EHB 335E HW45

· Assume Vint, Since Id, us and hence Id, us cannot drange

To accomodate the increase in Vin, Vss should increase

· Ass Vsst, Voutt (Vasiz = Vout - Uss), Iding is fred

· As Nort 1, Vo31 (Nos, 3 = Vo, 3 - Vout), Id, us is fixed

· As V₆,3 T, drain-to-source voltage of 1/2 increases and Id, 1/2 lends to increase but since

Id, 112 cannot increase Ngs,2 has to decrease to hold Id, 112 constant. Therefore would

has to decrease. Therefore this circuit employs negative feedback.

In Summary,

Wint a Vsst > North > North > North regotive Feedback

Jewond Method Vin=0

Av =
$$\frac{V_{out}}{V_{in}} = \frac{V_{x}}{V_{in}} \cdot \frac{V_{out}}{V_{x}}$$

Av = $\frac{V_{out}}{V_{in}} = \frac{V_{x}}{V_{x}} \cdot \frac{V_{out}}{V_{x}}$

Av = $\frac{V_{out}}{V_{in}} = \frac{V_{x}}{V_{x}} \cdot \frac{V_{out}}{V_{x}}$

Av = $\frac{V_{out}}{V_{in}} = \frac{V_{x}}{V_{in}} \cdot \frac{V_{out}}{V_{x}}$

Av = $\frac{V_{x}}{V_{in}} = \frac{V_{x}}{V_{in}} \cdot \frac{V_{out}}{V_{x}}$

Av = $\frac{V_{x}}{V_{in}} = \frac{V_{x}}{V_{in}} \cdot \frac{V_{x}}{V_{x}}$

Av = $\frac{V_{x}}{V_{in}} = \frac{V_{x}}{V_{in}} \cdot \frac{V_{x}}{V_{x}}$

Av = $\frac{V_{x}}{V_{in}} = \frac{V_{x}}{V_{x}} \cdot \frac{V_{x}}{V_{x}}$

Av = $\frac{V_{x}}{V_{x}} = \frac{V_{x}}{V_{x}} \cdot \frac{V_{x}}{V_{x}}$

Av = $\frac{V_{x}}{V_{x}}$

NOD

New

RF+RL In (current II = 1 + RE+RL

RO (RF+RL) I'm ; In + BE+BL

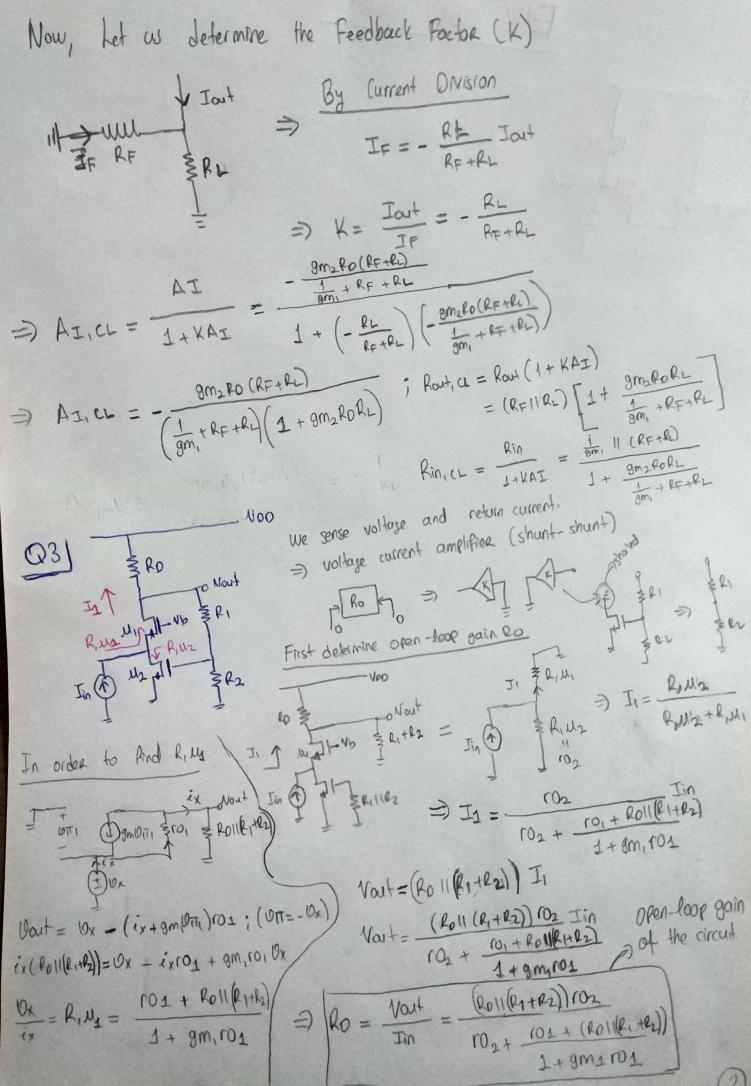
Figure

$$Vout = -8m_2 (RFIIRL) Vx$$

$$Jout = \frac{Vout}{Jaut} = -9m_2 Vx = -\frac{9m_2 Lo(RF+RL)}{8m_1 + RF+RL} Jin$$

