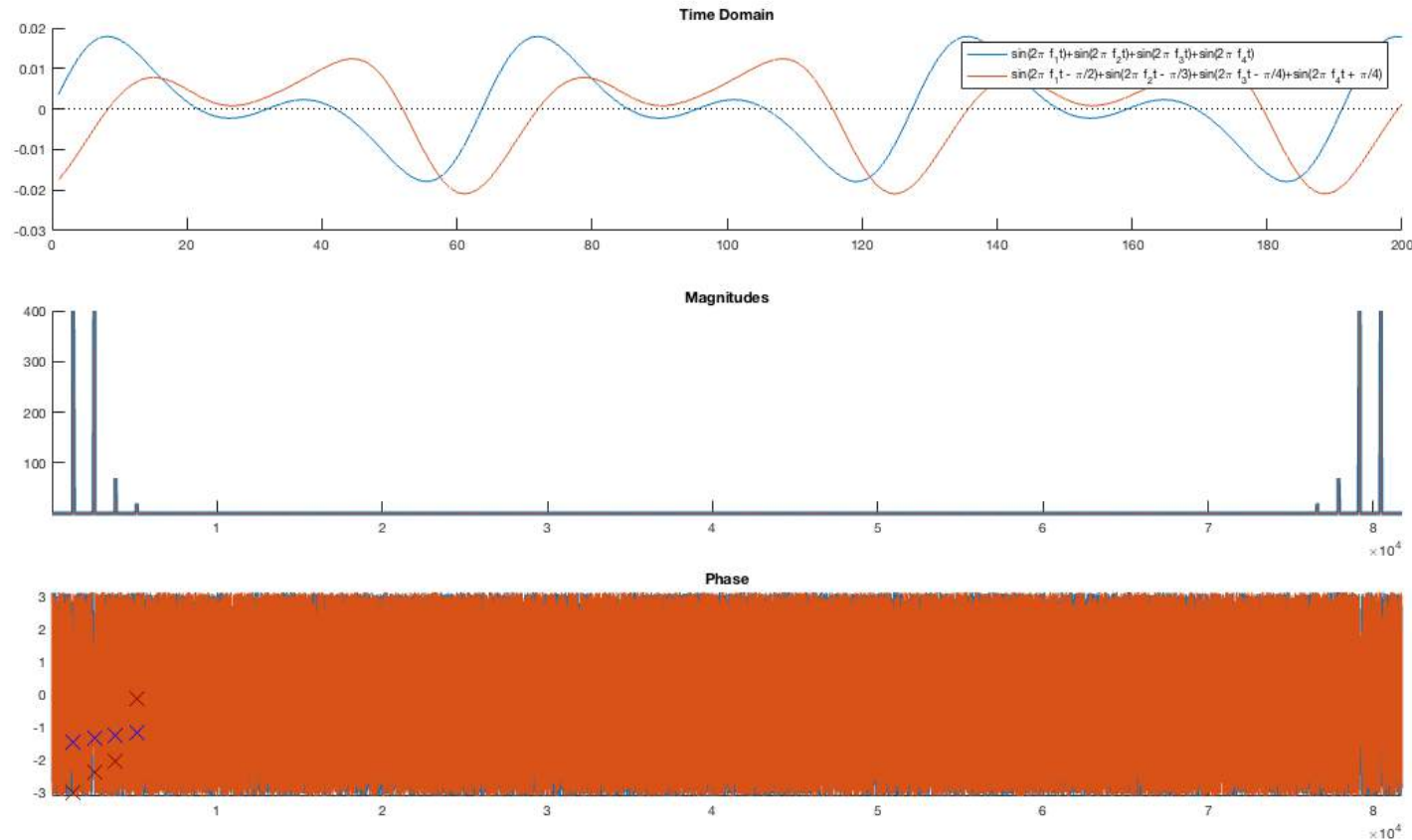


$$X_r(f) = \sum_n x[n] \cos(2\pi f \frac{n}{N})$$

$$X_i(f) = \sum_n x[n] \sin(2\pi f \frac{n}{N})$$

# Discrete Fourier Transform

- Magnitudes VS phase

