

# Lecture 03:

## Discrete Time Systems (Part II)

# Outlines

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- 1. Discrete-time systems
- 2. Convolution
- ✓ 3. Linear Constant-Coefficient Difference Equations  
(LCCDEs)
- ✓ 4. Correlation

### 3. Linear Constant-Coefficient Difference Equation (LCCDE)

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- General spec. of DT, LSI, finite-dim sys:

$$\sum_{k=0}^N d_k y[n-k] = \sum_{k=0}^M p_k x[n-k]$$

- defined by  $\{d_k\}, \{p_k\}$
- **order** =  $\max(N, M)$

- Rearrange for  $y[n]$  in **causal** form:

$$y[n] = -\sum_{k=1}^N \frac{d_k}{d_0} y[n-k] + \sum_{k=0}^M \frac{p_k}{d_0} x[n-k]$$

- WLOG, always have  $d_0 = 1$

# Solving LCCDEs

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- “Total solution”

$$y[n] = \underline{y_c[n]} + \underline{y_p[n]}$$

**Complementary Solution**

satisfies  $\sum_{k=0}^N d_k y[n-k] = 0$

**Particular Solution**  
for given forcing function  
 $x[n]$

# Complementary Solution

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- General form of unforced oscillation  
i.e. system's 'natural modes'

- Assume  $y_c$  has form  $y_c[n] = \lambda^n$

$$\Rightarrow \sum_{k=0}^N d_k \lambda^{n-k} = 0$$

$$\Rightarrow \lambda^{n-N} (d_0 \lambda^N + d_1 \lambda^{N-1} + \dots + d_{N-1} \lambda + d_N) = 0$$

$$\Rightarrow \sum_{k=0}^N d_k \lambda^{N-k} = 0$$

**Characteristic polynomial**  
of system - depends only on  $\{d_k\}$

# Complementary Solution

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- $\sum_{k=0}^N d_k \lambda^{N-k} = 0$  factors into **roots**  $\lambda_i$ , i.e.  
 $(\lambda - \lambda_1)(\lambda - \lambda_2)\dots = 0$
- Each/any  $\lambda_i$  satisfies eqn.
- Thus, **complementary solution**:  
$$y_c[n] = \alpha_1 \lambda_1^n + \alpha_2 \lambda_2^n + \alpha_3 \lambda_3^n + \dots$$

Any linear combination will work  
→  $\alpha_i$ s are free to match **initial conditions**

# Complementary Solution

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- Repeated roots in chr. poly:

$$(\lambda - \lambda_1)^L (\lambda - \lambda_2) \dots = 0$$

$$\Rightarrow y_c[n] = \alpha_1 \lambda_1^n + \alpha_2 n \lambda_1^n + \alpha_3 n^2 \lambda_1^n \\ + \dots + \alpha_L n^{L-1} \lambda_1^n + \dots$$

- Complex  $\lambda_i$ s  $\rightarrow$  sinusoidal  $y_c[n] = \alpha_i \lambda_i^n$

# Particular Solution

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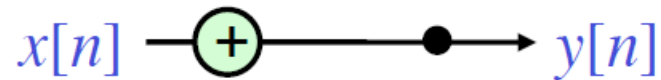
- Recall: Total solution  $y[n] = y_c[n] + y_p[n]$
- Particular solution reflects input
- ‘Modes’ usually decay away for large  $n$  leaving just  $y_p[n]$
- Assume ‘form’ of  $x[n]$ , scaled by  $\beta$ :  
e.g.  $x[n]$  constant  $\rightarrow y_p[n] = \beta$   
 $x[n] = \lambda_0^n \rightarrow y_p[n] = \beta \cdot \lambda_0^n \quad (\lambda_0 \notin \lambda_i)$   
or  $= \beta n^L \lambda_0^n \quad (\lambda_0 \in \lambda_i)$



# LCCDE example

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$$y[n] + y[n-1] - 6y[n-2] = x[n]$$



- Need **input**:  $x[n] = 8\mu[n]$
- Need **initial conditions**:  
 $y[-1] = 1, y[-2] = -1$

