

Lecture 02:

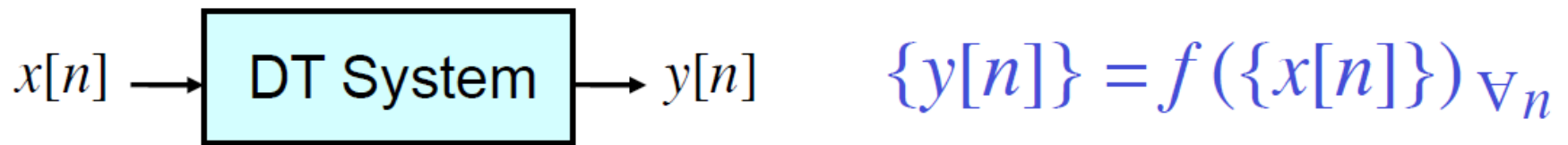
Discrete Time Systems (Part I)

Outlines

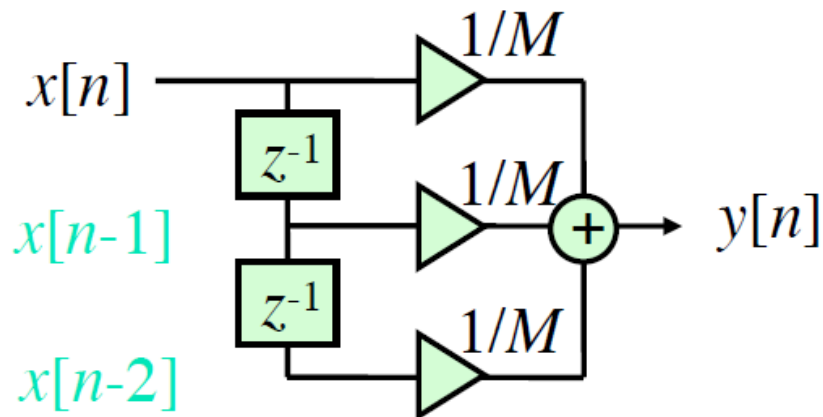
- ✓ 1. Discrete-time systems
- ✓ 2. Convolution
- 3. Linear Constant-Coefficient Difference Equations
(LCCDEs)
- 4. Correlation

1. Discrete-time systems

- A **system** converts input to output:



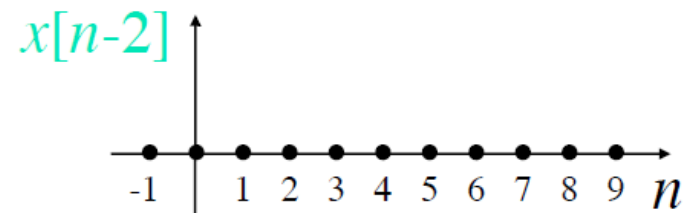
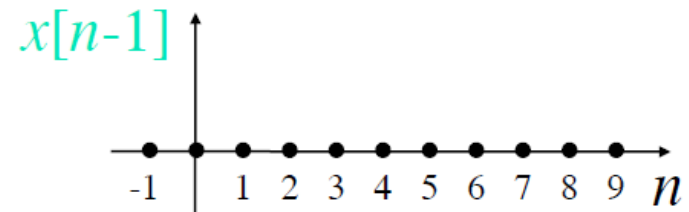
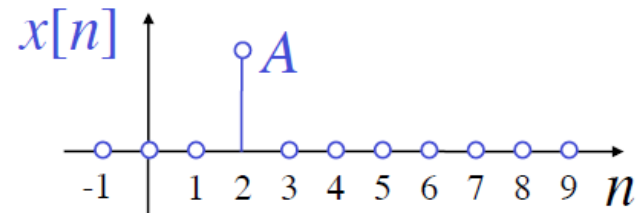
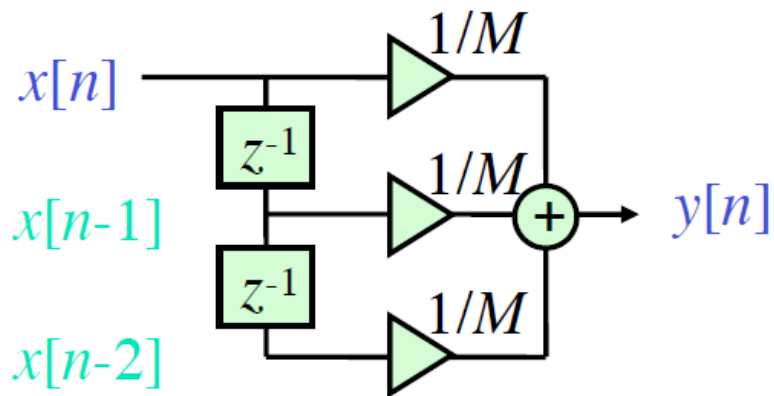
- E.g. **Moving Average (MA)**:



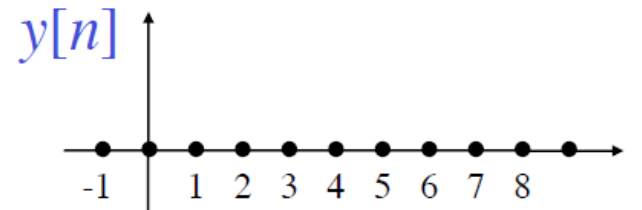
$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$$

$(M = 3)$

Moving Average (MA)



$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$$



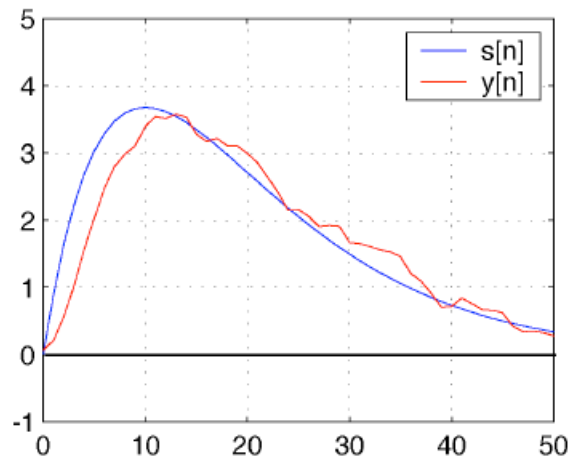
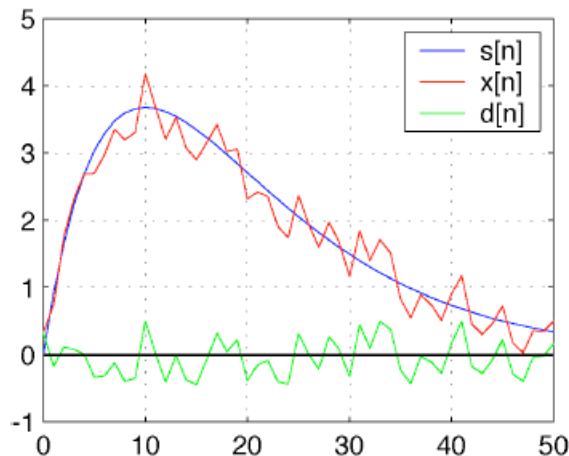
MA Smoother

- MA smoothes out rapid variations (e.g. “12 month moving average”)

- e.g. *signal noise*
 $x[n] = s[n] + d[n]$

$$y[n] = \frac{1}{5} \sum_{k=0}^4 x[n-k]$$

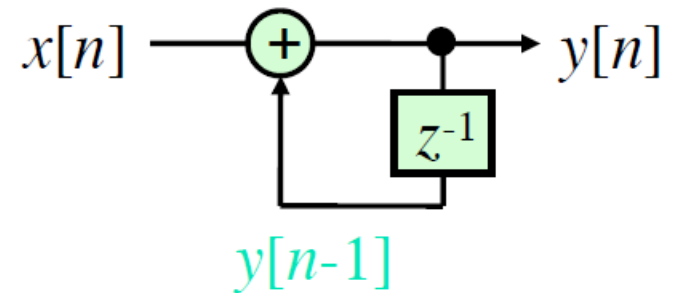
*5-pt
moving
average*



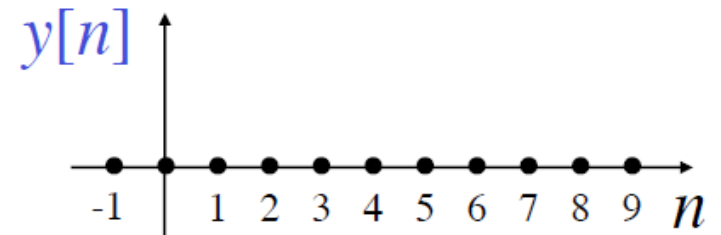
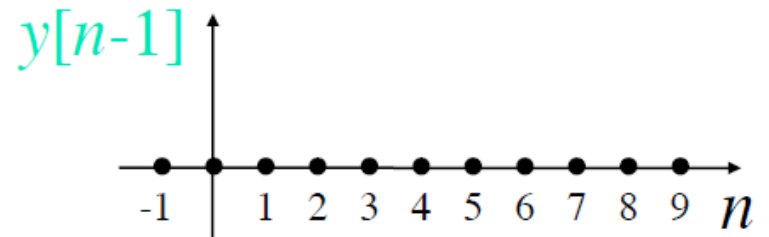
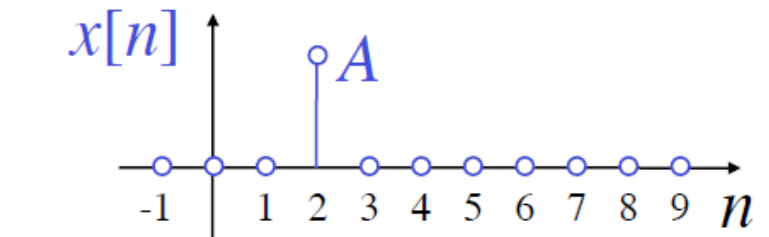
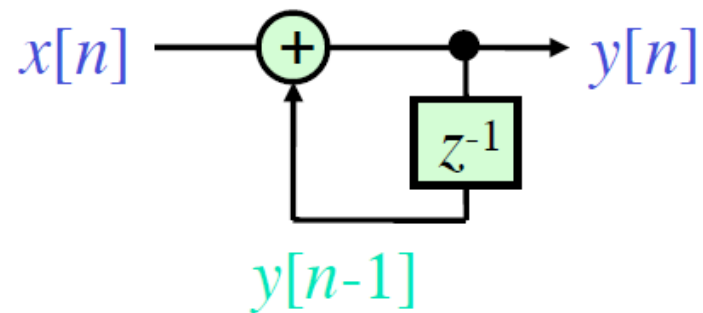
Accumulator

