

NANYANG TECHNOLOGICAL UNIVERSITY
SEMESTER 2 EXAMINATION 2020-2021
EE3001 – ENGINEERING ELECTROMAGNETICS

April / May 2021

Time Allowed: 2½ hours

INSTRUCTIONS

1. This paper contains 4 questions and comprises 7 pages.
 2. Answer all 4 questions.
 3. All questions carry equal marks.
 4. This is a closed book examination.
 5. Unless specifically stated, all symbols have their usual meanings.
 6. A list of formulae and physical constants is provided in Appendix A on pages 5 to 7.
 7. The Smith Chart may be used in the solution of Question 4. Please put the Smith Chart inside the answer script and tie it with a thread.
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1. (a) A line charge of uniform charge density ρ_l is located in free space along the z axis from $z = a$ to $z = b > a$.
 - (i) Using Coulomb's law, determine the electric field intensity \vec{E} at the point $(x, y, 0)$ due to the line charge.
 - (ii) Simplify your expression of \vec{E} above for the case of $a = -\infty$ and $b = \infty$. Give your answers for both Cartesian and cylindrical coordinate systems.

Note:
$$\int \frac{1}{(x^2 + u^2)^{3/2}} dx = \frac{x}{u^2 \sqrt{x^2 + u^2}}$$

(15 Marks)

Note: Question No. 1 continues on page 2.

- (b) Assume that the charges in part (a)(ii) are moving in the $+z$ direction to form an infinitely long direct current I . Determine the magnetic field intensity \vec{H} at the point $(x, y, 0)$ due to the line current. Give your answers for both Cartesian and cylindrical coordinate systems.

(10 Marks)

2. (a) Consider a square loop of area a^2 in the xy -plane in free space. At time $t = 0$, the loop has its center position at the origin, and is moving at constant velocity v along the $+x$ axis. The loop region is subjected to a spatially uniform but time-varying magnetic flux density of the form (for time $t \geq 0$)

$$\vec{B} = (C_1 t^2 + C_2 t) \vec{a}_z \text{ T},$$

where C_1 and C_2 are arbitrary constants.

- (i) Find the magnetic flux Φ_m passing through the loop and the induced voltage V_{emf} at time $t \geq 0$. State any assumption made.
- (ii) Assume that the loop has a uniform per-unit-length resistance of R_l (in Ω/m), determine the induced current I_{ind} at time $t \geq 0$.

(11 Marks)

- (b) A lossy medium is characterized by dielectric constant $\epsilon_r = 10$, loss tangent $\tan \delta = 8$, and relative permeability $\mu_r = 1$ at 500 MHz.

- (i) Comment whether the medium is a good conductor. Find the conductivity of the medium.
- (ii) Assume that a 500 MHz plane wave is propagating along $+z$ direction in the medium, calculate the complex intrinsic impedance η_c , the attenuation constant α and the phase constant β .

(14 Marks)

3. (a) The magnetic field of a uniform plane wave (UPW) travelling in a lossless non-magnetic medium occupying the region $z \leq 0$ is given as

$$\tilde{H}_i(z, t) = -\vec{a}_y 50 \cos(6 \times 10^9 t - 24.5z + 40^\circ) \text{ mA/m.}$$

The UPW is incident normally on a plane interface at $z = 0$ with a lossy medium having complex intrinsic impedance $\eta_c = 10 \angle 45^\circ \Omega$ and occupying the region $z \geq 0$.

Determine the following and state any assumption(s) made:

- (i) The phase velocity u_p of the UPW in the lossless medium.
- (ii) The permittivity of the lossless medium.
- (iii) The time-domain expression of the incident electric field $\tilde{E}_i(z, t)$.
- (iv) The percentage of average incident power reflected at the planar interface at $z = 0$.

(12 Marks)

- (b) A uniform plane wave (UPW) in free space occupying the region $z \leq 0$ is incident at a plane interface with a lossless dielectric medium having $\mu_r = 1$ and $\epsilon_r = 2.25$, occupying the region $z \geq 0$. The incident electric field of the UPW is given by

$$\vec{E}_i(x, z) = (20\vec{a}_x - 40\vec{a}_z) e^{-j(8x+4z)} \text{ V/m.}$$

Find the following:

- (i) The angle of incidence θ_i and the angle of transmission θ_t . Give both angles in degrees.
- (ii) The amplitude of the transmitted electric field at $z = 0$, i.e., E_{ot} .
- (iii) The time-average power transmitted through a 2-m^2 area at $z = 0$.

