

NOTE HARMONIC POWER FLUX METHOD EMPHASIS ON SUDDEN ATTACK INSTRUMENT FAMILY FOR MIREX 2012 ONSET DETECTION

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ABSTRACT

This extended abstract proposes a new method for onset detection using harmonic information on each note. Especially, this method is optimized on sudden attack instrument family. Even though there are some exceptions such as missing fundamentals in piano first octave notes or only existence of even harmonics in clarinet, each note in music has integer multiple harmonics in common. This method outperforms about 6 and 17% in F-measure than existing algorithms such as spectral flux or spectral difference.

1. INTRODUCTION

Onset detection is important preprocessing step for audio signal processing. In the automatic transcription system, many previous researches about this problem based on learning [1] and signal processing [2][3][4][5] has been reported. Masri[5] has proposed the spectrum based onset detection functions such as dissimilar function. Juan[2] has proved that many spectrum power flux based onset detection algorithms have robust and fast performance. Recently, Sebastian Bock's BLSTM-RNN[1] algorithm based on three dimensional recurrent neural networks learning algorithm has become the state of art performance in MIREX 2011 onset detection. However, this algorithm is time expensive owing to the multi-layer neural network learning. Therefore, it is required to build time inexpensive and robust working algorithms for onset detection. In Chapter 2, we will propose two onset detection functions based on music instrument acoustics. In Chapter 3, three parameters are introduced for improving the onset detection performance. Test environment and evaluation methods are explained in Chapter 4 and we will show the test results followed by discussion.

2. ONSET DETECTION FUNCTIONS

We propose two onset detection functions based on note-scale filter bank power spectrum.

2.1 Note-Scale Filter Bank

As Figure 1 shows, the note-scale spectrum is composed of five blocks; HPF (High Pass Filtering), Windowing, $|STFT|^2$ (Power Spectrum), Note-Scale Filter Banks, and Log. Firstly, the high pass filtering is the first order FIR filter with [1 -0.97] coefficients, which pre-emphasize the high frequency components. We do this process for compensating for the weak high frequency power in the note spectrum. The windowing is performed for framing the time domain signal and do the short time fourier transform square for building power spectrum. Then, we weight and sum with the triangular note scale filter bank on this power spectrum. Lastly, we convert the note-scale filter bank output with log, which is an inevitably required process to discriminate more between the weak and strong power component in the log spectrum.

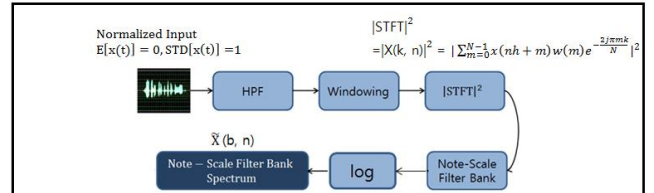


Figure 1. Note-Scale Filter Bank Spectrum Extraction Procedure

In Figure 2, the Note-scale filter banks have center frequencies as the musical note fundamental frequency, F0s. We designed each of the previous filter banks' center and end frequencies overlapping with the following filter banks' the start and center frequencies separately. The height has triangular shape with decreasing amplitude. It preserves the same energy on each filter bank. Lastly, we designed 103 filters, covering fully the weaker and higher power harmonics in music instruments.

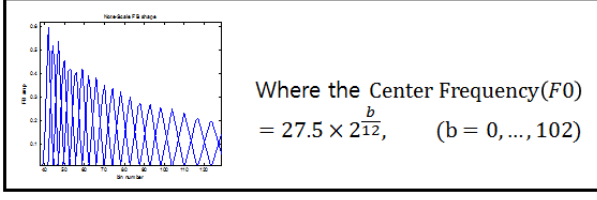


Figure 2. Note-Scale Filter Banks Design

2.2 Note Power Flux (NPF)

Note Power Flux is the column-wise sum of the two consecutive note-scale power spectrum difference. To detect only onset points except offsets, we passed half-wave rectifier to the spectrum difference as formula 1 defines.

$$NPF(n-1) = \sum_{b=1}^{103} H(\tilde{X}(b,n) - \tilde{X}(b,n-1)) \quad (1)$$

,where H represents half-wave rectifier.

2.3 NHPF (Note-Harmonic Power Flux)

For increasing the musical harmonic information in the feature, we add the first eight integer multiple frequency note filter bank bins on each pitch allowing repetition as formula 2 shows.

$$NHPF(n-1) = \sum_{P=1}^{88} \sum_{b \in B} H(\tilde{X}(b,n) - \tilde{X}(b,n-1)) \quad (2)$$

,where H represents half-wave rectifier.

