$$q_1 = \chi$$

$$q_2 = \theta$$

$$T = \frac{1}{2} m \left[\dot{y}^2 + \dot{l}^2 \dot{\theta}^2 \cos^2 \theta + 2 l \dot{\theta} \dot{x} \cos \theta + l^2 \dot{\theta}^2 \sin^2 \theta \right]$$

$$+ \underbrace{M \dot{x}^2}_{2}$$

$$\frac{22}{29} = 0 \qquad \frac{22}{29} = m \times + M \times + locosom$$

Q= \(\varphi\) cheial.
$$2\varphi_{10}$$
 $F_{M}=U$

$$\frac{2\vec{r}_{M/o}}{2q_1} = 1$$

$$\frac{2\vec{r}_{M/o}}{2q_2} = 0$$

$$\dot{o} = \chi$$

$$\frac{d}{dt}(m\dot{x} + M\dot{x} + ml\dot{\theta}\cos\theta) = U$$

$$U = M\ddot{x} + M\ddot{x} + ml\ddot{\theta}\cos\theta - ml\dot{\theta}^{2}\sin\theta \qquad (1)$$

$$\dot{s} = \theta$$

$$\frac{d}{dt}(ml^{2}\dot{\theta} + ml\dot{x}\cos\theta) - (-ml\dot{\theta}\dot{x}\sin\theta + mgl\sin\theta) = 0$$

$$ml^{2}\ddot{\theta} + ml\ddot{x}\cos\theta - ml\dot{x}\cos\theta + ml\dot{x}\sin\theta - msl\sin\theta = 0$$

$$ml^{2}\ddot{\theta} + ml\ddot{x}\cos\theta - msl\sin\theta = 0 \qquad (2)$$

$$\ddot{\theta} = -\ddot{\chi}\cos\theta + g\sin\theta$$

$$\dot{\chi} = U - ml\ddot{\theta}\cos\theta + ml\dot{\theta}^{2}\sin\theta$$

$$(m+M)$$

$$\dot{\chi} = U + m \dot{\chi} \cos \theta - m \cos \theta \sin \theta + m l \dot{\theta} \sin \theta$$
 $(m+m)$

$$\ddot{\chi} = U + mg \cos \theta \sin \theta + ml \theta^{2} \sin \theta$$
 (3)

Non-linear state space hodel:

m+M-mcos2(x3)

L(m+M-m cos2(x3))

$$A = \begin{pmatrix} \frac{2f}{2x_1} & \frac{1}{2x_2} & \frac{2f}{2x_1} & \frac{1}{2x_2} & \frac{2f}{2x_2} & \frac{1}{2x_2} & \frac{1}{2x$$

$$B = \begin{vmatrix} \frac{\partial f}{\partial v} & (x^*, v^*) \\ \frac{\partial f}{\partial v} & (x^*, v^*) \\ \frac{\partial f}{\partial v} & (x^*, v^*) \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ y_3 \end{vmatrix} = \begin{vmatrix} x_1 \\ x_3 \\ y_4 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix} = \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

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$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$= \begin{vmatrix} x_1 \\ x_2 \\ x_3 \\ y_5 \end{vmatrix}$$

$$A = \begin{pmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 11 & 0 \end{pmatrix} \qquad B = \begin{pmatrix} 0 \\ 10 \\ 0 \\ -10 \\ 0 \\ -10 \\ 0 \end{pmatrix}$$

$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \qquad D = \begin{pmatrix} 0 \\ 0 \end{pmatrix}$$

rank[B AB AB AB] = 4

Desired poles:
$$S_1 = -1$$

 $S_2 = -4$
 $S_3 = -4 + 1$
 $S_4 = -4 - 1$

$$8u = -1.8x$$
 $8x = X-x*$
 $8x = X-x*$

$$U = U^* + 8U = (x_3)$$
 $y = y^* + 8y = (x_3)$

$$8y = \begin{pmatrix} x_1 \\ \chi_3 \end{pmatrix} - \begin{pmatrix} x_1^* \\ \chi_3^* \end{pmatrix} = \begin{pmatrix} x_1 \\ \chi_3 \end{pmatrix} = y$$

