

Observer-Based Feedback Control Design

- **Objective:** Construct **Combined Observer-Controller Compensator** to control linear lumped continuous-time (CT) or discrete-time (DT) system using only input-output signals
- We consider linear time-varying (LTI) system

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t) \\ y(t) &= \mathbf{C}\mathbf{x}(t) + \mathbf{D}u(t)\end{aligned}$$

or

$$\begin{aligned}\mathbf{x}[k+1] &= \mathbf{A}\mathbf{x}[k] + \mathbf{b}u[k] \\ y[k] &= \mathbf{C}\mathbf{x}[k] + \mathbf{D}u[k]\end{aligned}$$

- We assume that the system at hand is both reachable and observable

Combined Observer-Controller Compensator

- The equation of the observer,

$$\dot{\tilde{\mathbf{x}}}(t) = \mathbf{A}\tilde{\mathbf{x}}(t) + \mathbf{B}\mathbf{u}(t) + \mathbf{L}(\mathbf{y}(t) - \tilde{\mathbf{y}}(t))$$

- In the stability analysis of the closed-loop system driven by the combined observer-controller compensator, take into account that $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t)$ and $\tilde{\mathbf{y}}(t) = \mathbf{C}\tilde{\mathbf{x}}(t)$
- Substituting the above expressions for $\mathbf{y}(t)$ and $\tilde{\mathbf{y}}(t)$ gives

$$\begin{aligned}\dot{\tilde{\mathbf{x}}}(t) &= \mathbf{A}\tilde{\mathbf{x}}(t) + \mathbf{B}\mathbf{u}(t) + \mathbf{L}(\mathbf{C}\mathbf{x}(t) - \mathbf{C}\tilde{\mathbf{x}}(t)) \\ &= (\mathbf{A} - \mathbf{LC})\tilde{\mathbf{x}}(t) + \mathbf{LC}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)\end{aligned}$$

- Note that we do not implement the observer using the above representation; it is only for the stability analysis of the closed-loop system

Combined Observer-Controller Compensator Analysis

- Equivalent observer implementation format

$$\dot{\tilde{x}}(t) = (\mathbf{A} - \mathbf{LC})\tilde{x}(t) + \mathbf{B}u(t) + \mathbf{L}y(t)$$

- Equations of the closed-loop system

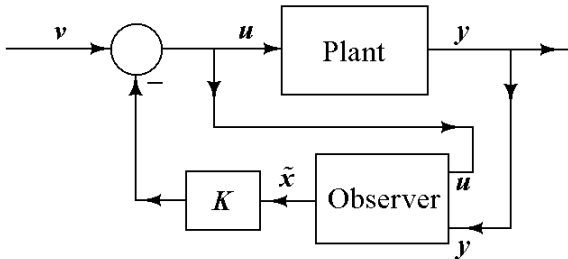
$$\left. \begin{aligned} \begin{bmatrix} \dot{x}(t) \\ \dot{\tilde{x}}(t) \end{bmatrix} &= \begin{bmatrix} \mathbf{A} & \mathbf{O} \\ \mathbf{LC} & \mathbf{A} - \mathbf{LC} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} + \begin{bmatrix} \mathbf{B} \\ \mathbf{B} \end{bmatrix} u(t) \\ y(t) &= \begin{bmatrix} \mathbf{C} & \mathbf{O} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} \end{aligned} \right\}$$

Controller implementation

- The control law

$$u(t) = -K\tilde{x}(t) + v(t)$$

instead of the actual state-feedback control law



Closed-loop system

- Closing the loop

$$\left. \begin{aligned} \begin{bmatrix} \dot{x}(t) \\ \dot{\tilde{x}}(t) \end{bmatrix} &= \begin{bmatrix} \mathbf{A} & \mathbf{O} \\ \mathbf{LC} & \mathbf{A} - \mathbf{LC} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} + \begin{bmatrix} \mathbf{B} \\ \mathbf{B} \end{bmatrix} (-\mathbf{K}\tilde{x}(t) + v(t)) \\ y(t) &= \begin{bmatrix} \mathbf{C} & \mathbf{O} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} \end{aligned} \right\}$$

- Closed-loop system,

$$\begin{aligned} \begin{bmatrix} \dot{x}(t) \\ \dot{\tilde{x}}(t) \end{bmatrix} &= \begin{bmatrix} \mathbf{A} & -\mathbf{BK} \\ \mathbf{LC} & \mathbf{A} - \mathbf{LC} - \mathbf{BK} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} + \begin{bmatrix} \mathbf{B} \\ \mathbf{B} \end{bmatrix} v(t) \\ y(t) &= \begin{bmatrix} \mathbf{C} & \mathbf{O} \end{bmatrix} \begin{bmatrix} x(t) \\ \tilde{x}(t) \end{bmatrix} \end{aligned}$$