In formula 2, the P is the pitch number from 1 to 88 representing A0 and C8 notes in music. B is the note filter bank bin set of the first eight integer multiple fundamental frequencies (Table 1).

| Octave # | Note # | Hz Ratio (Frequency/F0) |
|----------|--------|----------------------------|
| 1 | 1 (F0) | 1 |
| 2 | 13 | 2 |
| 3 | 20 | 2.996614≈3 |
| 4 | 25 | 4 |
| 5 | 29 | 5.039684≈5 |
| 6 | 32 | 5.993228≈6 |
| 7 | 35 | 7.12719≈7 |
| 8 | 37 | 8 |

Table 1. Note Number Frequency Ratio (Frequency and F0s are calculated through Center Frequency Equation)

In the high frequency notes, we add up to 103th note filter bank numbers.

3. PARAMETER SET

There are three parameters to improve the performance.

3.1 Threshold Value

The threshold value separates onset detection function's peak values into two groups; valid onset points and invalid onset points.

3.2 Time Delay Offset

On the time axis, the true positive points are detected points in the tolerance interval. We draw a histogram with the distribution of these true positive points and find the 50% existence point in the histogram. Then, we moved this point back and forth to locate on the same spot as the ground truth onset point. This time difference is called time delay offset.

3.3 Merge detected notes(Post Processing Process)

At the post processing, we merged onset detection functions' peaks, which are located in the pre-defined interval. Here, the shortest length note in the music is very important information for selecting the merging time.

4. TEST

4.1 TEST Procedure

| | |
|---------|--|
| Step 1. | Input Mixture Data Normalization |
| Step 2. | Transformation into Note-Scale Filter Bank |
| Step 3. | Calculate Onset Detection Function |
| Step 4. | Peak detection |
| Step 5. | Thresh-holding |
| Step 6. | Post Processing |
| Step 7. | Evaluate (with tolerance length, ±25msec) |

Table 2. Bello set NPP (non-pitched percussive) Test Procedure

We checked our algorithms with bello set non-pitched percussive instrument family such as drums (NPP). Test procedure is composed of seven steps. We preprocessed the 6 files chopped length music with mean zero and standard deviation one and transformed the time signal into note-scale filter bank output. We calculate NPF and NHPF onset detection value sequences and do the peak detection and lastly do the peak picking process through thresh-holding with the pre-optimized value. Comparing to the MIREX 2007 MFOD test, this experiment has half the delay offset value and 10% smaller threshold value. We evaluated the tolerance interval length with ±25msec.

4.2 Evaluation Method

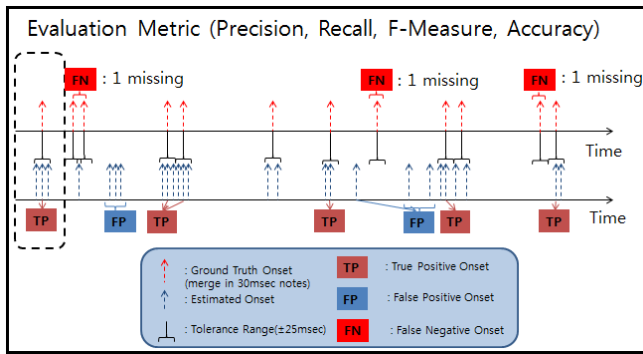


Figure 3. Evaluation Example

From the Figure 3, the first three detected onsets are true-positive points inside the tolerance range. In the second group, only one onset is detected and one is missing, which makes false negative point. All other remains in the third group points are false positives.

5. RESULTS

5.1 Experiment Result

As Table 3 shows, NHPF has the best F-measure values 0.9458 among three other onset detection functions. Also, the NHPF has the best recall value, 0.9320. Only SD has the best performance than NHPF in precision as 0.9906. However, comparing to the conventional method such as SF and SD, the NPH and NHPF has better performance.

| | Precision | Recall | F-measure |
|------|---------------|---------------|---------------|
| SF | 0.899 | 0.8641 | 0.8812 |
| SD | 0.9906 | 0.6364 | 0.7749 |
| NPF | 0.9545 | 0.9175 | 0.9356 |
| NHPF | 0.9600 | 0.9320 | 0.9458 |

Table 3. Bello set NPP (non-pitched percussive) test result

(SF and SD represent spectral flux and spectral difference each.)

6. DISCUSSION

If we change the two test processes; peak detection and thresh-holding, we will miss some peaks, which have only peak point over the thresh-holding line.

7. REFERENCES

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