$$\frac{2}{2} = (A - LC) \frac{1}{2} + BU + Ly$$

$$\frac{2}{2} = (A - LC) \frac{1}{2} - BK^{2} + Ly$$

$$\frac{1}{2} = (A - LC - BK) \frac{1}{2} + Ly$$

$$\frac{1}{2} = (A - LC - BK) \frac{1}{2} + Ly$$

$$\frac{1}{2} = (A - LC - BK) \frac{1}{2} + Ly$$

$$\frac{1}{2} = (A - LC - BK) \frac{1}{2} + Ly$$

$$\frac{1}{2} = (A - LC - BK) \frac{1}{2} + Ly$$

$$\frac{1}{2} = (A - LC) \frac{1}{2} + C$$

$$\frac{1}{2}$$

#6)
$$U = -Kz + \Gamma$$
 $\dot{x} = A \times + B0 = A \times - BNz + B\Gamma$
 $\dot{z} = Az + B0 + L(y = \hat{y}) = (A - BN - LC)z + LCX + B\Gamma$
 $\dot{c}z$
 $\dot{c$

$$\frac{y(s)}{R(s)} = \begin{pmatrix} \frac{y_1(s)}{R(s)} \\ \frac{y_2(s)}{R(s)} \end{pmatrix} = \begin{pmatrix} \frac{x_1(s)}{R(s)} \\ \frac{x_2(s)}{R(s)} \end{pmatrix}$$

$$\frac{\sqrt{1(s)}}{R(s)} = \frac{0.1 s^2 - 1}{s^4 + 13 s^3 + 61 s^2 + 117 s + 68}$$

$$\frac{Y_{2(s)}}{r(cs)} = \frac{-0.1s^2}{s^4 + 13s^3 + 61s + 117s + 68}$$

Contents

- Parameters
- Part 2 Linearize Model about Operating Point x = 0, u = 0
- Part 3 Stabilizing Feedback Controller for Linearized Model, u = -Kx
- Part 4 Feedback Controller on Nonlinear System
- Part 5 State Observer Design with Linear System
- Part 6 Closed Loop Transfer Function
- Part 7 State Observer with Non-Linear System
- Functions

```
clear
close all
clc
```

Parameters

```
= 1;
                   % [kg]
m
1
      = 1;
                    % [m]
      = 10;
                    % [m/s^2]
g
      = 10;
                    % [kg]
x_star = zeros(4,1); % State Operating Point
               % Input Operating Point
u_star = 0;
dt = .005;
                   % Time Step
time = (0:dt:5)'; % Time
```

Part 2 - Linearize Model about Operating Point x = 0, u = 0

```
syms x1 x2 x3 x4 u s real
% Create Symbolic Set of Equations
       = [x1;x2;x3;x4];
xdot = NonlinearCartPendulum([],x,u,m,M,g,1);
% Symbolic Jacobian Matrices
A_sym
       = [diff(xdot(1),x1), diff(xdot(1),x2), diff(xdot(1),x3), diff(xdot(1),x4);...
           diff(xdot(2),x1), diff(xdot(2),x2), diff(xdot(2),x3), diff(xdot(2),x4);...
           diff(xdot(3),x1), diff(xdot(3),x2), diff(xdot(3),x3), diff(xdot(3),x4);...
           diff(xdot(4),x1), diff(xdot(4),x2), diff(xdot(4),x3), diff(xdot(4),x4)];
B_{\underline{sym}} = [diff(xdot(1),u); diff(xdot(2),u); diff(xdot(3),u); diff(xdot(4),u)];
disp('--Part 2 -----')
disp(' ')
disp('Linearized State Space Model')
% Linearized State Space Model
       = double(subs(A_sym,[x1;x2;x3;x4;u],[x_star;u_star]))
В
       = double(subs(B_sym,[x1;x2;x3;x4;u],[x_star;u_star]))
C
       = [1 0 0 0; 0 0 1 0] % y = [x1;x3]
       = zeros(2,1)
% Verify System is Controllable and Observable
CO
    = ctrb(A,B);
0bs
       = obsv(A,C);
if rank(CO) == length(A)
```

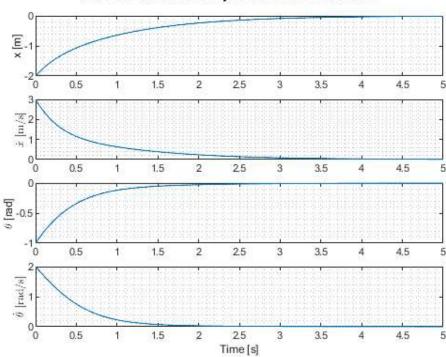
```
disp('System is Controllable');
end

if rank(Obs) == length(A)
    disp('System is Observable');
end
```

