

BLG 354E Signals & Systems

Week 7

12.04.2021

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Recap : Systems

- 1) Model physical phenomena : \rightarrow Inverse System Identification
- 2) Implement a desired effect \leftarrow Filtering
on a signal.

- Filtering : removing unwanted parts/features in a signal,
- compression
- feature extraction

LTI : Linear Time Inv. Systems.

Properties of System

- (1) Linearity : S obeys superposition
- (2) Time Inv : S ; does not change behaviour over time,
- (3) Causality : real-time / "online" implementation of systems.
- (4) Stability : today

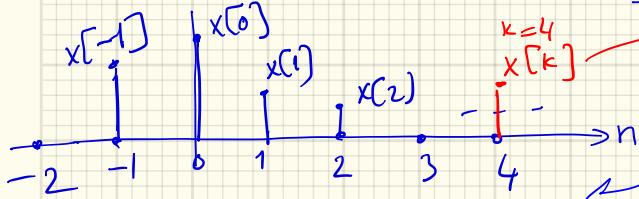
$$x(t) \rightarrow [S] \rightarrow y(t)$$
$$x[n] \rightarrow [S] \rightarrow y[n]$$

↑
System S

LTI systems \rightarrow have a simple I/O relationship depending on impulse-response

For any signal: $x(n)$

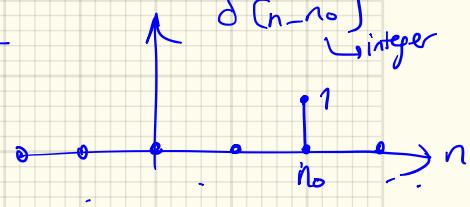
I can represent any DT signal $x[n]$



$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \cdot \delta[n-k]$$

$$\begin{aligned} x[n] &= x[0] \delta[n] + x[1] \delta[n-1] + x[2] \delta[n-2] + \dots \\ &+ x[-1] \delta[n+1] + x[-2] \delta[n+2] + \dots \\ &x[-1000] \delta[n+1000] + \dots \end{aligned}$$

Recall $\delta[n-n_0]$ integer



$$y[n] = T\{x[n]\} = T\left\{ \sum_{k=-\infty}^{\infty} x[k] \delta[n-k] \right\} = \sum_k T\{x[k] \delta[n-k]\}$$

↓
superposition of inputs
Use linearity of the System S

$$\rightarrow y[n] = \sum_{k=-\infty}^{\infty} x[k] T\{\delta[n-k]\}$$

Define impulse response

$$h[n] \triangleq T\{\delta[n]\}$$

we use time-invariance of LTI system S

$$\delta[n] \rightarrow [S] \rightarrow h[n]$$

$$\delta[n-k] \rightarrow [S] \rightarrow h[n-k]$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$\delta[n] \rightarrow [LTI System S] \rightarrow h[n]$$

$$\delta[n] \text{ is input } \rightarrow T\{\delta[n]\} \rightarrow h[n]$$

$h[n]$: impulse response of an LTI system.

$x[n]$ is convolved w/ $h[n]$

this def. defines Convolution operation

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k] = x[n] * h[n]$$

* For an LTI system S: $h[n]$: impulse response of the system S.

Given input $x[n] \rightarrow [S] \rightarrow y[n] = x[n] * h[n]$

$\xrightarrow{x} h[n] \quad \downarrow \quad \xrightarrow{\text{output.}}$

Q. Is the FIR filter (system) LTI ?

yes.
exercise : show
this

$$x[n] \xrightarrow[\{b_k\}]{} y[n] = \sum_{k=-\infty}^{\infty} b_k x[n-k]$$

$\triangleq h[k]$

FIR : filter coefficients

Properties of CONVOLUTION: $* \leftarrow$ convolution operator

① Commutative: $x_1[n] * x_2[n] = x_2[n] * x_1[n]$ ✓

show:

$$\sum_{k=-\infty}^{\infty} x_1[k] x_2[n-k] = \sum_{m=-\infty}^{\infty} x_1[n-m] x_2[m] = x_2 * x_1$$

Diagram illustrating the proof:

- Left sum: $\sum_{k=-\infty}^{\infty} x_1[k] x_2[n-k]$. A red arrow points from $x_1[k]$ to $x_2[n-k]$.
- Right sum: $\sum_{m=-\infty}^{\infty} x_1[n-m] x_2[m]$. A red arrow points from $x_1[n-m]$ to $x_2[m]$.
- A blue curved arrow shows the mapping $n-k=m \rightarrow m=-\infty$, with $k=n-m$ indicated below it.
- A red checkmark is at the end of the rightmost sum.

