Experiment 1: Transfer Characteristics of a CS Amplifier

January 28, 2016

Objective

- To obtain the transfer characteristics of a NMOS CS amplifier when (i) RD = 2.43 k Ω , (ii) RD = 4.58 k Ω and (iii) Coupled NMOS with RD = 4.58 k Ω .
- To plot the derivative (gain) in MATLAB and obtain the maximum gain and Vmax i.e. the input voltage VIN at which the maximum gain occurs.
- To use the techniques of polynomial fitting and fast fourier transform to find the frequency spectrum of amplitude of Vout and thus determine the Total Harmonic Distortion (THD).
- To check the THD value for a given input signal when we cascade multiple amplifier stages together. Also, to see whether we get any difference in THD in case we change the order of the amplifiers (high gain first vs low gain first).
- To check the THD degradation as the output bias point is skewed away from the mid value of VDD/2.
- To check the THD value for increasing signal magnitude and different gain values.

Components

- MOSFET
- Breadboard
- Resistors
- Potentiometers
- Connecting wires

Circuit Diagram

ckt.jpg

Circuit diagram

Observation Table

0.1 \mathbf{V}_{out} vs \mathbf{V}_{in} for \mathbf{R}_D =

1.53 12.02 1.78 12.01 1.97 11.83 2.14 11.54 2.24 11.29 2.31 11.05 2.36 10.88 2.45 10.54 2.55 10.13 2.63 9.78 2.7 9.38 2.77 9.04 2.81 8.86 2.87 8.57 2.91 8.19 2.98 7.81 3 7.67 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	$V_{out}(V)$	$V_{in}(V)$
1.78 12.01 1.97 11.83 2.14 11.54 2.24 11.29 2.31 11.05 2.36 10.88 2.45 10.54 2.55 10.13 2.63 9.78 2.7 9.38 2.77 9.04 2.81 8.86 2.87 8.57 2.91 8.19 2.98 7.81 3 7.67 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	1.53	12.02
1.97 11.83 2.14 11.54 2.24 11.29 2.31 11.05 2.36 10.88 2.45 10.54 2.55 10.13 2.63 9.78 2.7 9.38 2.77 9.04 2.81 8.86 2.87 8.57 2.91 8.19 2.98 7.81 3 7.67 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	1.78	
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2.81 8.86 2.87 8.57 2.91 8.19 2.98 7.81 3 7.67 3.06 7.27 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	2.11	9 04
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3 7.67 3.06 7.27 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	2.98	7.81
3.06 7.27 3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3	7.67
3.11 6.98 3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.06	7.27
3.17 6.55 3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.11	6.98
3.22 6.14 3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.17	6.55
3.3 5.64 3.4 4.99 3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.22	6.14
3.43 4.57 3.51 4.11 3.61 3.83 3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.3	5 64
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3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.43	4.57
3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.51	4.11
3.59 3.46 3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.61	3.83
3.64 2.94 3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.59	3.46
3.69 2.6 3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.64	2.94
3.74 2.08 3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.69	2.6
3.78 1.7 3.85 1.22 3.94 0.99 4.1 0.8	3.74	2.08
3.85 1.22 3.94 0.99 4.1 0.8	3.78	1.7
3.94 0.99 4.1 0.8	3.85	1.22
4.1 0.8	3.94	0.99
	4.1	0.8
4.88 0.54	4.88	0.54
5.86 0.41	5.86	0.41
6.3 0.38	6.3	0.38

0.2 $\mathbf{V}_{out} \mathbf{vs} \mathbf{V}_{in} \mathbf{for} \mathbf{R}_D =$

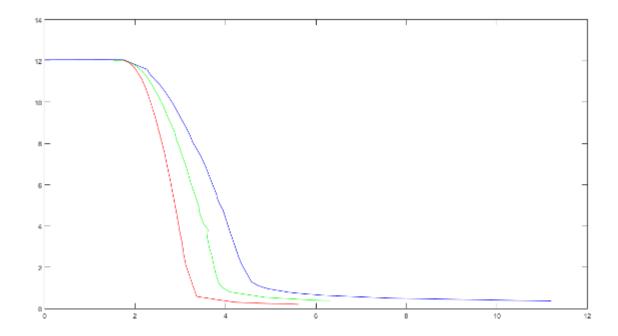
$\mathbf{V}_{\mathbf{V}}$	V (V)
$V_{out}(V)$	$V_{in}(V)$
0.0	12.06
1.33	12.07
1.74 2.27	12.06
2.27	12.06 11.58
2.27 2.31 2.42	11.4
2.42	11.13
2.54 2.54 2.67 2.83 3.03	11.58 11.4 11.13 10.86 10.48 9.95 9.15
2.67	10.48
2.83	9.95
3.03	9.15
3.03 3.19 3.27 3.45 3.56 3.68 3.8	8.5 8.07 7.39
3.27	8.07
3.45	7.39
3.56	6.86
3.68	6.18
3.8	5 53
3.81 3.86	5.37 5.1 4.78 3.89 3.63 3.13 2.48 2.16 1.29
3.86	5.1
2 04	4.78
4.07 4.11 4.19 4.29	3.89
4 1 1	3.63
4.19	3.13
4 29	2.48
4 1 1	2.16
4.57 4.73	1 29
4 73	1 1
4.93	1.1 0.96
5.42	0.78
5.69	0.72
5.69 6.15	0.64
7.63	0.5
10.14	0.4
10.14	0.36
11.2	0.50

