

## PROBLEM 1

The necessary and sufficient conditions for the existence of unknown input observer (UIO) for a plant model of the following form will be derived.

$$\begin{aligned}\dot{x} &= Ax + B_1 u_1 + B_2 u_2 \\ y &= Cx + D_1 u_1 + D_2 u_3\end{aligned}\tag{1}$$

In equation 1,  $u_2$  is the unknown input and  $u_3$  is the output disturbance. The dimensions of the states, outputs, and inputs are as follows. There are  $n$  states,  $m_1$  known inputs,  $m_2$  unknown inputs,  $p$  outputs, and  $r$  output disturbances. To begin the derivation, we can first represent  $x$  as:

$$x = x - M(Cx + D_1 u_1) + M(Cx + D_1 u_1)\tag{2}$$

Substituting the output equation from equation 1 into 2 yields:

$$\begin{aligned}x &= x - M(Cx + D_1 u_1) + M(y - D_2 u_3) \\ x &= (I - MC)x - MD_1 u_1 + My - MD_2 u_3\end{aligned}\tag{3}$$

The first condition for the existence of the UIO can be found by selecting  $M$  such that the effect of  $u_3$  is the zero matrix. Therefore the first necessary and sufficient condition is:

$$MD_2 = 0_{n \times r}\tag{4}$$

Applying the above condition to equation 3 gives the following:

$$x = (I - MC)x - MD_1 u_1 + My\tag{5}$$

Next, the variable  $z$  will be defined as,  $z = (I - MC)x$ . Then equation 5 becomes:

$$x = z - MD_1 u_1 + My\tag{6}$$

It should be noted that the state estimate,  $\hat{x}$ , can be found by replacing  $x$  with  $\hat{x}$  in equation 6. Using the definition of  $z$ , the dynamics of  $\dot{z}$  are given below.

$$\dot{z} = (I - MC)\dot{x} = (I - MC)(Ax + B_1 u_1 + B_2 u_2)\tag{7}$$

Combining equations 6 and 7 gives the following dynamics:

$$\begin{aligned}\dot{z} &= (I - MC)(Az - AMD_1 u_1 + AMy + B_1 u_1 + B_2 u_2) \\ \dot{z} &= (I - MC)[Az + AMy + (B_1 - AMD_1) u_1] + (I - MC)B_2 u_2\end{aligned}\quad (8)$$

The second condition for the existence of the UIO can be found by selecting M such that the effect of  $u_2$  is the zero matrix in equation 8. Therefore the second necessary and sufficient condition is:

$$(I - MC)B_2 = 0_{n \times m_2} \quad (9)$$

Applying this second condition to equation 8 and adding an innovation term to close the loop for the UIO, the dynamics for  $z$  becomes

$$\dot{z} = (I - MC)[Az + AMy + (B_1 - AMD_1) u_1] + L(y - \hat{y}) \quad (10)$$

The state estimate can then be determined with

$$\hat{x} = z - MD_1 u_1 + My \quad (11)$$

If M is set to 0, then the Luenberger observer is obtained, as shown below:

$$\dot{z} = Az + B_1 u_1 + L(y - \hat{y}) \quad (12)$$

Combining the two conditions for the existence of the UIO, M can be found by solving the following:

$$M \begin{bmatrix} CB_2 & D \end{bmatrix} = \begin{bmatrix} B_2 & 0_{n \times r} \end{bmatrix} \quad (13)$$

$$M = \begin{bmatrix} B_2 & 0_{n \times r} \end{bmatrix} \begin{bmatrix} CB_2 & D \end{bmatrix}^\dagger \quad (14)$$

The matrix M exists if and only if  $\begin{bmatrix} CB_2 & D \end{bmatrix}$  is full column rank, and is therefore left invertible. The observation error is defined as  $e = x - \hat{x}$ . Using equation 11, this is expressed as

$$e = x - (z - MD_1 u_1 + My) = x - z + MD_1 u_1 - M(Cx + D_1 u_1 + D_2 u_3) \quad (15)$$

Applying the first condition (equation 4) eliminates the  $u_3$  term in equation 15. Then the

observation error can be simplified to  $e = (I - MC)x - z$ . The error dynamics can be expressed as:

$$\dot{e} = (I - MC)(Ax + B_1 u_1 + B_2 u_2) - [(I - MC)(Az + AMy + (B_1 - AMD_1)u_1) + L(y - \hat{y})] \quad (16)$$

The estimated output is given by  $\hat{y} = C\hat{x} + D_1 u_1$ . Therefore the innovation term,  $L(y - \hat{y})$ , can be written as:

$$L(y - \hat{y}) = L[(Cx + D_1 u_1 + D_2 u_3) - (C(z - MD_1 u_1 + My) + D_1 u_1)] \quad (17)$$

Substituting in the output equation for  $y$  into the above equation, then simplifying yields:

$$L(y - \hat{y}) = L[Cx + D_2 u_3 - Cz - CMCx - CMD_2 u_3] \quad (18)$$

Again applying the first condition in equation 4 (eliminating the  $MD_2$  term), and using the definition of the error, equation 18 can be simplified to

$$L(y - \hat{y}) = LC(x - z - MCx) + LD_2 u_3 = LCe + LD_2 u_3 \quad (19)$$

Applying the second condition (equation 9) to equation 16, it can be simplified to:

$$\dot{e} = (I - MC)(Ax) - [(I - MC)(Az + AM(Cx + D_1 u_1 + D_2 u_3) - AMD_1 u_1) + L(y - \hat{y})] \quad (20)$$

Again applying the first condition, the  $u_3$  term can be eliminated. Then simplifying yields:

$$\begin{aligned} \dot{e} &= (I - MC)(Ax) - [(I - MC)(Az + AMCx) + LCe + LD_2 u_3] \\ \dot{e} &= (I - MC)(Ax - Az - AMCx) - LCe - LD_2 u_3 \end{aligned} \quad (21)$$

Applying the error definition to equation 21, the error dynamics can be simplified to the final following form.

$$\dot{e} = ((I - MC)A - LC)e - LD_2 u_3 \quad (22)$$

It can then be shown that the state observation error is  $l_\infty$  stable if there exists a matrix  $P = P^T > 0$  and an observer gain matrix  $L$  that satisfies the following matrix inequality

$$\begin{bmatrix} E^T P E - (1 - \alpha)P & E^T P N \\ N^T P E & N^T P N - \alpha I \end{bmatrix} \preceq 0 \quad (23)$$

where  $E = ((I - MC)A - LC)$ ,  $N = -LD$ , and  $0 < \alpha < 1$ . If the pair  $((I - MC)A, C)$  is detectable (all observable states are stable), then we can find an observer gain matrix  $L$  such that  $E$  is Schur stable