② Associative: $x_1[n] * (x_2[n] * x_3[n]) = (x_1[n] * x_2[n]) * x_3[n]$ ✓

(3) Distributive Over Addition: $x[n] * (\underbrace{h_1[n] + h_2[n]}_{\downarrow}) = x[n] * h_1[n] + x[n] * h_2[n]$

(4) $*$ operator
Has an identity element, $\delta[n]$: unit impulse sequence.

$x[n] * \delta[n] = x[n]$

show $\sum_{k=-\infty}^{\infty} x(k) \underbrace{\delta(n-k)}_{n=k} = x[n] * \delta[n] = \cancel{x(n)}$

Q: $x[n] * \delta[n-1] = ?$

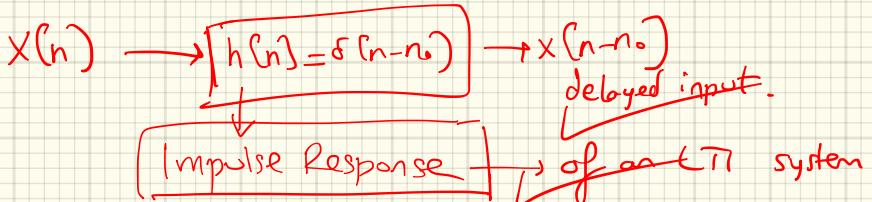
$$\left(\sum_{k=-\infty}^{\infty} x(k) \underbrace{\delta[(n-1)-k]}_{k=n-1} \right) = \underline{x[n-1]}$$

$$x(n) \rightarrow \boxed{\delta[n-1]} \rightarrow y(n) = x[n-1]$$

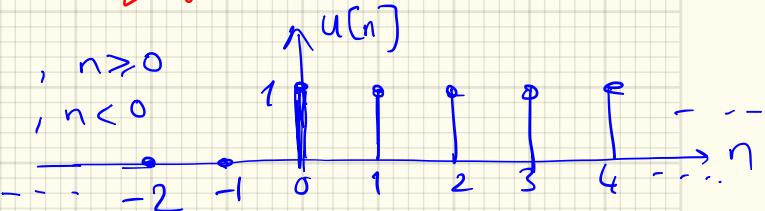
$h[n] = \delta[n-1]$ Delay system by 1 sample.

(5)

$$\delta(n-n_0) * x[n] = x[n-n_0]$$

Step Sequence:

$$u[n] = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$$



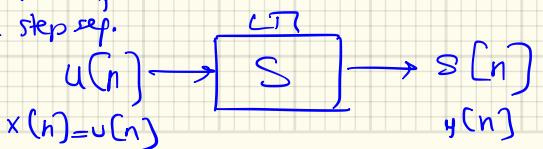
$$u[n] = \delta(n) + \delta(n-1) + \delta(n-2) + \dots$$

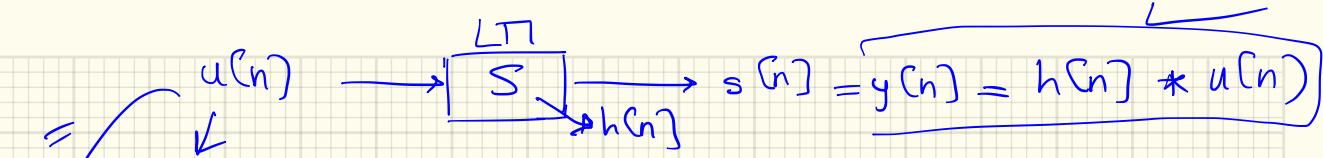
$$u[n] = \sum_{k=0}^{\infty} \delta(n-k) \Rightarrow u[n] = \sum_{l=-\infty}^{n} \delta(l)$$

$\hookrightarrow l=n-k$
change of
var

STEP RESPONSE
of an LTI System :

Set the input to
a step seq.





$$x[n] = \sum_{k=0}^{\infty} \delta[n-k] \xrightarrow{[S]} y[n] = h[n] * x[n]$$

$$y[n] = h[n] * \left(\sum_{k=0}^{\infty} \delta[n-k] \right) = \sum_{k=0}^{\infty} (h[n] * \delta[n-k])$$