0.3 V_{out} vs V_{in} for coupled FETs

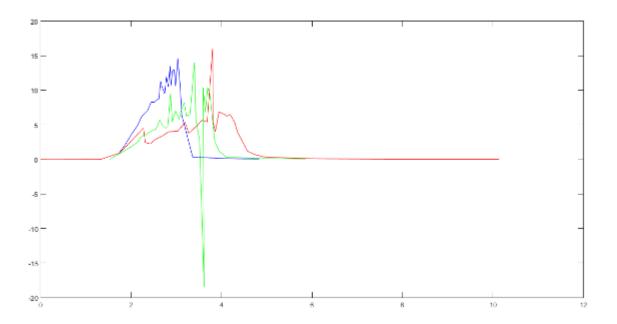
$V_{out}(V)$	$V_{in}(V)$
1.74	12.04
1.88	11.9 11.63
2.0	11.63
2.14	11.13
2.23	10.69
2.28	10.38
2.37	9.79
2.44	9.29
2.14 2.23 2.28 2.37 2.44 2.51 2.58 2.61 2.65 2.68	10.69 10.38 9.79 9.29 8.71 8.13 7.87 7.52 7.18 6.55
2.58	8.13
2.61	7.87
2.65	7.52
2.68	7.18
2.74	6.55
2.78 2.82 2.86 2.88 2.91 2.95 2.98 3.03 3.08 3.12	5.69 5.27 5.0 4.68 4.17 3.78 3.25 2.52 2.09
2.82	5.69
2.86	5.27
2.88	5.0
2.91	4.68
2.95	4.17
2.98	3.78
3.03	3.25
3.08	2.52
3.12	2.09
3.30	0.58
4.17 4.45	0.31
4.43	0.28
4.83 5.61	0.38 0.31 0.28 0.25 0.2
3.01	0.2

Plots

Plot of $V_{out}vsV_{in}$

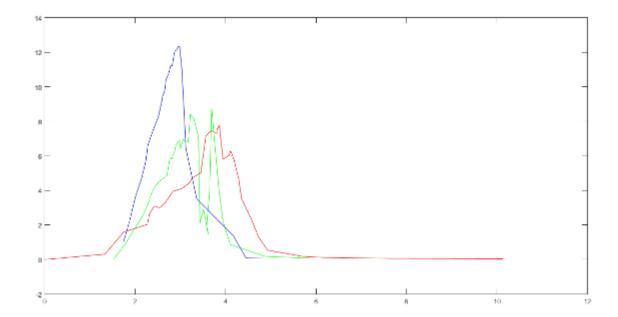


Gain versus $V_{\it in}$

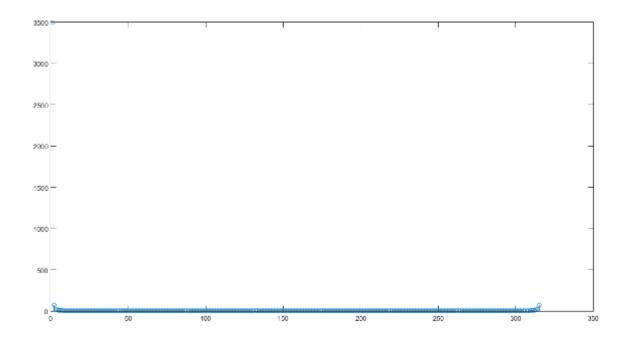


Gain versus \mathbf{V}_{in} with moving average

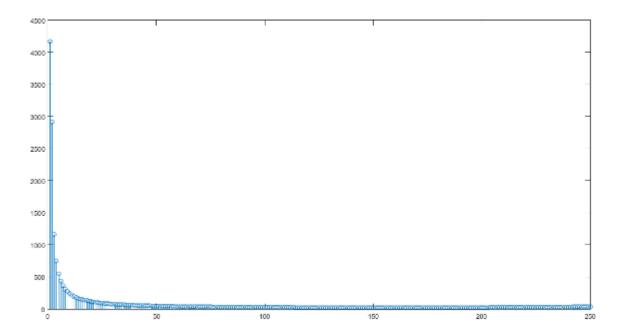
Then, we take a moving average smoothing filter (included in the MATLAB library as a built-in function) to smoothen it. Given below are the unsmoothed and smoothed plots respectively.



