# **Avalon Interface Specifications**





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# Introduction to the Avalon Interface Specifications

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Avalon® interfaces simplify system design by allowing you to easily connect components in an Altera® FPGA. The Avalon interface family defines interfaces appropriate for streaming high-speed data, reading and writing registers and memory, and controlling off-chip devices. These standard interfaces are designed into the components available in Qsys. You can also use these standardized interfaces in your custom components. By using these standard interfaces, you enhance the interoperability of your designs.

This specification defines all of the Avalon interfaces. After reading it, you should understand which interfaces are appropriate for your components and which signal roles to use for particular behaviors. This specification defines the following seven interfaces:

- Avalon Streaming Interface (Avalon-ST)—an interface that supports the unidirectional flow of data, including multiplexed streams, packets, and DSP data.
- Avalon Memory Mapped Interface (Avalon-MM)—an address-based read/write interface typical of master-slave connections.
- Avalon Conduit Interface— an interface type that accommodates individual signals or groups of
  signals that do not fit into any of the other Avalon types. You can connect conduit interfaces inside a
  Qsys system. Or, you can export them to make connections to other modules in the design or to FPGA
  pins.
- Avalon Tri-State Conduit Interface (Avalon-TC) —an interface to support connections to off-chip peripherals. Multiple peripherals can share pins through signal multiplexing, reducing the pin count of the FPGA and the number of traces on the PCB.
- Avalon Interrupt Interface—an interface that allows components to signal events to other components.
- Avalon Clock Interface—an interface that drives or receives clocks.
- Avalon Reset Interface—an interface that provides reset connectivity.

A single component can include any number of these interfaces and can also include multiple instances of the same interface type. For example, in the first figure below, the Ethernet Controller includes the following six different interface types:

- Avalon-MM
- Avalon-ST
- Avalon Conduit
- Avalon-TC
- Avalon Interrupt
- Avalon Clock.

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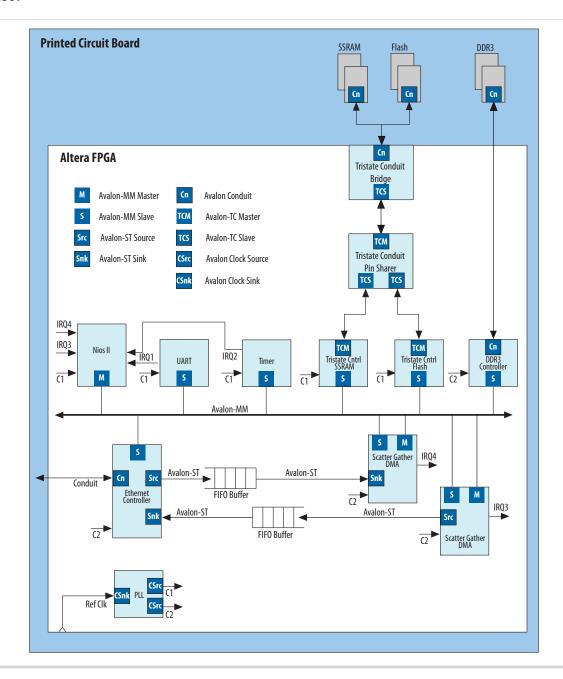
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**Note:** Avalon interfaces are an open standard. No license or royalty is required to develop and sell products that use, or are based on Avalon interfaces.

The following figures illustrate the use of the Avalon interfaces in system designs.

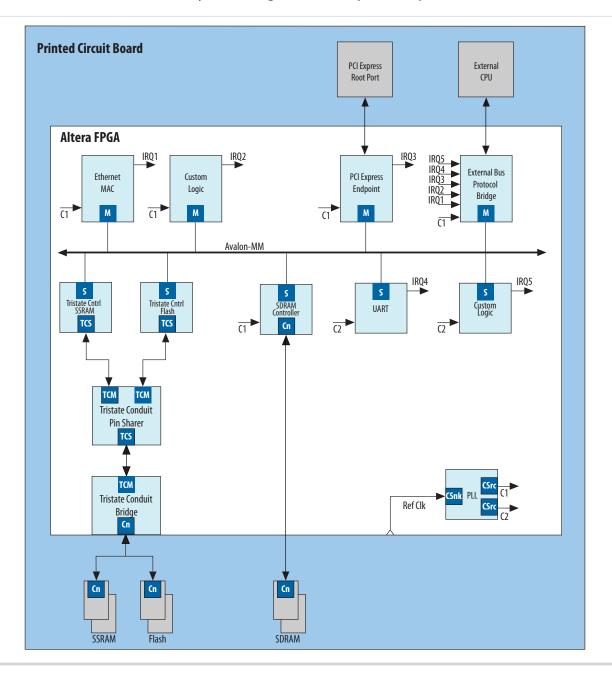
Figure 1-1: Avalon Interfaces in a System Design with Scatter Gather DMA Controller and Nios II Processor



In this figure, the Nios® II processor accesses the control and status registers of on-chip components using an Avalon-MM interface. The scatter gather DMAs send and receive data using Avalon-ST interfaces. Four components include interrupt interfaces serviced by software running on the Nios II processor. A PLL accepts a clock via an Avalon Clock Sink interface and provides two clock sources. Two components

include Avalon-TC interfaces to access off-chip memories. Finally, the DDR3 controller accesses external DDR3 memory using an Avalon Conduit interface.

Figure 1-2: Avalon Interfaces in a System Design with PCI Express Endpoint and External Processor



In the previous figure, an external processor accesses the control and status registers of on-chip components via an external bus bridge with an Avalon-MM interface. The PCI Express Root Port controls devices on the printed circuit board and the other components of the FPGA by driving an on-chip PCI Express Endpoint with an Avalon-MM master interface. An external processor handles interrupts from five components. A PLL accepts a reference clock via a Avalon Clock sink interface and provides two clock sources. The flash and SRAM memories use an Avalon-TC interface to share FPGA pins. Finally, an SDRAM controller accesses an external SDRAM memory using an Avalon Conduit interface.

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### 1.1 Avalon Properties and Parameters

Avalon interfaces use properties to describe their behavior. For example, the maxChannel property of Avalon-ST interfaces allows you to specify the number of channels supported by the interface. The clockRate property of the Avalon Clock interface provides the frequency of a clock signal. The specification for each interface type defines all of its properties and specifies the default values.

### 1.2 Signal Roles

Each of the Avalon interfaces defines a number of signal roles and their behavior. Many signal roles are optional. You have the flexibility to select only the signal roles necessary to implement the required functionality. For example, the Avalon-MM interface includes optional beginbursttransfer and burstcount signal roles for use in components that support bursting. The Avalon-ST interface includes the optional startofpacket and endofpacket signal roles for interfaces that support packets.

With the exception of Avalon Conduit interfaces, each interface may include only one signal of each signal role. Active-low signals are permitted for many signal roles. Active-high signals are generally used in this document.

### 1.3 Interface Timing

Subsequent chapters of this document include timing information that describes transfers for individual interface types. There is no guaranteed performance for any of these interfaces. Actual performance depends on many factors, including component design and system implementation.

Most Avalon interfaces must not be edge sensitive to signals other than the clock and reset. Other signals may transition multiple times before they stabilize. The exact timing of signals between clock edges varies depending upon the characteristics of the selected Altera device. This specification does not specify electrical characteristics. Refer to the appropriate device documentation for electrical specifications.

#### 1.4 Related Documents

For more information on related topics in the following documents and design examples, refer to the following documents:

#### **Related Information**

- Creating a System with Qsys.
   For an overview of the Qsys system integration tool
- Creating Qsys Components
   For information about creating Qsys components, composed components, and dynamic file generation
- Optimizing Qsys System Performance.
   For information about system design, including: hierarchy, concurrency, pipelining, throughput, reducing logic utilization and power consumption



- Component Interface Tcl Reference
  - For information about defining Qsys components and a reference for component Tool Command Language (Tcl)
- Qsys System Design Components
   For information about system design components available in the IP Catalog
- **Qsys Tutorial Design Example**For tutorial that shows how to build a memory test system using components with Avalon interfaces



### **Avalon Clock and Reset Interfaces**

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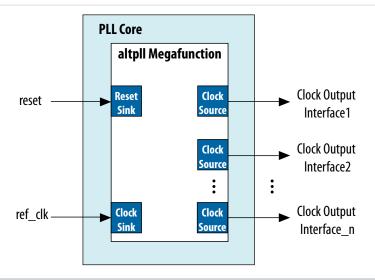




Avalon Clock interfaces define the clock or clocks used by a component. Components can have clock inputs, clock outputs, or both. A phase locked loop (PLL) is an example of a component that has both a clock input and clock outputs.

The following figure is a simplified illustration showing the most important inputs and outputs of a PLL component.

Figure 2-1: PLL Core Clock Outputs and Inputs



## 2.1 Avalon Clock Sink Signal Roles

A clock sink provides a timing reference for other interfaces and internal logic.

**Table 2-1: Clock Sink Signal Roles** 

Signal Role	Width	Direction	Required	Description
clk	1	Input	Yes	A clock signal. Provides synchronization for internal logic and for other interfaces.

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### 2.2 Clock Sink Properties

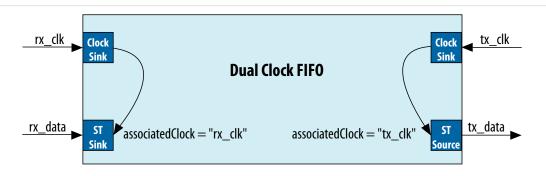
**Table 2-2: Clock Sink Properties** 

Name	Default Value	Legal Values	Description
clockRate	0	0-2 <sup>32</sup> -1	Indicates the frequency in Hz of the clock sink interface. If 0, the clock rate allows any frequency. If non-zero, Qsys issues a warning if the connected clock source is not the specified frequency.

### 2.3 Associated Clock Interfaces

All synchronous interfaces have an associatedClock property that specifies which clock source on the component is used as a synchronization reference for the interface. This property is illustrated in the following figure.

Figure 2-2: associatedClock Property



### 2.4 Avalon Clock Source Signal Roles

An Avalon Clock source interface drives a clock signal out of a component.

**Table 2-3: Clock Source Signal Roles** 

Signal Role	Width	Direction	Required	Description
clk	1	Output	Yes	An output clock signal.

Send Feedback

**Avalon Clock and Reset Interfaces** 

# **2.5 Clock Source Properties**

**Table 2-4: Clock Source Properties** 

Name	Default Value	Legal Values	Description
associatedDirectClock	N/A	an input clock name	The name of the clock input that directly drives this clock output, if any.
clockRate	0	0-2 <sup>32</sup> -1	Indicates the frequency in Hz at which the clock output is driven.
clockRateKnown	false	true, false	Indicates whether or not the clock frequency is known. If the clock frequency is known, this information can be used to customize other components in the system.

### 2.6 Reset Sink

#### **Table 2-5: Reset Input Signal Roles**

The reset\_req signal is an optional signal that you can use to prevent memory content corruption by performing reset handshake prior to an asynchronous reset assertion.

Signal Role	Width	Direction	Required	Description
reset, reset_n	1	Input	Yes	Resets the internal logic of an interface or component to a user-defined state. The synchronous properties of the reset are defined by the synchronousEdges parameter.
reset_req	1	input	No	Early indication of reset signal. This signal acts as a least a one-cycle warning of pending reset for ROM primitives. It is used to gate address wires.

