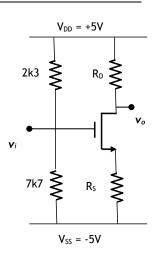
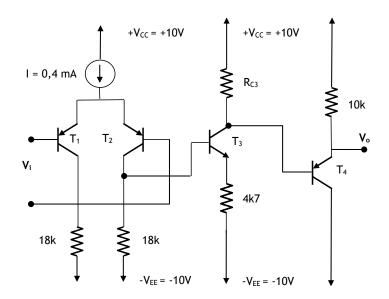
**IMPORTANT**: Besides your **calculator** and the sheets you use for calculations you are only allowed to have an A4 sized "copy sheet" during this exam. Notes, problems and alike are not permitted. **Please submit your "copy sheet" along with your solutions.** You may get your "copy sheet" back after your solutions have been graded. **Do not forget to write down units and convert units carefully! Cell phones are not allowed and should be placed on the front desk before the exam.** 

## ELE222E INTRODUCTION TO ELECTRONICS (21454) Midterm Exam #2 5 May 2014 9.30-11.30 inci ÇİLESİZ, PhD, Hacer ATAR YILDIZ, MSE EEF 2104 and 5201

- 1. You remember the MOS transistor shown on the circuit to the left from your first exam. The NMOS properties are  $V_T = 2 V$ ,  $V_A = \infty$ ,  $\mu_n C_{ox} \left( \frac{W}{L} \right) = 2 mA/V^2$ 
  - a. Find the missing resistor values for the NMOS to operate in saturation with  $I_D = 1 \text{ mA}$ . (10 points).
  - b. Calculate input and output resistances, gain v<sub>o</sub>/v<sub>i</sub> and v<sub>o</sub>/v<sub>s</sub> when the circuit is fed capacitively (i.e., using a capacitor to ensure that biasing conditions are not affected) with a source having an internal resistance of 1 k. (30 points).
  - c. How can you increase the gain <u>without changing biasing conditions</u>? Calculate the increased gain  $\mathbf{v_o/v_i}$  and  $\mathbf{v_o/v_s}$  (10 points).



- 2. You also know the 3-stage BJT amplifier circuit from your first exam.  $|V_{BE}| = 0,6 \text{ V}$ ,  $h_{FE} = h_{fe} = 200$ ,  $h_{re} = 0$ ,  $h_{oe} = 0$ , for all four transistors. Do NOT neglect base currents.
  - a. Determine the value of the emitter resistor that will provide 0,4 mA biasing current to the differential stage.
     (5 points)
  - b. Choose  $R_{C3}$  such that, waveform distortion at the output  $V_o$  is minimum and symmetrical, i.e.,  $V_o = 0V$ ? (10 points)
  - c. Find the input and output resistances, the gain v<sub>o</sub>/v<sub>i</sub> and CMRR.
     V<sub>T</sub> = 25 mV. (35 points)



## **SOLUTIONS:**

1. The MOS design looks very simple BUT as you know from your first exam, many make mistakes. 2k3 and 7k7 resistors divide the 10 V voltage difference, such that  $V_G = -5V + 7.7V = 2.7V$ .

From the given properties it is obvious that  $V_{\it GS}-V_t=1V$  .

Since we are working with NMOS  $V_{GS}=3V \Rightarrow V_S=V_G-V_{GS}=-0.3~V \Rightarrow$ 

$$R_S = \frac{V_S - V_{SS}}{I_D} = \frac{-0.3V - (-5V)}{1mA} = \frac{4k7}{1}$$

For the NMOS to operate in saturation  $V_{DS}=V_{GS}-V_t=1V \ \Rightarrow \ V_D=V_S+1V=-0.3+1=0.7 \ V_S$ 

$$R_D = \frac{V_{DD} - V_D}{I_D} = \frac{5V - 0.7V}{1mA} = \frac{4k3}{1mA}$$

$$Gain = \frac{v_o}{v_i} = \frac{-R_D}{1/g_m + R_S}$$
 with  $g_m = \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{in}) = 2(3-2) = \underline{\underline{2mA/V}} = \underline{15000}$ 

Thus 
$$\frac{v_o}{v_i} = \frac{-R_D}{1/g_m + R_S} = \frac{-4k3}{500\Omega + 4k7} = \frac{-0.826}{1/2000}$$

