

### Samples to Digital Symbols: Symbol Clock Recovery and Improved Symbol Synchronization Blocks

Andy Walls 12 Sept 2017

### Outline



- Advertisement
  - About SilverBlock
- Theoretical
  - Problem Statement
  - Symbol Synchronization Overview
  - PLL Symbol Synchronizer
  - Clock Tracking PLL Model
  - Timing Error Detector
  - Interpolating Resampler

### Outline (continued)



#### Practical

- GNURadio Sync Blocks
- Existing Blocks' Deficiencies
- New Symbol Sync Blocks Overview
- New Symbol Sync Blocks Architecture
- Adding a New Timing Error Detector
- Adding a New Resampler
- Using a Different Slicer
- Existing Blocks to New Blocks
- Usage Hints and Gotchas
- Experimental Tuning Example
- TED S-Curve Simulation

### SilverBlock Systems

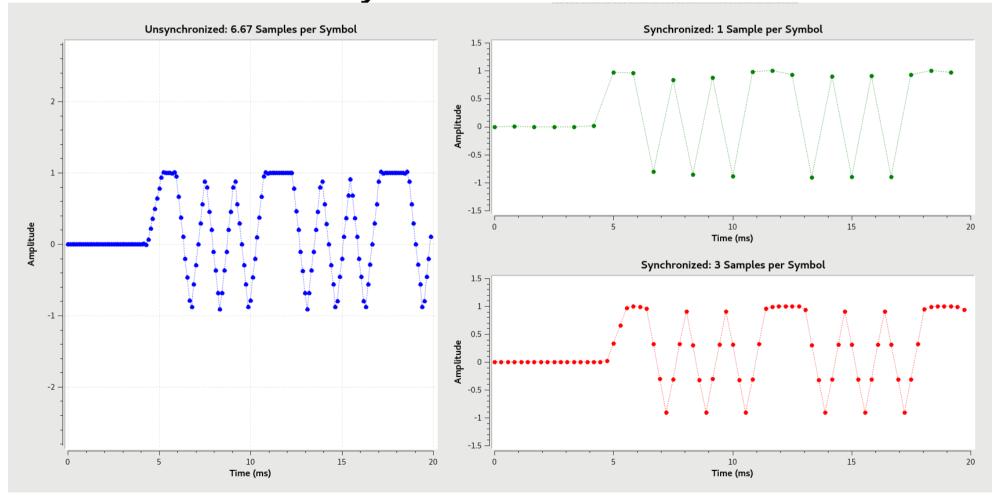


- Small company founded in 2009
  - Leonardtown, MD (0.3 miles from the waterfront)
     https://www.visitstmarysmd.com/see-do/towns-communities/leonardtown-area/
  - Rochester, NY
- What we build
  - Distributed Mission Computing and Display Software
  - RADAR Processing Algorithms and GPU Implementations
  - Experimental Communications Systems
  - Experimental Airborne Sensors and Systems
- Looking to hire 1 or 2 experienced software engineers
  - C++ required. US Citizenship required.
  - contact@silverblocksystems.net

### **Problem Statement**



Output a sample stream synchronized with the center of data symbols



# Symbol Synch Overview



- Two broad categories of algorithms
  - Feedfoward (Open Loop)
    - Block oriented
    - Operate on samples for a number of symbols at a time
    - Non-tracking, but can be computationally complex
    - Burst mode communications, or initial acquisition of synchronization
  - Feedback (Closed Loop)
    - Stream oriented
    - Operate on immediate incoming sample or symbol
    - Tracking, and not computationally complex for any 1 input
    - Continual stream of symbols, or tracking after initial acquisition

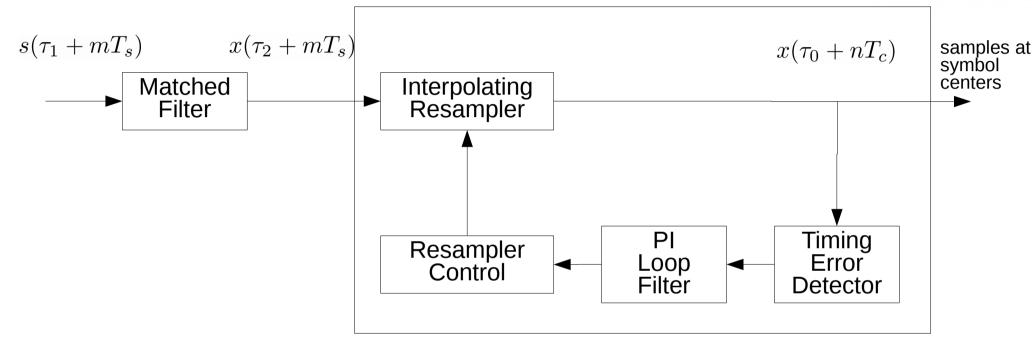
### Symbol Synch Overview



- Algorithm and Implementation Requirements
  - Resampling with interpolation
    - Imposes signal bandwidth requirements
  - Symbol clock timing estimation or timing error estimation
    - Imposes signal conditioning requirements
  - Offline modeling, simulation, and analysis by the designer!
- Correlation-based Feedforward Algorithm and Implementation
  - GNURadio has ~1 implementation
- PLL-based Feedback Algorithm and Implementation
  - GNURadio now has 4 implementations