# 2.7 Reset Sink Interface Properties

**Table 2-6: Reset Input Signal Roles** 

Name	Default Value	Legal Values	Description
associatedClock	N/A	a clock name	The name of a clock to which this interface is synchronized. Required if the value of synchronousEdges is DEASSERT or BOTH.

**Avalon Clock and Reset Interfaces** 

Name	Default Value	Legal Values	Description
synchronous-Edges	DEASSERT	NONE DEASSERT BOTH	<ul> <li>Indicates the type of synchronization the reset input requires. The following values are defined:</li> <li>NONE—no synchronization is required because the component includes logic for internal synchronization of the reset signal.</li> <li>DEASSERT—the reset assertion is asynchronous and deassertion is synchronous.</li> <li>BOTH—reset assertion and deassertion are synchronous.</li> </ul>

### 2.8 Associated Reset Interfaces

All synchronous interfaces have an associatedReset property that specifies which reset signal resets the interface logic.

### 2.9 Reset Source

#### **Table 2-7: Reset Output Signal Roles**

The reset\_req signal is an optional signal that you can use to prevent memory content corruption by performing reset handshake prior to an asynchronous reset assertion.

Signal Role	Width	Direction	Required	Description
reset reset_n	1	Output	Yes	Resets the internal logic of an interface or component to a user-defined state.
reset_req	1	Output	Optional	Enables reset request generation, which is an early signal that is asserted before reset assertion. Once asserted, this cannot be deasserted until the reset is completed.

# 2.10 Reset Source Interface Properties

**Table 2-8: Reset Interface Properties** 

Name	Default Value	Legal Values	Description
associatedClock	N/A	a clock name	The name of a clock to which this interface synchronized. Required if the value of synchronousEdges is DEASSERT or BOTH.

Avalon Clock and Reset Interfaces



Name	Default Value	Legal Values	Description
associatedDirectReset	N/A	a reset name	The name of the reset input that directly drives this reset source through a one-to-one link.
associatedResetSinks	N/A	a reset name	Specifies reset inputs which will eventually cause a reset source to assert reset. For example, a reset synchronizer ORS a number of reset inputs to generate a reset output.
synchronousEdges	DEASSERT	NONE DEASSERT BOTH	<ul> <li>Indicates the reset output's synchronization.</li> <li>The following values are defined:</li> <li>NONE-The reset interface is asynchronous.</li> <li>DEASSERT-the reset assertion is asynchronous and deassertion is synchronous.</li> <li>BOTH-reset assertion and deassertion are synchronous.</li> </ul>



## **Avalon Memory-Mapped Interfaces**

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### 3.1 Introduction to Avalon Memory-Mapped Interfaces

You can use Avalon Memory-Mapped (Avalon-MM) interfaces to implement read and write interfaces for master and slave components. The following are examples of component that typically include memory-mapped interfaces:

- Microprocessors
- Memories
- UARTs
- DMAs
- Timers

Avalon-MM interfaces range from simple to complex. For example, SRAM interfaces that have fixed-cycle read and write transfers have simple Avalon-MM interfaces. Pipelined interfaces capable of burst transfers are complex.

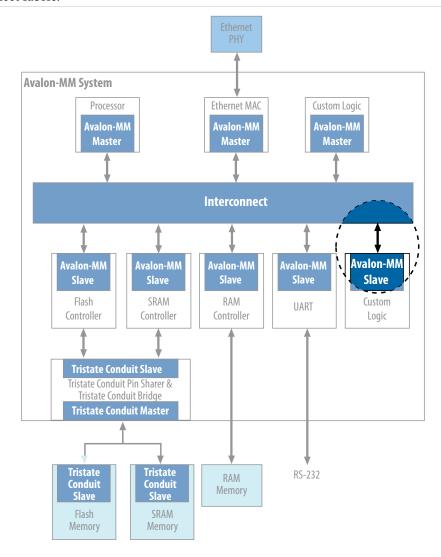
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#### Figure 3-1: Focus on Avalon-MM Slave Transfers

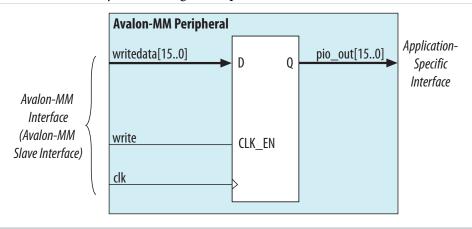
The following figure shows a typical system, highlighting the Avalon-MM slave interface connection to the interconnect fabric.



Avalon-MM components typically include only the signals required for the component logic.

#### Figure 3-2: Example Slave Component

The 16-bit general-purpose I/O peripheral shown in the following figure only responds to write requests. This component includes only the slave signals required for write transfers.



Each signal in an Avalon-MM slave corresponds to exactly one Avalon-MM signal role. An Avalon-MM interface can use only one instance of each signal role.

### 3.2 Avalon Memory-Mapped Interface Signal Roles

The following table lists signal roles for the Avalon-MM interface. Signal roles allow you to create masters that use bursts for read and write processing. When necessary, Qsys interconnect enables the connection between the master and slave pair. When the master and slave interfaces match, a direct connection is possible.

This specification does not require all signals to exist in an Avalon-MM interface. There is no one signal that is always required. The minimum requirements for an Avalon-MM interface are readdata for a read-only interface, or writedata and write for a write-only interface.

**Table 3-1: Avalon-MM Signal Roles** 

Avalon-MM signals can be active high or active low. When active low, the signal name ends with \_n.

Signal Role	Width	Direction	Description						
	Fundamental Signals								
address	1 - 64	Master → Slave	Masters: By default, the address signal represents a byte address. The value of the address must be aligned to the data width. To write to specific bytes within a data word, the master must use the byteenable signal. Refer to the addressUnits interface property for word addressing.  Slaves: By default, the interconnect translates the byte address into a word address in the slave's address space. Each slave access is for a word of data from the perspective of the slave. For example, address = 0 selects the first						



Signal Role	Width	Direction	Description
			word of the slave. address = 1 selects the second word of the slave. Refer to the addressUnits interface property for byte addressing.
byteenable_n	2, 4, 8, 16, 32, 64, 128	Master → Slave	Enables specific byte lane(s) during transfers on interfaces of width greater than 8 bits. Each bit in byteenable corresponds to a byte in writedata and readdata. The master bit <n> of byteenable indicates whether byte <n> is being written to. During writes, byteenables specify which bytes are being written to. Other bytes should be ignored by the slave. During reads, byteenables indicate which bytes the master is reading. Slaves that simply return readdata with no side effects are free to ignore byteenables during reads. If an interface does not have a byteenable signal, the transfer proceeds as if all byteenables are asserted.  When more than one bit of the byteenable signal is asserted, all asserted lanes are adjacent. The number of adjacent lines must be a power of 2. The specified bytes must be aligned on an address boundary for the size of the data. For example, the following values are legal for a 32-bit slave:  1111 writes full 32 bits 0011 writes lower 2 bytes 1100 writes byte 0 only 0010 writes byte 1 only 1000 writes byte 2 only 1000 writes byte 3 only To avoid unintended side effects, Altera strongly recommends that you use the byteenable signal in systems with different word sizes.  Note: The AXI interface supports unaligned accesses going from an AXI master to an Avalon-MM slave may result in an illegal transaction. To avoid this issue, only use aligned accesses to Avalon-MM slaves.</n></n>
debugaccess	1	Master → Slave	When asserted, allows the Nios II processor to write on-chip memories configured as ROMs.



Signal Role	Width	Direction	Description
read read_n	1	Master → Slave Asserted to indicate a read transfer. If present readdata is required.	
readdata	8,16, 32, 64,128, 256, 512, 1024	Slave → Master	The readdata driven from the slave to the master in response to a read transfer.
response [1:0]	2	Slave → Master	The response signal is an optional signal that carries the response status.  Note: Because the signal is shared, an interface cannot issue or accept a write response and a read response in the same clock cycle.  • 00: OKAY—Successful response for a transaction.  • 01: RESERVED—Encoding is reserved.  • 10: SLAVEERROR—Error from an endpoint slave. Indicates an unsuccessful transaction.  • 11: DECODEERROR—Indicates attempted access to an undefined location.  For read responses:  • One response is sent with each readdata. A read burst length of N results in N responses. It is not valid to produce fewer responses, even in the event of an error. It is valid for the response signal value to be different for each readdata in the burst.  • The interface must have read control signals. Pipeline support is possible with the readdatavalid signal.  • On read errors, the corresponding readdata is "don't care".  For write responses:  • The interface must have write control signals. Qsys completes all write commands with write response if the write signal is present. The interface must have a writeresponsevalid signal.  • One write response must be sent for each write command. A write burst results in only one response, which must be sent after the final write transfer in the burst is accepted.
write write_n	1	Master → Slave	Asserted to indicate a write transfer. If present, writedata is required.



Signal Role	Width	Direction	Description
writedata	8,16, 32, 64, 128, 256, 512, 1024	Master → Slave	Data for write transfers. The width must be the same as the width of readdata if both are present.
		Wait-State Si	gnals
lock	1	Master → Slave	lock ensures that once a master wins arbitration, it maintains access to the slave for multiple transactions. It is asserted coincident with the first read or write of a locked sequence of transactions. It is deasserted on the final transaction of a locked sequence of transactions. lock assertion does not guarantee that arbitration will be won. After the lock-asserting master has been granted, it retains grant until it is deasserted.  A master equipped with lock cannot be a burst master. Arbitration priority values for lock-equipped masters are ignored.  lock is particularly useful for read-modify-write (RMW) operations. The typical read-modify-write operation includes the following steps:  1. Master A asserts lock and reads 32-bit data that has multiple bit fields.  2. Master A deasserts lock, changes one bit field, and writes the 32-bit data back.  lock prevents master B from performing a write between Master A's read and write.

Signal Role	Width	Direction	Description
waitrequest waitrequest_n	1	Slave → Master	Asserted by the slave when it is unable to respond to a read or write request. Forces the master to wait until the interconnect is ready to proceed with the transfer. At the start of all transfers, a master initiates the transfer and waits until waitrequest is deasserted. A master must make no assumption about the assertion state of waitrequest when the master is idle: waitrequest may be high or low, depending on system properties.
			When waitrequest is asserted, master control signals to the slave are to remain constant with the exception of beginbursttransfer. For a timing diagram illustrating the beginbursttransfer signal, refer to the figure in <i>Read Bursts</i> .
			An Avalon-MM slave may assert waitrequest during idle cycles. An Avalon-MM master may initiate a transaction when waitrequest is asserted and wait for that signal to be deasserted. To avoid system lockup, a slave device should assert waitrequest when in reset.
		Pipeline Sig	nals

readdatavalid	1	Slave → Master	Used for variable-latency, pipelined read transfers. When asserted, indicates that the
readdatavalid_n			readdata signal contains valid data. A slave with readdatavalid must assert this signal for one cycle for each read access received. There must be at least one cycle of latency between acceptance of the read and assertion of readdatavalid. For a timing diagram illustrating the readdatavalid signal, refer to Pipelined Read Transfer with Variable Latency.  A slave may assert readdatavalid to transfer data to the master independently of whether or not the slave is stalling a new command with waitrequest.
			Required if the master supports pipelined reads. Bursting masters with read functionality must include the readdatavalid signal.

