NANYANG TECHNOLOGICAL UNIVERSITY

SEMESTER 2 EXAMINATION 2017-2018

EE3001 - ENGINEERING ELECTROMAGNETICS

April / May 2018

Time Allowed: 21/2 hours

INSTRUCTIONS

- 1. This paper contains 4 questions and comprises 7 pages.
- 2. Answer ALL questions.
- 3. All questions carry equal marks.
- 4. This is a closed book examination.
- 5. Unless specifically stated, all symbols have their usual meanings.
- 6. A list of formulae and physical constants is provided in Appendix A on pages 5 to 7.
- 7. The Smith Chart may be used in the solution of Question 4. Please put the Smith Chart inside the answer script and tie it with a thread.
- 1. (a) Three semi-infinitely long straight wires are shown in Figure 1 on page 2. The wires have an infinitesimally small gap between them at the origin. The wires are uniformly charged with line charge density of $+\rho_l$, $-\rho_l$, and $+\rho_l$, respectively.
 - What is the direction of the combined electric field intensity due to wires 1 and 2? Explain your answer briefly.
 - (ii) Determine the combined electric field intensity at point $(0, 0, z_0)$ $(z_0 > 0)$ due to the three wires.

It is given that
$$\int \frac{zdz}{(b+z^2)^{1.5}} = \frac{-1}{\sqrt{b+z^2}}$$

(13 Marks)

Note: Question No. 1 continues on page 2.

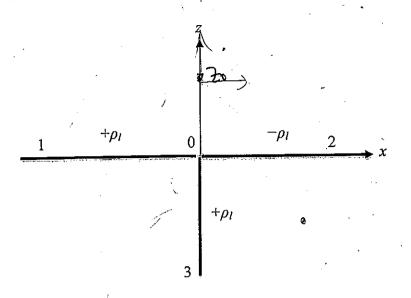


Figure 1

- (b) A thin conducting wire of length 4w forms a square loop in the xy-plane. A direct current I flows in the wire along the counter-clockwise direction looking from the +z direction.
 - (i) What is the direction of the overall magnetic field due to the loop? Explain your answer briefly.
 - (ii) Find the overall magnetic field intensity at point $(0, 0, z_0)$ $(z_0 > 0)$. It is given that $\int \frac{dz}{(a+z^2)^{1.5}} = \frac{z}{a\sqrt{a+z^2}}$ (12 Marks)
- 2. (a) A circular conducting loop of radius a = 0.2 m lies in the y = 0 plane with its centre at the origin.
 - (i) Find the voltage induced in the loop if it lies in a magnetic flux density $\vec{B} = 0.05 \cos(10^3 t) (\vec{a}_y + 2\vec{a}_z) \text{ T.}$
 - (ii) Using a diagram, show the direction of the induced current in the loop at t = 0. Give reasons for your answer.

 (12 Marks)

Note: Question No. 2 continues on page 3.

(b) A 4-GHz uniform plane wave is propagating in a lossless medium. The medium has a relative permittivity $\varepsilon_r = 8$ and a relative permeability $\mu_r = 2$. It is observed that the wave has the following electric field components:

$$E_y = 2\cos(\omega t - kx)$$
 V/m and $E_z = 3\cos(\omega t - kx)$ V/m.

- (i) Determine the magnetic field components of the wave.
- (ii) Determine the <u>angles</u> made by the electric field vector \vec{E} and the magnetic field vector \vec{H} with respect to the z-axis.
- Determine if Gauss' Law is satisfied by \vec{E} and \vec{H} . Give brief reasons for your answer.

 (13 Marks)

3. (a) The electric field of a uniform plane wave (UPW) in free space can be expressed as follows:

$$\vec{E} = (E_x \vec{a}_x + E_y \vec{a}_y + E_z \vec{a}_z) e^{-j(8x+6z)}$$
 V/m.