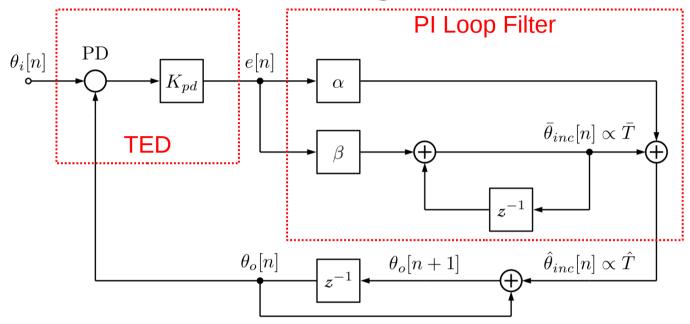
# PLL Symbol Synchronizer





- Timing Error Detector: analogous to PLL phase detector
  - Estimates the symbol clock timing error, a hidden quantity
- Resampler Control: analogous to PLL phase accumulator/NCO
- T<sub>s</sub> input sample clock period
- T<sub>c</sub> symbol clock period
- $\tau_0$  optimal symbol sampling offset





- $\bullet$  Timing Error Detector (TED) modeled by phase detector (PD) and  $K_{pd}$  gain
  - Outputs symbol clock timing error estimate, e[n], once per symbol
- Proportional-Integral (PI) Loop Filter
  - Proportional arm gain, α
  - Integral arm gain, β
  - Integral arm output: estimate of average period of symbol clock,  $T_{avg}$
  - Filter output: estimate of instantaneous period of symbol clock, T<sub>inst</sub>
- Resampler & Control maps to phase accumulator and phase signals



Loop Phase Transfer Function (PI gains)

$$H(z) = \frac{\Theta_o(z)}{\Theta_i(z)} = K_{pd}(\alpha + \beta)z^{-1} \cdot \frac{1 - \frac{\alpha}{\alpha + \beta}z^{-1}}{1 - 2\left(1 - K_{pd}\frac{\alpha + \beta}{2}\right)z^{-1} + (1 - K_{pd}\alpha)z^{-2}}$$

Zeros, Poles, and Critical Damping (PI gains)

$$\begin{split} z_1 &= \frac{\alpha}{\alpha + \beta} \\ z_2 &\to \infty \\ p_{1,2} &= \left(1 - K_{pd} \frac{\alpha + \beta}{2}\right) \pm \sqrt{\left(1 - K_{pd} \frac{\alpha + \beta}{2}\right)^2 - (1 - K_{pd} \alpha)} \\ \alpha &= -\beta \pm \frac{2}{|K_{pd}|} \sqrt{K_{pd} \beta} \quad \text{for critical damping} \end{split}$$



Mapping Poles of a 2<sup>nd</sup> Order Analog Control System to z-plane

$$s_{1,2} = -\zeta \omega_n \pm j\omega_d = -\zeta \omega_n \pm j\omega_n \sqrt{1-\zeta^2}$$

 $\zeta$ : damping factor (1.0 is critically damped)

 $\omega_n$ : natural frequency (radius from origin, approx 1-sided BW for underdamped system)

 $\omega_d$ : damped frequency of oscillation (ringing of underdamped impulse response)

$$z = e^{sT}$$

$$p_{1,2} = e^{-\zeta \omega_n T \pm j\omega_d T}$$

Loop Phase Transfer Function (2<sup>nd</sup> Order Control System)

$$H(z) = \begin{cases} \frac{[2 - 2\cos(\omega_d T)e^{-\zeta\omega_n T}]z - 2\sinh(\zeta\omega_n T)e^{-\zeta\omega_n T}}{z^2 - 2\cos(\omega_d T)e^{-\zeta\omega_n T}z + e^{-2\zeta\omega_n T}} & \text{for } \zeta < 1 & \text{with } \omega_d T = \omega_n T\sqrt{1 - \zeta^2} \\ \frac{[2 - 2(1)e^{-\zeta\omega_n T}]z - 2\sinh(\zeta\omega_n T)e^{-\zeta\omega_n T}}{z^2 - 2(1)e^{-\zeta\omega_n T}z + e^{-2\zeta\omega_n T}} & \text{for } \zeta = 1 & \text{with } \omega_d T = 0 \\ \frac{[2 - 2\cosh(\omega_d T)e^{-\zeta\omega_n T}]z - 2\sinh(\zeta\omega_n T)e^{-\zeta\omega_n T}}{z^2 - 2\cosh(\omega_d T)e^{-\zeta\omega_n T}z + e^{-2\zeta\omega_n T}} & \text{for } \zeta > 1 & \text{with } \omega_d T = \omega_n T\sqrt{\zeta^2 - 1} \end{cases}$$



 PI Gains from 2<sup>nd</sup> Order Digital Control Loop parameters:

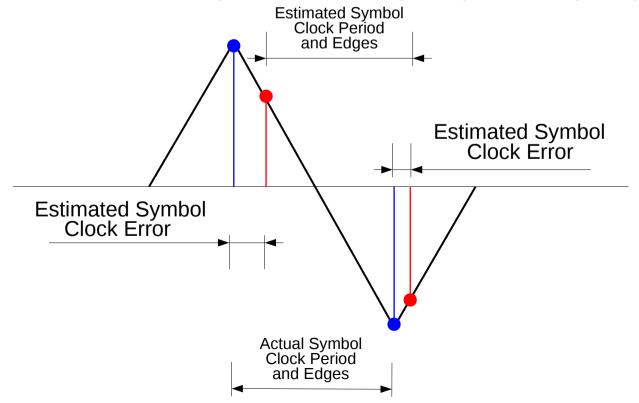
$$\alpha = \frac{2}{K_{pd}} e^{-\zeta \omega_n T} \sinh(\zeta \omega_n T)$$

$$\beta = \begin{cases} \frac{2}{K_{pd}} \left( 1 - e^{-\zeta \omega_n T} \left[ \sinh(\zeta \omega_n T) + \cos(\omega_d T) \right] \right) & \text{for } \zeta < 1 \quad (under \, damped) \\ \frac{2}{K_{pd}} \left( 1 - e^{-\zeta \omega_n T} \left[ \sinh(\zeta \omega_n T) + 1 \right] \right) & \text{for } \zeta = 1 \quad (critically \, damped) \\ \frac{2}{K_{pd}} \left( 1 - e^{-\zeta \omega_n T} \left[ \sinh(\zeta \omega_n T) + \cosh(\omega_d T) \right] \right) & \text{for } \zeta > 1 \quad (over \, damped) \end{cases}$$

### **Timing Error Detector**



- A TED emits an error value <u>proportional</u> to the time difference between
  - Optimal current symbol sampling time (blue)
  - Actual current symbol sampling time (red)



### Timing Error Detector

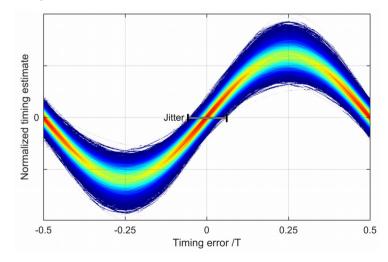


- A TED is just an expression for the error value, e[n]
  - Formally derived Pay attention to the assumptions!
  - Can include factors such as
    - Current symbol estimate
    - Nearby samples or symbol estimates
    - Current symbol decision (Decision Directed TEDs)
    - Nearby symbol decisions (Decision Directed TEDs)
    - Estimate of signal slope at symbol sampling time
    - Estimate of E<sub>s</sub>/N<sub>0</sub>
  - Usually a simple expression in the end
- Examples
  - Mueller and Müller  $e[n] = \operatorname{slice}(x[n-1])x[n] \operatorname{slice}(x[n])x[n-1]$
  - Small signal ML approximation e[n] = x[n]x'[n]

### **Timing Error Detector**

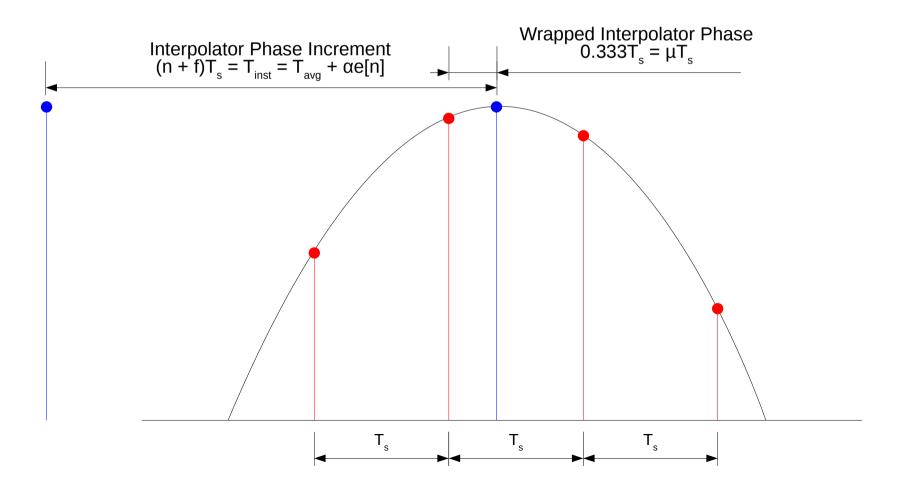


- A TED is characterized by its S-curve:  $S(\tau_{\epsilon}) = E\{e[n]|\tau_{\epsilon}\}$ 
  - y-axis: expected value of timing error output, given the normalized symbol clock timing offset
  - τ-axis: normalized symbol clock timing offset
  - Slope at timing offset  $\tau = 0$  is TED gain,  $K_{ted}$  (=  $K_{pd}$ )
    - Units of  $K_{ted}$  might not match the units required by the loop scaling needed
- S-Curve shape and central slope at  $\tau = 0$  depend on
  - TED's error estimator expression
  - Input signal amplitude
  - Pulse shaping filter / pulse shape
  - Input E<sub>s</sub>/N<sub>0</sub>
  - Other factors
- Simulation required to find TED gain
  - Octave, MatLab, R, Python, or whatever





- Symbol synchronization process reduces sample rate
- Input samples (red) not at optimal symbol sampling (blue)
- Need to resample input between samples





- Interpolation implemented with FIR filters
  - Fractional Delay (FD)
    - MMSE interpolator polyphase filterbank
    - Matched filter/interpolator polyphase filterbank
  - Polynomial Interpolation using Farrow structure
    - Lagrange
    - B-spline
- Practical interpolation filters impose bandwidth constraint
  - Input signal bandwidth must be some fraction of F<sub>s</sub>
  - Bounds the error in the interpolated output samples



- Ideal FD interpolation filter frequency responses
  - Interpolator

$$H(\omega) = 1 \cdot e^{j\omega\mu} \quad \text{for} \quad -\pi \le \omega \le \pi$$