# LCCDE example

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- Complementary solution:

$$y[n] + y[n-1] - 6y[n-2] = 0; \quad y[n] = \lambda^n$$

$$\Rightarrow \lambda^{n-2}(\lambda^2 + \lambda - 6) = 0$$

$$\Rightarrow (\lambda + 3)(\lambda - 2) = 0 \rightarrow \text{roots } \lambda_1 = -3, \lambda_2 = 2$$

$$\Rightarrow y_c[n] = \alpha_1(-3)^n + \alpha_2(2)^n$$

- $\alpha_1, \alpha_2$  are unknown at this point

# LCCDE example

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- Particular solution:
- Input  $x[n]$  is constant  $= 8\mu[n]$

assume  $y_p[n] = \beta$ , substitute in:

$$y[n] + y[n-1] - 6y[n-2] = x[n] \quad (\text{'large' } n)$$

$$\Rightarrow \beta + \beta - 6\beta = 8\mu[n]$$

$$\Rightarrow -4\beta = 8 \Rightarrow \beta = -2$$

# LCCDE example

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- Total solution  $y[n] = y_c[n] + y_p[n]$   
$$= \alpha_1(-3)^n + \alpha_2(2)^n + \beta$$
- Solve for unknown  $\alpha_i$ s by substituting *initial conditions* into DE at  $n = 0, 1, \dots$   
$$y[n] + y[n-1] - 6y[n-2] = x[n]$$
- $n = 0$   $y[0] + y[-1] - 6y[-2] = x[0]$   
$$\Rightarrow \alpha_1 + \alpha_2 + \beta + 1 + 6 = 8$$
  
$$\Rightarrow \alpha_1 + \alpha_2 = 3$$

from ICs

# LCCDE example

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- $n = 1$      $y[1] + y[0] - 6y[-1] = x[1]$   
 $\Rightarrow \alpha_1(-3) + \alpha_2(2) + \beta + \alpha_1 + \alpha_2 + \beta - 6 = 8$   
 $\Rightarrow -2\alpha_1 + 3\alpha_2 = 18$
- solve:  $\alpha_1 = -1.8, \alpha_2 = 4.8$
- Hence, system output:  
 $y[n] = -1.8(-3)^n + 4.8(2)^n - 2 \quad n \geq 0$
- **Don't** find  $\alpha_i$ s by solving with ICs at  
 $n = -1, -2$     *(ICs may not reflect natural modes;  
Mitra3 ex 2.37-8 (4.22-3) is wrong)*

# LCCDE solving summary

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- Difference Equation (DE):  
 $Ay[n] + By[n-1] + \dots = Cx[n] + Dx[n-1] + \dots$   
Initial Conditions (ICs):  $y[-1] = \dots$
- DE RHS = 0 with  $y[n] = \lambda^n \rightarrow$  roots  $\{\lambda_i\}$   
gives complementary soln  $y_c[n] = \sum \alpha_i \lambda_i^n$
- Particular soln:  $y_p[n] \sim x[n]$   
solve for  $\beta \lambda_0^n$  “at large  $n$ ”
- $\alpha_i$ s by substituting DE at  $n = 0, 1, \dots$   
ICs for  $y[-1], y[-2]$ ;  $y_t = y_c + y_p$  for  $y[0], y[1]$

# LCCDEs: zero input/zero state

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- Alternative approach to solving LCCDEs is to solve two subproblems:
  - $y_{zi}[n]$ , response with zero input (just ICs)
  - $y_{zs}[n]$ , response with zero state (just  $x[n]$ )
- Because of linearity,  $y[n] = y_{zi}[n] + y_{zs}[n]$
- Both subproblems are ‘fully realized’
- But, have to solve for  $\alpha_i$ s twice (then sum them)

# Impulse response of LCCDEs

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- Impulse response:  $\delta[n] \rightarrow \boxed{\text{LCCDE}} \rightarrow h[n]$

i.e. solve with  $x[n] = \delta[n] \rightarrow y[n] = h[n]$   
(zero ICs)