(13 Marks)

4. (a) A generator having an open-circuit voltage $V_g(t) = 96 \cos(2.5\pi \times 10^8 t)$ V and an internal impedance $Z_g = 100 \Omega$ is connected to a $Z_o = 100 \Omega$ lossless air-filled transmission line of length $\ell = 0.64$ m. The phase velocity on the line is $u_p = 3 \times 10^8$ m/s and the line is terminated in a complex load $Z_L = 140 - j64 \Omega$. Assuming that the load end is at $z = 0$ and the source end is at $z = -\ell$, find the following and state any assumption(s) made:

- (i) The electrical length $\frac{\ell}{\lambda}$ of the transmission line.
- (ii) The reflection coefficient $\Gamma(z)$ in polar form, i.e., $|\Gamma| \angle \theta_\Gamma$ at $z = 0$.
- (iii) The input impedance $Z_{in}(z)$ in polar form at $z = -\ell$.
- (iv) The amplitude of the incident voltage wave at $z = 0$, i.e., V_o^+ .
- (v) The time-domain expression for the voltage at $z = 0$, i.e., $V(z = 0, t)$.

(20 Marks)

- (b) Find the position z of maximum voltage for the transmission line in part (a).

(5 Marks)

Note: The Smith chart may be used in the solution of one or both parts of this question. Please put the Smith chart inside the answer script and tie it with a thread.

Appendix A

Physical Constants

Permittivity of free space $\epsilon_0 = \frac{1}{36\pi} \times 10^{-9} \text{ F/m}$

Permeability of free space $\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$

∇ Operator

$$\nabla V = \vec{a}_x \frac{\partial V}{\partial x} + \vec{a}_y \frac{\partial V}{\partial y} + \vec{a}_z \frac{\partial V}{\partial z}$$

$$\nabla \cdot \vec{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$$

$$\nabla \times \vec{A} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix}$$

$$\nabla V = \vec{a}_r \frac{\partial V}{\partial r} + \vec{a}_\phi \frac{\partial V}{r \partial \phi} + \vec{a}_z \frac{\partial V}{\partial z}$$

$$\nabla \cdot \vec{A} = \frac{\partial(rA_r)}{r \partial r} + \frac{\partial A_\phi}{r \partial \phi} + \frac{\partial A_z}{\partial z}$$

$$\nabla \times \vec{A} = \frac{1}{r} \begin{vmatrix} \vec{a}_r & r\vec{a}_\phi & \vec{a}_z \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ A_r & rA_\phi & A_z \end{vmatrix}$$

Appendix A (continued)

Electric and Magnetic Fields

$$\vec{E} = \frac{1}{4\pi\epsilon} \iiint_V \frac{\rho_v \vec{a}_R}{R^2} dv = \frac{1}{4\pi\epsilon} \iiint_V \frac{\rho_v \vec{R}}{R^3} dv$$

$$V = \frac{1}{4\pi\epsilon} \iiint_V \frac{\rho_v}{R} dv$$

$$\vec{H} = \frac{1}{4\pi} \int_C \frac{I \vec{dl} \times \vec{a}_R}{R^2} = \frac{1}{4\pi} \int_C \frac{I \vec{dl} \times \vec{R}}{R^3}$$

$$\oint_C \vec{H} \cdot d\vec{l} = I = \iint_S \vec{J} \cdot d\vec{s}$$

$$emf = \oint_C \vec{E} \cdot d\vec{l} = -\frac{d\Phi_m}{dt} = -\frac{d}{dt} \iint_S \vec{B} \cdot d\vec{s}$$

Maxwell's Equations

$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

$$\nabla \times \vec{H} = \frac{\partial \vec{D}}{\partial t} + \vec{J}$$

$$\nabla \cdot \vec{D} = \rho$$

$$\nabla \cdot \vec{B} = 0$$

Complex Propagation Constant

$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\epsilon_c} = j\omega\sqrt{\mu(\epsilon - j\sigma/\omega)}$$