- To analyze the above closed-loop system, it is convenient to perform a change of coordinates

Closed-loop system analysis

- Use the transformation

$$\begin{bmatrix} \mathbf{x} \\ \mathbf{x} - \tilde{\mathbf{x}} \end{bmatrix} = \begin{bmatrix} \mathbf{I}_n & \mathbf{O} \\ \mathbf{I}_n & -\mathbf{I}_n \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ \tilde{\mathbf{x}} \end{bmatrix}$$

- Note that

$$\begin{bmatrix} \mathbf{I}_n & \mathbf{O} \\ \mathbf{I}_n & -\mathbf{I}_n \end{bmatrix}^{-1} = \begin{bmatrix} \mathbf{I}_n & \mathbf{O} \\ \mathbf{I}_n & -\mathbf{I}_n \end{bmatrix}$$

Closed-loop system in the new coordinates

- The closed-loop system in the new coordinates

$$\begin{aligned} \begin{bmatrix} \dot{\mathbf{x}}(t) \\ \dot{\mathbf{x}}(t) - \dot{\tilde{\mathbf{x}}}(t) \end{bmatrix} &= \begin{bmatrix} \mathbf{A} - \mathbf{BK} & \mathbf{BK} \\ \mathbf{O} & \mathbf{A} - \mathbf{LC} \end{bmatrix} \begin{bmatrix} \mathbf{x}(t) \\ \mathbf{x}(t) - \tilde{\mathbf{x}}(t) \end{bmatrix} \\ &\quad + \begin{bmatrix} \mathbf{B} \\ \mathbf{O} \end{bmatrix} \mathbf{v}(t) \\ \mathbf{y}(t) &= \begin{bmatrix} \mathbf{C} & \mathbf{O} \end{bmatrix} \begin{bmatrix} \mathbf{x}(t) \\ \mathbf{x}(t) - \tilde{\mathbf{x}}(t) \end{bmatrix} \end{aligned}$$

- Note that $\mathbf{e}(t) = \mathbf{x}(t) - \tilde{\mathbf{x}}(t)$ is the estimation error

Transfer function of the closed-loop system

- The subsystem corresponding to the error component $\mathbf{e}(t) = \mathbf{x}(t) - \tilde{\mathbf{x}}(t)$ is unreachable
- The $2n$ poles of the closed-loop system are equal to the individual eigenvalues of both $\mathbf{A} - \mathbf{LC}$ and $\mathbf{A} - \mathbf{BK}$
- Thus the design of the observer is separated from the construction of the controller—the **separation principle**
- The closed-loop transfer function relating $\mathbf{Y}(s)$ and $\mathbf{V}(s)$ is

$$\begin{aligned}\mathbf{Y}(s) &= \begin{bmatrix} \mathbf{C} & \mathbf{O} \end{bmatrix} \begin{bmatrix} s\mathbf{I}_n - \mathbf{A} + \mathbf{BK} & -\mathbf{BK} \\ \mathbf{O} & s\mathbf{I}_n - \mathbf{A} + \mathbf{LC} \end{bmatrix}^{-1} \begin{bmatrix} \mathbf{B} \\ \mathbf{O} \end{bmatrix} \mathbf{V}(s) \\ &= \mathbf{C}(s\mathbf{I}_n - \mathbf{A} + \mathbf{BK})^{-1} \mathbf{B} \mathbf{V}(s)\end{aligned}$$

- The closed-loop system driven by the combined observer-controller compensator has the same transfer function as the system driven by the state-feedback control law $\mathbf{u}(t) = -\mathbf{K}\mathbf{x}(t) + \mathbf{v}(t)$

Observer pole selection

- The combined observer-controller compensator yields the same closed-loop transfer function as the actual state-feedback control law
- It is recommended that the real parts of the observer poles, that is, the real parts of the eigenvalues of the matrix $\mathbf{A} - \mathbf{LC}$, be a factor of 2 to 6 times deeper in the open left-half plane than the real parts of the controller poles which are the eigenvalues of the matrix $\mathbf{A} - \mathbf{BK}$
- Such a choice ensures a faster decay of the observer error $\mathbf{e}(t) = \mathbf{x}(t) - \tilde{\mathbf{x}}(t)$ compared with the desired controller dynamics
- This in turn causes the controller poles to dominate the closed-loop system response

Combined observer-controller compensator design final comments

- The observer poles represent a measure of the speed with which the estimation error $e(t) = x(t) - \tilde{x}(t)$ decays to zero, one would tend to assign observer poles deep in the left-hand plane
- However, fast decay requires large gains which may lead to saturation of some signals and unpredictable nonlinear effects
- If the observer poles were slower than the controller poles, the closed-loop system response would be dominated by the observer, which is undesirable
- As it is usual in engineering practice, the term compromise could be used to describe the process of constructing the final compensator structure