Signal Role	Width	Direction	Description
writeresponse- valid			An optional signal. If present, the interface issues write responses for write commands.
			When asserted, the value on the response signal is a valid write response.
			Writeresponsevalid is only asserted one clock cycle or more after the write command is accepted. There is at least a one clock cycle latency from command acceptance to assertion of writeresponsevalid.
		Burst Sign	als
burstcount	1 – 11	Master → Slave	Used by bursting masters to indicate the number of transfers in each burst. The value of the maximum burstcount parameter must be a power of 2. A burstcount interface of width <n> can encode a max burst of size 2<sup>(<n>-1)</n></sup>. For example, a 4-bit burstcount signal can support a maximum burst count of 8. The minimum burstcount is 1. The constantBurstBehavior property controls the timing of the burstcount signal. Bursting masters with read functionality must include the readdatavalid signal.  For bursting masters and slaves using byte addresses, the following restriction applies to the width of the address:  <address_w> = <burstcount_w> +log<sub>2</sub>(<symbols_per_word_of_interface>).  For bursting masters and slaves using word addresses, the log<sub>2</sub> term above is omitted.</symbols_per_word_of_interface></burstcount_w></address_w></n>
beginburst- transfer	1	Interconnect → Slave	Asserted for the first cycle of a burst to indicate when a burst transfer is starting. This signal is deasserted after one cycle regardless of the value of waitrequest. For a timing diagram illustrating beginbursttransfer, refer to the figure in <i>Read Bursts</i> .  beginbursttransfer is optional. A slave can always internally calculate the start of the next write burst transaction by counting data transfers.  Altera recommends that you do not use this signal. This signal exists to support legacy memory controllers.

#### **Related Information**

• Typical Read and Write Transfers on page 3-12



- **Read Bursts** on page 3-18
- Read and Write Responses on page 3-19
- Pipelined Read Transfer with Variable Latency on page 3-15

# 3.3 Interface Properties

**Table 3-2: Avalon-MM Interface Properties** 

Name	Default Value	Legal Values	Description
addressUnits	Master - symbols	words, symbols	Specifies the unit for addresses. A symbol is typically a byte.
	Slave - words		Refer to the definition of address in the Avalon Memory-Mapped Interface Signal Types table for the typical use of this property.
burstCountUnits	words	words, symbols	This property specifies the units for the burstcount signal. For symbols, the burstcount value is interpreted as the number of symbols (bytes) in the burst. For words, the burstcount value is interpreted as the number of word transfers in the burst.
burstOnBurstBoundarie- sOnly	false	true, false	If true, burst transfers presented to this interface begin at addresses which are multiples of the burst size in bytes.
constantBurstBehavior	Master - false Slave -false	true, false	Masters: When true, declares that the master holds address and burstcount constant throughout a burst transaction. When false (default), declares that the master holds address and burstcount constant only for the first beat of a burst.
			Slaves: When true, declares that the slave expects address and burstcount to be held constant throughout a burst. When false (default), declares that the slave samples address and burstcount only on the first beat of a burst.
holdTime(1)	0	0 – 1000 cycles	Specifies time in timingUnits between the deassertion of write and the deassertion of address and data. (Only applies to write transactions.)

Name	Default Value	Legal Values	Description
linewrapBursts	false	true, false	Some memory devices implement a wrapping burst instead of an incrementing burst. When a wrapping burst reaches a burst boundary, the address wraps back to the previous burst boundary. Only the low-order bits are required for address counting. For example, a wrapping burst to address 0xC with burst boundaries every 32 bytes across a 32-bit interface writes to the following addresses:  • 0xC  • 0x10  • 0x14  • 0x18  • 0x1C  • 0x0  • 0x4  • 0x8
maximumPendingReadTran- sactions (1)	1(2)	1 – 64	Slaves: This parameter is the maximum number of pending reads that the slave can queue. For a timing diagram that illustrates this property, refer to Figure 3-5. Do not set this parameter to 0. (For backwards compatibility, the software supports a parameter setting of 0. However, you should not use this setting in new designs). Refer to Pipelined Read Transfer with Variable Latency on page 3-15 for additional information about using waitrequest and readdatavalid with multiple outstanding reads.  Masters: This property is the maximum number of outstanding read transactions that the master can generate. Do not set this parameter to 0. (For backwards compatibility, the software supports a parameter setting of 0. However, you should not use this setting in new designs).
maximumPendingWriteTran- sactions			The maximum number of pending non-posted writes that a slave can accept or a master can issue. A slave asserts waitrequest once the interconnect reaches this limit, and the master stops issuing commands.  The default value is 0, which allows unlimited pending write transactions for a master that supports write responses. A slave that supports write responses must set this to a non-zero value.

Name	Default Value	Legal Values	Description
readLatency(1)	0	0 - 63	Read latency for fixed-latency Avalon-MM slaves. For a timing diagram that uses a fixed latency read, refer to <b>Figure 3-6</b> .
			Avalon-MM slaves that are fixed latency must provide a value for this interface property. Avalon-MM slaves that are variable latency use the readdatavalid signal to specify valid data.
readWaitTime(1)	1	0 – 1000 cycles	For interfaces that don't use the waitrequest signal. readWaitTime indicates the timing in timingUnits before the slave accepts a read command. The timing is as if the slave asserted waitrequest for readWaitTime cycles.
setupTime(1)	0	0 – 1000 cycles	Specifies time in timingUnits between the assertion of address and data and assertion of read or write.
timingUnits(1)	cycles	cycles, nanosec- onds	Specifies the units for setupTime, holdTime, writeWaitTime and readWaitTime. Use cycles for synchronous devices and nanoseconds for asynchronous devices. Almost all Avalon-MM slave devices are synchronous.
			An Avalon-MM component that bridges from an Avalon-MM slave interface to an off-chip device may be asynchronous. That off-chip device might have a fixed settling time for bus turnaround.
writeWaitTime(1)	0	0 – 1000 Cycles	For interfaces that do not use the waitrequest signal, writeWaitTime specifies the timing in timingUnits before a slave accepts a write. The timing is as if the slave asserted waitrequest for writeWaitTime cycles or nanoseconds.
			For a timing diagram that illustrates the use of writeWaitTime, refer to Figure 3-4.
	Interfac	e Relationshi	ip Properties
associatedClock	N/A	N/A	Name of the clock interface to which this Avalon-MM interface is synchronous.
associatedReset	N/A	N/A	Name of the reset interface which resets the logic on this Avalon-MM interface.
bridgesToMaster	0	Avalon- MM Master name on the same component	An Avalon-MM bridge consists of a slave and a master, and has the property that an access to the slave requesting a particular byte or bytes will cause the same byte or bytes to be requested by the master. The Avalon-MM Pipeline Bridge in the Qsys component library implements this functionality.



Name	Default	Legal	Description
	Value	Values	

#### Notes:

- 1. Although this property characterizes a slave device, masters can declare this property to enable direct connections between matching master and slave interfaces.
- 2. If a slave interface accepts more read transfers than allowed, the interconnect pending read FIFO may overflow with unpredictable results. The slave may lose readdata or route readdata to the wrong master interface. Or, the system may lock up. The slave interface must assert waitrequest to prevent this overflow.

#### **Related Information**

- Avalon Memory-Mapped Interface Signal Roles on page 3-3
- Read and Write Responses on page 3-19

### 3.4 Timing

The Avalon-MM interface is synchronous. Each Avalon-MM interface is synchronized to an associated clock interface. Signals may be combinational if they are driven from the outputs of registers that are synchronous to the clock signal. This specification does not dictate how or when signals transition between clock edges. Timing diagrams are devoid of fine-grained timing information.

### 3.5 Transfers

This section defines two basic concepts before introducing the transfer types:

- Transfer—A transfer is a read or write operation of a word or one or more symbol of data. Transfers occur between an Avalon-MM interface and the interconnect. Transfers take one or more clock cycles to complete.
  - Both masters and slaves are part of a transfer. The Avalon-MM master initiates the transfer and the Avalon-MM slave responds to it.
- Master-slave pair—This term refers to the master interface and slave interface involved in a transfer. During a transfer, the master interface control and data signals pass through the interconnect fabric and interact with the slave interface.

### 3.5.1 Typical Read and Write Transfers

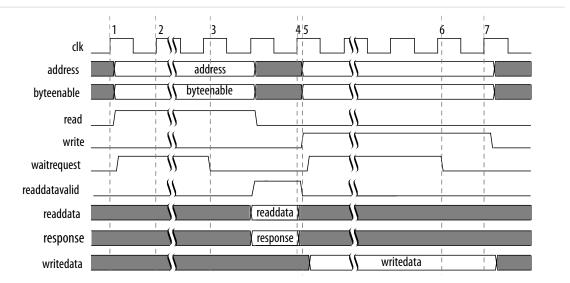
This section describes a typical Avalon-MM interface that supports read and write transfers with slave-controlled waitrequest. The slave can stall the interconnect for as many cycles as required by asserting the waitrequest signal. If a slave uses waitrequest for either read or write transfers, it must use waitrequest for both.

A slave typically receives address, byteenable, read or write, and writedata after the rising edge of the clock. A slave asserts waitrequest before the rising clock edge to hold off transfers. When the slave asserts waitrequest, the transfer is delayed. While waitrequest is asserted, the address and other control signals are held constant. Transfers complete on the rising edge of the first clk after the slave interface deasserts waitrequest.

There is no limit on how long a slave interface can stall. Therefore, you must ensure that a slave interface does not assert waitrequest indefinitely. The following figure shows read and write transfers using waitrequest. In this example, the master and slave both have a readdatavalid signal.

**Note:** waitrequest can be decoupled from the read and write request signals. waitrequest may be asserted during idle cycles. An Avalon-MM master may initiate a transaction when waitrequest is asserted and wait for that signal to be deasserted. Decoupling waitrequest from read and write requests may improve system timing. Decoupling eliminates a combinational loop including the read, write, and waitrequest signals.

Figure 3-3: Read and Write Transfers with Waitrequest



The numbers in this timing diagram, mark the following transitions:

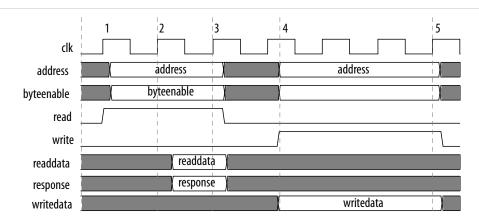
- 1. address, byteenable, and read are asserted after the rising edge of clk. The slave asserts waitrequest, stalling the transfer.
- 2. waitrequest is sampled. Because waitrequest is asserted, the cycle becomes a wait-state. address, read, write, and byteenable remain constant.
- 3. The slave deasserts waitrequest after the rising edge of clk.
- **4.** readdata, response and deasserted waitrequest are sampled, completing the transfer.
- 5. address, writedata, byteenable, and write signals are asserted after the rising edge of clk. The slave asserts waitrequest stalling the transfer.
- **6.** The slave deasserts waitrequest after the rising edge of clk.
- 7. The slave captures write data ending the transfer.