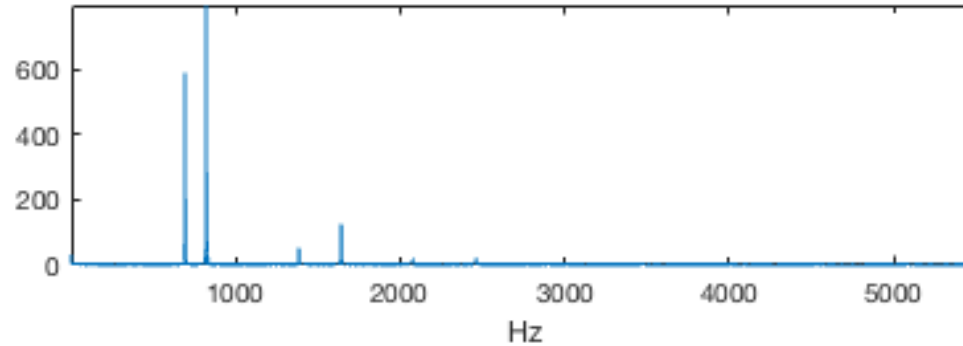


# Short-Time Fourier Transform

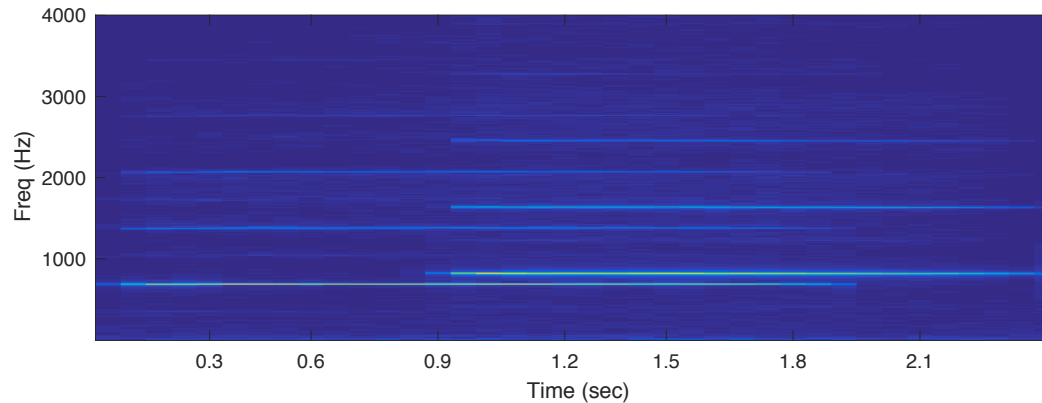
- Time-frequency representation

○ Two notes: 

○ Do you like this representation?



