# ELEC 221 Lecture 25 The *z*-transform

Tuesday 6 December 2022

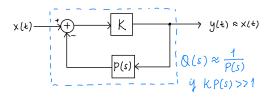
#### Announcements

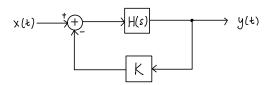
- Last class!
- Please come pick up your midterms
- Assignment 7 due tonight at 23:59 (hard deadline, no extensions)
- Details for final exam to be posted on Piazza when available

#### Last time

We saw how knowledge of the Laplace transform can help us:

- analyze feedback systems
- stabilize unstable systems
- find inverse systems





#### Last time

We introduced the DT counterpart, the *z*-transform:

We represented its region of convergence on the z-plane

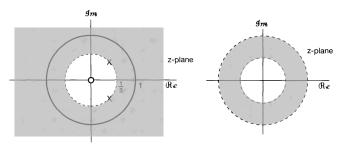


Image credit: Oppenheim 11.1, 11.2

## Today

#### Learning outcomes:

- use the *z*-transform to determine whether a system is causal or stable
- apply the *z*-transform to systems described by difference equations
- analyze simple feedback systems with the *z*-transform

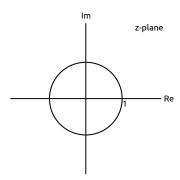
# Regions of convergence

Exercise: how many signals could have produced the z-transform

$$X(z) = \frac{1}{(1 - \frac{1}{3}z^{-1})(1 - 2z^{-1})}$$

Draw the pole-zero plot and determine the possible ROCs.

Hint: this function has 2 zeros; express it in a different way to find them.

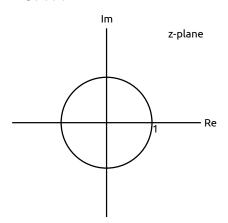


# Regions of convergence

Exercise: how many signals could have produced the z-transform

$$X(z) = \frac{1}{(1 - \frac{1}{3}z^{-1})(1 - 2z^{-1})}$$

Solution:



When the *z*-transform can be expressed as a rational function, we can compute the inverse using partial fractions. We still need the ROC to help us.

Exercise: compute the inverse z-transform of

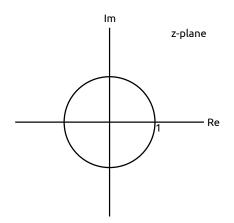
$$X(z) = \frac{1}{(1 - \frac{1}{3}z^{-1})(1 - 2z^{-1})}$$

if ROC is specified to be |z| > 2.

Exercise: compute the inverse z-transform of

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if ROC is specified to be |z| > 2.



Use partial fractions:

From ROC, signal is right-sided:

Take a closer look at the structure of X(z):

$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$$

This is a *power series in z*. If we can do the expansion, we can recover x[n] from the coefficients.

Exercise 1: what is the inverse z-transform of

$$X(z) = 3z^2 - 1 + 2z^{-3}, \quad 0 < |z| < \infty$$

Solution:

Particularly helpful for non-linear cases.

Exercise 2 (Oppenheim 10.63a): what is the inverse z-transform of

$$X(z) = \log(1-2z), \quad |z| < \frac{1}{2}$$

Hint:

$$\log(1-w) = -\sum_{i=1}^{\infty} \frac{w^i}{i}, \ |w| < 1$$

Solution:

# Properties of the z-transform

$$x_1[n] \stackrel{\mathcal{Z}}{\longleftrightarrow} X_1(z)$$
 w/ROC  $R_1$   
 $x_2[n] \stackrel{\mathcal{Z}}{\longleftrightarrow} X_2(z)$  w/ROC  $R_2$ 

#### Linearity:

Example:  $a^n u[n]$ ,  $a^n u[n-1]$  both have ROC of |z| > |a|. What is the ROC of z-transform of

Solution: .

# Properties of the z-transform

$$x[n] \stackrel{\mathcal{Z}}{\longleftrightarrow} X(z)$$
 w/ROC R

Time shift:

Time reversal:

**Time expansion** (zero-insertion of k-1 zeros):

# The z-transform and causality

$$x[n] \stackrel{\mathcal{Z}}{\longleftrightarrow} X(z)$$
 w/ROC R

Scaling in z:

Conjugation:

If x[n] is real, the poles and zeros come in *conjugate pairs*.

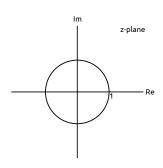
# The z-transform and causality

The convolution property of the z-transform tells us that

Remember how we previously tested causality: h[n] = 0 for all n < 0 (it is right-sided).

