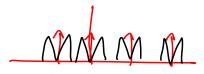
ELEC 221 Lecture 19 Amplitude modulation

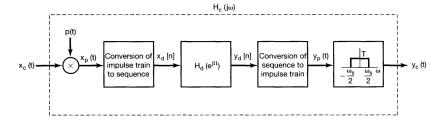
Tuesday 19 November 2024

Announcements

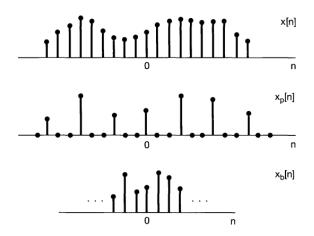


- Quiz 8 today
- Assignment 4 due Saturday 23:59
- Tutorial assignment 4 due Monday 23:59
- Friday office hour cancelled
- Class next Tuesday (26 Nov) on Zoom

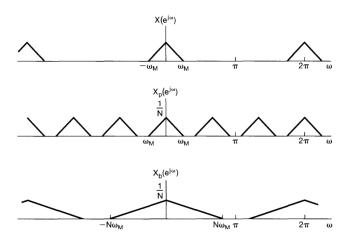
We performed CT signal processing by sampling, converting to DT, processing, then converting back.



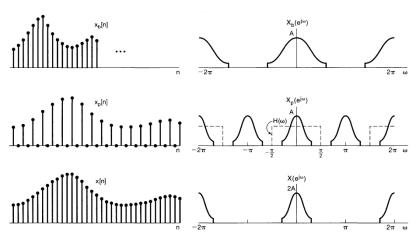
We sampled DT signals (decimation). If original signal was CT, this is called *downsampling*.



Decimation spreads out the spectrum.



The opposite of decimation is interpolation (upsampling).



Today

First: finish example from last time.

Learning outcomes:

- perform sinusoidal amplitude modulation (AM) and demodulation in CT and DT
- differentiate between synchronous and asynchronous modulation techniques, and identify pros/cons of each method
- carry out frequency-division multiplexing

Example: down/upsampling

Oppenheim problem 7.19,

x[n]

7.19. Consider the system shown in Figure P7.19, with input x[n] and the corresponding output y[n]. The zero-insertion system inserts two points with zero amplitude between each of the sequence values in x[n]. The decimation is defined by

$$y[n] = w[5n],$$

where w[n] is the input sequence for the decimation system. If the input is of the form

$$x[n] = \frac{\sin \omega_1 n}{\pi n},$$

$$\text{determine the output } y[n] \text{ for the following values of } \omega_1:$$

$$\text{(a) } \omega_1 \leq \frac{3\pi}{5}$$

$$\text{(b) } \omega_1 > \frac{3\pi}{5}$$

$$\text{(b) } \omega_1 > \frac{3\pi}{5}$$

$$\text{(b) } \frac{1}{5} S[n]$$

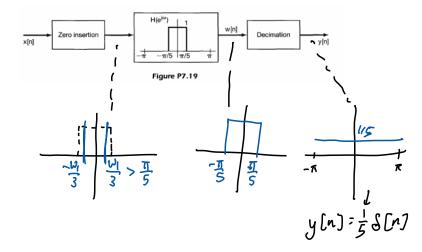


Example: down/upsampling First case: $\omega_1 \leq \frac{3\pi}{5}$ H(e^{jω}) w[n] Zero insertion Decimation $-\pi/5$ $\pi/5$ Figure P7.19 45 $y(n) = \frac{\sin\left(\frac{6}{3}w_{1}n\right)}{5\pi n}$

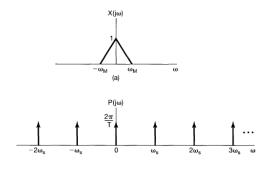
Example: down/upsampling

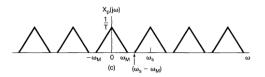
Second case:
$$\omega_1 > \frac{3\pi}{5}$$

$$f[n] \stackrel{F}{\longleftrightarrow} 1$$



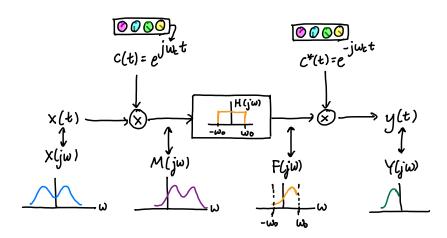
Impulse train sampling





Frequency-selective filtering with variable centre frequency

Recall from Assignment 3:



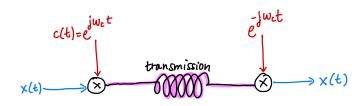
Multiplication property

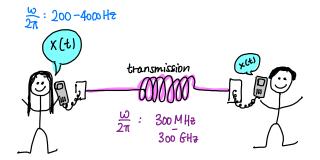
Both results come from the multiplication property:

$$y(t) = x(t) p(t)$$
 $\rightarrow Y(jw) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j0) P(j(w-0)) dw$

This is also known as the modulation property.

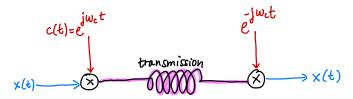
Motivation: communication systems





Modulation

The process of embedding an information-bearing signal into a second signal. (Extracting the signal: demodulation)



Two main types (we will only discuss AM):

- amplitude modulation (AM)
- frequency modulation (FM)

We focus on two types of **carrier signal**, c(t):

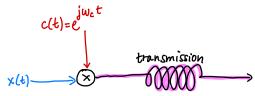
complex exponential signal $C(t) = e^{j(w_c t + \theta_c)}$

sinusoidal signal

$$c(t) = cos(\omega_c t + \theta_c)$$

Complex exponential amplitude modulation

We've already seen what happens with the first one.



In practice:

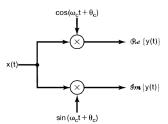
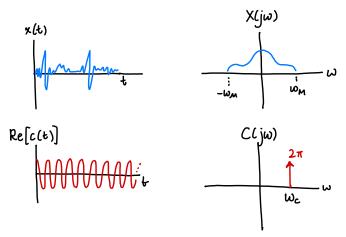


Figure 8.2 Implementation of amplitude modulation with a complex exponential carrier $c(t) = e^{j(\omega_c t + \theta_c)}$.

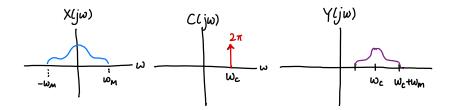
Complex exponential amplitude modulation

Consider the Fourier spectrum of both signals:

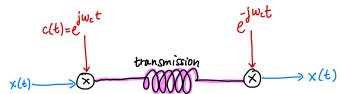


Complex exponential amplitude modulation

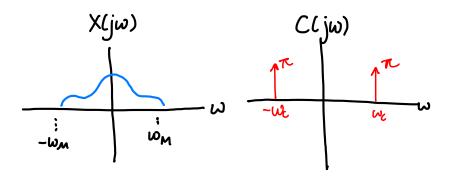
Multiplication by c(t) shifts spectrum to different frequency range.

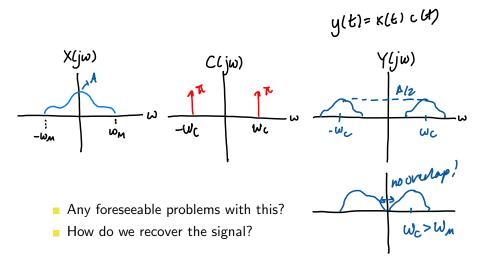


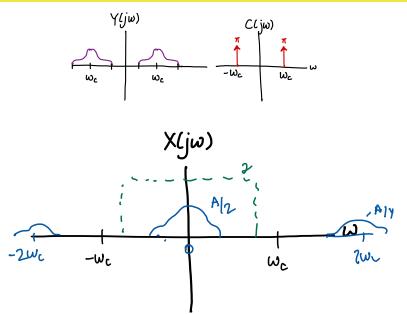
Demodulation is straightforward.

