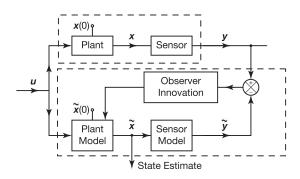


ECE 68000: MODERN AUTOMATIC CONTROL

Professor Stan Żak

Observers for Systems With Unknown Inputs

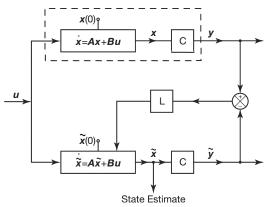
Closed-loop observer



• Luenberger's Innovation to obtain the closed-loop observer

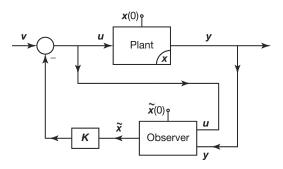
$$\dot{ ilde{x}} = A ilde{x} + Bu + L(y - ilde{y})$$

Luenberger's closed-loop observer



- Luenberger's observer, $\dot{\tilde{x}} = A\tilde{x} + Bu + L(y \tilde{y})$
- Observation error dynamics, $(\dot{x} \dot{\tilde{x}}) = (A LC)(x \tilde{x})$

Combined observer-controller compensator



- Works well for systems without uncertainties
- What about systems with uncertainties?

Plant Model

Standard linear dynamical system model:

$$\begin{array}{rcl}
\dot{x} & = & Ax + Bu \\
y & = & Cx,
\end{array}$$

where $\boldsymbol{B} \in \mathbb{R}^{n \times m}$, $\boldsymbol{C} \in \mathbb{R}^{p \times n}$

- Parameters A, B, C are known
- m_1 of inputs are known and $m_2 = m m_1$ are unknown
- Re-arrange the order of the inputs if necessary, partition the input matrix \boldsymbol{B} corresponding to the known, \boldsymbol{u}_1 , and unknown inputs, \boldsymbol{u}_2 , as

$$\boldsymbol{B} = \left[\begin{array}{cc} \boldsymbol{B}_1 & \boldsymbol{B}_2 \end{array} \right],$$

with $\boldsymbol{B}_1 \in \mathbb{R}^{n \times m_1}$ and $\boldsymbol{B}_2 \in \mathbb{R}^{n \times m_2}$ and

$$oldsymbol{u} = \left[egin{array}{c} oldsymbol{u}_1 \ oldsymbol{u}_2 \end{array}
ight]$$

System Model—Contd.

The system model

$$egin{array}{lll} \dot{m{x}} &=& m{A}m{x} + m{B}_1m{u}_1 + m{B}_2m{u}_2 \ m{y} &=& m{C}m{x} \end{array}$$

- The vector function u_2 may also model lumped uncertainties or nonlinearities in the plant
- ullet Similar notation as in Basile and Marro, where $oldsymbol{u}_2$ is called the disturbance vector
- The output matrix is $C \in \mathbb{R}^{p \times n}$
- The pair (A, C) detectable

G. Basile and G. Marro, On the observability of linear, time-invariant systems with unknown inputs, Journal of Optimization Theory and Applications, Vol. 3, No. 6, pp. 410–415, Nov. 1969

Second page of the 1969 Basile and Marro's paper

We deal with a linear, purely dynamical, time-invariant system described by the equations

$$\dot{x} = Ax + B_1 u_1 + B_2 u_2 \tag{1}$$

$$v = Cx$$
 (2)

where $x \in R^n$ is the state vector, $u_1 \in R^n$ is the control vector, $u_2 \in R^1$ is the disturbance vector, $y \in R^n$ is the output vector, and A, P_1, B_2, C are real, constant matrices of proper sizes. We call $\mathcal{F}_1 = \mathcal{B}(B_2)$ the subspace of control actions and $\mathcal{F}_2 = \mathcal{B}(B_2)$ the subspace of disturbance actions.