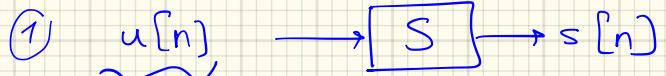
$*$: distrib. over addition

$$\rightarrow \text{Step response: } s[n] = y[n] = \sum_{k=0}^{\infty} h[n-k]$$

* Note: $\underbrace{\delta[n]}_{\substack{\text{another} \\ \text{property}}} * \underbrace{\delta[n-n_0]}_{\substack{\text{of convolution.}}} = \delta[n-n_0]$

$$\underbrace{\delta[n-n_1]}_{\substack{\text{.}}} * \underbrace{\delta[n-n_0]}_{\substack{\text{.}}} = \delta[n - (n_1 + n_0)]$$

Ex: Given step response $s[n] \rightarrow h[n] = ?$

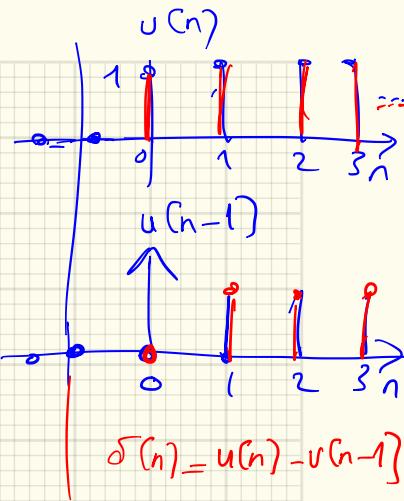


Q: $\underbrace{\delta[n]}_{\text{write } \delta[n]} \rightarrow [S] \rightarrow h[n] = ?$

use superposition
of inputs.

$u[n-1] \xrightarrow{\text{Time Inv.}} [S] \rightarrow s[n-1]$

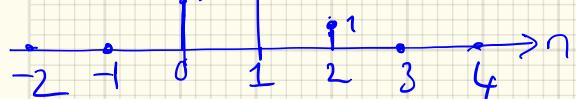
$\underbrace{u[n] - u[n-1]}_{\text{linearity.}} = \underbrace{\delta[n]}_{\text{in } [S]} \rightarrow [S] \rightarrow h[n] = \underbrace{s[n]}_{\text{in } [S]} - \underbrace{s[n-1]}_{\text{in } [S]}$



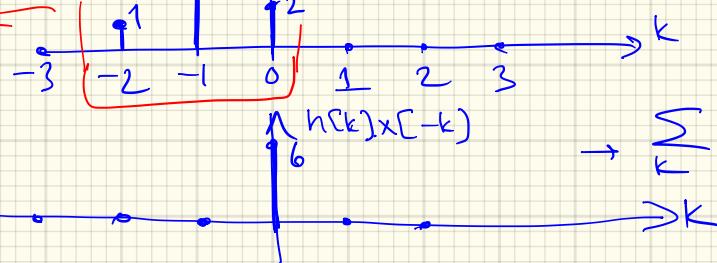
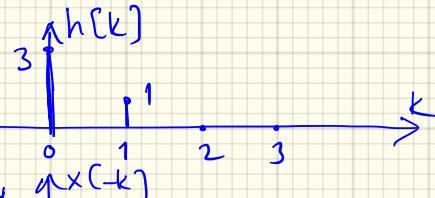
Computation of Convolution:

Given $x[n]$ (input) \times $h[n]$ (impulse response)

$$\text{Ex: } x[n] = 2\delta[n] + 4\delta[n-1] + \delta[n-2]$$



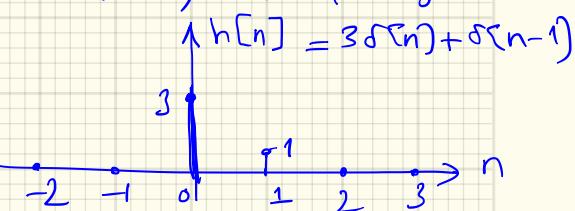
$$y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$



$$\rightarrow \sum_k h(k)x(-k) = 6 = y[n=0]$$

$$x[n] \rightarrow \boxed{S_{h[n]}} \rightarrow x[n] * h[n] = y[n]$$

\rightarrow compute $y[n]$.