- With  $x[n] = \delta[n]$ , 'form' of  $y_p[n] = \beta \delta[n]$

$\rightarrow$  solve  $y[n]$  for  $n = 0, 1, 2, \dots$  to find  $\alpha_i$ s



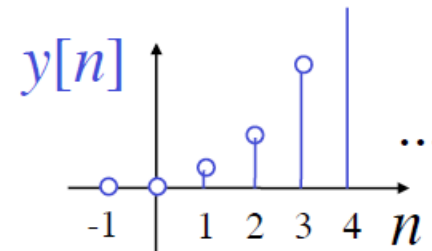
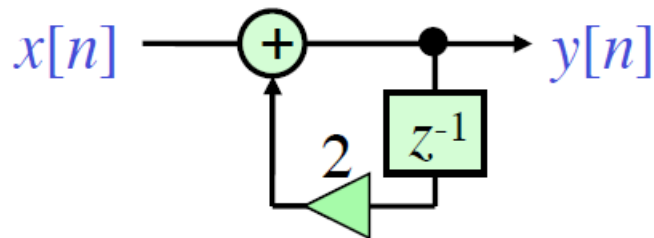
# LCCDE IR example

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- e.g.  $y[n] + y[n-1] - 6y[n-2] = x[n]$   
(from before);  $x[n] = \delta[n]$ ;  $y[n] = 0$  for  $n < 0$
- $y_c[n] = \alpha_1(-3)^n + \alpha_2(2)^n$        $y_p[n] = \beta\delta[n]$
- $n = 0$ :  $y[0] + y[-1] - 6y[-2] = x[0]$  <sup>1</sup>  
 $\Rightarrow \alpha_1 + \alpha_2 + \beta = 1$
- $n = 1$ :  $\alpha_1(-3) + \alpha_2(2) + 1 = 0$
- $n = 2$ :  $\alpha_1(9) + \alpha_2(4) - 1 - 6 = 0$   
 $\Rightarrow \alpha_1 = 0.6, \alpha_2 = 0.4, \beta = 0$
- thus  $h[n] = 0.6(-3)^n + 0.4(2)^n$   <sup>$n \geq 0$</sup>  **Infinite length**

# System property: Stability

- Certain systems can be **unstable** e.g.



Output grows without limit in some conditions

# Stability

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- Several definitions for stability; we use **Bounded-input, bounded-output (BIBO) stable**
- For every bounded input  $|x[n]| < B_x \quad \forall n$  output is also subject to a finite bound,  
 $|y[n]| < B_y \quad \forall n$

# Stability example

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- MA filter:  $y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$

$$|y[n]| = \left| \frac{1}{M} \sum_{k=0}^{M-1} x[n-k] \right|$$

$$\leq \frac{1}{M} \sum_{k=0}^{M-1} |x[n-k]|$$

$$\leq \frac{1}{M} M \cdot B_x \leq B_y \quad \rightarrow \text{BIBO Stable}$$

# Stability & LCCDEs

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- LCCDE output is of form:

$$y[n] = \alpha_1 \lambda_1^n + \alpha_2 \lambda_2^n + \dots + \beta \lambda_0^n + \dots$$

- $\alpha$ s and  $\beta$ s depend on input & ICs,  
*but* to be bounded for **any** input  
we need  $|\lambda| < 1$

## 4. Correlation

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- **Correlation** ~ identifies similarity between sequences:

Cross correlation of  $x$  against  $y$

$$r_{xy}[\ell] = \sum_{n=-\infty}^{\infty} x[n]y[n - \ell]$$

“lag”

- **Note:**  $r_{yx}[\ell] = \sum_{n=-\infty}^{\infty} y[n]x[n - \ell]$  call  $m = n - \ell$

$$= \sum_{m=-\infty}^{\infty} y[m + \ell]x[m] = r_{xy}[-\ell]$$

# Correlation and convolution

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- Correlation: 
$$r_{xy}[n] = \sum_{k=-\infty}^{\infty} x[k]y[k-n]$$
- Convolution: 
$$x[n] \circledast y[n] = \sum_{k=-\infty}^{\infty} x[k]y[n-k]$$
- Hence: 
$$r_{xy}[n] = x[n] \circledast y[-n]$$

Correlation may be calculated by  
convolving with time-reversed sequence

# Autocorrelation

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- **Autocorrelation** (AC) is correlation of signal with itself:

$$r_{xx}[\ell] = \sum_{n=-\infty}^{\infty} x[n]x[n-\ell] = r_{xx}[-\ell]$$

- **Note:**  $r_{xx}[0] = \sum_{n=-\infty}^{\infty} x^2[n] = \varepsilon_x$  **Energy of sequence  $x[n]$**