Complex Intrinsic Impedance

$$\eta_c = \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu}{\epsilon - j\sigma/\omega}}$$

Appendix A (continued)

Reflection and Transmission of Electromagnetic Wave

$$\frac{\sin \theta_t}{\sin \theta_i} = \sqrt{\frac{\mu_1 \varepsilon_1}{\mu_2 \varepsilon_2}} \quad \tan \theta_{B\parallel} = \sqrt{\frac{\varepsilon_2}{\varepsilon_1}} \quad \sin \theta_c = \sqrt{\frac{\varepsilon_2}{\varepsilon_1}}$$

$$\Gamma_{\perp} = \frac{\eta_2 \cos(\theta_i) - \eta_1 \cos(\theta_t)}{\eta_2 \cos(\theta_i) + \eta_1 \cos(\theta_t)} \quad \tau_{\perp} = \frac{2\eta_2 \cos(\theta_i)}{\eta_2 \cos(\theta_i) + \eta_1 \cos(\theta_t)}$$

$$\Gamma_{\parallel} = \frac{\eta_2 \cos(\theta_t) - \eta_1 \cos(\theta_i)}{\eta_2 \cos(\theta_t) + \eta_1 \cos(\theta_i)} \quad \tau_{\parallel} = \frac{2\eta_2 \cos(\theta_i)}{\eta_2 \cos(\theta_t) + \eta_1 \cos(\theta_i)}$$

Transmission Line

$$V(z) = V_o^+ e^{-j\beta z} + V_o^- e^{+j\beta z}$$

$$I(z) = \frac{1}{Z_o} \{ V_o^+ e^{-j\beta z} - V_o^- e^{+j\beta z} \}$$

$$\beta = \frac{2\pi}{\lambda}$$

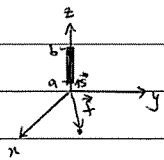
$$\Gamma_L = \frac{V_o^-}{V_o^+} = \frac{Z_L - Z_o}{Z_L + Z_o}$$

$$\Gamma(z) = \Gamma_L e^{+j2\beta z} \quad -\ell \leq z \leq 0$$

$$Z_{in}(-\ell) = \frac{Z_L + jZ_o \tan(\beta\ell)}{Z_o + jZ_L \tan(\beta\ell)} Z_o$$

END OF PAPER

i) a)



i) $\vec{f} = x\vec{a}_x + y\vec{a}_y$, $\vec{z} = z\vec{a}_z$ for $a \leq z \leq b$, $dl = dz$

$\vec{R} = \vec{f} - \vec{z} = x\vec{a}_x + y\vec{a}_y - z\vec{a}_z$, $R = \sqrt{(x^2 + y^2) + z^2}$

$\vec{E}(x, y) = \frac{1}{4\pi\epsilon_0} \int_c \frac{\rho_L \vec{R}}{R^3} dl$

$= \frac{1}{4\pi\epsilon_0} \int_a^b \frac{\rho_L (x\vec{a}_x + y\vec{a}_y - z\vec{a}_z)}{[(x^2 + y^2) + z^2]^{3/2}} dz$

Note: $\int \frac{1}{(x^2 + b)^{3/2}} dx = \frac{x}{b\sqrt{x^2 + b}}$
 $\int \frac{x}{(x^2 + b)^{3/2}} dx = \frac{-1}{\sqrt{x^2 + b}}$

$= \frac{\rho_L}{4\pi\epsilon_0} \left[\frac{z(x\vec{a}_x + y\vec{a}_y)}{[(x^2 + y^2) + z^2]^{3/2}} + \frac{\vec{a}_z}{\sqrt{(x^2 + y^2) + z^2}} \right]_a^b$

$= \frac{\rho_L}{4\pi\epsilon_0} \left[\frac{1}{\sqrt{(x^2 + y^2) + b^2}} \left(\frac{b(x\vec{a}_x + y\vec{a}_y)}{x^2 + y^2} + \vec{a}_z \right) - \frac{1}{\sqrt{(x^2 + y^2) + a^2}} \left(\frac{a(x\vec{a}_x + y\vec{a}_y)}{x^2 + y^2} + \vec{a}_z \right) \right] \text{ V/m}$

ii) Note: For the next part, $\frac{\alpha}{\sqrt{(x^2 + y^2) + \alpha^2}} = \text{sign}(\alpha) \frac{1}{\sqrt{(x/a)^2 + (y/b)^2 + 1}}$ where $\text{sign}(\alpha) = \begin{cases} 1 & \text{for } \alpha > 0 \\ -1 & \text{for } \alpha < 0 \end{cases}$