- Differentiating interpolator  $H(\omega) = j\omega \cdot e^{j\omega\mu}$  for  $-\pi \le \omega \le \pi$
- $\mu$  is intersample interpolation fraction in [0.0, 1.0]
- Ideal FD interpolation filter impulse responses
  - Interpolator

$$h[n] = \operatorname{sinc}(n + \mu)$$

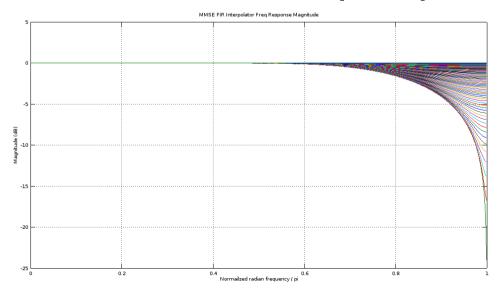
Differentiating interpolator

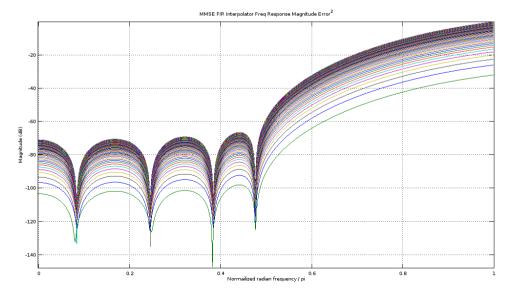
$$h[n] = \frac{1}{n+\mu} \left[ \cos \left( \pi[n+\mu] \right) - \operatorname{sinc}(n+\mu) \right]$$

- For a particular  $\mu$  in [0.0, 1.0]
- These ideal responses from IDTFT are infinite length



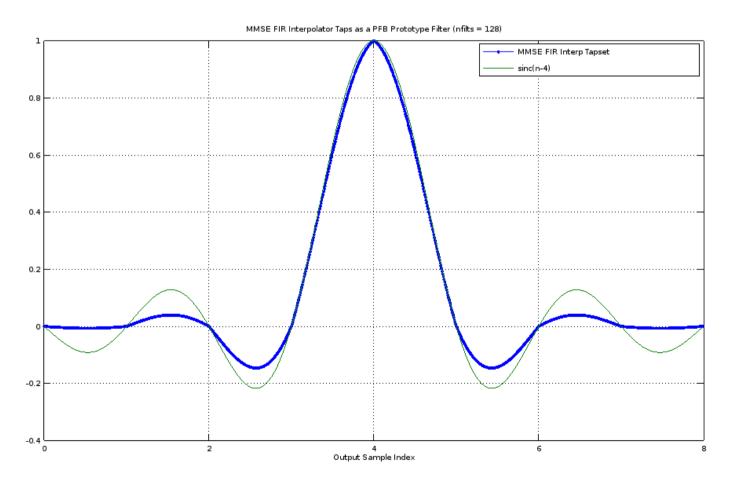
- GR Minimum Mean Squared Error interpolators
  - Truncated, MMSE version of ideal filters
  - Bank of 129, 8 tap FIR filters
  - For  $\mu$  in {0/128, 1/128, 2/128, ..., 127/128, 128/128}
  - Each filter has a MMSE  $H(\omega)$  only in  $[-F_s/4, F_s/4]$
- One sided freq response and error<sup>2</sup> of the filters







- GR's MMSE interpolator is a polyphase filter bank
- Equivalent PFB prototype filter vs truncated ideal



### GNURadio Sync Blocks



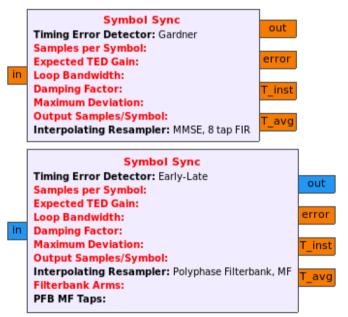
#### **Existing**

Omega: Gain Omega: Mu: Gain Mu: Omega Relative Limit: Clock Recovery MM MSK Timing Recovery out Omega: Gain: Gain Omega: Samples per symbol: err Error limit: Gain Mu: Output samples per symbol: omega Omega Relative Limit:

Clock Recovery MM

Polyphase Clock Sync out Samples/Symbol: Loop Bandwidth: err Taps: Filter Size: rate Initial Phase: Maximum Rate Deviation: phase Output SPS: Polyphase Clock Sync out Samples/Symbol: Loop Bandwidth: err Taps: in Filter Size: rate Initial Phase: Maximum Rate Deviation: phase Output SPS:

#### <u>New</u>



# Existing Blocks' Deficiencies

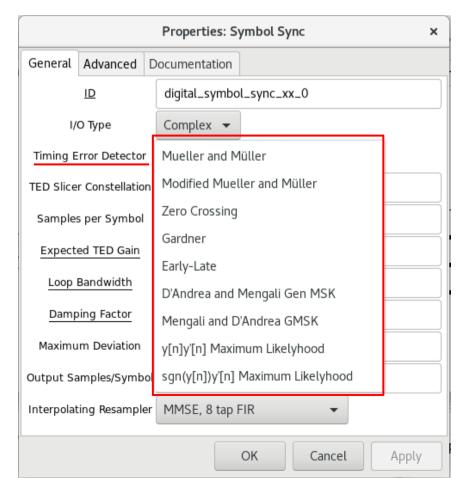
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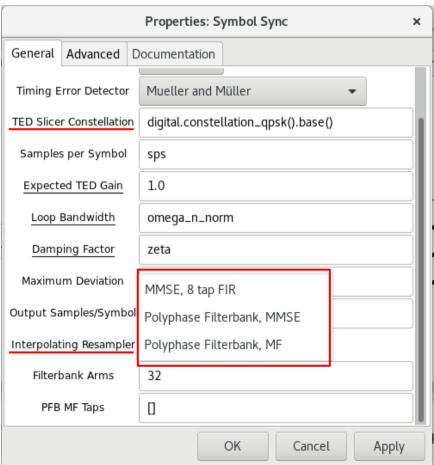
- Incorrect tag propagation
- Conflates symbol clock phase and interpolator phase
  - Self noise & Unable to stay locked on a clock pattern
- Incorrect decision slicer constellation
- Drops some input, when > 8 samples/symbol
- No reset on receipt of time\_est tag
- No way to change TED, slicer, or resampler
  - Whole new blocks needed bringing new bugs
- Initializes to very overdamped loop filter
- PI filter gain computations ignored TED gain
- Restricted to 1 or 2 samples/symbol on output

## New Symbol Sync Blocks



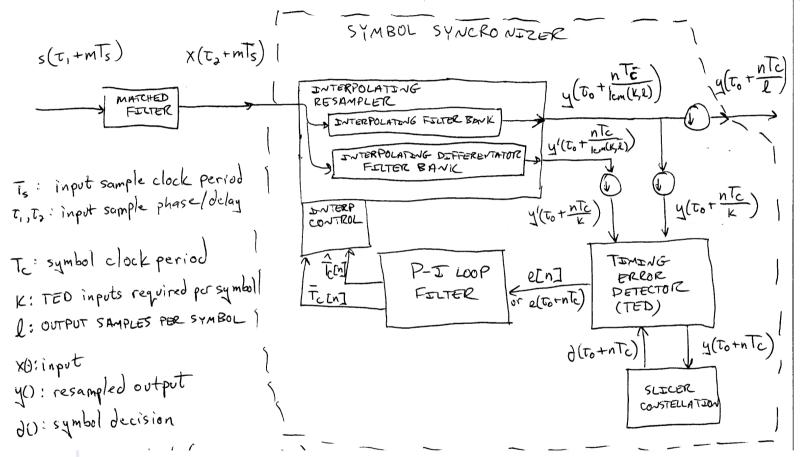
- Fixed, replacement superset of existing blocks
  - Except can't change MF taps on the fly (yet)
- Selectable TED, slicer, and resampler





### New Symbol Sync Blocks





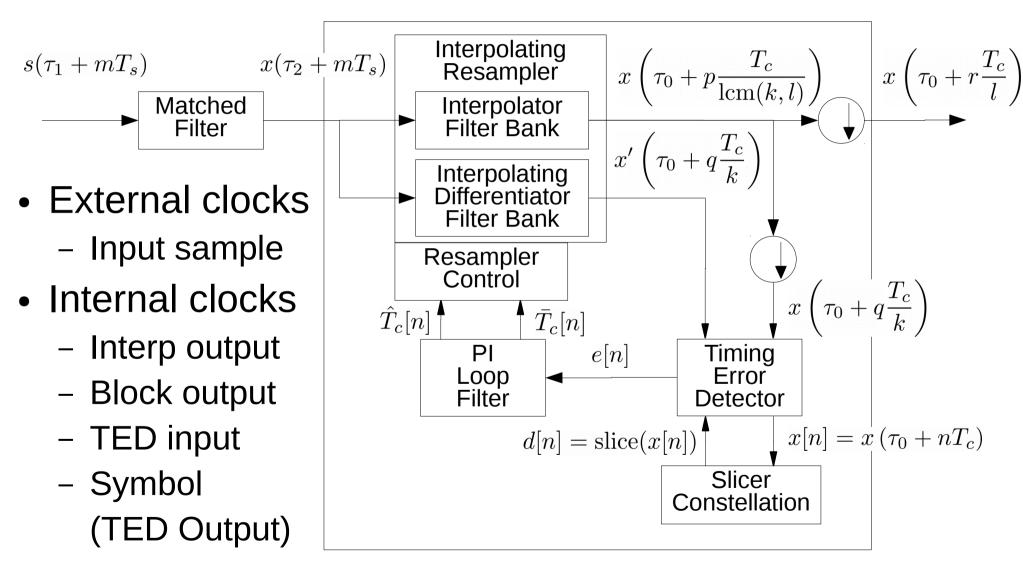


jmcorgan commented on Apr 23

@awalls-cx18 序 for not using an actual napkin for that diagram.

### New Symbol Sync Blocks





### Adding a New TED



- Modify the following files
  - gr-digital/include/.../timing\_error\_detector\_type.h
  - gr-digital/grc/digital\_symbol\_sync\_xx.xml
  - gr-digital/lib/timing\_error\_detector.\*
- Your new derived TED class only needs
  - A simple constructor
  - Two methods to compute the error output term
    - Complex input
    - Float input
- Leave the symbol sync blocks' code alone
  - tag handling, slicer, resampler, & loop filter all done!