- (i) Determine the direction of time-average Poynting vector.
- (ii) Write down the corresponding magnetic field expression.
- (iii) For the electric field to represent a linearly polarized UPW, state all the pertaining conditions in terms of its components (i.e., $E_x = |E_x| \angle \phi_x$, $E_y = |E_y| \angle \phi_y$, $E_z = |E_z| \angle \phi_z$). You should consider all possibilities of having none, one or two of the components being zero.

(14 Marks)

- (b) A uniform plane wave in air (occupying the region $z \le 0$) is obliquely incident at an angle $\theta_i = 30^\circ$ onto a lossless medium (occupying the region $z \ge 0$) with $\varepsilon = 2.1\varepsilon_0$ and $\mu = \mu_0$.
 - (i) Determine the angle of transmission θ_t .
 - (ii) Calculate the reflection and transmission coefficients for parallel polarization, i.e., Γ_{\parallel} and τ_{\parallel} .
 - (iii) An analytical relation between the coefficients above is given by

$$1 + \Gamma_{\parallel} = \tau_{\parallel} \left(\frac{\cos \theta_t}{\cos \theta_i} \right)$$

Check whether this relation is satisfied using your answers above.

(iv) Discuss briefly how the relation above may be derived analytically and whether it is applicable for a lossy medium.

(11 Marks)

A transmission line in (air) has characteristic impedance $Z_0 = 50 \Omega$ and 4. length l = 0.42 (with λ being the wavelength). It is connected to a generator at z = -l and terminated with an unknown load at z = 0. The generator has an open-circuit voltage $V_g = 10 \text{ V}$ and an internal impedance $Z_g = 50 \Omega$. The ratio between the power delivered to the load P_L and the maximum available power from the generator P_{av} is given as

$$\frac{P_L}{P_{av}} = \frac{8}{9}$$

- What are the maximum available power from the generator P_{av} and the power delivered to the load P_L ?
 - Determine the magnitude of load reflection coefficient $|\Gamma_L|$ and the standing wave ratio (SWR). (ii)
 - It is found that the first voltage maximum occurs at $z_{\text{max}} = -0.35\lambda$. Determine the phase angle of load reflection coefficient $\theta_0 = \angle \Gamma_L$ (in degrees) and the load impedance Z_L .
 - (iv) Determine the input reflection coefficient Γ_{in} at z = -l.

(16 Marks)

- The load in part (a) is subsequently removed and the transmission line is left (b) open-circuited.
 - What are the input reflection coefficient Γ_{in} and the input impedance (i) Z_{in} at z = -l?
 - Determine all the positions of voltage minima z_{\min} on the (ii) transmission line.

(9 Marks)

State any assumption made in the above. The Smith chart may be used in the solutions for one or both parts of this question. Please put the Smith chart inside (not outside) the answer script and tie it with a thread.



Physical Constants

Permittivity of free space
$$\varepsilon_0 = \frac{1}{36\pi} \times 10^{-9} \,\text{F/m}$$

Permeability of free space
$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

∇ Operator

$$\nabla V = \vec{a}_x \frac{\partial V}{\partial x} + \vec{a}_y \frac{\partial V}{\partial y} + \vec{a}_z \frac{\partial V}{\partial z}$$

$$\nabla \cdot \vec{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$$

$$\nabla \times \vec{A} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix}$$

$$\nabla V = \vec{a}_{r} \frac{\partial V}{\partial r} + \vec{a}_{\phi} \frac{\partial V}{r \partial \phi} + \vec{a}_{z} \frac{\partial V}{\partial z}$$

$$\nabla \cdot \vec{A} = \frac{\partial (rA_{r})}{r \partial r} + \frac{\partial A_{\phi}}{r \partial \phi} + \frac{\partial A_{z}}{\partial z}$$

$$\nabla \times \vec{A} = \frac{1}{r} \begin{vmatrix} \vec{a}_{r} & r\vec{a}_{\phi} & \vec{a}_{z} \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ A_{r} & rA_{\phi} & A_{z} \end{vmatrix}$$