#### 3.5.2 Read and Write Transfers with Fixed Wait-States

A slave can specify fixed wait-states using the readWaitTime and writeWaitTime properties. Using fixed wait-states is an alternative to using waitrequest to stall a transfer. The address and control signals (byteenable, read, and write) are held constant for the duration of the transfer. Setting readWaitTime or writeWaitTime to <n> is equivalent to asserting waitrequest for <n> cycles per transfer.

In the following figure, the slave has a writeWaitTime = 2 and readWaitTime = 1.

Figure 3-4: Read and Write Transfer with Fixed Wait-States at the Slave Interface



The numbers in this timing diagram mark the following transitions:

- 1. The master asserts address and read on the rising edge of clk.
- 2. The next rising edge of clk marks the end of the first and only wait-state cycle. The readWaitTime is 1.
- 3. The slave asserts readdata and response on the rising edge of clk. The read transfer ends.
- 4. writedata, address, byteenable, and write signals are available to the slave.
- **5.** The write transfer ends after 2 wait-state cycles.

Transfers with a single wait-state are commonly used for multicycle off-chip peripherals. The peripheral captures address and control signals on the rising edge of clk. The peripheral has one full cycle to return data.

Components with zero wait-states are allowed. However, components with zero wait-states may decrease the achievable frequency. Zero wait-states require the component to generate the response in the same cycle that the request was presented.

### 3.5.3 Pipelined Transfers

Avalon-MM pipelined read transfers increase the throughput for synchronous slave devices that require several cycles to return data for the first access. Such devices can typically return one data value per cycle for some time thereafter. New pipelined read transfers can start before readdata for the previous transfers is returned. Write transfers cannot be pipelined.

A pipelined read transfer has an address phase and a data phase. A master initiates a transfer by presenting the address during the address phase. A slave fulfills the transfer by delivering the data during the data phase. The address phase for a new transfer (or multiple transfers) can begin before the data phase of a previous transfer completes. The delay is called pipeline latency. The pipeline latency is the duration from the end of the address phase to the beginning of the data phase.

Transfer timing for wait-states and pipeline latency have the following key differences:

- Wait-states—Wait-states determine the length of the address phase. Wait-states limit the maximum throughput of a port. If a slave requires one wait-state to respond to a transfer request, the port requires two clock cycles per transfer.
- Pipeline Latency—Pipeline latency determines the time until data is returned independently of the address phase. A pipelined slave with no wait-states can sustain one transfer per cycle. However, it may require several cycles of latency to return the first unit of data.

Wait-states and pipelined reads can be supported concurrently. Pipeline latency can be either fixed or variable.

#### 3.5.3.1 Pipelined Read Transfer with Variable Latency

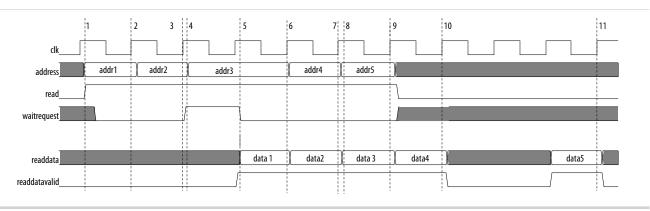
After capturing address and control signals, an Avalon-MM pipelined slave takes one or more cycles to produce data. A pipelined slave may have multiple pending read transfers at any given time. Variable-latency pipelined read transfers require one additional signal, readdatavalid. readdatavalid indicates when read data is valid. Variable-latency pipelined read transfers also include the same set of signals as non-pipelined read transfers. Slave peripherals that use readdatavalid are considered pipelined with variable latency. The readdata and readdatavalid signals corresponding to a particular read command can be asserted the cycle after that read command is asserted, at the earliest.

The slave must return readdata in the same order that it accepted the read commands. Pipelined slave ports with variable latency must use waitrequest. The slave can assert waitrequest to stall transfers to maintain an acceptable number of pending transfers. A slave may assert readdatavalid to transfer data to the master independently of whether or not the slave is stalling a new command with waitrequest.

**Note:** The maximum number of pending transfers is a property of the slave interface. The interconnect fabric builds logic to route readdata to requesting masters using this number. The slave interface, not the interconnect fabric, must track the number of pending reads. The slave must assert waitrequest to prevent the number of pending reads from exceeding the maximum number.

#### Figure 3-5: Pipelined Read Transfers with Variable Latency

The following figure shows several slave read transfers. The slave is pipelined with variable latency. In this figure, the slave can accept a maximum of two pending transfers. The slave uses waitrequest to avoid overrunning this maximum.



The numbers in this timing diagram, mark the following transitions:

- 1. The master asserts address and read, initiating a read transfer.
- 2. The slave captures addr1.
- 3. The slave captures addr2.
- **4.** The slave asserts waitrequest because it has accepted a maximum of two pending reads, causing the third transfer to stall.
- **5.** The slave asserts data1, the response to addr1. It deasserts waitrequest.
- **6.** The slave captures addr3.
- 7. The slave captures addr4. The interconnect captures data2.
- 8. The slave drives readdatavalid and readdata in response to the third read transfer.



- **9.** The slave captures addr5. The interconnect captures data3. The read signal is deasserted. The value of waitrequest is no longer relevant.
- **10.**The interconnect captures data4.
- 11. The slave drives data5 and asserts readdatavalid completing the data phase for the final pending read transfer.

If the slave cannot handle a write transfer while it is processing pending read transfers, the slave must assert its waitrequest and stall the write operation until the pending read transfers have completed. The Avalon-MM specification does not define the value of readdata in the event that a slave accepts a write transfer to the same address as a currently pending read transfer.

#### 3.5.3.2 Pipelined Read Transfers with Fixed Latency

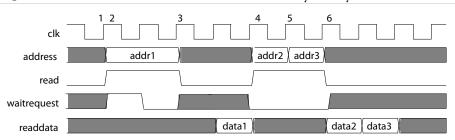
The address phase for fixed latency read transfers is identical to the variable latency case. After the address phase, a pipelined slave with fixed read latency takes a fixed number of clock cycles to return valid readdata. The readwaitTime property specifies the number of clock cycles to return valid readdata. The interconnect captures readdata on the appropriate rising clock edge, ending the data phase.

During the address phase, the slave can assert waitrequest to hold off the transfer. Or, the slave specifies the readWaitTime for a fixed number of wait states. The address phase ends on the next rising edge of clk after wait states, if any.

During the data phase, the slave drives readdata after a fixed latency. For a read latency of < n >, the slave must present valid readdata on the < nth > rising edge of clk after the end of the address phase.

#### Figure 3-6: Pipelined Read Transfer with Fixed Latency of Two Cycles

The following figure shows multiple data transfers between a master and a pipelined slave. The slave drives waitrequest to stall transfers. and has a fixed read latency of 2 cycles.



The numbers in this timing diagram, mark the following transitions:

- 1. A master initiates a read transfer by asserting read and addr1.
- **2.** The slave asserts waitrequest to hold off the transfer for one cycle.
- 3. The slave captures addr1 at the rising edge of clk. The address phase ends here.
- **4.** The slave presents valid readdata after 2 cycles, ending the transfer.
- 5. addr2 and read are asserted for a new read transfer.
- **6.** The master initiates a third read transfer during the next cycle, before the data from the prior transfer is returned.

#### 3.5.4 Burst Transfers

A burst executes multiple transfers as a unit, rather than treating every word independently. Bursts may increase throughput for slave ports that achieve greater efficiency when handling multiple words at a time, such as SDRAM. The net effect of bursting is to lock the arbitration for the duration of the burst. A

bursting Avalon-MM interface that supports both reads and writes must support both read and write bursts.

Bursting Avalon-MM interfaces include a burstcount output signal. If a slave has a burstcount input, it is considered burst capable.

The burstcount signal behaves as follows:

- At the start of a burst, burstcount presents the number of sequential transfers in the burst.
- For width  $\langle n \rangle$  of burstcount, the maximum burst length is  $2^{(\langle n \rangle 1)}$ . The minimum legal burst length is one.

To support slave read bursts, a slave must also support:

- Wait states with the waitrequest signal.
- Pipelined transfers with variable latency with the readdatavalid signal.

At the start of a burst, the slave sees the address and a burst length value on burstcount. For a burst with an address of < a > and a burstcount value of < b >, the slave must perform < b > consecutive transfers starting at address < a >. The burst completes after the slave receives (write) or returns (read) the  $< b^{\text{th}} >$  word of data. The bursting slave must capture address and burstcount only once for each burst. The slave logic must infer the address for all but the first transfers in the burst. A slave can also use the input signal beginbursttransfer, which the interconnect asserts on the first cycle of each burst.

#### 3.5.4.1 Write Bursts

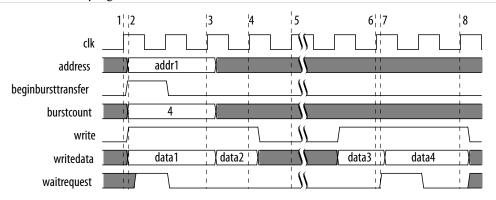
These rules apply when a write burst begins with burstcount greater than one:

- When a burstcount of <n> is presented at the beginning of the burst, the slave must accept <n> successive units of writedata to complete the burst. Arbitration between the master-slave pair is locked until the burst completes. This lock guarantees that no other master can execute transactions on the slave until the write burst is completes.
- The slave must only capture writedata when write is asserted. During the burst, the master can deassert write indicating that writedata is invalid. Deasserting write does not terminate the burst. It delays it. When a burst is delayed, no other masters can access the slave, reducing the transfer efficiency.
- The constantBurstBehavior property specifies the behavior of the burst signals. When constant-BurstBehavior is true for a master, it indicates that the master holds address and burstcount stable throughout a burst. When constantBurstBehavior is false, it indicates that the master holds address and burstcount stable only for the first transaction of a burst. When true for a slave, constantBurstBehavior declares that the slave expects address and burstcount to be held stable throughout a burst. When constantBurstBehavior is false, it indicates that the slave samples address and burstcount only on the first transaction of a burst.
- The slave can delay a transfer by asserting waitrequest forcing writedata, write, burstcount, and byteenable to be held constant.
- The functionality of the byteenable signal is the same for bursting and non-bursting slaves. For a 32-bit master burst-writing to a 64-bit slave, starting at byte address 4, the first write transfer seen by the slave is at its address 0, with byteenable = 8'bl1110000. The byteenables can change for different words of the burst.
- The byteenable signals do not all have to be asserted. A burst master writing partial words can use the byteenable signal to identify the data being written.



#### Figure 3-7: Write Burst with constantBurstBehavior Set to False for Master and Slave

The following figure demonstrates a slave write burst of length 4. In this example, the slave asserts waitrequest twice delaying the burst.



The numbers in this timing diagram, mark the following transitions:

- 1. The master asserts address, burstcount, write, and drives the first unit of writedata.
- 2. The slave immediately asserts waitrequest, indicating that it is not ready to proceed with the transfer.
- **3.** waitrequest is low. The slave captures addr1, burstcount, and the first unit of writedata. On subsequent cycles of the transfer, address and burstcount are ignored.
- **4.** The slave captures the second unit of data at the rising edge of clk.
- **5.** The burst is paused while write is deasserted.
- **6.** The slave captures the third unit of data at the rising edge of clk.
- 7. The slave asserts waitrequest. In response, all outputs are held constant through another clock cycle.
- **8.** The slave captures the last unit of data on this rising edge of clk. The slave write burst ends.