# Correlation maxima

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- Note:  $r_{xx}[\ell] \leq r_{xx}[0] \Rightarrow \left| \frac{r_{xx}[\ell]}{r_{xx}[0]} \right| \leq 1$
- Similarly:  $r_{xy}[\ell] \leq \sqrt{\varepsilon_x \varepsilon_y} \Rightarrow \frac{r_{xy}[\ell]}{\sqrt{r_{xx}[0] r_{yy}[0]}} \leq 1$
- From geometry,  
$$\langle \mathbf{x} \mathbf{y} \rangle = \sum_i x_i y_i = \underbrace{\sqrt{\sum_i x_i^2}}_{|\mathbf{x}|} |\mathbf{y}| \cos \theta$$

angle  
between  
x and y
- when  $\mathbf{x} // \mathbf{y}$ ,  $\cos \theta = 1$ , else  $\cos \theta < 1$

# AC of a periodic sequence

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- Sequence of period  $N$ :  $\tilde{x}[n] = \tilde{x}[n + N]$
- Calculate AC over a finite window:

$$\begin{aligned} r_{\tilde{x}\tilde{x}}[\ell] &= \frac{1}{2M+1} \sum_{n=-M}^M \tilde{x}[n] \tilde{x}[n - \ell] \\ &= \frac{1}{N} \sum_{n=0}^{N-1} \tilde{x}[n] \tilde{x}[n - \ell] \quad \text{if } M \gg N \end{aligned}$$

# AC of a periodic sequence

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$$r_{\tilde{x}\tilde{x}}[0] = \frac{1}{N} \sum_{n=0}^{N-1} \tilde{x}^2[n] = P_{\tilde{x}} \leftarrow \begin{array}{l} \text{Average energy per} \\ \text{sample or } \mathbf{Power} \text{ of } x \end{array}$$

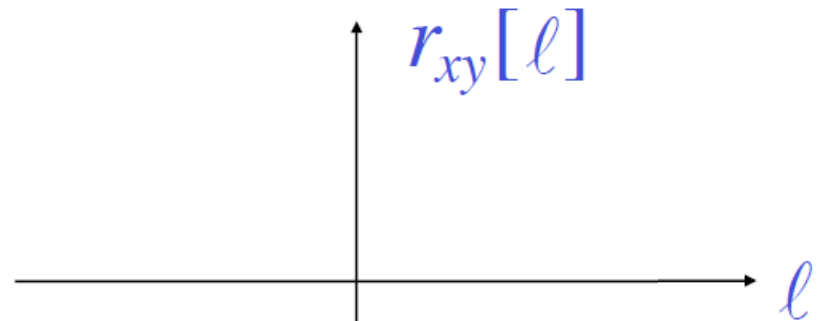
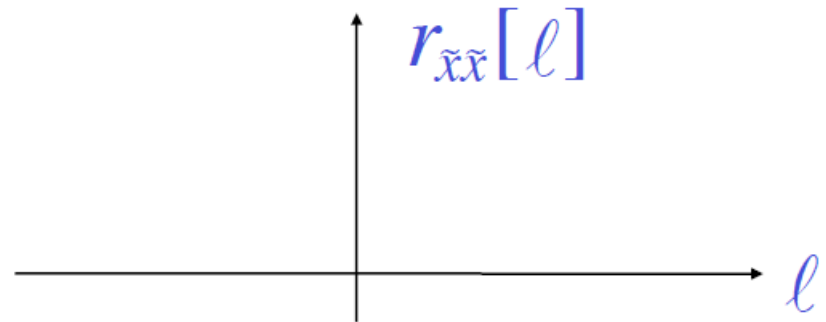
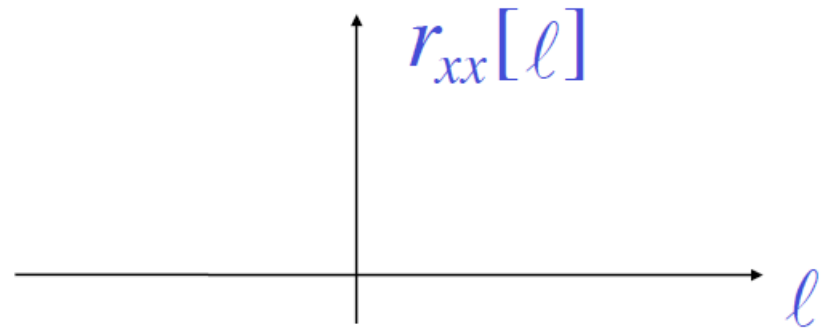
$$r_{\tilde{x}\tilde{x}}[\ell + N] = \frac{1}{N} \sum_{n=0}^{N-1} \tilde{x}[n] \tilde{x}[n - \ell - N] = r_{\tilde{x}\tilde{x}}[\ell]$$