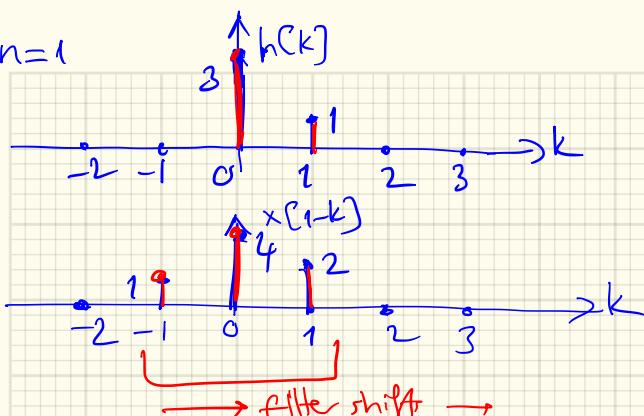
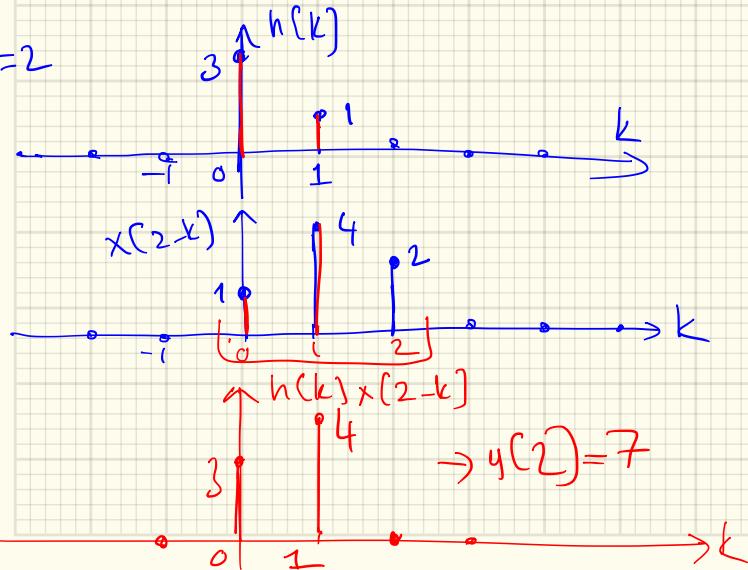
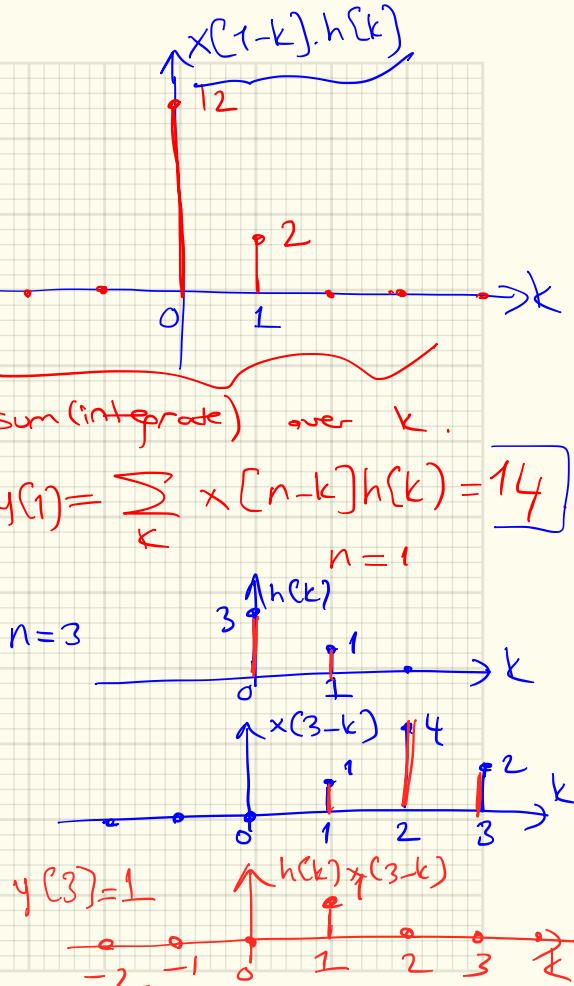