Electric and Magnetic Fields

$$\vec{E} = \frac{1}{4\pi\varepsilon} \iiint_{V} \frac{\rho_{v} \vec{a}_{R}}{R^{2}} dv = \frac{1}{4\pi\varepsilon} \iiint_{V} \frac{\rho_{v} \vec{R}}{R^{3}} dv$$

$$V = \frac{1}{4\pi\varepsilon} \iiint_V \frac{\rho_v}{R} dv$$

$$\vec{H} = \frac{1}{4\pi} \int_{C} \frac{\vec{ldl} \times \vec{a}_{R}}{R^{2}} = \frac{1}{4\pi} \int_{C} \frac{\vec{ldl} \times \vec{R}}{R^{3}}$$

$$\oint_C \vec{H} \cdot \vec{dl} = I = \iint_S \vec{J} \cdot \vec{ds}$$

$$emf = \oint_C \vec{E} \cdot \vec{dl} = -\frac{d\Phi_m}{dt} = -\frac{d}{dt} \iint_S \vec{B} \cdot \vec{ds}$$

Maxwell's Equations

$$\nabla \times \tilde{E} = -\frac{\partial \tilde{B}}{\partial t}$$

$$\nabla \times \tilde{H} = \frac{\partial \tilde{D}}{\partial t} + \tilde{J}$$

$$\nabla \cdot \tilde{D} = \rho$$

$$\nabla \cdot \tilde{B} = 0$$

Complex Propagation Constant

$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\varepsilon_c} = j\omega\sqrt{\mu(\varepsilon - j\sigma/\omega)}$$

Complex Intrinsic Impedance

$$\eta_c = \sqrt{\frac{\mu}{\varepsilon_c}} = \sqrt{\frac{\mu}{\varepsilon - j\,\sigma/\omega}}$$

Reflection and Transmission of Electromagnetic Wave

$$\frac{\sin\theta_t}{\sin\theta_i} = \sqrt{\frac{\mu_1\varepsilon_1}{\mu_2\varepsilon_2}} \qquad \tan\theta_{B\parallel} = \sqrt{\frac{\varepsilon_2}{\varepsilon_1}} \qquad \sin\theta_c = \sqrt{\frac{\varepsilon_2}{\varepsilon_1}}$$

$$\Gamma_{\perp} = \frac{\eta_2 \cos(\theta_i) - \eta_1 \cos(\theta_t)}{\eta_2 \cos(\theta_i) + \eta_1 \cos(\theta_t)} \qquad \qquad \tau_{\perp} = \frac{2\eta_2 \cos(\theta_i)}{\eta_2 \cos(\theta_i) + \eta_1 \cos(\theta_t)}$$

$$\Gamma_{\parallel} = \frac{\eta_2 \cos(\theta_t) - \eta_1 \cos(\theta_i)}{\eta_2 \cos(\theta_t) + \eta_1 \cos(\theta_i)} \qquad \qquad \tau_{\parallel} = \frac{2\eta_2 \cos(\theta_t)}{\eta_2 \cos(\theta_t) + \eta_1 \cos(\theta_i)}$$

Transmission Line

$$V(z) = V_o^+ e^{-j\beta z} + V_o^- e^{+j\beta z}$$

$$I(z) = \frac{1}{Z_o} \{ V_o^+ e^{-j\beta z} - V_o^- e^{+j\beta z} \}$$

$$\beta = \frac{2\pi}{\lambda}$$

$$\Gamma_L = \frac{V_o^-}{V_o^+} = \frac{Z_L - Z_o}{Z_L + Z_o}$$

$$\Gamma(z) = \Gamma_L e^{+j2\beta z} \qquad -\ell \le z \le 0$$

$$Z_{in}(-\ell) = \frac{Z_L + jZ_o \tan(\beta \ell)}{Z_o + jZ_L \tan(\beta \ell)} Z_o$$