In the figure above, the beginbursttransfer signal is asserted for the first clock cycle of a burst and is deasserted on the next clock cycle. Even if the slave asserts waitrequest, the beginbursttransfer signal is only asserted for the first clock cycle.

For information about Avalon-MM properties, refer to **Table 3-2**.

#### **Related Information**

**Interface Properties** on page 3-9

#### 3.5.4.2 Read Bursts

Read bursts are similar to pipelined read transfers with variable latency. A read burst has distinct address and data phases. readdatavalid indicates when the slave is presenting valid readdata. Unlike pipelined read transfers, a single read burst address results in multiple data transfers.

These rules apply to read bursts:

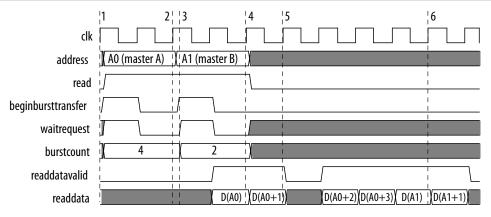
- When burstcount is  $\langle n \rangle$ , the slave must return  $\langle n \rangle$  words of readdata to complete the burst.
- The slave presents each word by providing readdata and asserting readdatavalid for a cycle. Deassertion of readdatavalid delays but does not terminate the burst data phase.
- The byteenables presented with a read burst command apply to all cycles of the burst. A byteenable value of 1 means that the least significant byte is being read across all of the read cycles.

**Note:** Altera recommends that burst capable slaves not have read side effects. (This specification does not guarantee how many bytes will be read from the slave in order to satisfy a request.)



#### Figure 3-8: Read Burst

The following figure illustrates a system with two bursting masters accessing a slave. Note that Master B can drive a read request before the data has returned for Master A.



The numbers in this timing diagram, mark the following transitions:

- 1. Master A asserts address (A0), burstcount, and read after the rising edge of clk. The slave asserts waitrequest, causing all inputs except beginbursttransfer to be held constant through another clock cycle.
- 2. The slave captures AO and burstcount at this rising edge of clk. A new transfer could start on the next cycle.
- 3. Master B drives address (A1), burstcount, and read. The slave asserts waitrequest, causing all inputs except beginbursttransfer to be held constant. The slave could have returned read data from the first read request at this time, at the earliest.
- **4.** The slave presents valid readdata and asserts readdatavalid, transferring the first word of data for master A.
- **5.** The second word for master A is transferred. The slave deasserts readdatavalid pausing the read burst. The slave port can keep readdatavalid deasserted for an arbitrary number of clock cycles.
- **6.** The first word for master B is returned.

#### 3.5.4.3 Line-Wrapped Bursts

Processors with instruction caches gain efficiency by using line-wrapped bursts. When a processor requests data that is not in the cache, the cache controller must refill the entire cache line. For a processor with a cache line size of 64 bytes, a cache miss causes 64 bytes to be read from memory. If the processor reads from address 0xC when the cache miss occurred, then an inefficient cache controller could issue a burst at address 0, resulting in data from read addresses 0x0, 0x4, 0x8, 0xC, 0x10, 0x14, 0x18, . . . 0x3C. The requested data is not available until the fourth read. With line-wrapping bursts, the address order is 0xC, 0x10, 0x14, 0x18, . . . 0x3C, 0x0, 0x4, and 0x8. The requested data is returned first. The entire cache line is eventually refilled from memory.

### 3.6 Read and Write Responses

For any Avalon-MM slave, commands must be processed in a hazard-free manner. Read and write responses are issued following the order in which commands are accepted.

Send Feedback

# 3.6.1 Transaction Order for Avalon-MM Read and Write Responses (Masters and Slaves)

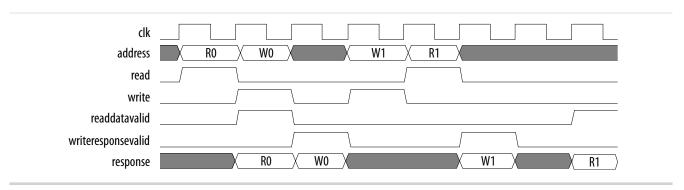
For any Avalon-MM master:

- Commands to the same slave are guaranteed to reach the slave in command issue order, and the slave responds in command issue order.
- Commands to different slaves may reach and be responded to by the slaves in a different order than the order in which the commands were issued. When successful, the slave responds in command issue order.
- Responses (if present) return in command issue order, regardless of whether the read or write commands are directed to the same or different slaves.
- There is no guaranteed transaction order between different masters.

#### 3.6.2 Avalon-MM Read and Write Responses Timing Diagram

The following diagram shows command acceptance and command issue order for Avalon-MM read and write responses.

Figure 3-9: Avalon-MM Read and Write Responses Timing Diagram



### 3.7 Address Alignment

The interconnect only supports aligned accesses. A master can only issue addresses that are a multiple of its data width in symbols. A master can write partial words by deasserting some byteenables. For example, a write of 2 bytes at address 2 would have 4'b1100 for the byteenables.

### 3.8 Avalon-MM Slave Addressing

Dynamic bus sizing manages data during transfers between master-slave pairs of differing data widths. Slave data are aligned in contiguous bytes in the master address space.

If the master data width is wider than the slave data width, words in the master address space map to multiple locations in the slave address space. For example, a 32-bit master read from a 16-bit slave results in two read transfers on the slave side. The reads are to consecutive addresses.

If the master is narrower than the slave, then the interconnect manages the slave byte lanes. During master read transfers, the interconnect presents only the appropriate byte lanes of slave data to the

narrower master. During master write transfers, the interconnect automatically asserts the byteenable signals to write data only to the specified slave byte lanes.

Slaves must have a data width of 8, 16, 32, 64, 128, 256, 512 or 1024 bits. The following table shows how slave data of various widths is aligned within a 32-bit master performing full-word accesses. In this table, OFFSET[N] refers to a slave word size offset into the slave address space.

Table 3-3: Dynamic Bus Sizing Master-to-Slave Address Mapping

Master Byte	Access	32-Bit Master Data					
Address (1)		When Accessing an 8-Bit Slave Interface	When Accessing a 16-Bit Slave Interface	When Accessing a 64-Bit Slave Interface			
	1	OFFSET[0] <sub>70</sub>	OFFSET[0] <sub>150</sub> (2)	OFFSET[0] <sub>310</sub>			
0x00	2	OFFSET[1] <sub>70</sub>	OFFSET[1] <sub>150</sub>	_			
0.000	3	OFFSET[2] <sub>70</sub>	_	_			
	4	OFFSET[3] <sub>70</sub>	_	_			
	1	OFFSET[4] <sub>70</sub>	OFFSET[2] <sub>150</sub>	OFFSET[0] <sub>6332</sub>			
0x04	2	OFFSET[5] <sub>70</sub>	OFFSET[3] <sub>150</sub>	_			
0X04	3	OFFSET[6] <sub>70</sub>	_	_			
	4	OFFSET[7] <sub>70</sub>	_	_			
	1	OFFSET[8] <sub>70</sub>	OFFSET[4] <sub>150</sub>	OFFSET[1]310			
0x08	2	OFFSET[9] <sub>70</sub>	OFFSET[5] <sub>150</sub>	_			
0.000	3	OFFSET[10] <sub>70</sub>	_	_			
	4	OFFSET[11] <sub>70</sub>	_	_			
	1	OFFSET[12] <sub>70</sub>	OFFSET[6] <sub>150</sub>	OFFSET[1] <sub>6332</sub>			
0x0C	2	OFFSET[13] <sub>70</sub>	OFFSET[7] <sub>150</sub>	_			
UXUC	3	OFFSET[14] <sub>70</sub>	_	_			
	4	OFFSET[15] <sub>70</sub>	_	_			
•••							

#### Notes:

- 1. Although the master is issuing byte addresses, it is accessing full 32-bit words.
- **2.** For all slave entries, [< n>] is the word offset and the subscript values are the bits in the word.

**Send Feedback** 

## **Avalon Interrupt Interfaces**

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Avalon Interrupt interfaces allow slave components to signal events to master components. For example, a DMA controller can interrupt a processor when it has completed a DMA transfer.

### 4.1 Interrupt Sender

An interrupt sender drives a single interrupt signal to an interrupt receiver. The timing of the <code>irq</code> signal must be synchronous to the rising edge of its associated clock. <code>irq</code> has no relationship to any transfer on any other interface. <code>irq</code> must be asserted until acknowledged on the associated Avalon-MM slave interface.

Interrupts are component specific. The receiver typically determines the appropriate response by reading an interrupt status register from an Avalon-MM slave interface.

### 4.1.1 Avalon Interrupt Sender Signal Roles

**Table 4-1: Interrupt Sender Signal Roles** 

Signal Role	Width	Direction	Required	Description
irq irq_n	1	Output	Yes	Interrupt Request. A slave asserts irq when it needs service. The interrupt receiver determines the relative priority of the interrupts.

### 4.1.2 Interrupt Sender Properties

#### **Table 4-2: Interrupt Sender Properties**

Property Name	Default Value	Legal Values	Description
associatedAddres- sablePoint	N/A	Name of Avalon- MM slave on this component.	The name of the Avalon-MM slave interface that provides access to the registers to service the interrupt.

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Property Name	Default Value	Legal Values	Description
associatedClock	N/A	Name of a clock interface on this component.	The name of the clock interface to which this interrupt sender is synchronous. The sender and receiver may have different values for this property.
associatedReset	N/A	Name of a reset interface on this component.	The name of the reset interface to which this interrupt sender is synchronous.

### 4.2 Interrupt Receiver

An interrupt receiver interface receives interrupts from interrupt sender interfaces. Components with Avalon-MM master interfaces can include an interrupt receiver to detect interrupts asserted by slave components with interrupt sender interfaces. The interrupt receiver accepts interrupt requests from each interrupt sender as a separate bit.

### 4.2.1 Avalon Interrupt Receiver Signal Roles

**Table 4-3: Interrupt Receiver Signal Roles** 

Signal Role	Width	Direction	Required	Description
irq	1–32	Input	Yes	irq is an < <i>n</i> >-bit vector, where each bit corresponds directly to one IRQ sender with no inherent assumption of priority.

### **4.2.2 Interrupt Receiver Properties**

**Table 4-4: Interrupt Receiver Properties** 

Property Name	Default Value	Legal Values	Description
associatedAddressable Point	N/A	Name of Avalon- MM master interface	The name of the Avalon-MM master interface used to service interrupts received on this interface.
associatedClock	N/A	Name of an Avalon Clock interface	The name of the Avalon Clock interface to which this interrupt receiver is synchronous. The sender and receiver may have different values for this property.

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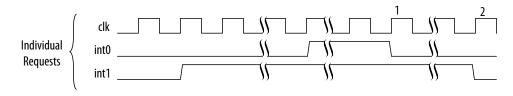
Property Name	Default Value	Legal Values	Description
associatedReset	N/A	Name of an Avalon Reset interface	The name of the reset interface to which this interrupt receiver is synchronous.