### Adding a New Resampler



- Modify the following files
  - gr-digital/include/.../interpolating\_resampler\_type.h
  - gr-digital/grc/digital\_symbol\_sync\_xx.xml
  - gr-digital/lib/interpolating\_resampler.\*
- Your two (1 float, 1 complex) new derived resampler classes each need
  - A constructor
  - A simple ntaps() method
  - An interpolate() method
  - A differentiate() method (interpolating differentiator)
- Leave the symbol sync blocks' code alone
  - tag handling, slicer, TED, & loop filter all done!

### Using a Different Slicer

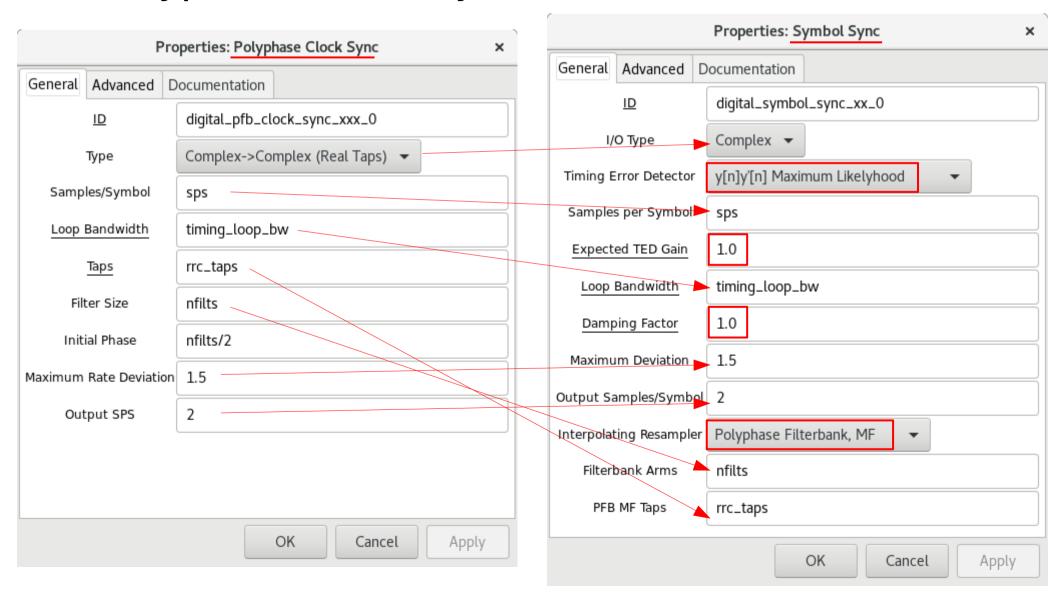


- Instantiate a custom Constellation Object
  - Slicer only needed for decision directed TEDs though
    - M&M, Modified M&M, Zero Crossing
- Pass in the constellation object as the TED slicer

- Leave the symbol sync blocks' code alone
  - tag handling, resampler, TED, & loop filter all done!

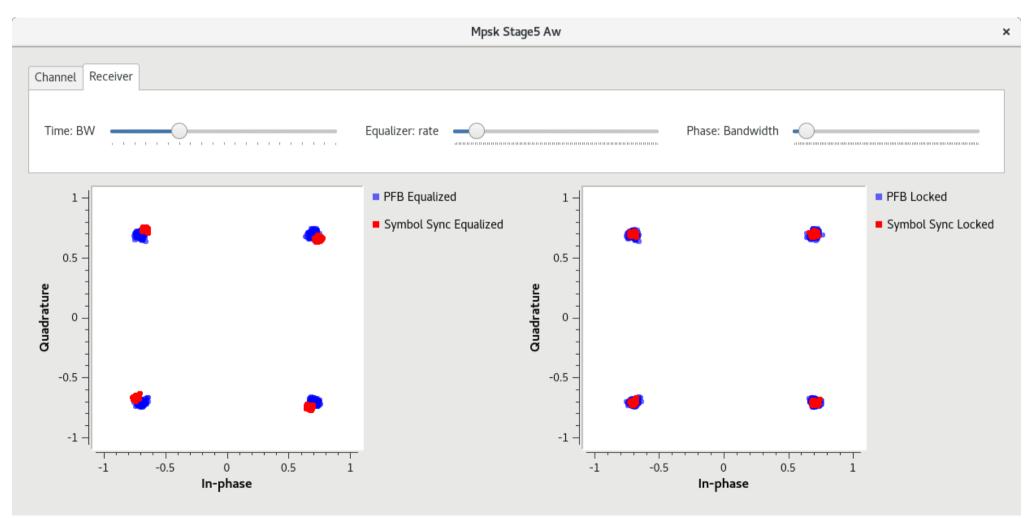
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Polyphase Clock Sync