$$y(0) = \sum_{k=-\infty}^{\infty} h(k)x(-k)$$

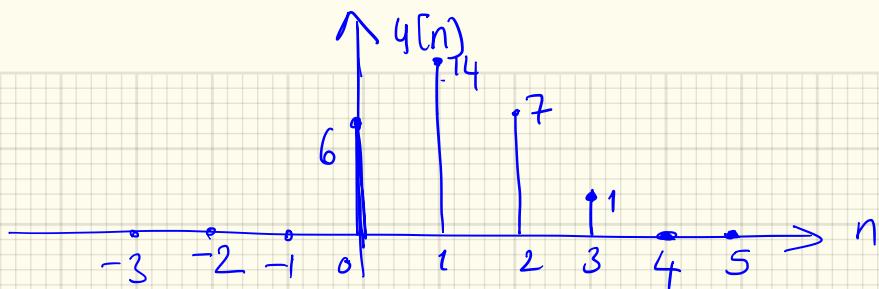
$$y(1) = \sum_k h(k)x(1-k)$$

$$y[0]=6$$

signal
slide this flipped filter
over the other signal

$n=1$  $n=2$  $n=3$ 

Result
→



Due to commutativity of convolution:

→ We would arrive at the same result if we worked w/
up

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$



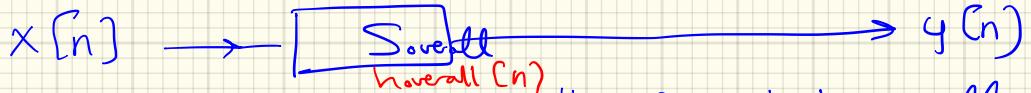
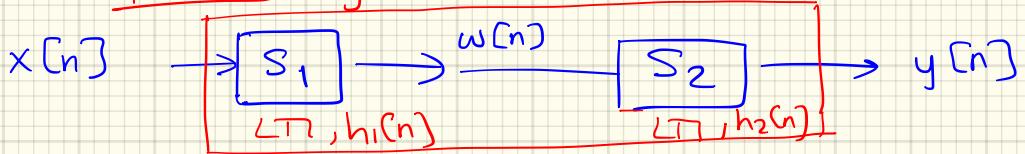
Another way for DT convolution:

$$\begin{aligned} y(n) &= (x(n) = 2\delta[n] + 4\delta[n-1] + \delta[n-2]) * (h[n] = 3\delta[n] + \delta[n-1]) \\ &= 6\delta[n] + \underbrace{(2\delta[n-1] + 3\delta[n-2])}_{+} + \underbrace{(2\delta[n-1] + 4\delta[n-2] + \delta[n-3])}_{+} \\ y(n) &= 6\delta[n] + 14\delta[n-1] + 7\delta[n-2] + \delta[n-3] \end{aligned}$$



The same result as in the 1st way of convolution: using a graph method

Cascaded LTI Systems :



Q: What is the equivalent overall system?

$$y[n] = h_2(n) * (x[n] * h_1(n)) = h_2(n) * (h_1(n) * x[n])$$

(commutative)

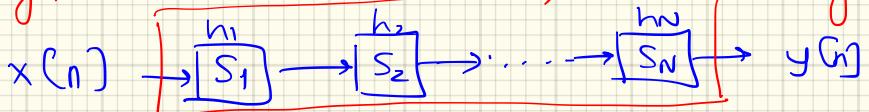
$$y[n] = (h_2(n) * h_1(n)) * x[n]$$

(associative)

$\underbrace{h_{\text{overall}}[n]}_{= h_1[n] * h_2[n]}$

Overall System
impulse

response for Cascaded (in series) connection of LTI systems

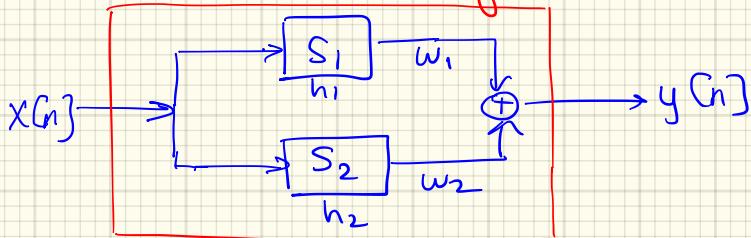


$$x[n] \xrightarrow{\text{h overall}} y[n]$$

$\underbrace{h_{\text{overall}}[n]}_{= h_1[n] * \dots * h_N[n]}$

Valid
For N
LTI systems

Parallel Connection of LTI Systems:



$h[n]$: overall system S

$h[n] = ?$

i.e. $h_1[n]$, $h_2[n]$

$$\begin{aligned} y[n] &= \underbrace{(x[n] * h_1[n])}_{w_1(n)} + \underbrace{(x[n] * h_2[n])}_{w_2(n)} \\ &= x[n] * \underbrace{(h_1[n] + h_2[n])}_{h[n]} \end{aligned}$$

$$y[n] = x[n] * h[n]$$

addition



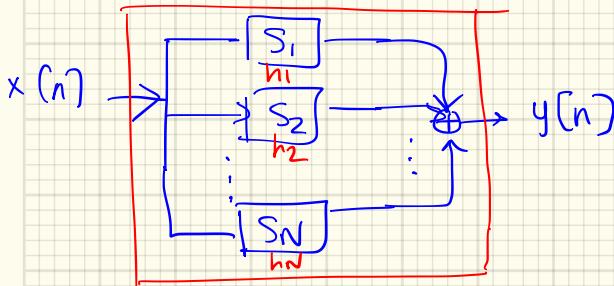
Overall system in Parallel connection has : $h[n] = h_1[n] + h_2[n]$

((((Cascaded)))) : $h[n] = h_1[n] * h_2[n]$



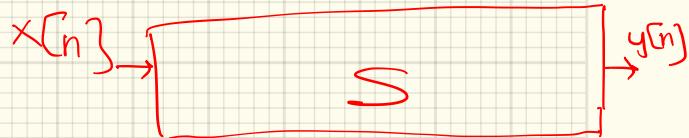
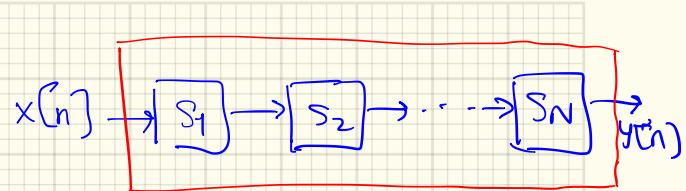
Convolution

Extension to N systems :



$$h[n] = \sum_{k=1}^N h_k[n] : \text{Addition of all impulse responses}$$

Parallel System :

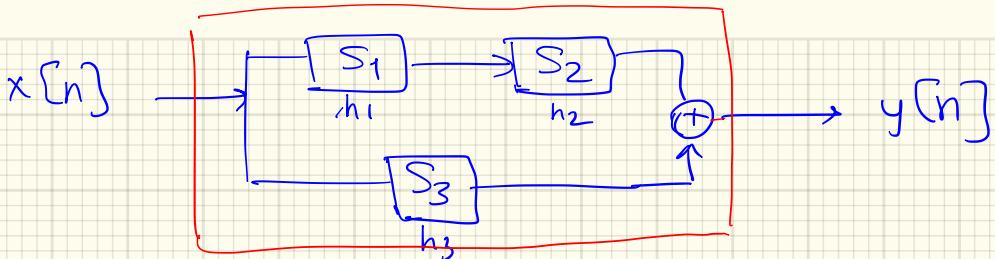


$$h[n] = \underbrace{h_1[n] * h_2[n] * \dots * h_N[n]}$$

Cascaded System:

Convolution of all impulse responses

Ex:



$$h[n] = ? \left(h_1[n] * h_2[n] \right) + h_3[n]$$

You can think of many different combinations and calculate their overall system response.

Recall Causality of a System :

General Defn: A system is causal if the output $y(n)$ only depends on current & past values of $x(n)$.

ex: $y(n) = x(n) + (x(n-1))^2$: Is this system causal?

To relation
of a system

$$\underbrace{y[2]}_{n=2} = x[2] + (x[1])^2 \quad \text{Yes} \checkmark$$

Checking Causality for LTI systems: → check the impulse response.

$$y(n) = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

for $k \geq 0$

$y(n) \rightarrow$ depends $x(n)$ & $x[n-k]$

past values

for $k < 0$

$x[n+k']$: future values of $x(n)$.

Def (Causality): An LTI system is causal iff (if and only if)
 $h[n] = 0$ for $n < 0$.