### 4.2.3 Interrupt Timing

The Avalon-MM master services the priority 0 interrupt before the priority 1 interrupt.

### Figure 4-1: Interrupt Timing

In the following figure, interrupt 0 has higher priority. The interrupt receiver is in the process of handling int1 when int0 is asserted. The int0 handler is called and completes. Then, the int1 handler resumes. The diagram shows int0 deasserts at time 1. int1 deasserts at time 2.



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## **Avalon Streaming Interfaces**

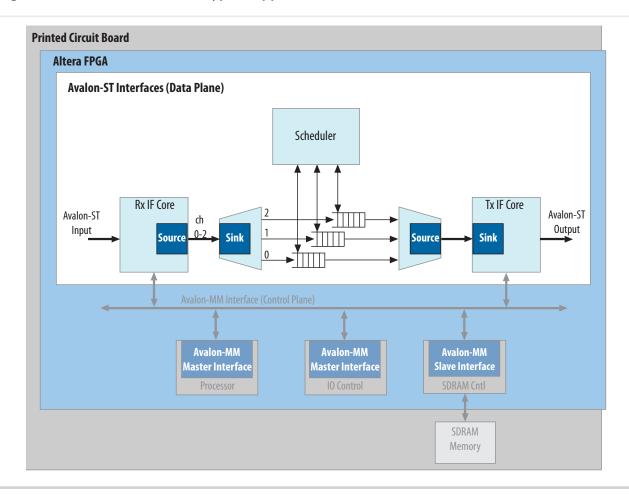
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You can use Avalon Streaming (Avalon-ST) interfaces for components that drive high bandwidth, low latency, unidirectional data. Typical applications include multiplexed streams, packets, and DSP data. The Avalon-ST interface signals can describe traditional streaming interfaces supporting a single stream of data without knowledge of channels or packet boundaries. The interface can also support more complex protocols capable of burst and packet transfers with packets interleaved across multiple channels.

Figure 5-1: Avalon-ST Interface - Typical Application of the Avalon-ST Interface





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#### 5.1 Terms and Concepts

All Avalon-ST source and sink interfaces are not necessarily interoperable. However, if two interfaces provide compatible functions for the same application space, adapters are available to allow them to interoperate.

Avalon-ST interfaces support datapaths requiring the following features:

- Low latency, high throughput point-to-point data transfer
- Multiple channel support with flexible packet interleaving
- Sideband signaling of channel, error, and start and end of packet delineation
- Support for data bursting
- Automatic interface adaptation

### 5.1 Terms and Concepts

The Avalon-ST interface protocol defines the following terms and concepts:

- Avalon Streaming System—An Avalon Streaming system contains one or more Avalon-ST connections that transfer data from a source interface to a sink interface. The system shown above consists of Avalon-ST interfaces to transfer data from the system input to output. Avalon-MM control and status register interfaces provide for software control.
- Avalon Streaming Components—A typical system using Avalon-ST interfaces combines multiple functional modules, called components. The system designer configures the components and connects them together to implement a system.
- Source and Sink Interfaces and Connections—When two components are connected, the data flows from the source interface to the sink interface. The combination of a source interface connected to a sink interface is referred to as a connection.
- Backpressure—Backpressure allows a sink to signal a source to stop sending data. Support for backpressure is optional. The sink uses backpressure to stop the flow of data for the following reasons:
  - When its FIFOs are full
  - When there is congestion on its output interface.
- Transfers and Ready Cycles—A transfer results in data and control propagation from a source interface to a sink interface. For data interfaces, a ready cycle is a cycle during which the sink can accept a transfer.
- Symbol—A symbol is the smallest unit of data. For most packet interfaces, a symbol is a byte. One or more symbols make up the single unit of data transferred in a cycle.
- Channel—A channel is a physical or logical path or link through which information passes between two ports.
- Beat—A beat is a single cycle transfer between a source and sink interface made up of one or more symbols.
- Packet—A packet is an aggregation of data and control signals that is transmitted together. A packet may contain a header to help routers and other network devices direct the packet to the correct destination. The packet format is defined by the application, not this specification. Avalon-ST packets can be variable in length and can be interleaved across a connection. With an Avalon-ST interfaces, the use of packets is optional.

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## **5.2 Avalon Streaming Interface Signal Roles**

Each signal in an Avalon-ST source or sink interface corresponds to one Avalon-ST signal role. An Avalon-ST interface may contain only one instance of each signal role. All Avalon-ST signal roles apply to both sources and sinks and have the same meaning for both.

**Table 5-1: Avalon-ST Interface Signals** 

In the following table, all signal roles are active high.

In the following table, all signal Role	Width	Direction	Description		
Fundamental Signals					
channel	1 – 128	Source → Sink	The channel number for data being transferred on the current cycle.		
			If an interface supports the channel signal, it must also define the maxChannel parameter.		
data	1 - 4,096	Source → Sink	The data signal from the source to the sink, typically carries the bulk of the information being transferred.  The contents and format of the data signal is further defined by parameters.		
error	1 – 256	Source → Sink	A bit mask used to mark errors affecting the data being transferred in the current cycle. A single bit in error is used for each of the errors recognized by the component, as defined by the errorDescriptor property.		
ready	1	Sink → Source	Asserted high to indicate that the sink can accept data. ready is asserted by the sink on cycle <n> to mark cycle <n +="" readylatency=""> as a ready cycle. The source may only assert valid and transfer data during ready cycles.</n></n>		
			Sources without a ready input cannot be backpressured. Sinks without a ready output never need to backpressure.		
valid	1	Source → Sink	Asserted by the source to qualify all other source to sink signals. The sink samples data and other source-to-sink signals on ready cycles where valid is asserted. All other cycles are ignored.		
			Sources without a valid output implicitly provide valid data on every cycle that they are not being backpressured. Sinks without a valid input expect valid data on every cycle that they are not backpressuring.		
De drot Transfer Signals					

**Packet Transfer Signals** 

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Signal Role	Width	Direction	Description
empty	1 – 5	Source → Sink	Indicates the number of symbols that are empty, that is, do not represent valid data. The empty signal is not used on interfaces where there is one symbol per beat.
endofpacket	1	Source → Sink	Asserted by the source to mark the end of a packet.
startofpacket	1	Source → Sink	Asserted by the source to mark the beginning of a packet.

## 5.3 Signal Sequencing and Timing

### 5.3.1 Synchronous Interface

All transfers of an Avalon-ST connection occur synchronous to the rising edge of the associated clock signal. All outputs from a source interface to a sink interface, including the data, channel, and error signals, must be registered on the rising edge of clock. Inputs to a sink interface do not have to be registered. Registering signals at the source facilitates high frequency operation.

### 5.3.2 Clock Enables

Avalon-ST components typically do not include a clock enable input. The Avalon-ST signaling itself is sufficient to determine the cycles that a component should and should not be enabled. Avalon-ST compliant components may have a clock enable input for their internal logic. But, they must ensure that the timing of the interface control signals still adheres to the protocol.

## **5.4 Avalon-ST Interface Properties**

**Table 5-2: Avalon-ST Interface Properties** 

Property Name	Default Value	Legal Values	Description
symbolsPerBeat	1	1 – 32	The number of symbols that are transferred on every valid cycle.
associatedClock	1	Clock interface	The name of the Avalon Clock interface to which this Avalon-ST interface is synchronous.
associatedReset	1	Reset interface	The name of the Avalon Reset interface to which this Avalon-ST interface is synchronous.

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Property Name	Default Value	Legal Values	Description
beatsPerCycle	8		Specifies the number of beats that are transferred in a single cycle. This property allows you to transfer 2 separate, but correlated streams using the same start_of_packet, end_of_packet, ready and valid signals.  beatsPerCycle is a rarely used feature of the Avalon-ST protocol.
dataBitsPerSymbol	8	1 – 512	Defines the number of bits per symbol. For example, byte-oriented interfaces have 8-bit symbols. This value is not restricted to be a power of 2.
emptyWithinPacket	false	true, false	When true, empty is valid for the entire packet.
errorDescriptor	0	List of strings	A list of words that describe the error associated with each bit of the error signal. The length of the list must be the same as the number of bits in the error signal. The first word in the list applies to the highest order bit. For example, "crc, overflow" means that bit[1] of error indicates a CRC error. Bit[0] indicates an overflow error.
firstSymbolInHigh OrderBits	true	true, false	When true, the first-order symbol is driven to the most significant bits of the data interface. The highest-order symbol is labeled D0 in this specification. When this property is set to false, the first symbol appears on the low bits. D0 appears at data[7:0]. For a 32-bit bus, if true, D0 appears on bits[31:24].
maxChannel	0	0 – 255	The maximum number of channels that a data interface can support.
readyLatency	0	0 - 8	Defines the relationship between assertion and deassertion of ready and cycles which are considered to be available for data transfer. The readyLatency is defined separately for each interface.

# **5.5 Typical Data Transfers**

This section defines the transfer of data from a source interface to a sink interface. In all cases, the data source and the data sink must comply with the specification. It is not the responsibility of the data sink to detect source protocol errors.

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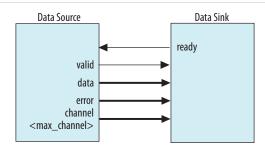
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## 5.6 Signal Details

The following figure shows the signals that are typically included in an Avalon-ST interface. As this figure indicates, a typical Avalon-ST source interface drives the valid, data, error, and channel signals to the sink. The sink can apply backpressure using the ready signal.

Figure 5-2: Typical Avalon-ST Interface Signals



Here are more details about these signals:

- ready—On interfaces supporting backpressure, the sink asserts ready to mark the cycles where transfers may take place. If ready is asserted on cycle <n>, cycle <n + readyLatency> is considered a ready cycle.
- valid—The valid signal qualifies valid data on any cycle where data is being transferred from the source to the sink. On each valid cycle the data signal and other source to sink signals are sampled by the sink.
- data—The data signal typically carries the bulk of the information being transferred from the source to the sink. The data signal consists of one or more symbols being transferred on every clock cycle. The dataBitsPerSymbol parameter defines how the data signal is divided into symbols.
- error—Errors are signaled with the error signal, where each bit in error corresponds to a possible error condition. A value of 0 on any cycle indicates the data on that cycle is error-free. The action that a component takes when an error is detected is not defined by this specification.
- channel—The optional channel signal is driven by the source to indicate the channel to which the data belongs. The meaning of channel for a given interface depends on the application. Some applications use channel as an interface number indication. Other applications use channel as a page number or timeslot indication. When the channel signal is used, all of the data transferred in each active cycle belongs to the same channel. The source may change to a different channel on successive active cycles.

An interface that uses the channel signal must define the maxChannel parameter to indicate the maximum channel number. If the number of channels an interface supports changes dynamically, maxChannel is the maximum number the interface can support.

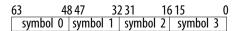
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### 5.7 Data Layout

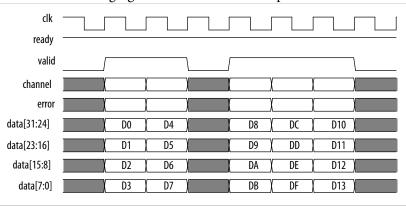
#### Figure 5-3: Data Symbols

The following figure shows a 64-bit data signal with dataBitsPerSymbol=16. Symbol 0 is the most significant symbol.