Hence, for $a = -\infty$, $b = \infty$,

$\vec{E}(x, y) = \frac{\rho_L}{4\pi\epsilon_0} \left\{ \lim_{a \rightarrow -\infty, b \rightarrow \infty} \left[\left(\frac{\text{sign}(b)}{\sqrt{(x/b)^2 + (y/b)^2 + 1}} - \frac{\text{sign}(a)}{\sqrt{(x/a)^2 + (y/a)^2 + 1}} \right) \left(\frac{x\vec{a}_x + y\vec{a}_y}{x^2 + y^2} \right) + \left(\frac{1}{\sqrt{x^2 + y^2 + b^2}} - \frac{1}{\sqrt{x^2 + y^2 + a^2}} \right) \vec{a}_z \right] \right\}$

$= \frac{\rho_L}{4\pi\epsilon_0} \left[\left(\frac{1}{\sqrt{0^2 + 0^2 + 1}} - \frac{(-1)}{\sqrt{0^2 + 0^2 + 1}} \right) \left(\frac{x\vec{a}_x + y\vec{a}_y}{x^2 + y^2} \right) + (0 - 0) \vec{a}_z \right]$

$= \frac{\rho_L}{2\pi\epsilon_0} \left(\frac{x\vec{a}_x + y\vec{a}_y}{x^2 + y^2} \right) \text{ V/m}$

For cylindrical coordinate system,

$x = r \cos \phi$, $y = r \sin \phi$, $x^2 + y^2 = r^2$, $\vec{a}_x = \cos \phi \vec{a}_r - \sin \phi \vec{a}_\phi$, $\vec{a}_y = \sin \phi \vec{a}_r + \cos \phi \vec{a}_\phi$

$\vec{E}(r, \phi) = \frac{\rho_L}{2\pi\epsilon_0 r^2} \left[(r \cos \phi)(\cos \phi \vec{a}_r - \sin \phi \vec{a}_\phi) + (r \sin \phi)(\sin \phi \vec{a}_r + \cos \phi \vec{a}_\phi) \right]$

$= \frac{\rho_L}{2\pi\epsilon_0 r^2} \left[(r \cos^2 \phi + r \sin^2 \phi) \vec{a}_r + (-r \sin \phi \cos \phi + r \sin \phi \cos \phi) \vec{a}_\phi \right]$

$= \frac{\rho_L}{2\pi\epsilon_0 r^2} (r \vec{a}_r) = \frac{\rho_L}{2\pi\epsilon_0 r} \vec{a}_r \text{ V/m}$

$\therefore \vec{E}(x, y) = \frac{\rho_L}{2\pi\epsilon_0} \left(\frac{x\vec{a}_x + y\vec{a}_y}{x^2 + y^2} \right) \text{ V/m}$ and $\vec{E}(r, \phi) = \frac{\rho_L}{2\pi\epsilon_0 r} \vec{a}_r \text{ V/m}$

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1) b) Method 1: As this question deals with current of infinite extent, Ampere's law can be used.

Using RHR, $\vec{H} = H_\phi \vec{a}_\phi$ only has an \vec{a}_ϕ component.

$\oint_C \vec{H} \cdot d\vec{l} = I$

$\int_0^{2\pi} H_\phi \vec{a}_\phi \cdot r d\phi \vec{a}_\phi = I$

$\int_0^{2\pi} H_\phi r d\phi = I \iff$ As the Ampere loop is symmetrical/circular, H_ϕ is same at every part of the loop.

$H_\phi \int_0^{2\pi} r d\phi = I$ Hence it is a constant and can be taken out of the integral.