END OF PAPER

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EE3001 April/May 2018 PXP Solution

		Date	No.
] (a)	$\frac{1}{2}$ $\frac{1}{2}$ $\frac{1}{2}$ $\frac{1}{2}$ on	Z-axis: points	to the right
(i)		, 1	U
		2 (4) quadrant : p	points to the upper night
	200	(3) quadrant: p	joints to the lower right
(1.1	§ / (9)		
(ji)		D - T) 7	7 7 7
	$ \begin{array}{c c} F = (0, 0, \overline{2}_0) \overline{S} = (X, 0, 0) \\ A(= dX) \overline{R} = \overline{(X^2 + 2)^2} \end{array} $	120 R=F-3	$3 = -x\overrightarrow{ax} + z\overrightarrow{az}$
	$\overline{E}_{1} = \frac{1}{4720} \int_{-\infty}^{\infty} \frac{\rho_{1}R}{k^{2}} dt = \frac{\rho_{1}}{4720}$	$\int_{-\infty}^{0} \frac{(-x \cdot 0x + 2x)}{(x^2 + 2x)}$	az) dv
	7,000		2) 2 0 1
	- Pi (0 -xax dx +)0 -xax - 1 -a	3, az (x+3) = dx)	
	4x20) -0 (x2+20)=) -a	(x+3) = 1	
	= P. / [= xax 1 + 10	70 al	
	= Pr ([~ xax dx +] ~ dx + [~ dx + dx	(X2+22)= dx)	
	- D. IF - Mr 1 = T =	57 5 70 7	
		23 \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	·
	Pi / ox , ,)		
	line 2 : P = 10.0570 $S' = (X.0.0)$ $R' =$	T 7 - VA	7 + 2002 X20
	$dl = dX$ $ R = \frac{1}{12^{2} + 3^{2}}$	J=3 NOV	1 7 50 UZ ///
	$ \longrightarrow $	- Pi 100 1-xax	1+3a2) du
	E2 = 4750 0 R3 M -	4750 JO 1x2	十六十
	$=\frac{-P_1}{(1/2)^2} \left(\frac{\infty}{1} \frac{-X\alpha x}{(1/2)^2} \right) \frac{1}{2} dx + \left(\frac{\alpha}{1} \frac{-X\alpha x}{(1/2)^2} \right) \frac{1}{2} dx$	20 Zaz (X)	
	- 425 () (X720) 2 T 02 X.	(XTZ) > -	P1 1 - 02 1 1
	$= \overline{4750} \left(\overline{Jx^4z^3} \right) 0 \left(\overline{z_0} \overline{x^4z^3} \right)$	_	75. (-M-+)
	line 3:		
	F = (0, 0, 2) $S = (0, 0, 2)$	R'=f-y=1	72) 02 2/0
<u> </u>	$ \alpha = \alpha + R = (3 - 2)$		· · · · · · · · · · · · · · · · · · ·