Part 3 Stabilizing Feedback Controller for Linearized Model, u = -Kx

```
del\_xdot = A*del\_x + B*del\_u, del\_u = -K*del\_x, u = u\_star + del\_u, x = x\_star + del\_x, u = u\_star - K*del\_x => u = -K*x, del\_x => u = -K*x, del_x == -K*x, del_x =
```

```
disp('--Part 3 -----')
disp(' ')
% desired poles
s_{desired} = [-1; -4; -4 + 1i; -4 - 1i];
% Gain Matrix via ackerman formula
           = acker(A,B,s_desired)
disp('Closed Loop Poles')
disp(eig(A - B*K))
% Simulate Controlled Linear System
                           % Initial Conditions [m, ms/, rad, rad/s]
           = [-2; 3;-1;2];
% ODE45 solver options
           = odeset('AbsTol',1e-8,'RelTol',1e-8);
options
% ODE45 Function call
[T, X_{lin}] = ode45(@(t,x) ControlledLinearCartPendulum(t,x,A,B,K),time,x0,options);
figure
subplot(4,1,1)
sgtitle('Part 3: Linearized System State Variables')
plot(T,X_lin(:,1))
ylabel('x [m]')
grid minor
subplot(4,1,2)
plot(T,X_lin(:,2))
ylabel('$\dot{x}$ [m/s]','Interpreter','latex')
grid minor
subplot(4,1,3)
plot(T,X_lin(:,3))
grid minor
ylabel('\theta [rad]')
subplot(4,1,4)
plot(T,X_lin(:,4))
grid minor
ylabel('$\dot{\theta}$ [rad/s]','Interpreter','latex')
xlabel('Time [s]')
```

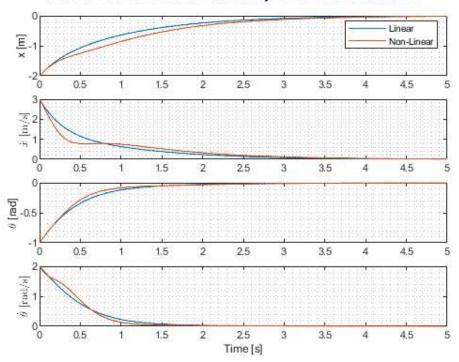


Part 3: Linearized System State Variables

Part 4 Feedback Controller on Nonlinear System

```
% ODE45 Function call
[T, X]
            = ode45(@(t,x) ControlledCartPendulum(t,x,K,m,M,g,l),time,x0,options);
figure
subplot(4,1,1)
sgtitle('Part 4: Linear vs Non-Linear System State Variables')
plot(T,X_lin(:,1),T,X(:,1))
legend('Linear','Non-Linear')
ylabel('x [m]')
grid minor
subplot(4,1,2)
plot(T,X_lin(:,2),T,X(:,2))
ylabel('$\dot{x}$ [m/s]','Interpreter','latex')
grid minor
subplot(4,1,3)
plot(T,X_lin(:,3),T,X(:,3))
grid minor
ylabel('\theta [rad]')
subplot(4,1,4)
plot(T,X_lin(:,4),T,X(:,4))
grid minor
ylabel('$\dot{\theta}$ [rad/s]','Interpreter','latex')
xlabel('Time [s]')
```

Part 4: Linear vs Non-Linear System State Variables



Part 5 State Observer Design with Linear System

```
disp('--Part 5 -----
disp(' ')
% desired observer poles
s_obs_desired = [-3;-12;-12 + 1i;-12 - 1i];
% Observer Gain Matrix
                = place(A',C',s_obs_desired)'
disp('Observer Poles')
disp(eig(A - L*C))
% Observer Initial Conditions
                = zeros(4,1);
z0
% ODE45 Function call
[T, X_{lin\_obs}] = ode45(@(t,x) ControllerEstimatorLinearCartPendulum(t,x,A,B,C,K,L),time,[x0;z0],options);
figure
subplot(4,1,1)
sgtitle('Part 5: System and Observer States')
plot(T,X_lin_obs(:,1),T,X_lin_obs(:,5))
legend('States','Observer States')
ylabel('x [m]')
grid minor
subplot(4,1,2)
plot(T,X_lin_obs(:,2),T,X_lin_obs(:,6))
ylabel('$\dot{x}$ [m/s]','Interpreter','latex')
grid minor
subplot(4,1,3)
plot(T,X_lin_obs(:,3),T,X_lin_obs(:,7))
grid minor
ylabel('\theta [rad]')
subplot(4,1,4)
plot(T,X_lin_obs(:,4),T,X_lin_obs(:,8))
```

```
grid minor
ylabel('$\dot{\theta}$ [rad/s]','Interpreter','latex')
xlabel('Time [s]')
```