#### Figure 5-4: Layout of Data

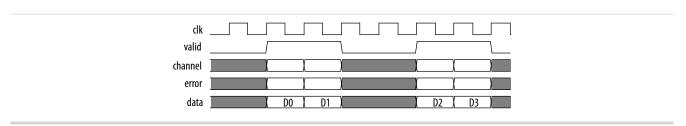
The timing diagram in the following figure shows a 32-bit example where dataBitsPerSymbol=8.



## 5.8 Data Transfer without Backpressure

The data transfer without backpressure is the most basic of Avalon-ST data transfers. On any given clock cycle, the source interface drives the data and the optional channel and error signals, and asserts valid. The sink interface samples these signals on the rising edge of the reference clock if valid is asserted.

Figure 5-5: Data Transfer without Backpressure



## 5.9 Data Transfer with Backpressure

The sink asserts ready for a single clock cycle to indicate it is ready for an active cycle. Cycles during which the sink is ready for data are called ready cycles. During a ready cycle, the source may assert valid and provide data to the sink. If it has no data to send, it deasserts valid and can drive data to any value.

Each interface that supports backpressure defines the readyLatency parameter to indicate the number of cycles from the time that ready is asserted until valid data can be driven. If the readyLatency is nonzero,



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cycle < n + readyLatency> is a ready cycle if ready is asserted on cycle < n>. Any interface that includes the ready signal and defines the readyLatency parameter supports backpressure.

When readyLatency = 0, data is transferred only when ready and valid are asserted on the same cycle. In this mode, the source does not receive the sink's ready signal before it begins sending valid data. The source provides the data and asserts valid whenever it can. The source waits for the sink to capture the data and assert ready. The source can change the data it is providing at any time. The sink only captures input data from the source when ready and valid are both asserted.

When readyLatency >= 1, the sink asserts ready before the ready cycle itself. The source can respond during the appropriate cycle by asserting valid. It may not assert valid during a cycle that is not a ready cycle.

Figure 5-6: Avalon-ST Interface with readyLatency = 4

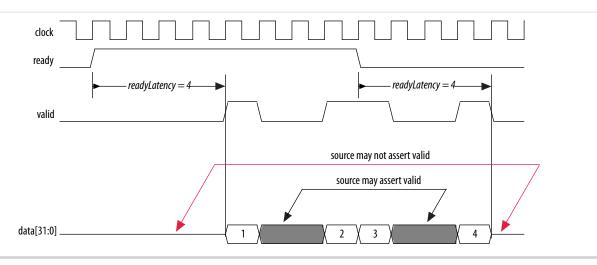
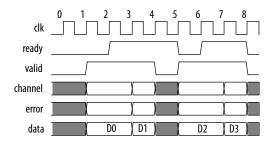


Figure 5-7: Transfer with Backpressure, readyLatency=0

The following figure illustrates these events:

- 1. The source provides data and asserts valid on cycle 1, even though the sink is not ready.
- 2. The source waits until cycle 2, when the sink does assert ready, before moving onto the next data cycle.
- **3.** In cycle 3, the source drives data on the same cycle and the sink is ready to receive it. The transfer occurs immediately.
- **4.** In cycle 4, the sink asserts ready, but the source does not drive valid data.



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### Figure 5-8: Transfer with Backpressure, readyLatency=1

The following figures show data transfers with readyLatency=1 and readyLatency=2, respectively. In both these cases, ready is asserted before the ready cycle, and the source responds 1 or 2 cycles later by providing data and asserting valid. When readyLatency is not 0, the source must deassert valid on non-ready cycles.

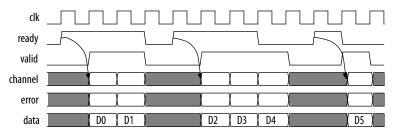
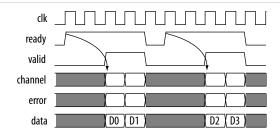


Figure 5-9: Transfer with Backpressure, readyLatency=2

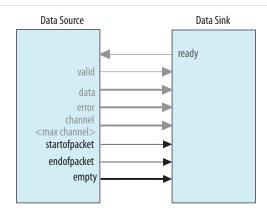


### 5.10 Packet Data Transfers

The packet transfer property adds support for transferring packets from a source interface to a sink interface. Three additional signals are defined to implement the packet transfer. Both the source and sink interfaces must include these additional signals to support packets. You can only connect source and sink interfaces with matching packet properties. Qsys does not automatically add the startofpacket, endofpacket, and empty signals to source or sink interfaces that do not include these signals.

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#### Figure 5-10: Avalon-ST Packet Interface Signals



## 5.11 Signal Details

- startofpacket—All interfaces supporting packet transfers require the startofpacket signal. startofpacket marks the active cycle containing the start of the packet. This signal is only interpreted when valid is asserted.
- endofpacket—All interfaces supporting packet transfers require the endofpacket signal. endofpacket marks the active cycle containing the end of the packet. This signal is only interpreted when valid is asserted. startofpacket and endofpacket can be asserted in the same cycle. No idle cycles are required between packets. The startofpacket signal can follow immediately after the previous endofpacket signal.
- empty—The optional empty signal indicates the number of symbols that are empty the endofpacket cycle. The sink only checks the value of the empty during active cycles that have endofpacket asserted. The empty symbols are always the last symbols in data, those carried by the low-order bits when firstSymbolInHighOrderBits = true. The empty signal is required on all packet interfaces whose data signal carries more than one symbol of data and have a variable length packet format. The size of the empty signal in bits is ceil[log2(<symbols per cycle>)].

### 5.12 Protocol Details

Packet data transfer follows the same protocol as the typical data transfer with the addition of the startofpacket, endofpacket, and empty.

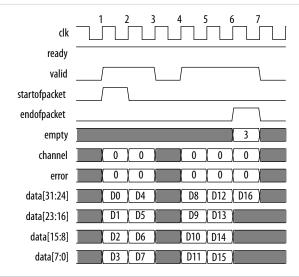
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#### Figure 5-11: Packet Transfer

The following figure illustrates the transfer of a 17-byte packet from a source interface to a sink interface, where readyLatency=0. It illustrates the following events:

- 1. Data transfer occurs on cycles 1, 2, 4, 5, and 6, when both ready and valid are asserted.
- 2. During cycle 1, startofpacket is asserted. The first 4 bytes of packet are transferred.
- **3.** During cycle 6, endofpacket is asserted. empty has a value of 3. This value indicates that this is the end of the packet and that 3 of the 4 symbols are empty. In cycle 6, the high-order byte, data[31:24] drives valid data.





### **Avalon Conduit Interfaces**

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Avalon Conduit interfaces group an arbitrary collection of signals. You can specify any role for conduit signals. However, when you connect conduits, the roles and widths must match and the directions must be opposite. An Avalon Conduit interface can include input, output, and bidirectional signals. A module can have multiple Avalon Conduit interfaces to provide a logical signal grouping. Conduit interfaces can declare an associated clock. When connected conduit interfaces are in different clock domains, Qsys generates an error message.

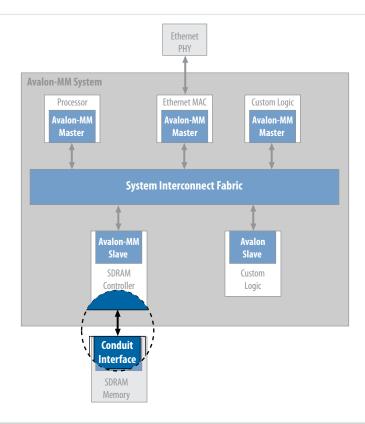
**Note:** If possible, you should use the standard Avalon-MM or Avalon-ST interfaces instead of creating an Avalon Conduit interface. Qsys provides validation and adaptation for these interfaces. It cannot provide validation or adaptation for Avalon Conduit interfaces.

Conduit interfaces typically used to drive off-chip device signals, such as an SDRAM address, data and control signals.

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Figure 6-1: Focus on the Conduit Interface



# **6.1 Avalon Conduit Signal Roles**

**Table 6-1: Conduit Signal Roles** 

Signal Role	Width	Direction	Description
<any></any>	<n></n>	In, out, or bidirectional	A conduit interface consists of one or more input, output, or bidirectional signals of arbitrary width. Conduits can have any user-specified role. You can connect compatible Conduit interfaces inside a Qsys system provided the roles and widths match and the directions are opposite.

# **6.2 Conduit Properties**

There are no properties for conduit interfaces.

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## **Avalon Tristate Conduit Interface**

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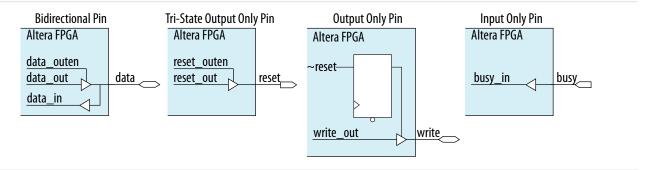


The Avalon Tristate Conduit Interface (Avalon-TC) is a point-to-point interface designed for on-chip controllers that drive off-chip components. This interface allows data, address, and control pins to be shared across multiple tristate devices. Sharing conserves pins in systems that have multiple external memory devices.

The Avalon-TC interface restricts the more general Avalon Conduit Interface in two ways:

- The Avalon-TC requires request and grant signals. These signals enable bus arbitration when multiple Tristate Conduit Masters (TCM) are requesting access to a shared bus.
- The pin type of a signal must be specified using suffixes appended to a signal's role. The three suffixes are: \_out, \_in, and \_outen. Matching role prefixes identify signals that share the same I/O Pin. The following illustrates the naming conventions for Avalon-TC shared pins.

Figure 7-1: Shared Pin Types



The next figure illustrates pin sharing using Avalon-TC interfaces. This figure illustrates the following points.

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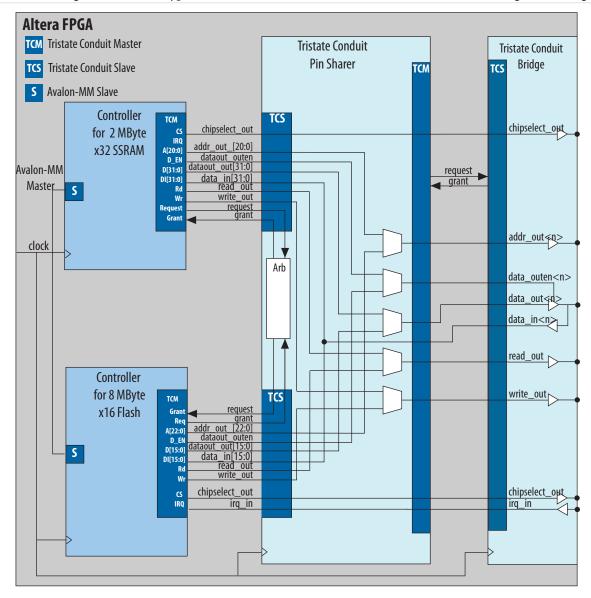
- The Tristate Conduit Pin Sharer includes separate Tristate Conduit Slave Interfaces for each Tristate Conduit Master. Each master and slave pair has its own request and grant signals.
- The Tristate Conduit Pin Sharer identifies signals with identical roles as tristate signals that share the same FPGA pin. In this example, the following signals are shared: addr\_out, data\_out, data\_in, read\_out, and write\_out.
- The Tristate Conduit Pin Sharer drives a single bus including all of the shared signals to the Tristate Conduit Bridge. If the widths of shared signals differ, the Tristate Conduit Pin Sharer aligns them on their 0<sup>th</sup> bit. It drives the higher-order pins to 0 whenever the smaller signal has control of the bus.
- Signals that are not shared propagate directly through the Tristate Conduit Pin Sharer. In this example, the following signals are not shared: chipselect0\_out, irq0\_out, chipselect1\_out, and irq1\_out.
- All Avalon-TC interfaces connected to the same Tristate Conduit Pin Sharer must be in the same clock domain.