It is well known that, in the particular case where $B_1 \neq 0, B_2 = 0$, from the observation of input and output functions in a finite interval of time, it is possible to recognize the orthogonal projection of the state on the least subspace which is invariant under A^T and contains $\mathcal{B}(C^T)$. The word least is justified because the intersection of two invariants is an invariant. This subspace is sometimes called observability subspace and its orthogonal complement unobservability subspace.

In this particular case, when the input functions are completely known, the observation of the system (1)-(2) reduces to the observation of the corresponding autonomous system; that is, since

$$y(t) = C\Phi(t, 0) x_0 + C \int_{t}^{t} \Phi(t, \tau) B_1 u_1(\tau) d\tau$$
 (3)

where $\Phi(t, \tau)$ is the state-transition matrix, it is possible to determine by a simple subtraction the output functions of the corresponding autonomous system, namely, the zero-input output functions.

By similar reasoning, the general case in which a part of the input is known and a part is unknown can be reduced to the case of completely unknown input. Thus, it is sufficient to consider only this last case. In the next section, we state a theorem that provides the observability subspace as the least conditioned invariant under the matrix A^r , with respect to the subspace \mathscr{F}_{r-1}^+ , containing $\mathscr{B}(C^r)$, and which includes the previous results, corresponding to $B_r = 0$.

2. Observability Subspace for Systems with Unknown Inputs

First, we recall some definitions and results given in a previous paper (Ref. 6) which provide a background for the analysis presented here. Consider

Projection Operator UIO—Idea

 \bullet Decompose the state x as

$$egin{array}{lll} oldsymbol{x} &=& oldsymbol{x} - oldsymbol{M} oldsymbol{y} + oldsymbol{M} oldsymbol{y} \ &=& (oldsymbol{I} - oldsymbol{M} oldsymbol{C}) oldsymbol{x} + oldsymbol{M} oldsymbol{y}, \ &=& (oldsymbol{I} - oldsymbol{M} oldsymbol{C}) oldsymbol{x} + oldsymbol{M} oldsymbol{y}, \end{array}$$

where $M: n \times p$ real matrix to be determined

• q = (I - MC)x: Unknown part of the decomposition

S. Hui and S. H. Żak, Observer design for systems with unknown inputs, International Journal of Applied Mathematics and Computer Science, Vol. 15, No. 4, pp. 431–446, 2005

Projection Operator UIO—Decomposed Dynamics

• Some manipulations:

$$egin{array}{lll} \dot{m{q}} &=& (m{I}-m{M}m{C})\dot{m{x}} \ &=& (m{I}-m{M}m{C})(m{A}m{x}+m{B}_1m{u}_1+m{B}_2m{u}_2) \ &=& (m{I}-m{M}m{C})(m{A}m{x}+m{B}_1m{u}_1)+(m{I}-m{M}m{C})m{B}_2m{u}_2 \end{array}$$

• Recall: x = q + My

$$\dot{q} = (I - MC)(Aq + AMy + B_1u_1) + (I - MC)B_2u_2$$

- Choose M to make $(I MC)B_2 = O$
- Then

$$\boxed{\dot{oldsymbol{q}} = (oldsymbol{I} - oldsymbol{M}oldsymbol{C})(oldsymbol{A}oldsymbol{q} + oldsymbol{A}oldsymbol{M}oldsymbol{y} + oldsymbol{B}_1oldsymbol{u}_1)}$$

• Important: u_1 and y are known

Projection Operator UIO—The Rank Condition

• Need:

$$(\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{B}_2 = \boldsymbol{O}$$

• Linear Algebra:

$$\operatorname{rank} (\boldsymbol{MCB}_2) \leq \operatorname{rank} (\boldsymbol{CB}_2) \leq \operatorname{rank} (\boldsymbol{B}_2)$$

• Necessary and Sufficient Condition:

$$rank (\mathbf{C}\mathbf{B}_2) = rank(\mathbf{B}_2)$$

• Implication: At least as many independent outputs as unknown inputs for the method to work