$H_\phi (2\pi r) = I$

$H_\phi = \frac{I}{2\pi r} \text{ A/m} \Rightarrow \vec{H}(r, \phi) = H_\phi \vec{a}_\phi = \frac{I}{2\pi r} \vec{a}_\phi \text{ A/m}$

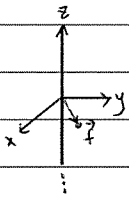
For cartesian coordinate system,

$$\begin{aligned} x &= r \cos \phi, \quad y = r \sin \phi, \quad x^2 + y^2 = r^2, \quad \vec{a}_\phi = -\sin \phi \vec{a}_x + \cos \phi \vec{a}_y \\ \vec{H}(x, y) &= \frac{I}{2\pi r^2} r \vec{a}_\phi \\ &= \frac{I}{2\pi (x^2 + y^2)} r (-\sin \phi \vec{a}_x + \cos \phi \vec{a}_y) \\ &= \frac{I}{2\pi} \left(\frac{-y \vec{a}_x + x \vec{a}_y}{x^2 + y^2} \right) \text{ A/m} \end{aligned}$$

Method 2: Integration in cartesian coordinate system

$$\begin{aligned} \vec{r} &= x \vec{a}_x + y \vec{a}_y, \quad \vec{z} = z \vec{a}_z \quad \text{for } -\infty < z < \infty, \quad d\vec{z} = \vec{a}_z dz \\ \vec{R} &= \vec{r} - \vec{z} = x \vec{a}_x + y \vec{a}_y - z \vec{a}_z, \quad R = \sqrt{x^2 + y^2 + z^2} \end{aligned}$$

$$\vec{H}(x, y) = \frac{1}{4\pi} \int_C \frac{I d\vec{l} \times \vec{R}}{R^3}$$



$$= \frac{1}{4\pi} \int_{-\infty}^{\infty} \frac{I (\vec{a}_z dz) \times (x \vec{a}_x + y \vec{a}_y - z \vec{a}_z)}{[(x^2 + y^2) + z^2]^{3/2}}$$

Note: $\int \frac{1}{(x^2 + b)^{3/2}} dx = \frac{x}{b \sqrt{x^2 + b}}$

$$= \frac{I}{4\pi} \int_{-\infty}^{\infty} \frac{x \vec{a}_y - y \vec{a}_x}{[(x^2 + y^2) + z^2]^{3/2}} dz$$

$$= \frac{I}{4\pi} \left[\frac{z(x \vec{a}_y - y \vec{a}_x)}{(x^2 + y^2) \sqrt{(x^2 + y^2) + z^2}} \right]_{-\infty}^{\infty}$$

$$= \frac{I}{4\pi} \left[\frac{\text{sign}(z)(x \vec{a}_y - y \vec{a}_x)}{(x^2 + y^2) \sqrt{(x/z)^2 + (y/z)^2 + 1}} \right]_{-\infty}^{\infty}$$

$$= \frac{I}{4\pi} \left(\frac{(1)(x \vec{a}_y - y \vec{a}_x)}{(x^2 + y^2) \sqrt{0^2 + 0^2 + 1}} - \frac{(-1)(x \vec{a}_y - y \vec{a}_x)}{(x^2 + y^2) \sqrt{0^2 + 0^2 + 1}} \right)$$

$$= \frac{I}{2\pi} \left(\frac{x \vec{a}_y - y \vec{a}_x}{x^2 + y^2} \right) \text{ A/m}$$

For cylindrical coordinate system,

$$x = r \cos \phi, \quad y = r \sin \phi, \quad x^2 + y^2 = r^2, \quad \vec{a}_x = \cos \phi \vec{a}_r - \sin \phi \vec{a}_\phi, \quad \vec{a}_y = \sin \phi \vec{a}_r + \cos \phi \vec{a}_\phi$$

$$\begin{aligned} \vec{H}(r, \phi) &= \frac{I}{2\pi r} [(r \cos \phi)(\sin \phi \vec{a}_r + \cos \phi \vec{a}_\phi) - (r \sin \phi)(\cos \phi \vec{a}_r - \sin \phi \vec{a}_\phi)] \\ &= \frac{I}{2\pi r} [(r \sin \phi \cos \phi - r \sin \phi \cos \phi) \vec{a}_r + (r \cos^2 \phi + r \sin^2 \phi) \vec{a}_\phi] \\ &= \frac{I}{2\pi r} (r \vec{a}_\phi) = \frac{I}{2\pi r} \vec{a}_\phi \text{ A/m} \end{aligned}$$

$$\vec{H}(x, y) = \frac{I}{2\pi} \left(\frac{x \vec{a}_y - y \vec{a}_x}{x^2 + y^2} \right) \text{ A/m} \quad \text{and} \quad \vec{H}(r, \phi) = \frac{I}{2\pi r} \vec{a}_\phi \text{ A/m} \quad \#$$