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#### Figure 7-2: Tristate Conduit Interfaces

The following illustrates the typical use of Avalon-TC Master and Slave interfaces and signal naming.



For more information about the Generic Tristate Controller and Tristate Conduit Pin Sharer, refer to the *Avalon Tristate Conduit Components User Guide*.

#### **Related Information**

**Avalon Tristate Conduit Components User Guide** 

## 7.1 Avalon Tristate Conduit Signal Roles

The following table lists the signal defined for the Avalon Tristate Conduit interface. All Avalon-TC signals apply to both masters and slaves and have the same meaning for both

**Avalon Tristate Conduit Interface** 

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**Table 7-1: Tristate Conduit Interface Signal Roles** 

Signal Role	Width	Direction	Required	Description
request	1	Master → Slave	Yes	The meaning of request depends on the state of the grant signal, as the following rules dictate.
				When request is asserted and grant is deasserted, request is requesting access for the current cycle.
				When request is asserted and grant is asserted, request is requesting access for the next cycle. Consequently, request should be deasserted on the final cycle of an access.
				The request is deasserted in the last cycle of a bus access. It can be reasserted immediately following the final cycle of a transfer. This protocol makes both rearbitration and continuous bus access possible if no other masters are requesting access.
				Once asserted, request must remain asserted until granted. Consequently, the shortest bus access is 2 cycles. Refer to <i>Tristate Conduit Arbitration Timing</i> for an example of arbitration timing.
grant	1	Slave → Master	Yes	When asserted, indicates that a tristate conduit master has been granted access to perform transactions. grant is asserted in response to the request signal. It remains asserted until 1 cycle following the deassertion of request.
<name>_in</name>	1 – 1024	Slave → Master	No	The input signal of a logical tristate signal.
<name>_out</name>	1 – 1024	Master → Slave	No	The output signal of a logical tristate signal.
<name>_ outen</name>	1	Master → Slave	No	The output enable for a logical tristate signal.

## 7.2 Tristate Conduit Properties

There are no special properties for the Avalon-TC Interface.

## 7.3 Tristate Conduit Timing

The following illustrates arbitration timing for the Tristate Conduit Pin Sharer. Note that a device can drive or receive valid data in the granted cycle.

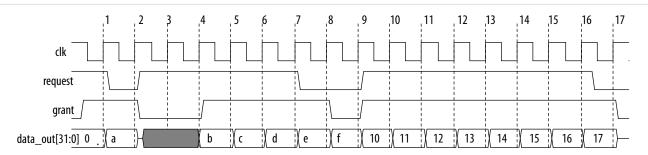
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### Figure 7-3: Tristate Conduit Arbitration Timing

This figure shows the following sequence of events:

- 1. In cycle 1, the tristate conduit slave asserts grant. The slave drives valid data in cycles1 and 2.
- 2. In cycle 4, the tristate conduit slave asserts grant. The slave drives valid data in cycles 5–8.
- 3. In cycle 9, the tristate conduit slave asserts grant. The slave drives valid data in cycles 10–17.
- **4.** Cycles 3, 4 and 9 do not contain valid data.





## **Deprecated Signals**



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Deprecated signals implement functionality that is no longer required or has been superseded.

### begintransfer

An output of Avalon-MM masters. Asserted for a single cycle at the beginning of a transfer. This is signal is not used and not necessary.

#### chipselect or chipselect\_n

chipselect or chipselect\_n: The chip select signal as described below was deprecated with the release of the Avalon Tristate Conduct (Avalon-TC) interface type which includes a chip select signal.

Formerly chipselect was a 1-bit input to an Avalon Memory-Mapped (Avalon-MM) slave interface signaling the beginning of a read or write transfer. The current Qsys interconnect filters read and write signals from masters according to the address and address map. The Qsys interconnect only drives read and write signals to the appropriate Avalon-MM slave, making a chip select unnecessary.

This signal dates from very early microprocessor designs. CPLDs decoded microprocessor addresses and generated chip selects for peripherals that typically were frequently asynchronous. With synchronous systems this signal is typically unnecessary.

#### flush

This signal was removed version 1.2 of the *Avalon Interface Specifications*. Formerly available to masters to clear pending transfers for pipelined reads.

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## **Additional Information**

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**Table B-1: Document Revision History** 

Date	Rev	Changes
December 2015	15.1	Made the following changes:
		<ul> <li>Changed the width of the empty signal from a maximum of 8 bits to a maximum of 5 bits.</li> <li>Improved the definition of the reset_req signal.</li> <li>Removed the readdatavalid signal from the <i>Pipelined Read Transfer with Fixed Latency of Two Cycles</i> timing diagram. This signal is not relevant for fixed latency transfers.</li> <li>Corrected equation defining the empty signal.</li> <li>Made the following changes in the <i>Pipelined Read Transfers with Variable Latency</i> timing diagram:</li> </ul>
		<ul> <li>Moved the deassertion of the read signal to cycle 9</li> <li>Changed waitrequest to don't care in cycle 9.</li> </ul>
March 2015	14.1	Fixed typo in Figure 1-1.

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Date	Rev	Changes
January 2015	14.1	Made the following changes:
		<ul> <li>Clarified address alignment example. The Avalon-MM master and slave interfaces are different widths.</li> <li>Improved discussion of <i>Pipelined read Transfers with Variable Latency</i>. Corrected timing marker 2 which should be exactly on the rising edge of clock.</li> <li>Improved discussion of <i>Pipelined Read Transfer with Fixed Latency of Two Cycles</i>.</li> <li>Clarified use of beatsPerCycle property.</li> <li>Corrected the address range for line-wrapped bursts. The correct address range for a 64-byte burst is 0x0-0x3C, not 0x0-0x1C.</li> <li>Corrected description of the <i>Tristate Conduit Arbitration Timing</i> diagram in the following ways:</li> </ul>
		<ul> <li>The tristate conduit slave asserts grant, not the tristate conduit master.</li> <li>The final grant comes in cycle 9, not cycle 8.</li> <li>Added a <i>Deprecated Signals</i> appendix.</li> <li>Added read response signal.</li> <li>Improved definitions of clock and reset signal types.</li> <li>Corrected definition of clock sink properties.</li> <li>Corrected definition of synchronousEdges for reset source interface.</li> <li>Clarified the Avalon-MM response signal type.</li> <li>Updated definition of empty. The signal must be interpreted emptyWithinPacket is true.</li> <li>Edited for clarity and consistency.</li> </ul>
June 2014	14.0	<ul> <li>Updated the Avalon-MM Signals table, begintransfer, readdatavalid, and readdatavalid_n.</li> <li>Updated the Read and Write Transfers with Waitrequest figure:         <ul> <li>Moved deassertion of write to cycle 6.</li> <li>Moved assertion of readdatavalid and readdata to cycle 4.</li> </ul> </li> <li>Updated the Pipelined Read Transfers with Variable Latency figure:         <ul> <li>Moved assertion of datal to just after cycle 5, and assertion of datal to cycle 6.</li> <li>Moved assertion of readdatavalid to match datal and datal.</li> </ul> </li> </ul>

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Date	Rev	Changes
April 2014	13.01	Corrected Read and Write Transfers with Waitrequest In Avalon Memory-Mapped Interfaces chapter .
May 2013	13.0	<ul> <li>Made the following changes:</li> <li>Minor updates to Avalon Memory-Mapped Interfaces.</li> <li>Minor updates to Avalon Streaming Interfaces.</li> <li>Updated Avalon Conduit Interfaces to describe the signal roles supported by Avalon conduit interfaces.</li> <li>Updated Shared Pin Types figure in the Avalon Tristate Conduit Interface chapter.</li> </ul>
May 2011	11.0	Initial release of the <i>Avalon Interface Specifications</i> supported by Qsys.

### **B.1 How to Contact Altera**

To locate the most up-to-date information about Altera products, refer to the following table.

**Table B-2: Altera Contact Information** 

Contact (1)	Contact Method	Address
Technical support	Website	www.altera.com/support
Tachnical training	Website	www.altera.com/training
Technical training	Email	custrain@altera.com
Product literature	Website	www.altera.com/literature
Nontechnical support (general)	Email	nacomp@altera.com
Nontechnical support (software licensing)	Email	authorization@altera.com

#### Note to Table:

1. You can also contact your local Altera sales office or sales representative.

#### **Related Information**

- www.altera.com/support
- www.altera.com/training
- custrain@altera.com
- www.altera.com/literature
- nacomp@altera.com
- authorization@altera.com

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# **B.2 Typographic Conventions**

The following table shows the typographic conventions this document uses.

Visual Cue	Meaning
Bold Type with Initial Capital Letters	Indicate command names, dialog box titles, dialog box options, and other GUI labels. For example, <b>Save As</b> dialog box. For GUI elements, capitalization matches the GUI.
bold type	Indicates directory names, project names, disk drive names, file names, file names, file name extensions, software utility names, and GUI labels. For example,  \text{qdesigns} \text{ directory, D: drive, and chiptrip.gdf} \text{ file.}
Italic Type with Initial Capital Letters	Indicate document titles. For example, Stratix IV Design Guidelines.
italic type	Indicates variables. For example, n + 1.
	Variable names are enclosed in angle brackets (< >). For example, < file name > and < project name > .pof file.
Initial Capital Letters	Indicate keyboard keys and menu names. For example, the Delete key and the Options menu.
"Subheading Title"	Quotation marks indicate references to sections within a document and titles of Quartus II Help topics. For example, "Typographic Conventions."
Courier type	Indicates signal, port, register, bit, block, and primitive names. For example, data1, tdi, and input. The suffix n denotes an active-low signal. For example, reset_n.
	Indicates command line commands and anything that must be typed exactly as it appears. For example, c:\qdesigns\tutorial\chiptrip.gdf.
	Also indicates sections of an actual file, such as a Report File, references to parts of files (for example, the AHDL keyword SUBDESIGN), and logic function names (for example, TRI).
r	An angled arrow instructs you to press the Enter key.
1., 2., 3., and a., b., c., and so on	Numbered steps indicate a list of items when the sequence of the items is important, such as the steps listed in a procedure.
•••	Bullets indicate a list of items when the sequence of the items is not important.
1	The hand points to information that requires special attention.
h	A question mark directs you to a software help system with related information.
f	The feet direct you to another document or website with related information.
С	A caution calls attention to a condition or possible situation that can damage or destroy the product or your work.

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Visual Cue	Meaning
	A warning calls attention to a condition or possible situation that can cause you injury.

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