$$Jout = \frac{Vout}{Jout} = -9m_2 Vx = -\frac{1}{8m_1 + RF+RL}$$

AI = Jout 3m, + PF+ P,



we draw

Tin (a)
$$= R_1 U_1 \parallel R_1 U_2$$

$$= R_1 U_2 \parallel R_$$

Now we can calculate the (losed-loop parameters

Now we can calculate the (losed-loop parameters

$$\frac{Ro \, \text{M(Links)} \, (O_2)}{(Ro \, \text{M(Links)}) \, (O_2)} = \text{closed-loop gain at transimpedance amplified} \\
Ro = \frac{(O_2 + \frac{1}{2} + 900 \cdot (O_2))}{(O_2 + \frac{1}{2} + 900 \cdot (O_3))} = \text{closed-loop input imperior}$$

$$\frac{Ro \, \text{M(Links)} \, (O_2)}{(O_2 + \frac{1}{2} + 900 \cdot (O_3))} = \text{closed-loop input imperior}$$

Por
$$\alpha = \frac{(R_0 \parallel (\ln k_1))(O_2)}{(R_0 \parallel (\ln k_1))(O_2)}$$
 $\alpha = \frac{(O_2 + \frac{(O_2 + \frac{(P_0 \parallel (\ln k_1))(O_2)}{2 + o(m_1 O_1)})}{(O_2 + \frac{(O_1 + \frac{(P_0 \parallel (\ln k_1))(O_2)}{2 + o(m_1 O_1)})}{(O_2 + \frac{(O_2 + \frac{(O_1 + \frac{(P_0 \parallel (\ln k_1))(O_2)}{2 + o(m_1 O_1)})}{(O_2 + \frac{(O_2 + \frac{(P_0 \parallel (\ln k_1))(O_2)}{2 + o(m_1 O_1)})}}$

Rola =
$$\frac{1 + \frac{gm_2 P_2}{p_1 + p_2}}{\frac{p_1 + p_2}{p_1 + p_2}} = \frac{\frac{p_2 P_2}{p_1 + p_2}}{\frac{p_1 + p_2}{p_2 + p_2}} = \frac{\frac{p_2 P_2}{p_1 + p_2}}{\frac{p_2 P_2}{p_2 + p_2}} = \frac{\frac{p_2 P_2}{p_2 + p_2}}{\frac{p_2 P_2}{p_2}} = \frac{\frac{p_2 P_2}{p_2}}{\frac{p_2 P_2}{p_2}$$

=) Negative feedback makes the amplifier close to the ideal by reducing Rout and hin of the cost of reducing the gain.