Recall General system of const. Linear Coeff. Difference Equations

→ an important class of LTI systems:

I/O relation: $\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^m b_k x[n-k]$

$$N=0 \rightarrow a_0 = 1 : \boxed{y[n] = \sum_{k=0}^m b_k x[n-k]} \quad \begin{array}{l} \text{FIR System} \\ \sum b_k \\ \text{Finite } \# b_k \\ h[k] \end{array}$$

ex FIR filter:
 $y[n] = \delta(n) - \frac{3}{2}\delta(n-1)$
 $\hookrightarrow h(n) = \frac{1}{2} \delta(n) \Rightarrow$

$$\text{ex: } y[n] = \frac{1}{2}x[n] - \frac{3}{2}x[n-1] \Rightarrow b_0 = \frac{1}{2} = h[0]$$

$$b_1 = -\frac{3}{2} = h[1]$$

Extra: Infinite Impulse Response (IIR) :

I/O: $y[n] - \frac{1}{2}y[n-1] = x[n] \rightarrow y[n] = \frac{1}{2}y[n-1] + x[n]$

Find impulse response: Let $x[n] = \delta(n)$

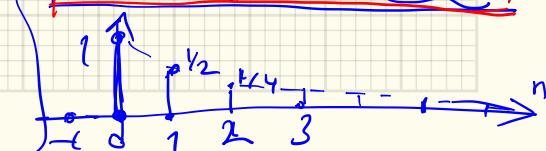
System is at rest $h[-1] = 0$
 $y[-1] = 0$

$$h[0] = h[-1] + x[0] = 0 + 1 = 1 = y[0]$$

$$h[1] = \frac{1}{2}h[0] + x[1] = \frac{1}{2}$$

$$h[2] = \frac{1}{2}h[1] + x[2] = \frac{1}{4} ;$$

$$\boxed{h[n] = \left(\frac{1}{2}\right)^n \cdot u[n]}$$



Ex: Is this \sum system causal? $h[n] = \left(\frac{1}{2}\right)^n u[n]$

Check whether $h[n] = 0$ for $n < 0$

? ✓

This IIR filter is causal.

(4)th property of Systems

STABILITY

General defn.
robustness against
perturbations

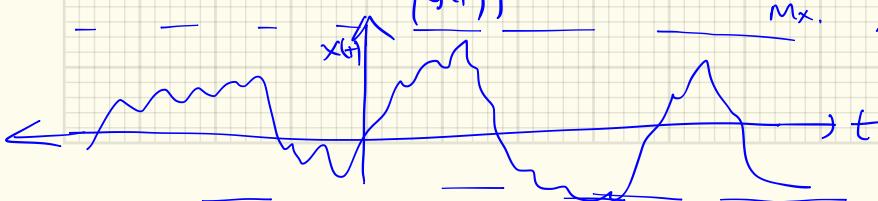
Def: (Bounded Input Bounded Output \Rightarrow BIBO) Stability:

Def: System S is stable iff every bounded input produces a bounded output.

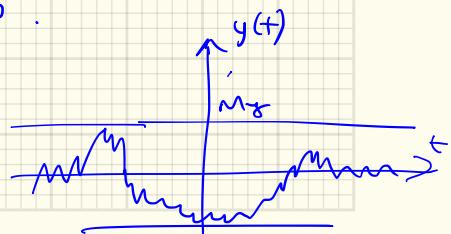
In math terms:

$$\text{If } |x(n)| \leq M_x, \quad M_x < \infty.$$

then $|y[n]| \leq M_y, \quad M_y < \infty.$



BIBO
 \Rightarrow



If the system is LTI : Stability is defined to the impulse response

Let $|x[n]| \leq M_x$

$$|y[n]| = \left| \sum_{k=-\infty}^{\infty} h[k] x[n-k] \right|$$

$$\leq \sum_k |h[k] x[n-k]|$$

$$\leq \sum_k |h[k]| \cdot \underbrace{|x[n-k]|}_{\text{a const.}}$$

$$\leq \sum_k |h[k]| \cdot \underbrace{M_x}_{= \text{a const.}}$$

if this is bounded

$$|y[n]| \leq M_y. \quad , \quad M_y = M_x \sum |h[k]|$$

Def (Stability for LTI system) : LTI system is stable if $\sum_{k=-\infty}^{\infty} |h[k]| < \infty$

Ex: $y[n] = x[-n]$ → How would you check (BIBO) stability?