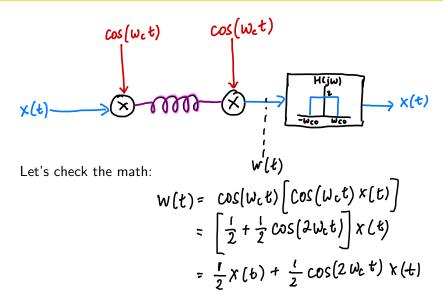


What if we use a sinusoidal signal instead?









Exercise: sinusoidal amplitude modulation



8.3. Let x(t) be a real-valued signal for which $X(j\omega) = 0$ when $|\omega| > 2,000\pi$. Amplitude modulation is performed to produce the signal

$$g(t) = x(t)\sin(2,000\pi t).$$

A proposed demodulation technique is illustrated in Figure P8.3 where g(t) is the input, y(t) is the output, and the ideal lowpass filter has cutoff frequency $2,000\pi$ and passband gain of 2. Determine y(t).

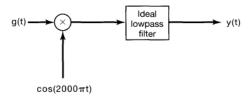


Figure P8.3

Exercise: sinusoidal amplitude modulation

Solution option 1: mathematical

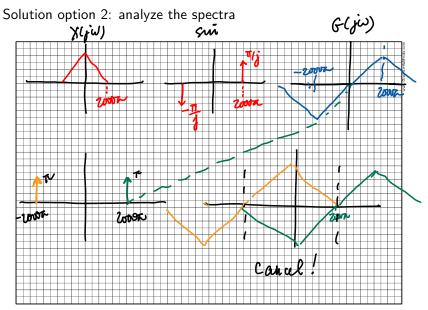
Evaluate
$$w(t) = g(t) \cos(2000\pi t)$$
 (input to filter):

$$= \chi(t) \sin(2000\pi t) \cos(2000\pi t)$$

$$= \chi(t) \cdot \frac{1}{2} \sin(4000\pi t)$$

$$y(t) = 0$$

Exercise: sinusoidal amplitude modulation



More generally, must consider phases in both modulating and demodulating signals:

$$w(t) = \cos(\omega_c t + \theta_c) \cos(\omega_c t + \phi_c) \times (t)$$

$$= \left[\frac{1}{2}\cos(\theta_c - \phi_c) + \frac{1}{2}\cos(2\omega_c t + \theta_c + \phi_c)\right] \times (t)$$

Output after the lowpass filter is

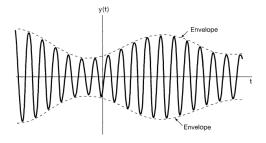
$$x'(\ell) = \frac{1}{2} \cos(\theta_c - \phi_c) x(\ell)$$

Synchronous demodulation: $\phi_c = \theta_c$. What could go wrong?

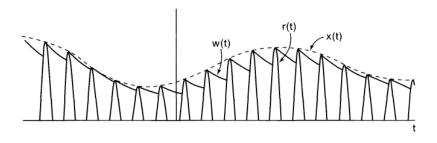
Suppose the following is true:

- x(t) is positive
- lacksquare ω_c is much larger than ω_m

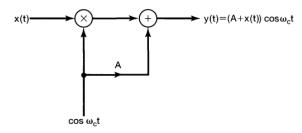
The transmitted signal will look something like this:



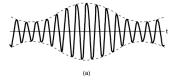
Design a system to track the envelope (we won't go into details).

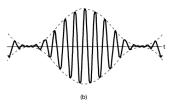


If x(t) is not positive, choose A sufficiently large and add to signal:



Suppose $|x(t)| \le K$. Must have A > K. Then m = K/A is known as the *modulation index*.





You will explore this in Assignment 4.

For next time

Content:

- single-sideband modulation
- pulse amplitude modulation and time-division multiplexing
- cellphone communication systems

Action items:

- 1. Assignment 4 due Saturday 23:59
- 2. Tutorial assignment 4 due Monday 23:59

Recommended reading:

- From this class: Oppenheim 8.0-8.3
- Suggested problems: 8.1, 8.2, 8.4-8.6, 8.8, 8.21-8.23, 8.40
- For next class: Oppenheim 8.4-8.6