○ We need some temporal structure in the figure



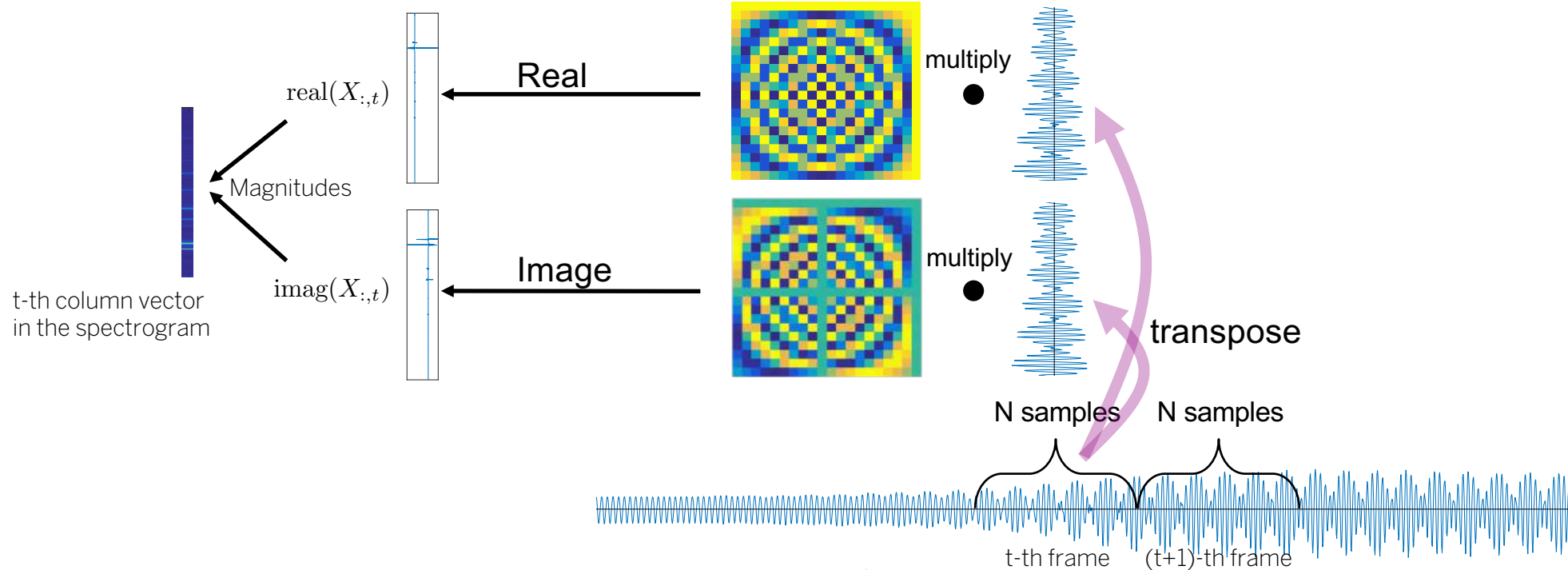


# Short-Time Fourier Transform

## - The mechanics

### ○ What's the magic?