- i.e **AC** of periodic sequence is **periodic**

# What correlations look like

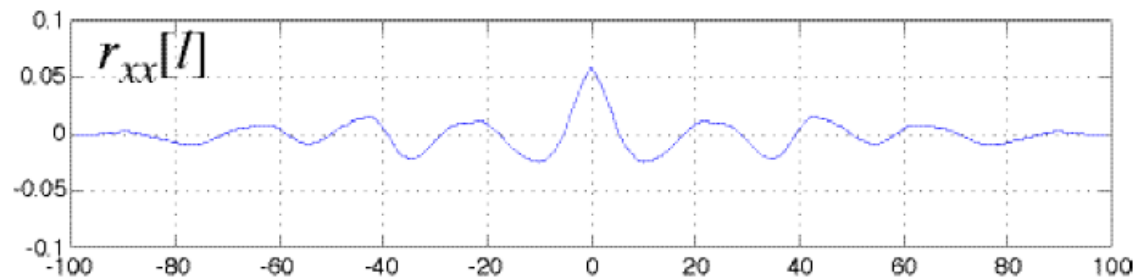
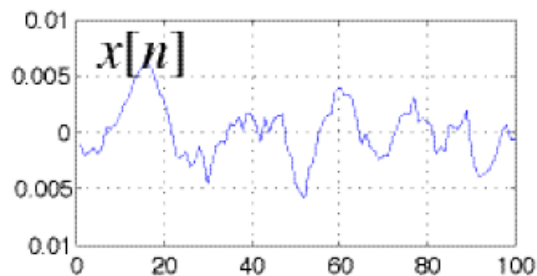
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- AC of any  $x[n]$
- AC of periodic
- Cross correlation

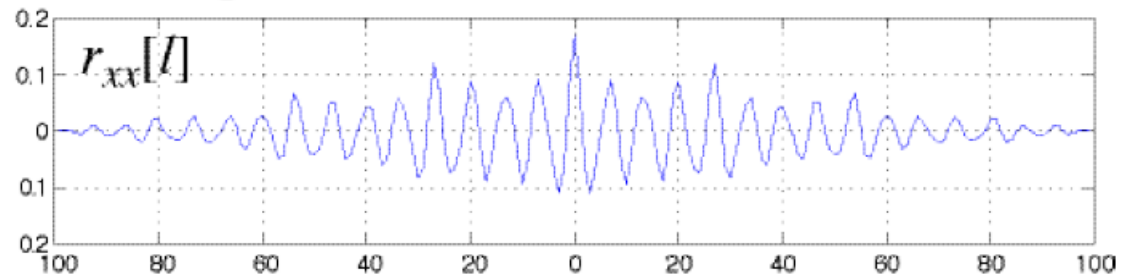
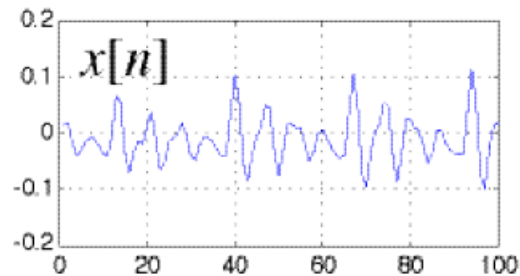