2) a) i) $\iint_S dx dy = a^2 = \text{total area of open surface}$

$$\Phi_m = \iint_S \vec{B} \cdot d\vec{z}$$



Assume surface normal is \vec{a}_z and the corresponding closed contour in the CCW direction.

$$= \iint_S (C_1 t^2 + C_2 t) \vec{a}_z \cdot \vec{a}_z dx dy$$

$$= \iint_S (C_1 t^2 + C_2 t) dx dy \quad \text{Note: } C_1 t^2 + C_2 t \text{ is not dependent on } x \text{ and } y \Rightarrow \text{can be taken out of integral}$$

$$= (C_1 t^2 + C_2 t) \iint_S dx dy$$

$$= (C_1 t^2 + C_2 t) a^2 \quad \text{Wb} \quad \#$$

$$V_{\text{emf}} = -\frac{d}{dt}(\Phi_m) = -a^2(2C_1 t + C_2) \quad \text{V} \quad \# \quad (\text{Assuming CCW loop})$$

ii) Length of square = $\sqrt{a^2} = a \text{ m} \Rightarrow \text{Perimeter/Total length} = 4a \text{ m} \Rightarrow \text{Resistance, } R = 4aR_L \Omega$

$$I_{\text{ind}} = \frac{V_{\text{emf}}}{R} = -\frac{q}{4R_L} (2C_1 t + C_2) A \quad (\text{in CW direction}) \#$$

$$\text{OR } \frac{q}{4R_L} (2C_1 t + C_2) A \quad (\text{in CW direction}) \#$$

2) b) i) loss tangent = $\frac{\sigma}{\epsilon \omega} = 8 < 20 \Rightarrow \text{so it is not a good conductor} \#$

$$\sigma = 8 \epsilon \omega = 8 (\epsilon_r \epsilon_0) (2\pi f)$$

$$\sigma = 8 (10) \left(\frac{1}{36\pi} \times 10^{-9} \right) (2\pi) (500 \times 10^6)$$

$$\sigma = \frac{22}{9} \approx 2.222 \text{ S/m} \#$$

ii) * Note: To calculate the square root of a complex number in a scientific calculator, we can use this:

$$\text{Let the complex number} = A, \sqrt{A} = \sqrt{\text{Abs}(A)} \angle \left(\frac{\text{Arg}(A)}{2} \right)$$

$$\epsilon_c = \epsilon - j \frac{\sigma}{\omega} = \epsilon_r \epsilon_0 - j \frac{\sigma}{2\pi f} = (8.85419 - j 77.8091) \times 10^{-11} \text{ F/m}$$

$$\eta_c = \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu_0}{\epsilon_c}} = 41.986 \angle 0.7232 \Omega \#$$

$$\gamma = j\omega \sqrt{\mu \epsilon_c} = j(2\pi f) \sqrt{\mu_0 \epsilon_c} = 62.228 + j 70.4907$$

$$\therefore \alpha = 62.228 \text{ Np/m} \#, \beta = 70.4907 \text{ rad/m} \#$$

3) a) Normally incident wave

(i)

$$k = \frac{\omega}{v_p}$$

$$\therefore v_p = \frac{\omega}{k} = \frac{6 \times 10^9}{24.5} = 2.45 \times 10^8 \text{ m/s} \#$$

$$(ii) v_p = \frac{c}{\sqrt{\mu_r \epsilon_r}}$$

$$\mu_r \epsilon_r = \frac{c^2}{v_p^2}, \mu_r = 1 \text{ [Non-magnetic medium]}$$

$$\epsilon_r = \frac{c^2}{\mu_r v_p^2} = \frac{(3 \times 10^8)^2}{(2.45 \times 10^8)^2} = 1.5$$

$$\therefore \text{Permittivity, } \epsilon = \epsilon_0 \epsilon_r = 1.5 \times \frac{1}{36\pi} \times 10^{-9} = 1.327 \times 10^{-11} \text{ F/m} \#$$