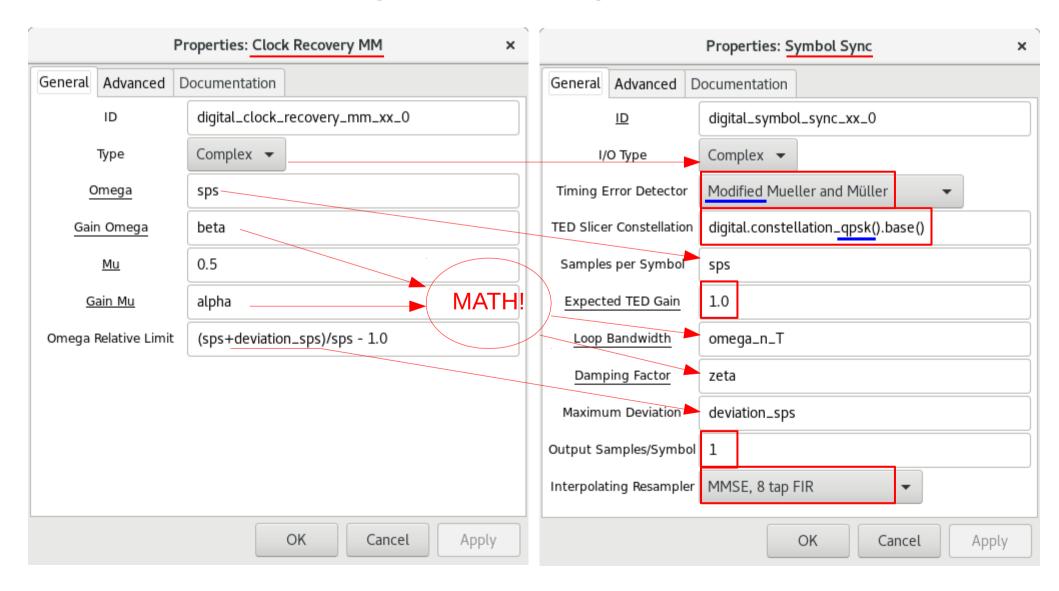
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• Polyphase Clock Sync comparison (MPSK tutorial)



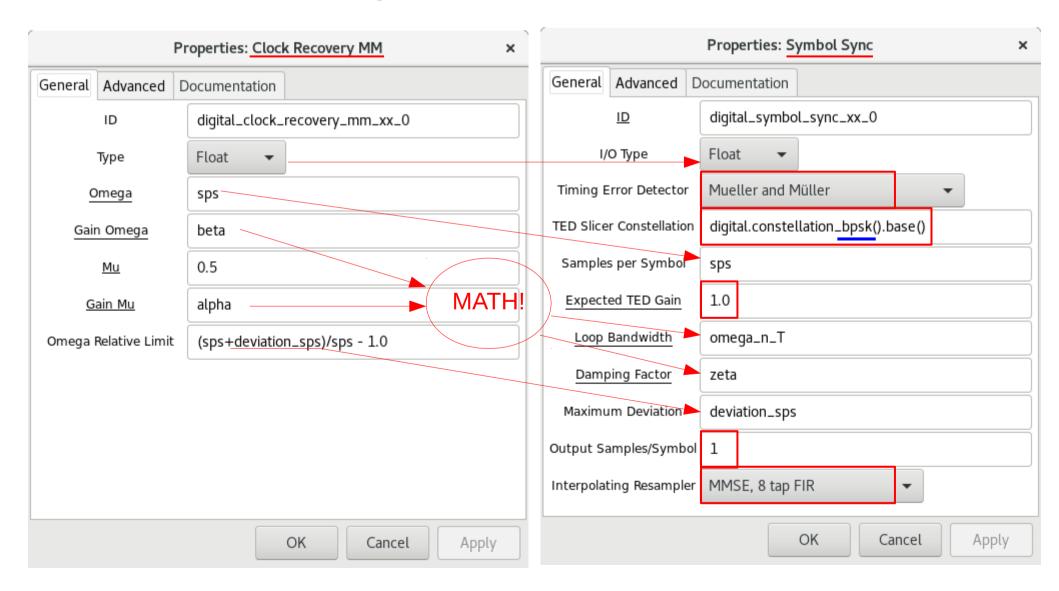


Clock Recovery MM, Complex I/O



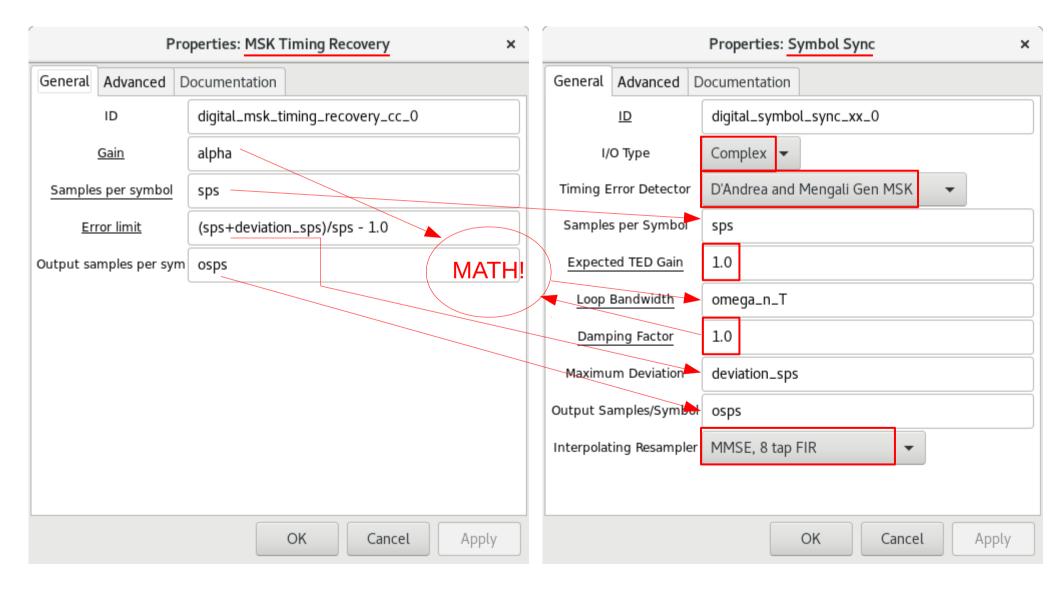
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Clock Recovery MM, Float I/O