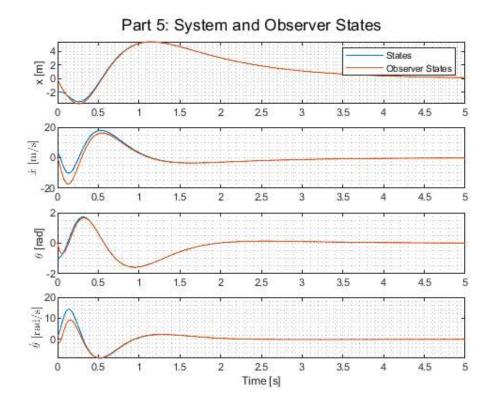
```
--Part 5

L =

15.4624   3.7298

43.7948   43.0396
   0.6685   23.5376
   16.4663   146.7505

Observer Poles
   -3.0000 + 0.0000i
   -12.0000 + 1.0000i
   -12.0000 + 0.0000i
   -12.0000 + 0.0000i
```



Part 6 Closed Loop Transfer Function

```
disp('--Part 6
disp('')

% Closed Loop Transfer Function Equatiom: Y(s)/R(s) = C*(sI - A + BK)^-1*B
Y_R = simplify(C*(inv(s*eye(size(A)) - A + B*K))*B);

% Closed loop matrices with states [x,e], where e is x - z
A_tilde = [(A - B*K), B*K; zeros(size(A)), (A - L*C)];
B_tilde = [B;zeros(size(B))];
C_tilde = [C,zeros(size(C))];
```

```
% Minimum realization for each output
[num,den] = ss2tf(A_tilde,B_tilde,C_tilde,zeros(2,1),1);
Y1 R
           = minreal(tf(num(1,:),den(1,:)));
Y2 R
           = minreal(tf(num(2,:),den(1,:)));
% Verify both methods give same output
disp('Closed Loop Transfer Function: Y_1(s)/R(s): ss2tf output')
disp('Closed Loop Transfer Function: Y_1(s)/R(s): syms output')
disp(Y_R(1))
disp('Closed Loop Transfer Function: Y_2(s)/R(s): ss2tf output')
Y2_R
disp('Closed Loop Transfer Function: Y_2(s)/R(s): syms output')
disp(Y_R(2))
--Part 6 -----
Closed Loop Transfer Function: Y_1(s)/R(s): ss2tf output
Y1 R =
     0.1 s^2 - 1.732e-15 s - 1
  s^4 + 13 s^3 + 61 s^2 + 117 s + 68
Continuous-time transfer function.
Closed Loop Transfer Function: Y_1(s)/R(s): syms output
(s^2 - 10)/(10*(s^4 + 13*s^3 + 61*s^2 + 117*s + 68))
Closed Loop Transfer Function: Y_2(s)/R(s): ss2tf output
Y2_R =
  -0.1 s^2 + 2.512e-16 s + 4.179e-30
  s^4 + 13 s^3 + 61 s^2 + 117 s + 68
Continuous-time transfer function.
```

Part 7 State Observer with Non-Linear System

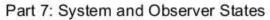
 $-s^2/(10*(s^4 + 13*s^3 + 61*s^2 + 117*s + 68))$

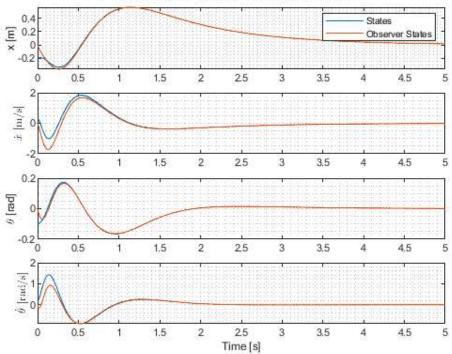
Closed Loop Transfer Function: Y_2(s)/R(s): syms output

```
% ODE45 Function call
[T, X_obs] = ode45(@(t,x) ControllerEstimatorCartPendulum(t,x,A,B,C,K,L,m,M,g,l),time,[-.2;.3;-.1;.2;z0],options);

figure
subplot(4,1,1)
sgtitle('Part 7: System and Observer States')
plot(T,X_obs(:,1),T,X_obs(:,5))
legend('States','Observer States')
ylabel('x [m]')
grid minor
subplot(4,1,2)
plot(T,X_obs(:,2),T,X_obs(:,6))
ylabel('$\dot{x}$ [m/s]','Interpreter','latex')
grid minor
subplot(4,1,3)
```

```
plot(T,X_obs(:,3),T,X_obs(:,7))
grid minor
ylabel('\theta [rad]')
subplot(4,1,4)
plot(T,X_obs(:,4),T,X_obs(:,8))
grid minor
ylabel('$\dot{\theta}$ [rad/s]','Interpreter','latex')
xlabel('Time [s]')
```