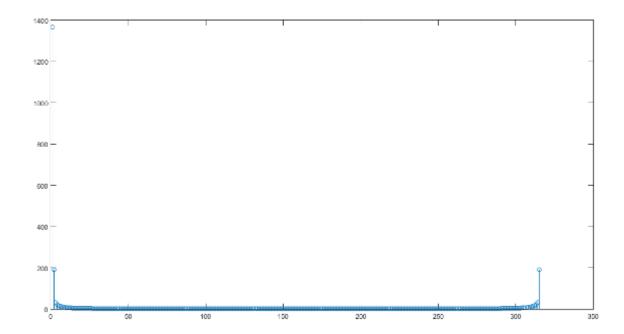
Gain versus



Gain versus



Gain versus



Results

The bias point selected are as follows:

- i. Vm = 3.726 V
- ii. Vm = 3.120 V
- iii. Vm = 2.717 V

From the plots, we infer that Vm value decreases and slop increases from case (i) to case (iii). This is the direct consequence of the formula Vout=VDD KnRD(Vin-Vtn). It is apparent that if either KN or RD increase, slope increases and Vm decreases.

SMALL SIGNAL OUTPUT CHARACTERISTICS (CALCULATION OF THD)

Now, we shift our origin to (Vm,VDD/2) and clip out the cut-off and triode parts and take only the region where the output is more or less linear. Following this, we use a degree 3 polynomial to fit the data (using inbuilt function polyfit), give a sinusoidal input array (Sin(t)) and then take the fast-Fourier transform (using inbuilt function fft) of the obtained output to study the

non-linearity of the amplifier. Following are the amplitude spectrum plots obtained: The given plots are for Case 1, Case 2 and Case 3 respectively.

Now, let us calculate the values of Total Harmonic Distortion for the three different gain values (the three different cases)

THD_1	THD_2	THD_3	Ī
-9.59	0.104	0.72	-

ADDITIONAL LINEARITY TESTS

We calculate the Total Harmonic Distortion (THD) by taking inputs of different peak to peak values. THD1(for amplitude 0.5 Vp-p) = 8.58% THD1(for amplitude 1.0 Vp-p) = 11.26% THD1(for amplitude 2.0 Vp-p) = 21.42%

We can see that as input signal amplitude is increased, the total harmonic distortion increases. It is obvious because the input signal is affected by non-linear gain area due to change in curvature of fitted curve (In this case we used 6th degree polynomial) and hence gets distorted. Now, we check the degradation of THD as the bias point is skewed away from (Vm, VDD/2). It is expected that closer the bias point is to VDD/2 the lesser the THD, since the entire input signal we lie in the linear range.

We perform this check for again the first set of data. We take the bias point as 4.16V skewing away from 3.727V.

THD1 = 17.01% (skewed case)

Next we try to see the THD values when we cascade amplifiers of different gains. Then we try to observe the difference between the cases when low gain amplifier is kept at first stage versus when high gain amplifier is kept at second stage. Following are the amplitude spectrum plots for the two cases:

 $THD_{1->2}(low-\dot{c}_{i}high\ gain) = 23.27\%\ THD_{2->1}(high-\dot{c}_{i}low\ gain) = 7.87\%$

So, we see that when the high gain amplifier is taken as the first stage, we get a lesser value of THD. It is because the output of the first amplifier (high gain) is fed into the low gain amplifier. There is a fair chance that the signal may enter the non-linear region. But due to low gain the effect of non-linearity is quite suppressed as compared to the other case.

Discussion

14EC10043

From the characteristics we could observe that in the output characteristics as Vds increases the Id increases upto the transition point.

After the transition the curve didnt remain perfectly horizontal as expected and this due to the channel width modulation effect.

While measuring Id with respect to Vgs we must keep Vds constant in our experiment we must adjust the value of drain resistance to keep Vds constant it was found roughly constant

14EC10049

- When giving the input signal, it is important to ensure that the input span is within the linear region. For this purpose, input had to be scaled, especially when the gain is very high, as in the case of cascade amplifier configuration. For cascade amplifier, we have used input voltage of 0.1Vp-p.
- For completely linear output, THD=0% and the amplifier is said to be a linear amplifier.
- As bias point is skewed away from (Vm,VDD/2), the new dc biasing is closer to the non-linear region and the same Vp-p causes more distortion. As a result, the THD obtained is higher.
- We get a higher value of THD for higher value of gain. This depicts the gain-linearity tradeoff.
- It is important to note that we do not consider the dc component of output signal while calculating the THD, as it is only because of the harmonics of the fundamental frequency.
- To find Vm, rather than using the discrete derivative curve or its smoothened counterpart (using moving average smoothing), we prefer taking the derivative of a Fourier or Gaussian fit as it is more exact and gives us a more accurate value of Vm. In our case we went on with Gaussian fit because it was observed to be more accurate of the two. It is worth noting that for the optimum performance of these fitting techniques the input needs to be equi-spaced or they tend to diverge at places where concentration of input is low.