- Slice the signal into pieces and then apply DFT one by one



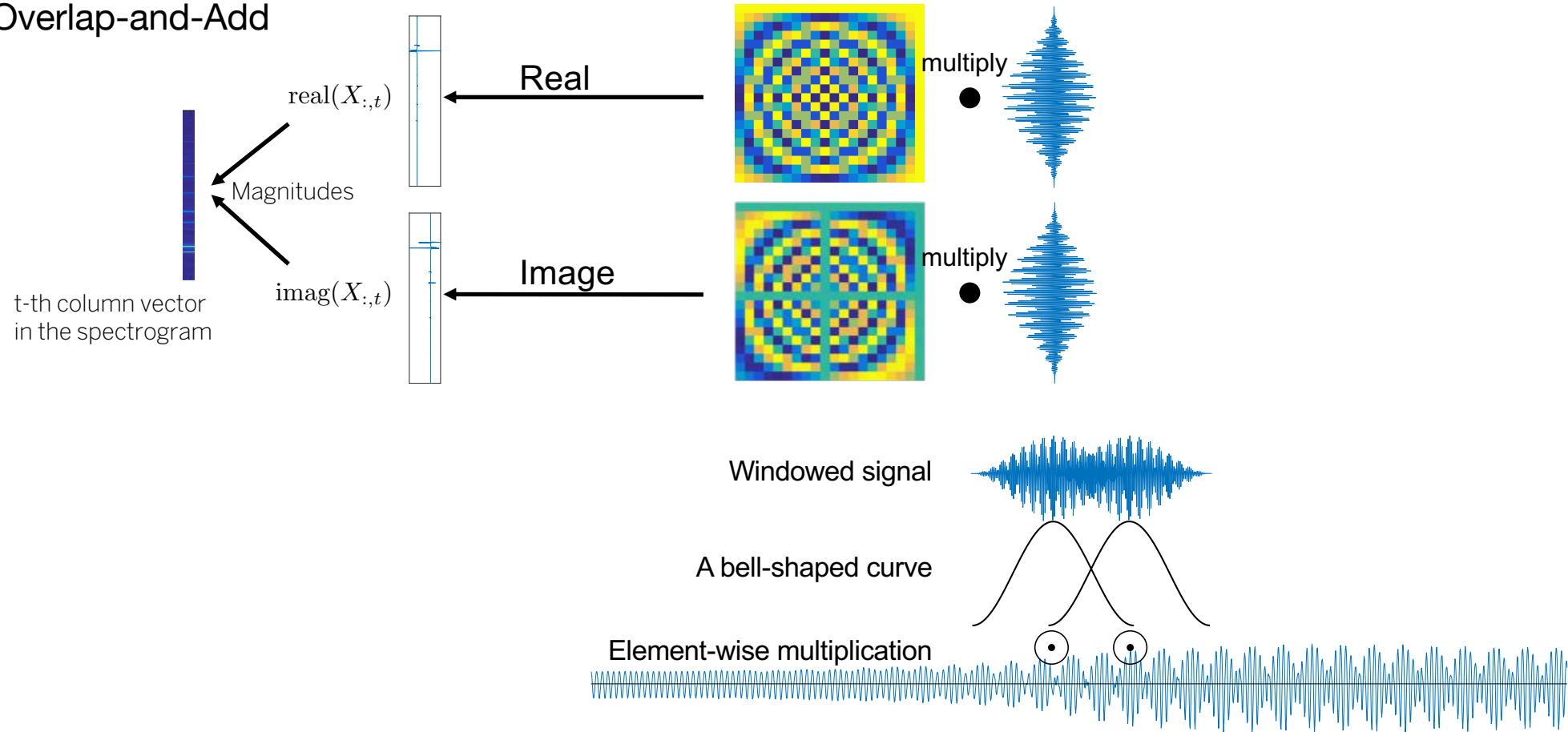
### ○ There's discontinuity we need to take care of