SILVERBLOCK
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MSK Timing Recovery



### Usage Hints and Gotchas



- Easy stuff
  - Output samples/symbol can be in [1, 2, 3, 4, 5, 6, ...]
    - Normally set to 1; or to 2, if upstream from an equalizer block
  - Maximum deviation is in units of samples/symbol
    - Smaller is better for acquiring lock at start of burst
    - Too small misses data when symbol clock is far from nominal
  - Tracking resets on a "time\_est" or "clock\_est" tag
    - time\_est tag value is a PMT double
      - Sample offset estimate, in [-1.0, 1.0] samples, relative to tagged sample
    - clock\_est tag value is a PMT 2-tuple of doubles
      - Sample offset estimate, in [-1.0, 1.0] samples, relative to tagged sample
      - Symbol clock period estimate, in samples/symbol
    - clock est tag has priority over time\_est tag

### Usage Hints and Gotchas



- Input signal conditioning and filtering
  - MSK signals and MSK TEDs don't use matched filters
    - But a narrow IF filter can be beneficial
  - Input signal should be at a consistent amplitude (e.g. +/- 1.0)
    - AGC
    - TEDs have specific assumptions about input amplitudes !!!
  - Input signal amplitude should match constellation
    - Only for decision directed TEDs: M&M, Modified M&M, Zero Crossing
    - GNURadio's Constellation Object silently scales your constellation !!!
  - Input signal should normally be NRZ (no DC offset)
  - Input signal should be peaked at symbol centers
    - Except for MSK signals and MSK TEDs
    - Normally accomplished with a matched filter
  - Sync block's "PFB, MF" resampler can do the matched filtering
    - Except for rectangular pulse filter and a TED that needs a derivative !!!

### Usage Hints and Gotchs

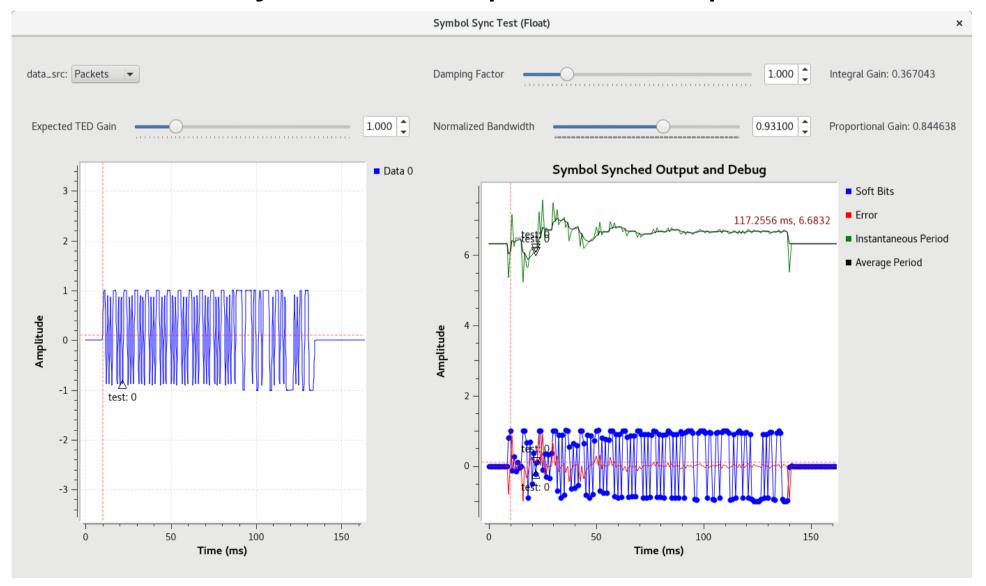


- Loop parameters and tuning
  - Use simulation to determine TED gain, K<sub>ted</sub>
    - Cannot know damping regime without it !!!
    - Ensure TED gain is scaled to the proper units for the loop !!!
  - Start with a critically damped, or over damped, loop
    - Damping factor, ζ, of 1.0, or greater than 1.0
    - An under damped loop *usually* isn't desirable for timing recovery
  - Use a Loop BW,  $\omega_n$ T, in [0.0,  $\pi$ (?)], usually closer to 0.0
  - Use simulation to determine optimal  $\zeta \& \omega_n T$  for best BER vs. Es/No
  - If you just want to play around and can accept suboptimal results
    - Start with  $K_{ted}$  = 1.0,  $\zeta$  = 1.0,  $\omega_n T$  = a number close to 0.0
    - Use GUI sliders to control all 3 of those values
    - Send all 4 outputs of the block to a single Time Sink/Scope
    - Adjust  $\omega_n T$  slider first, observing the primary,  $T_i$  inst, and  $T_i$  avg traces
    - Adjust  $\zeta$  to 1.3 or 1.5 or 2.0 (or 0.707 or 0.5), and try adjusting  $\omega_n T$  again
    - See gr-digital/examples/demod/symbol\_sync\_test\_float.grc

# Experimental Tuning Example

SYSTEMS

• Intentionally terrible loop BW example



### **TED S-Curve Simulation**



- gr-digital/examples/demod/\*\_ted\_gain.m
  - M&M TED gain:  $K_{ted} = 0.28271 \text{ sample}^{-1}$
  - Gardner TED gain:  $K_{ted} = 0.11810 \text{ sample}^{-1}$