Functions

```
function xdot = NonlinearCartPendulum(t,state,u,m,M,g,l)
% Part 2 Function - nonlinear model
            = state(1,1);
x1
x2
            = state(2,1);
x3
            = state(3,1);
            = state(4,1);
xdot(1,1)
           = x2;
            = (u - m*g*cos(x3)*sin(x3) + m*1*x4^2*sin(x3))/(m + M - m*cos(x3)^2);
xdot(2,1)
xdot(3,1)
           = x4;
           = (-u*\cos(x3) - m*1*x4^2*\sin(x3)*\cos(x3) + m*g*\sin(x3) + M*g*\sin(x3))/(1*(m + M - m*\cos(x3)^2));
xdot(4,1)
end
function [xdot] = ControlledLinearCartPendulum(t,state,A,B,K)
% Part 3 Function - Linear Model with state feedback controller
% State Vector
                = state(1:4,1);
Х
% Feedback Control Law
                = -K*x;
% State Space Equation
```

```
xdot
               = A*x + B*u;
end
function [xdot] = ControlledCartPendulum(t, state, K, m, M, g, 1)
% Part 4 Function - Non-linear model with state feedback controller
% State Vector
           = state(1,1);
           = state(2,1);
x2
х3
           = state(3,1);
х4
           = state(4,1);
           = state(1:4,1);
Х
% Control Law
           = -K*x;
% Non linear model
xdot(1,1)
          = x2;
xdot(2,1)
          = (u - m*g*cos(x3)*sin(x3) + m*1*x4^2*sin(x3))/(m + M - m*cos(x3)^2);
xdot(3,1)
           = x4;
xdot(4,1) = (-u*cos(x3) - m*1*x4^2*sin(x3)*cos(x3) + m*g*sin(x3) + M*g*sin(x3))/(1*(m + M - m*cos(x3)^2));
end
function [out] = ControllerEstimatorLinearCartPendulum(t,state,A,B,C,K,L)
% Part 5 Function - Linear Model with Luenberger Observer
% and Estimated State Feedback compensator
% State Vector
      = state(1:4,1);
% Observer State Vector
        = state(5:8,1);
% Control Law
       = -K*z;
% Output, Y = Cx = [x1, x3]
        = [state(1,1);state(3,1)];
% Observer Dynamics -> dz/dt = Az + Bu + L(Y - C*z), u = -Kz
zdot
       = (A - L*C - B*K)*z + L*Y;
% State Dynamics
xdot
      = A*x + B*u;
out
        = [xdot;zdot];
end
function [out] = ControllerEstimatorCartPendulum(t,state,A,B,C,K,L,m,M,g,l)
% Part 7 Function Non-Linear Model with Luenberger Observer
% and Estimated State Feedback compensator
% State Vector
x1
           = state(1,1);
x2
           = state(2,1);
x3
           = state(3,1);
           = state(4,1);
x4
% Observer State Vector
           = state(5:8,1);
% Control Law
            = -K*z;
u
```

```
% Output Y = [x1 x3]'
          = [x1;x3];
% Observer Dynamics
       = (A - L*C - B*K)*z + L*Y;
% Non linear model
xdot(1,1) = x2;
xdot(2,1) = (u - m*g*cos(x3)*sin(x3) + m*1*x4^2*sin(x3))/(m + M - m*cos(x3)^2);
xdot(3,1) = x4;
x dot(4,1) = (-u*cos(x3) - m*l*x4^2*sin(x3)*cos(x3) + m*g*sin(x3) + M*g*sin(x3))/(l*(m + M - m*cos(x3)^2));
out
          = [xdot;zdot];
end
Linearized State Space Model
A =
    0
       1 0
                  0
                  0
       0 -1
    0
         0
            0
    0
            11
B =
        0
   0.1000
  -0.1000
C =
    1
        0 1 0
    0
D =
    0
    0
System is Controllable
System is Observable
```

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