Is this system LTI? No.

$$\text{Let } |x(n)| < M_x, M_x < \infty.$$

$$|y(n)| = |x[-n]| < M_x \quad \checkmark \quad \text{stable.}$$

Ex: $y(n) = \underbrace{\frac{1}{x[n]}}$: Is this LTI? No

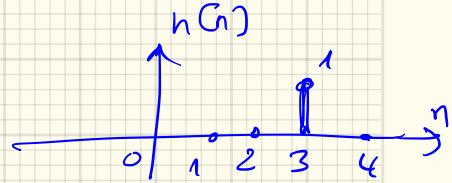
Is this stable? No \checkmark .

Let $x(n) = 0$ a finite $y(n) \rightarrow \infty$. Not a stable system.

Ex: $y(n) = x(n-3)$: Delay by 3 system.

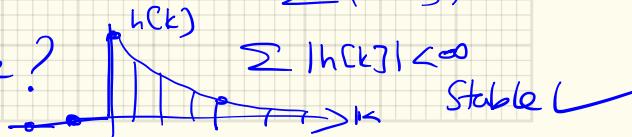
LTI system: $h[n] = \delta[n-3]$

Stable \checkmark



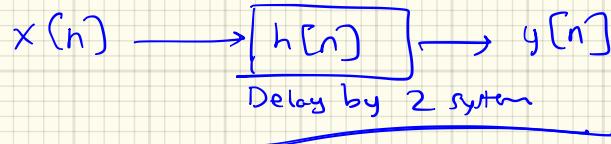
$$\sum (h[n]) = 1.$$

Ex: $h[n] = \left(\frac{1}{3}\right)^n u[n]$: Is this stable?
IIR:



$$\sum |h[k]| < \infty \quad \text{Stable} \quad \checkmark$$

Ex: Given $h[n] = \delta[n-2]$ & } Input ?
 Output $y[n] = u[n-3] - u[n-6]$ } $x[n]$



$$x[n] = u[n-1] - u[n-4]$$

Ex: Given LTI system: $u[n] \rightarrow [S] \rightarrow \delta[n] + 2\delta[n-1] - \delta[n-2]$

$$x[n] = 3u[n] - 2u[n-4] \rightarrow [S] ? \quad y[n] = ?$$

Use LTI properties

$$3u[n] \rightarrow [S] \rightarrow 3\delta[n] + 6\delta[n-1] - 3\delta[n-2]$$

$$-2u[n-4] \rightarrow [S] \rightarrow -2\delta[n-4] - 4\delta[n-5] + 2\delta[n-6]$$

+

$$x[n] \rightarrow [S] \rightarrow 3\delta[n] + 6\delta[n-1] - 3\delta[n-2] - 2\delta[n-4] - 4\delta[n-5] + 2\delta[n-6]$$

Homework exercise:

$$y[n] = \sum_{k=0}^{6} B^k x[n-k], \quad B \text{ a real number.}$$

IIo

$$x[n] \xrightarrow{\boxed{S}} y[n]$$

$$\text{Let } x[n] = \delta(n) - B \delta(n-1) \rightarrow \text{Calculate } y[n].$$

$$h[n] = ? \text{ calculate } h.$$

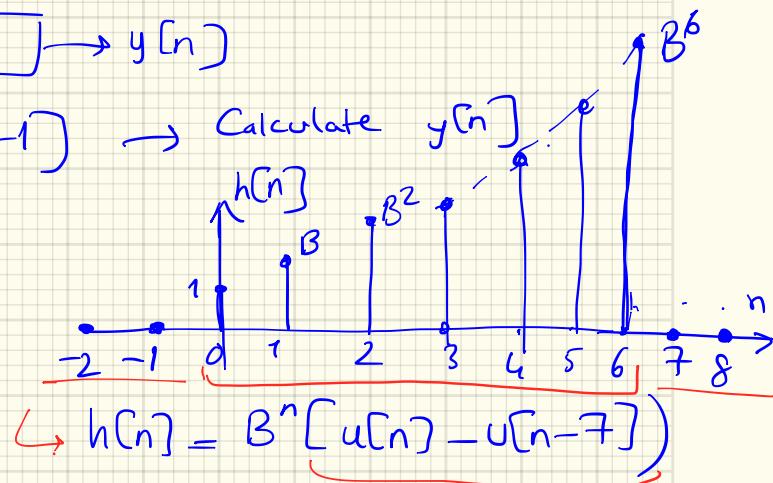
FIR filter.

.

:

:

!



Result. $y[n] = \sum_{k=0}^6 B^k \delta(n) - B^7 \delta(n-7)$