$$(iii) \eta_1 = 120\pi \sqrt{\frac{\mu_r}{\epsilon_r}} = 120\pi \sqrt{\frac{1}{1.5}} = 307.8 \Omega$$

$$\vec{a}_E = \vec{a}_H \times \vec{a}_k = -\vec{a}_y \times \vec{a}_z = -\vec{a}_x$$

$$\vec{H}_i(z) = -\vec{a}_y 50 \angle 40^\circ e^{-24.5z} \times 10^{-3} \text{ A/m}$$

$$\vec{E}_i(z) = \vec{a}_E |E_{oi}| e^{-24.5z} \times 10^{-3}, |E_{oi}| = \eta_1 |H_{oi}| = 15.4$$

$$\vec{E}_i(z) = (-\vec{a}_x) 15.4 \angle 40^\circ e^{-24.5z}$$

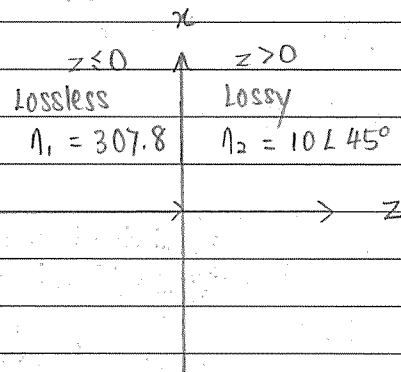
$$\therefore \vec{E}_i(z, t) = (-\vec{a}_x) 15.4 \cos(6 \times 10^9 t - 24.5z + 40^\circ) \text{ V/m} \#$$

$$(iv) \Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{10 \angle 45^\circ - 308}{10 \angle 45^\circ + 308} = 0.9551 \angle 177.4^\circ$$

$$\therefore \text{Percentage of average incident power reflected, } \frac{P_r}{P_i} = |\Gamma|^2$$

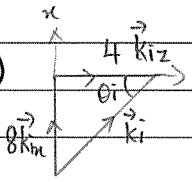
$$= 0.9551^2$$

$$= 91.22 \% \#$$



Q3.b) oblique incident wave

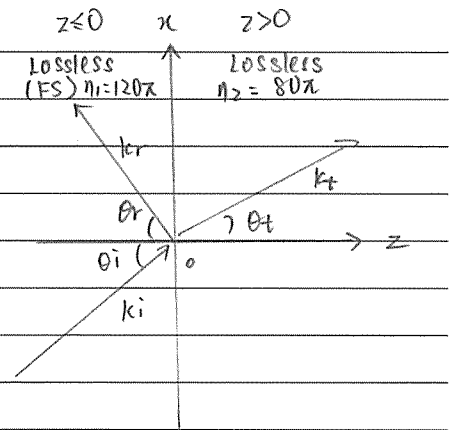
$\therefore \vec{a}_E = \alpha \vec{a}_x + \beta \vec{a}_y$ $\alpha, \beta \in \mathbb{R}$
the UPW is parallelly polarized.

(i)  $\tan \theta_i = \frac{8}{4}$
 $\therefore \theta_i = 63.43^\circ$ #

Using Snell's Law :

$$\sin \theta_t = \sin \theta_i \sqrt{\frac{\mu_r \epsilon_1}{\mu_2 \epsilon_2}} = \sin 63.43^\circ \times \sqrt{\frac{1 \times 1}{1 \times 2.25}}$$

$$\therefore \theta_t = 36.60^\circ$$
 #



(ii) $\theta_i = 63.43^\circ$, $\theta_t = 36.60^\circ$, $\eta_1 = 120\pi$, $\eta_2 = 120\pi \sqrt{\frac{1}{2.25}} = 80\pi$
(Free space)

$$\tau_{11} = \frac{2\eta_2 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i} = 0.607$$