A DT LTI system with rational *z*-transform is causal if:

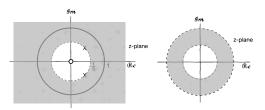
- the ROC is the exterior of a circle outside the outermost pole (including infinity)
- with H(z) expressed in polynomials of z, order of numerator does not exceed order of the denominator



## The z-transform and stability

Previously, to compute stability, we checked if the impulse response was absolutely summable:

This was also a condition required for the DTFT to exist.



An LTI system is stable if ROC includes the unit circle |z| = 1.

#### The initial value theorem

If 
$$x[n] = 0$$
 for all  $n < 0$ , then

How? Look again at expression for X(z):

#### Consequences:

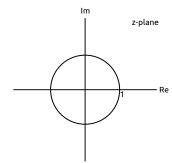
- if x[n] is causal,  $\lim_{z\to\infty} X(z)$  is finite
- if X(z) is a ratio of polynomials, order of numerator cannot be greater than order of denominator (cannot have more finite zeros than finite poles)

# Leveraging z-transform properties

- **10.17.** Suppose we are given the following five facts about a particular LTI system S with impulse response h[n] and z-transform H(z):
  - 1. h[n] is real.
  - **2.** h[n] is right sided.
  - 3.  $\lim_{z \to z} H(z) = 1$ .
  - **4.** H(z) has two zeros.
  - 5. H(z) has one of its poles at a nonreal location on the circle defined by |z| = 3/4. Answer the following two questions:
  - (a) Is S causal? (b) Is S stable?

## Leveraging z-transform properties

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  - (a) Is S causal? (b) Is S stable?



# Systems described by difference equations

Consider LTI system described by a DT difference equation

Using properties of the DTFT (convolution, time shift, linearity):

Using analogous properties of the z-transform,

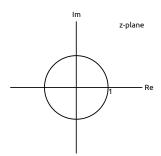
# Systems described by difference equations

Exercise (Oppenheim 10.36): consider LTI system described by a DT difference equation

Suppose the system is stable; what is its impulse response?

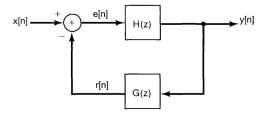
# Systems described by difference equations

Solution:



## Feedback systems

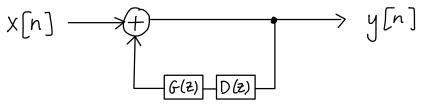
The z-transform can help us with the analysis of feedback systems (using them for stabilization, etc.) like we did in CT with the Laplace transform.



The closed-loop system function has the same form:

#### Example: comb filters

One type of system with this structure is called the **comb filter** 



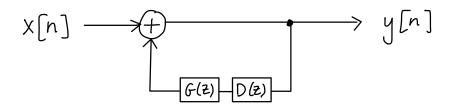
#### Suppose:

- D(z) is a system that causes a delay of K steps
- G(z) is a system with gain g

#### Exercise:

- what is the difference equation that describes the entire system?
- what is the closed-loop system function? (hint: you can compute it in two ways!)

# Example: comb filters

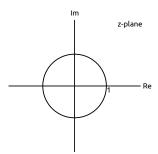


Difference equation:

System function:

### Example: comb filters

What are the poles and zeros?



Why is it called the comb filter? Let's look at its frequency response (take  $z=e^{j\omega}$ ).

#### Example: Karplus-Strong

Another example of this is the Karplus-Strong algorithm!

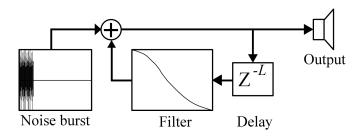
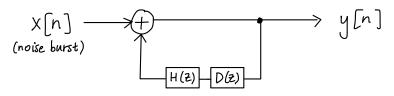


Image credit: https://commons.wikimedia.org/wiki/File:Karplus-strong-schematic.svg Author: PoroCYon CC BY-SA 3.0

## Example: Karplus-Strong



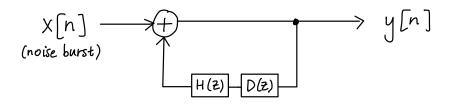
#### Suppose:

- D(z) is a system that causes a delay of K steps
- H(z) is a lowpass filter described by DE  $y[n] = \frac{1}{2}(x[n] + x[n-1])$

#### Exercise:

- what is the difference equation that describes the entire system?
- what is the closed-loop system function?

# Example: Karplus-Strong



Difference equation:

System function:

## Today

#### Learning outcomes:

- use the *z*-transform to determine whether a system is causal or stable
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Oppenheim practice problems: 10.13-10.16, 10.25-10.27, 10.31, 10.33-10.35, 11.1

#### For next time

#### Action items:

1. Assignment 7 due tonight at 23:59

Recommended reading: 10.5-10.7, 11.2