# Short-Time Fourier Transform

## - Windowing and overlap-and-add

### ○ Overlap-and-Add



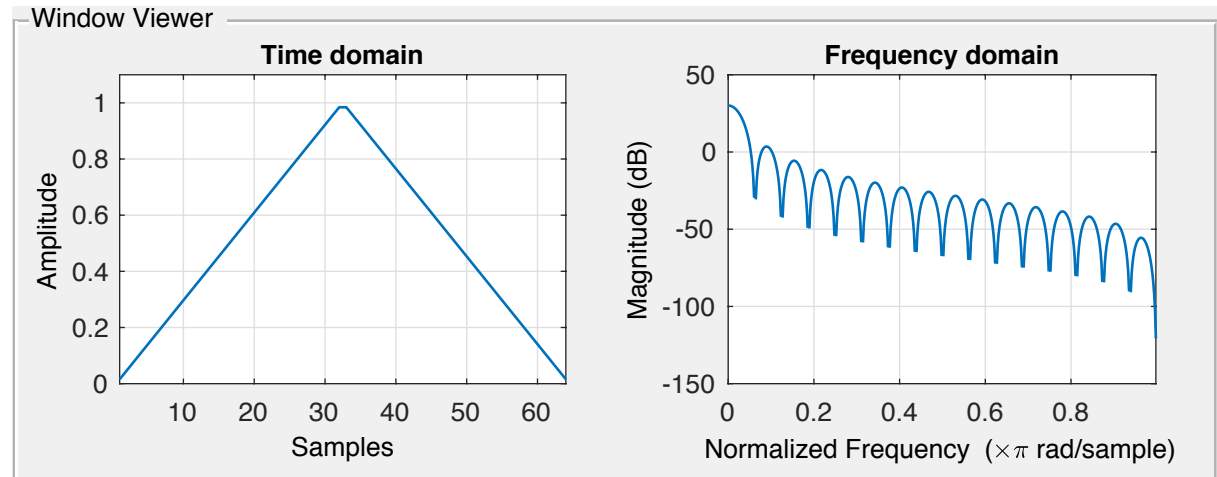
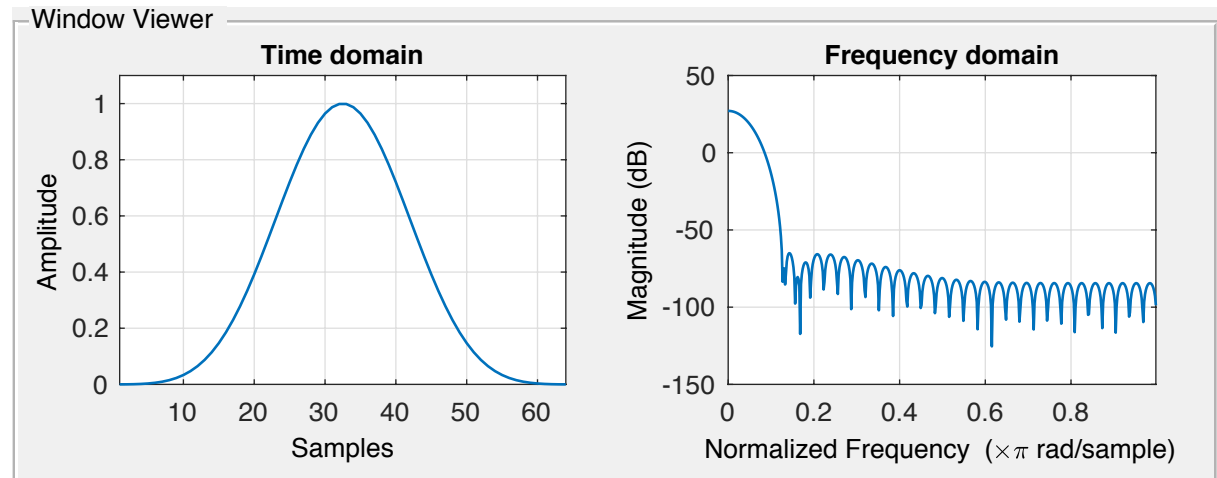
# Short-Time Fourier Transform