	Date No.
	E= 1 [0 PiR A1 = Pi [0 (20-2) Q2 A)
-	4750 J-20 R2 Olt = 4750 Ja . (70-7)3
	$=\frac{\ell_1\Omega_2}{4\chi_{\Sigma_0}}\int_{-\infty}^{0}\frac{1}{(Z_0-Z_0)^2}dz=\frac{\ell_1\Omega_2}{4\chi_{\Sigma_0}}\left[\frac{-1}{Z_0-Z_0}\right]_{-\infty}^{0}$
• • •	Piaz (-1) -Piaz
	- 425° (30) - 425° Zo
	$\rightarrow \rightarrow \rightarrow \rightarrow \leftarrow \rightarrow \leftarrow \rightarrow \rightarrow$
	Etot = E' + E' + E' = \(\frac{1}{47} \cdots \)
	- Pi (20x - TB) /m
(d) (d)	
<u> </u>	is to the 77 direction.
(ii)	The loop is at the position shown below
	2 (Xo, Yno) / (Xo, Yotw, O)
	(Xo, Mo) (Xo, MotW) (Yo)
	2 4
	2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2 2
	X (X, tw, y, o) (X, tw, o)
line 1	$\vec{T} = (0,0,Z_0)$ $\vec{S} = (X_0,Y_0)$ $\vec{R} = \vec{T} - \vec{J} = (-X_0,-Y_0,Z_0)$
(like)	
	$R = \sqrt{x^2 + y^2 + 2\delta^2} dT = -dy dy$
	$d\vec{l} \times \vec{R} = -dy \vec{a}_{1} \times (-X - X - y \vec{a}_{2} + 3 \vec{a}_{2}) = -(X - x \vec{a}_{2} + 3 \vec{a}_{3} + 3 \vec{a}_{2})$
	1
	Mo (RTZDTY)
	= '-I(Xaz+zax) (Moth dy.,
	4x Jyo (x'+20+4)=
	$= -I(\sqrt[3]{3} + 3\sqrt[3]{4}) I I I I I I I I I I I I I I I I I I I$
	47. L (X2+32)J:X3+32+42-19.
	= -I(x2+30x) T y.+w
	47 [[X2+25] [x2+26+ [4,+w]) (X2+26) [x2+22+42]



	<u> </u>
	Date No. 5
line 2	$f=(0,0,\overline{20})$ $S=(X,Y_0,0)$ $R=P-\overline{S}=(-X,-Y_0,\overline{20})$
	$R' = (X^2 + y^3 + Z^2) \qquad dT = dX dX$
	$II \times R = d \times \overline{a} \times (-X \overline{a} - Y \cdot \overline{a} + Z \cdot \overline{a}) = -(Y \cdot \overline{a} + Z \cdot \overline{a}) d x$
	I Notward -I Notw (Your + Zoay)
	$H_2 = 4\sqrt{1 + 1 + 2 + 2 + 2} = 41 + 2 + 2 + 2 + 2 + 2 + 2 + 2 + 2 + 2 + $
	= - I (YO W + ZO QY) / X / X+W
····	4x [(4)+2,2) [4,2+2,4x2] X6
	$= -I(\sqrt{0} + 2 \sqrt{3}) / \sqrt{x_0 + w} $ Xo
	47 (1/3+26) (1/3+26+W)2 (1/3+26) (1/3+26+X6)
line 3	$F' = (0,0,2)$ $S' = (x,+w,y,0)$ $R = F - S' = (-(x,+w), -y, z_0)$
,t. 	
	$ \mathcal{R} = \int (x_0 + w)^2 + y^2 + Z_0^2 \qquad d\overline{U} = dy dy$
	$d\vec{r} \times \vec{R} = (x_0 + w) \vec{Q}_0 + z_0 \vec{Q}_0$
	I Mitw (Xo+W B + Zo B)
	H3 = 42 Jy - Tixofw - [4] + 202) = 0
	$-\sqrt{1}(x+w)\omega + 2\omega x$
	- / (N) VE 12-M/
· · · · · · · · · · · · · · · · · · ·	$4z$ $((x_0+w)^2+2b^2)\sqrt{(x_0+w)^2+2b^2+4b^2}$
	7 ((X.+W) 02 + 200x) [40+W
	= 42 [[(X,+w)'+2,2)] (X,+w)'+2,+(y,+w)'
	T > 1 > 2 > 2 > 1 = 2 = 2 = 2 = 2 = 2 = 2 = 2 = 2 = 2 =
11 (4	$[(x_0+w_0^2+z_0^2)_1(x_0+w_0^2+z_0^2+y_0^2)]$
line4	$f = (0,0,\overline{20})$ $S = (X, Y,tw, 0)$ $R = \overline{f} - \overline{S} = (-x, -(Y,tw), \overline{20})$
	$ R = \int x^2 + (y_0 + w)^2 + 2\delta^2 dT = -dx dx $
	H4 = 47 /x 1 x 1/x 1/x 1/x 1/x 1/x 1/x 1/x 1/x
	= T[(4,+w) 02 + 2, ay) / x.+w
	A State of the Control of the Contro
	Tou 11 42 1 22 1 1 1 1 1 1 2 2 2 2 2 2 2 2
	Fton = Hi + Hi+Hi+Ho [14.+W'+Zo'] [X-+14.+w'+Zo'] [14.+W'+Zo'] [X-+14.+w'+Zo']
·	HIM - HI T IN T IN