# What correlation looks like

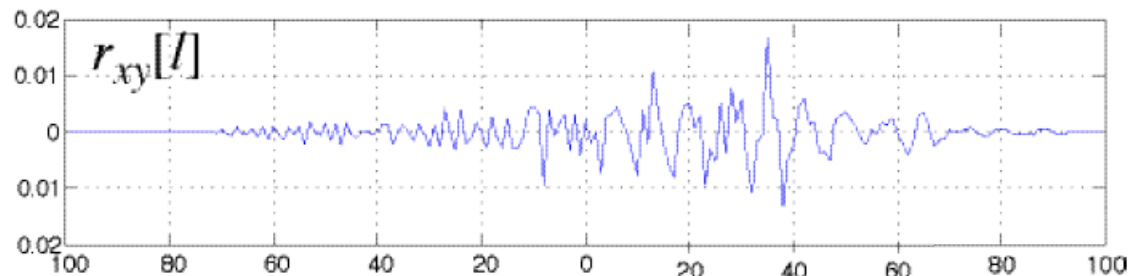
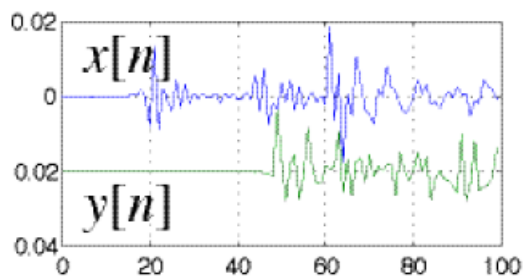
## Autocorrelation of generic signal



## Autocorrelation of near-periodic signal

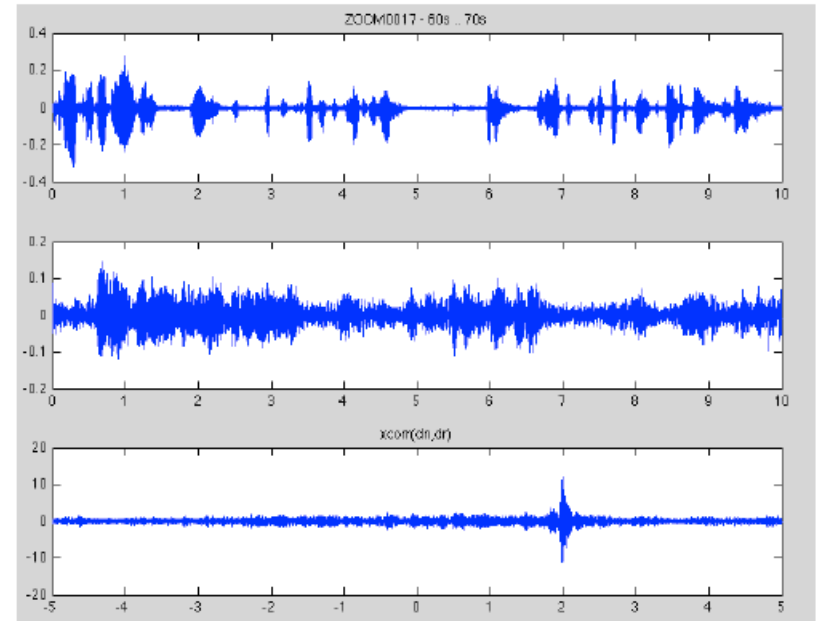
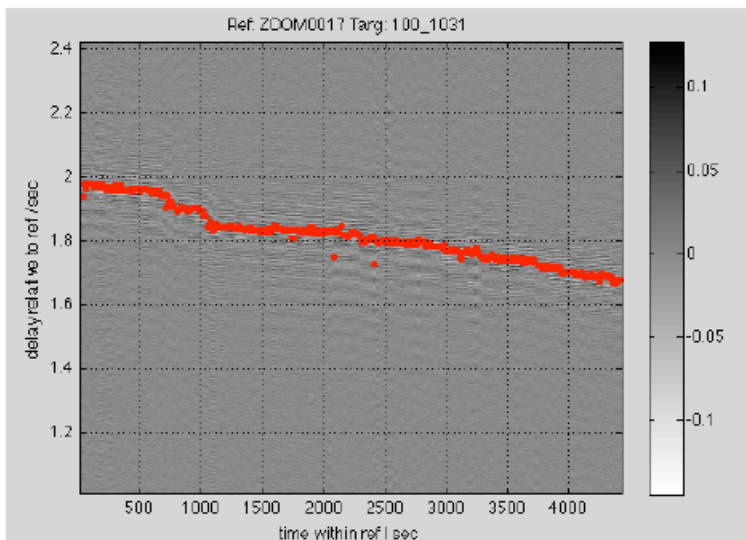


## Cross-correlation



# Correlation in action

- Close mic vs. video camera mic



- Short-time cross-correlation