## - Windowing and overlap-and-add

- The windows need to have some desired properties
  - Their sum should be 1 after the overlap-and-add

$$w_t[512 + n] + w_{t+1}[n] = 1$$
$$\forall n = \{0, 1, \dots, 511\}$$

- Windowing should not affect too much on the DFT



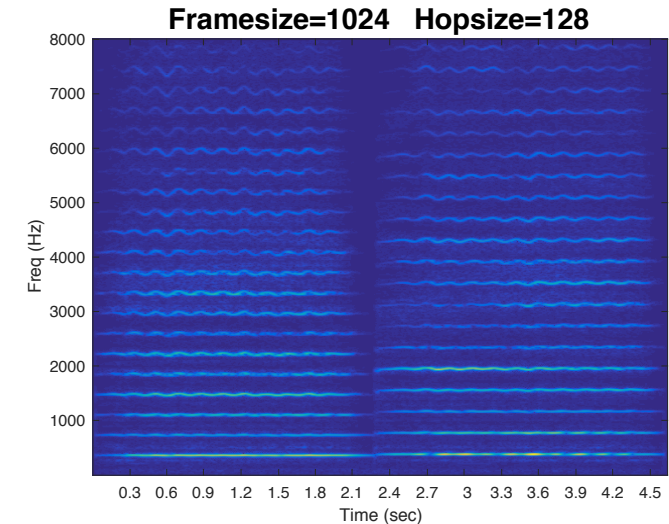
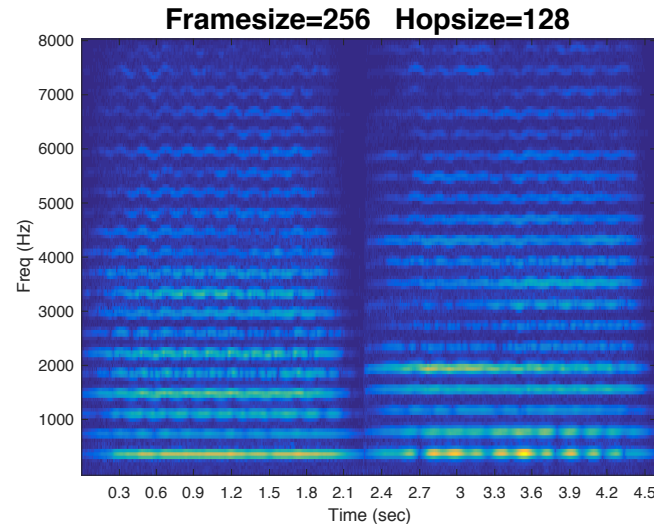
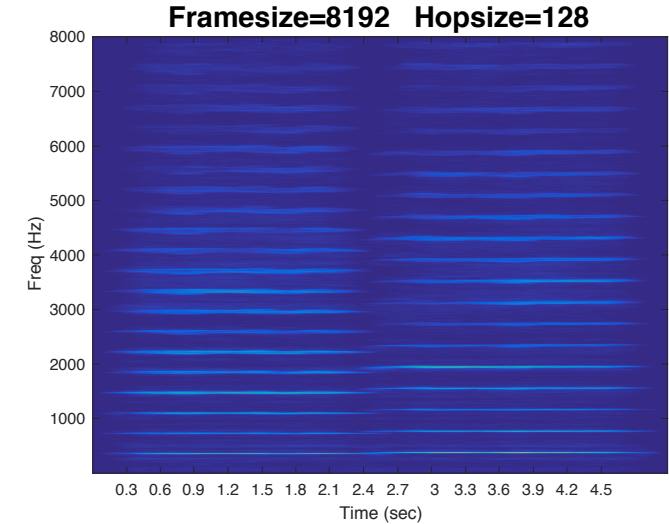
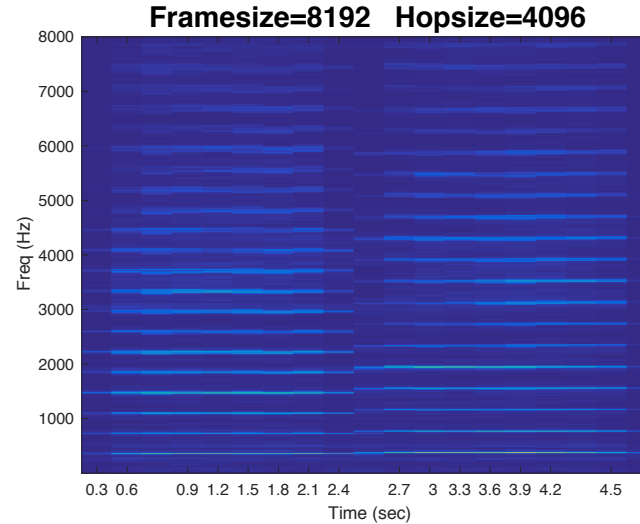
# Short-Time Fourier Transform

## - Resolution control

- Trade-off between time and frequency resolutions



- Which one do you like the best?



# Discrete Cosine Transform

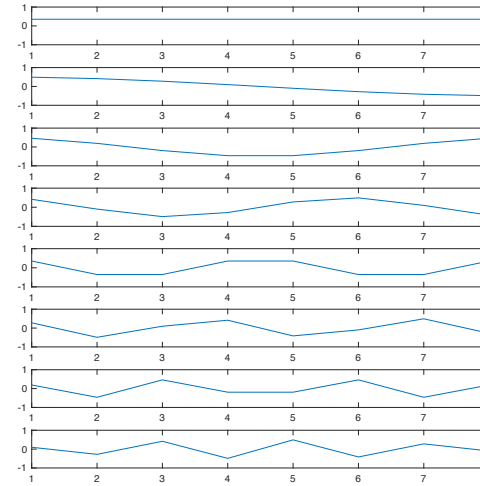
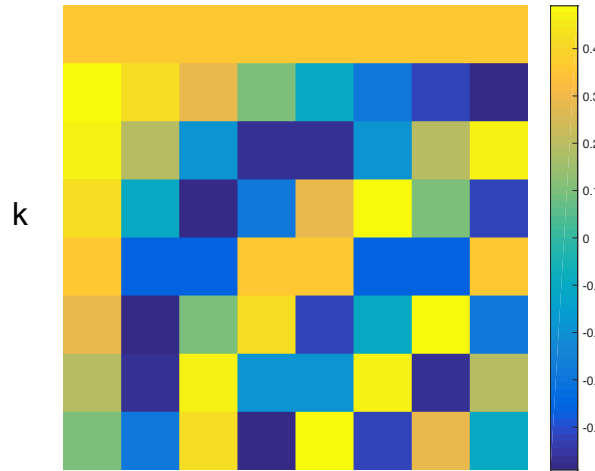
## - An alternative