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~ () () \	Date No.
2(a) (i)	
	$\frac{1}{3} - \frac{1}{2} \times \frac{10^{-3} \cdot \cos(10^{3} + 1)}{1}$
	$\frac{1}{2} \frac{1}{2} \frac{1}$
	$emf = -dom = 2xN^{-3}x/o^{3}sin(10) = 2sin(10) t)$
(ij)	B A to to in manual, the
*	t=0 : no magnetic flux t=ot: magnetic flux increase from D in ay direction, according to Lenz's Law, the
(not sure)	divorting according to lent (law, the
	direction of the induced current is in
· ·	Clockwise direction looking from y axis
2.bli)	1= 1 = 1 = 120x = 60x SZ
	1 JE - JET NIZON -0012 32
	$E = (2\alpha y + 3\alpha z) \omega s(wb - kx) V/m$
	$\overline{H} = \overline{Cx} \times \overline{E} = \frac{1}{300} \overline{Cx} - \frac{1}{2000} \overline{Cx} + \frac{1}{2000$
(ii)	
<u>U1/-</u>	angle of \mathbb{R}^{2} : $\tan^{-1}(\frac{2}{3}) = 33.69^{\circ}$. Angle of \mathbb{R}^{2} : $\tan^{-1}(\frac{1/20x}{1/30x}) = 56.31^{\circ}$
	Angr of 1 1/300 / = > 612 /
(ii)	Craves's Law is not satisfied by It because VII +0 =>
(not sure)	$\nabla \cdot \mathcal{B}' \neq 0$
,	not for \vec{E} because $\vec{\nabla} \cdot \vec{D} = \vec{\nabla} (\vec{\Sigma} \vec{E}) \neq \vec{P}$
3(a)(i)	$p.8ab \cdot + a.6ab = ab$
(ii)	y=120人几 中于18xxxxxxxxxxxxxxxxxxxxxxxxxxxxxxxxxxxx
	$= e^{-3(8\pi a)} \cdot (0.8 \text{ Ey } a) - 0.8 \text{ Ez } a) + (0.6 \text{ Ey } a)$
(50)	20T
(iii)	plane of Incidence $\equiv x \neq -plane$.
	$\frac{11 - P_0 \text{div} \text{ sation}}{2} = \frac{1}{2} = \frac{1} = \frac{1}{2} = \frac{1}{2} = \frac{1}{2} = \frac{1}{2} = \frac{1}{2} = \frac{1}{2} = $
	E11=(Ex e11 0x + Ez e 12 0z)
	$E_{11} = \int E_{x}^{2} e^{j2\phi x} + E_{z} ^{2} e^{j2\phi z}$ (not sure)
	L-Polarization, iverto
	$\overline{E}_{1}^{2} = E_{1} e^{\int d^{2}z} e^{-\int t8 \times t6z}$
	[E] = [E]/2