Realize 
$$r_i = \frac{v_i}{i_i} \to \infty$$
 for the NMOS, thus  $r_i^* = 2k3 \parallel 7k7 = \underline{\underline{1k77}}$  , also  $r_o = R_D = \underline{\underline{4k3}}$ 

Since 
$$\frac{v_s}{v_i} = \frac{r_i^*}{R_{in} + r_i^*} = \frac{1k77}{1k + 1k77} = 0,639, \frac{v_o}{v_s} = \frac{v_o}{v_i} \cdot \frac{v_i}{v_s} = \frac{-0,528}{1k + 1k77}$$

To increase gain R<sub>s</sub> should be shorted at AC using a bridging capacitor resulting in

$$\frac{v_o}{v_i} = \frac{R_D}{1/g_m} = \frac{-4k3}{1/2mA/V} = \frac{-4k3}{500} = \frac{-8.6}{m} \text{ and } \frac{v_o}{v_s} = \frac{v_o}{v_i} \cdot \frac{v_i}{v_s} \cong \frac{-5.5}{m}$$

2. DC characteristics are to be studied

a. 
$$R_E = \frac{V_{CC} - V_E}{I} = \frac{10V - 0.6V}{0.4mA} = \underline{23k5}$$

b. Without neglecting the base currents of the differential (the very first) stage, for  $V_i = 0 \text{ V}$ 

$$I_{C1} = I_{C2} = \frac{h_{FE}}{h_{FE} + 1} \cdot \frac{I_{ref}}{2} = \frac{200}{200 + 1} \cdot \frac{0.4mA}{2} \Rightarrow I_{C1} = I_{C2} = \underbrace{0.199mA}_{C1}$$

$$-(I_{C2} - I_{B3})18k + V_{BE3} + (h_{FE} + 1)I_{B3}4k7 = 0$$

$$I_{C3} = h_{FE} \frac{18k * I_{C2} - V_{BE3}}{(h_{FE} + 1)4k7 + 18k} = 200 \frac{18k * 0,199mA - 0,6V}{(200 + 1)4k7 + 18k} = \underbrace{0,62mA}_{C3} = \underbrace{0,62$$

That the output voltage is in the middle of the power supply range (+10 V and -10 V),  $V_0 = 0$  V

$$(h_{FE}+1)I_{B3}R_{E4}+V_{EB4}-(I_{C3}-I_{B4})R_{C3}=0$$
 , with  $(h_{FE}+1)I_{B4}10k=10V \Longrightarrow I_{E4}=\underline{1}$ 

$$(I_{C3} - I_{B4})R_{C3} = 10V + V_{EB4} \Rightarrow R_{C3} = \frac{10V + V_{EB4}}{I_{C3} - I_{B4}} = \underline{\underline{17k25}}$$

Now the gain is 
$$\frac{v_o}{v_i} = \frac{v_o}{v_{b3}} \cdot \frac{v_{b4}}{v_{b3}} \cdot \frac{v_{b3}}{v_i} = \frac{10k}{10k + r_{e4}} \cdot \frac{-R_{C3} \parallel r_{i4}}{4k7 + r_{e3}} \cdot \frac{18k \parallel r_{i3}}{2r_{e1/2}}$$
 with

$$r_i = r_{i1/2} = 2h_{fe}r_{e1/2} = 2 \cdot 200 \cdot \frac{25mV}{0.199mA} = \frac{50k25}{0.199mA}$$

$$r_{e1/2} = 125,6\Omega$$

$$r_{e^3} = \frac{25mV}{0.62mA} = 40.3\Omega;$$

$$r_{e4} = \frac{25mV}{1mA} = 25\Omega$$

$$r_{i3} = h_{fe}(r_{e3} + 4k7) = 948k$$

$$r_{i4} = h_{fe}(r_{e4} + 10k) = 2M$$

Resulting in 
$$\frac{v_o}{v_i} = \frac{10k}{10k + r_{e4}} \cdot \frac{-17k25 \parallel r_{i4}}{4k7 + r_{e3}} \cdot \frac{18k \parallel r_{i3}}{2r_{e1/2}} = 0,997 \cdot (-3,608) \cdot 70,307 \cong \underline{-253}$$

$$CMRR = 20\log_{10}\left|\frac{2R_E + r_{e1/2}}{r_{e1/2}}\right| = 20\log_{10}\left|\frac{2 \cdot 23k5 + 1256}{1256}\right| = \underbrace{51,5dB}_{======}$$