For E_{oi} : $|E_{oi}| = \sqrt{20^2 + (-40)^2} = 44.72$

$$\therefore |E_{ot}| = |E_{oi}| \tau_{11} = 27.15 \text{ V}$$
 #

(iii) For \vec{a}_{kt} : $\vec{a}_{kt} = |\vec{a}_{kt}| \cos \theta_t \vec{a}_z + |\vec{a}_{kt}| \sin \theta_t \vec{a}_x$
 $\vec{a}_{kt} = \frac{|E_{ot}|}{\eta_2} \cos \theta_t \vec{a}_z + \frac{|E_{ot}|}{\eta_2} \sin \theta_t \vec{a}_x$

$$\vec{S}_t = \vec{a}_{kt} \frac{|E_{ot}|^2}{2\eta_2} = \left[(\cos \theta_t) \vec{a}_z + \sin \theta_t \vec{a}_x \right] \times \frac{27.15^2}{2(80\pi)}$$

$$= (0.803 \vec{a}_z + 0.592 \vec{a}_x) \times 1.466 \text{ W/m}^2$$

At $z=0$, time-average power is transmitted only normal to the boundary, i.e. \vec{a}_z :

$$\therefore \text{Power} = 0.803 \times 1.466 \times 2$$

$$= 2.3539 \text{ W}$$
 #

$$Q4.a) (i) \beta = \frac{W}{up} = \frac{2.5\pi \times 10^8}{3 \times 10^8} = \frac{5}{6}\pi$$

$$\beta = \frac{2\pi}{\lambda} \Rightarrow \lambda = \frac{2\pi}{\frac{5}{6}\pi} = 2.4 \text{ m}$$

$$\therefore \frac{l}{\lambda} = \frac{0.64}{2.4}$$

$$= 0.2667 \quad \#$$

(ii) At $z=0$,

$$\therefore \Gamma(z) = \Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{140 - 64j - 100}{140 - 64j + 100}$$

$$= 0.304 \angle -0.752$$

#

(iii) At $z=-l$, $\tan(\beta l) = \tan\left(\frac{5\pi}{6}\right) = 1.676$

$$Z_{in}(-l) = \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)}$$

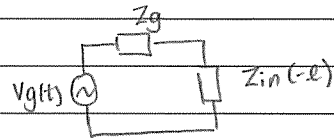
$$= \frac{140 - 64j + j(100)(1.676)}{100 + j(140 - 64j)(1.676)} (100) = 71.89 \angle 0.502 \Omega \quad \#$$

(iv) For V_o^+ , first we find $v(-l)$:

$$v(-l) = V_g(t) \times \frac{Z_{in}(-l)}{Z_{in}(-l) + Z_g}$$

$$= 96 \times \frac{71.89 \angle 0.502}{71.89 \angle 0.502 + 100}$$

$$= 41.41 \angle 0.293 \text{ V}$$



$$v(-l) = 40.15 \angle 0.292 = V_o^+ e^{-j\beta(-l)} + V_o^- e^{j\beta(-l)}$$

$$41.41 \angle 0.292 = V_o^+ \left[e^{j \times \frac{5\pi}{6} \times 0.64} + \Gamma_L e^{j \times \frac{5\pi}{6} \times -0.64} \right], \quad \Gamma_L = 0.304 \angle -0.752$$

$$= V_o^+ (0.863 \angle 1.969)$$

$$\therefore V_o^+ = 47.98 \angle -1.68$$

$$\therefore |V_o^+| = 48 \text{ V} \quad \#$$

$$(v) \quad v(z=0) = V_o^+ e^{-j\beta(0)} + V_o^- e^{j\beta(0)}$$

$$= V_o^+ + V_o^- = 48 \angle -1.675 + 48 \angle -1.675 \times 0.304 \angle -0.752$$

$$\therefore v(z=0, t) = 48 \cos(2.5\pi \times 10^8 t - 1.675) + 14.58 \cos(2.5\pi \times 10^8 t - 2.43) \text{ V} \quad \#$$

b) For max V : $\theta_r = \theta_0 + 2\beta z = 0, 2\pi, \dots$

$$Z_{max} = \frac{0 + 0.752}{2\beta} = 0.144 \text{ m} > 0, \quad Z_{max} = \frac{-2\pi + 0.752}{2\beta} = -1.056 \text{ m} < -0.64 \text{ m}$$

There are no global maximum. However there are local maximum voltage:

$$|v(-l)| = 41.41 \text{ V}, \quad |v(0)| = 59.46$$

$$\therefore v(0) > v(-l), \quad \therefore \text{The maximum voltage is at } z=0 \quad \#$$