- What if you don't really care about the perfect reconstruction?
  - As in lossy compression techniques: JPEG and AAC

$$C(k, n) = \sqrt{\frac{1}{N}} \cos \left( \frac{\pi(2n+1)k}{2N} \right)$$

n

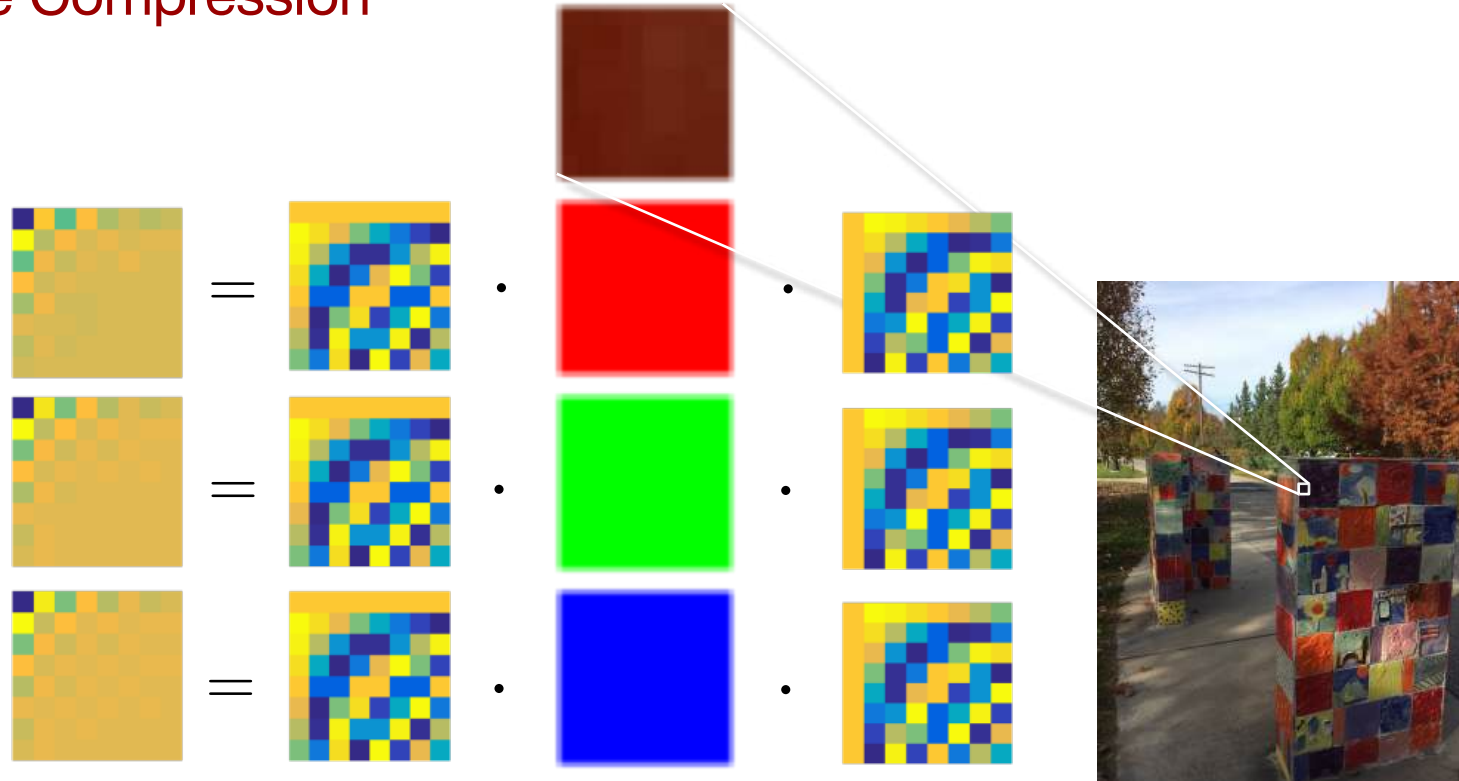
Annotations: "Basis vectors" points to  $k$ , "Samples" points to  $n$ .