	Date	No. 5
D Dne component i's D		
5=0 F "= [Ez <02 E =] [- 1 LON	
=) E2 =0 0R Ey =0 0F	$2 \phi_2-\phi_1 =00$	R / 102 - Out 180°
=> linear		
Ey=0 linear	• ,	
$E_{2}=0 E'' = E_{X} \angle b_{X} E^{\perp} = E_{Z}$	y) < 0 y	
=> Ex 20 OR Ey =0 DR	9x-9y) =0 OR/	$\phi_{X} - \phi_{Y} = 180^{\circ}$
=) linear		·
D two components are o	· · · · · · · · · · · · · · · · · · ·	
5x=5y=0 E+ = 0 => linear	, 	
$Ex=E_{2}=0$ $ E'' =0$ $=0$ linear		. ,
Ey=E>=0 E' =0 => linear	· ·	
3 none components 15 zero) 1\	
$\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{\underline{$	o =) linear	• ?
	=0 => linear :	p
$\frac{1}{3(b)(1)} \frac{1}{\sin \theta t} \frac{1}{\sqrt{4}} \frac{1}{\sqrt{2}} \frac{1}{$	=> linear	1 2 2 1 - 2 10 P
$\frac{3(b)(1)}{\sin \theta_i} = \frac{\ln \xi_1}{\ln \xi_2} = \frac{1}{21}$	> Dt = sh-1 (] 211	Xs11180/ = 2018
(ii) Fir = M2 COS LOE) -M, COSLOI)	- N1111	
12 costot) + Micustoi)	=-0144	,
N2=1202 1 = 260,1522	COS (DE) = CUS (2a	18% = a939
Cosloi) = Cosl300) = 13/2		
$T_{ii} = 2 n_2 \cos(\theta_i)$	= 0.7895	
1/2 CUSLOT) + 1/1 COS (D1)		
(ii) 1+ [ii = 0.85b	Power Conserva	itibn
Ty (cosot) = 0.816	$P_1'' = P_k'' + P_k$	<u> </u>
(iV) medium 2 Kt	of 1= St COSOt	
17=Stadoi		
	$S_i^{ij}(\omega \theta) = S_i^{ij}(\omega \theta)$	C110: + (1' CUSD+
Oi V	SI WO 1 - 37 (701 1 37 ° 20 (
D('=si wibi	S1 Cospi + (1) to	i Casor = Si Til WOH
		1 0001 5 010001



	Date No.
	It's applicable for lossy medium because Power conservation is still
	hold at the intertace.
4 (a)	I am not sure for O4
(i)	$P_2 - Z_{11} - 8 \rightarrow 7 - 6000$
	Pav zg +zin
	P1 = 2 70 + 2in Re(Zin) = 2 10 X400 = 0.0988 W
	1 - 2 50 + 51 1 = 5 = 1 70 1 = 5 = 1
	Pay = 0.111'W Zin(-() = 400 = Zo. 1+re+2)\$2
	, , , , , , , , , , , , , , , , , , ,
	$=) \Gamma_{L} = \frac{7}{9} e^{jkb} \sqrt{\frac{-1}{1+ \Gamma }} \sqrt{\frac{-1}{1+ \Gamma }}$
	$ T_{L} = 7/q = SWR = 1- R = 8$
<u>(ii)</u>	Dot2 & Zmax = Do + 2.27/2 (-a352) =0 =0 =1.42
**************************************	Do -147 = -27 => 100 = -106 T = -108°
	Z = 1+1- = 0.387-1/4123
N	$\frac{20}{2L} = 50(0387 - 1/4123) = 19.35 - 170.615$
(jv)	17121 - 1 0 + 3232
	= 1/2 e J LOO + 2 & 2) = 7 e - J2,27
(d)	$\frac{20}{1000} = \frac{-50!}{1000} = \frac{-68(82)}{1000}$
(j)	$\angle oc = tan \beta tan (a K)$
	201-20 -j1-257
	$l^2 = 2pc + 2o$
	$\Gamma(-1) = \Gamma_1 e^{-\frac{1}{2}\beta^2} = e^{-\frac{1}{2}\frac{28}{5}\Gamma/3}$
(ji)	I Company
—— <u>(,)</u>	Or = Do + 2BZmin = -T => Zmin = -D(5)
	· · · · · · · · · · · · · · · · · · ·
	Good tuck for your exams 1 10
	·
	<u> </u>