Projection Operator UIO Dynamics

• If we know q and the initial condition

$$\boldsymbol{q}(0) = (\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{x}(0),$$

then

$$x = q + MCx = q + My$$

is known for all $t \geq 0$

ullet Indeed, integrate both sides of $\dot{q}=(I-MC)\dot{x}$ to obtain

$$\boldsymbol{q}(t)\text{-}\boldsymbol{q}(0)\text{=}\;(\boldsymbol{I}\text{-}\boldsymbol{M}\boldsymbol{C})(\boldsymbol{x}(t)\text{-}\boldsymbol{x}(0))$$

Hence

$$\boldsymbol{q}(t) = (\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{x}(t) - (\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{x}(0) + \boldsymbol{q}(0)$$

• If q(0) = (I - MC)x(0), then

$$x = q + MCx = q + My$$

is known for all t > 0

Projection Operator UIO Dynamics—Contd.

- But we do not know $\boldsymbol{x}(0)$
- We have

$$\boldsymbol{q}(t) = (\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{x}(t) - (\boldsymbol{I} - \boldsymbol{M}\boldsymbol{C})\boldsymbol{x}(0) + \boldsymbol{q}(0)$$

• So we only get an approximation

$$| ilde{m{x}} = m{q} + m{M}m{y}|$$

where q is obtained from

$$\left|\dot{oldsymbol{q}}=(oldsymbol{I}-oldsymbol{M}oldsymbol{C})(oldsymbol{A}oldsymbol{q}+oldsymbol{A}oldsymbol{M}oldsymbol{y}+oldsymbol{B}_1oldsymbol{u}_1)
ight|$$

Open-Loop UIO Analysis

- Let $e(t) = x(t) \tilde{x}(t)$ be the estimation error
- Recall $(I MC)B_2 = O$ and y = Cx

Then we have

we have
$$\frac{de}{dt} = \frac{d}{dt}(x - \tilde{x})$$

$$= \frac{d}{dt}(x - q - MCx)$$

$$= \frac{d}{dt}((I - MC)x - q)$$

$$= (I - MC)(Ax + B_1u_1 + B_2u_2)$$

$$-(I - MC)(Aq + AMy + B_1u_1)$$

$$= (I - MC)(Ax + B_1u_1) + (I - MC)B_2u_2$$

$$-(I - MC)(Aq + AMCx + B_1u_1)$$

$$= (I - MC)A(x - q - MCx)$$

$$= (I - MC)Ae$$

Closed-Loop UIO

• We add the innovation term to obtain the closed-loop UIO:

$$egin{array}{lll} \dot{m{q}} &=& (m{I} - m{M} m{C}) ((m{A} m{q} + m{A} m{M} m{y} + m{B}_1 m{u}_1) + m{L} (m{y} - m{ ilde{y}})) \ &=& (m{I} - m{M} m{C}) ((m{A} m{q} + m{A} m{M} m{y} + m{B}_1 m{u}_1) \ &=& (m{I} - m{M} m{C}) ((m{A} m{q} + m{A} m{M} m{y} + m{B}_1 m{u}_1) \ &+ m{L} m{C} (m{x} - m{q} - m{M} m{y})) \end{array}$$

• State estimate is

$$\tilde{m{x}} = m{q} + m{M}m{y}$$

Closed-Loop UIO Analysis

- Let $\boldsymbol{e} = \boldsymbol{x} \tilde{\boldsymbol{x}}$
- We will show that

$$\dot{e} = (I - MC)(A - LC)e$$

and $e(t) \to 0$ as $t \to \infty$ under mild conditions

- Note that (A-LC) asymptotically stable does not guarantee that (I-MC)(A-LC) is asymptotically stable
- It is possible for a product of a projection matrix and an asymptotically stable matrix to be unstable

Example

• Let

$$\mathbf{\Pi} = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \text{ and } \mathbf{A} = \begin{bmatrix} 1 & -3 \\ 3 & -2 \end{bmatrix}$$

- \bullet **A** is asymptotically stable
- ΠA is unstable
- The system $\dot{x} = Ax$ restricted to the range of Π is governed by $\dot{z} = z$, which is also unstable