# Discrete Cosine Transform

## - JPEG Image Compression

- The 2D DCT



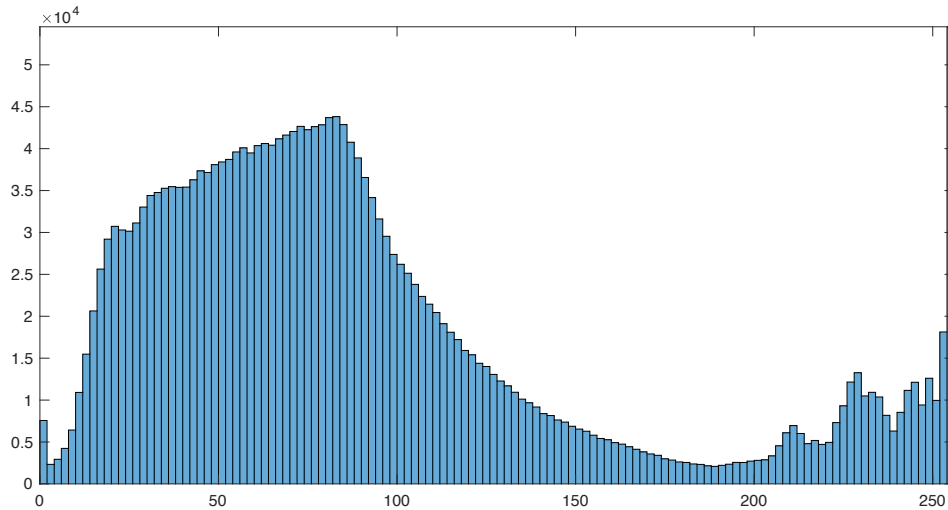
- Small patches (8X8) are with very similar values

- A high sum
- Low freq. components are stronger than high freq.

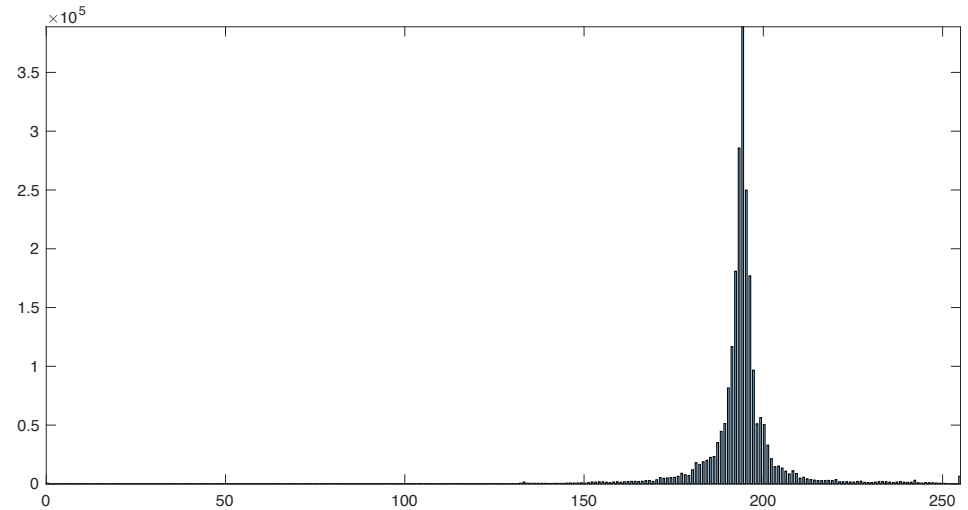
# Discrete Cosine Transform

## - JPEG Image Compression

- DCT coefficients are more centralized than the original pixels
  - Huffman coding can benefit



Sample distribution of the patch  
Entropy: 6.53



Sample distribution of the DCT coefficients  
Entropy: 4.73

- (Remember that entropy can be seen as the average number of bits to encode the signal)



# Open Questions

- Does this have anything to do with machine learning?





# Reading

- Textbook Chapter 6.8 and 6.9





**Thank You!**