- Output accumulates all past inputs:

$$\begin{aligned} y[n] &= \sum_{\ell=-\infty}^n x[\ell] \\ &= \sum_{\ell=-\infty}^{n-1} x[\ell] + x[n] \\ &= y[n-1] + x[n] \end{aligned}$$

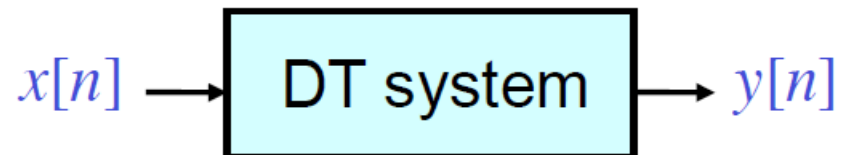


Accumulator



Classes of DT systems

- **Linear** systems obey **superposition**:



- if input $x_1[n] \rightarrow$ output $y_1[n]$, $x_2 \rightarrow y_2 \dots$
- given a linear combination of **inputs**: $x[n] = \alpha x_1[n] + \beta x_2[n]$
- then **output** $y[n] = \alpha y_1[n] + \beta y_2[n]$
for *all* α, β, x_1, x_2
i.e. same linear combination of **outputs**

Linearity: Example 1

■ Accumulator: $y[n] = \sum_{\ell=-\infty}^n x[\ell]$

$$x[n] = \alpha \cdot x_1[n] + \beta \cdot x_2[n]$$

$$\begin{aligned} \rightarrow y[n] &= \sum_{\ell=-\infty}^n (\alpha x_1[\ell] + \beta x_2[\ell]) \\ &= \sum_{\ell=-\infty}^n (\alpha x_1[\ell]) + \sum_{\ell=-\infty}^n (\beta x_2[\ell]) \\ &= \alpha \sum_{\ell=-\infty}^n x_1[\ell] + \beta \sum_{\ell=-\infty}^n x_2[\ell] \\ &= \alpha \cdot y_1[n] + \beta \cdot y_2[n] \end{aligned}$$

✓ Linear

Linearity Example 2:

- “Teager Energy operator”:

$$y[n] = x^2[n] - x[n-1] \cdot x[n+1]$$

$$x[n] = \alpha \cdot x_1[n] + \beta \cdot x_2[n]$$

$$\begin{aligned} \rightarrow y[n] &= (\alpha x_1[n] + \beta x_2[n])^2 \\ &\quad - (\alpha x_1[n-1] + \beta x_2[n-1]) \\ &\quad \cdot (\alpha x_1[n+1] + \beta x_2[n+1]) \\ &\neq \alpha \cdot y_1[n] + \beta \cdot y_2[n] \quad \mathbf{\times \textit{Nonlinear}} \end{aligned}$$

Linearity Example 3:

■ ‘Offset’ accumulator: $y[n] = C + \sum_{\ell=-\infty}^n x[\ell]$

$$\Rightarrow y_1[n] = C + \sum_{\ell=-\infty}^n x_1[\ell]$$

but $y[n] = C + \sum_{\ell=-\infty}^n (\alpha x_1[\ell] + \beta x_2[\ell])$

$$\neq \alpha y_1[n] + \beta y_2[n] \quad \text{X Nonlinear}$$

.. unless $C = 0$

Property: Shift (time) invariance

- **Time-shift** of input causes same shift in output
- i.e. if $x_1[n] \rightarrow y_1[n]$
then $x[n] = x_1[n - n_0]$
 $\Rightarrow y[n] = y_1[n - n_0]$
- i.e. process doesn't depend on absolute value of n

Shift-invariance counterexample

■ Upsampler: $x[n] \longrightarrow \boxed{\uparrow L} \longrightarrow y[n]$

$$y[n] = \begin{cases} x[n/L] & n = 0, \pm L, \pm 2L, \dots \\ 0 & \text{otherwise} \end{cases}$$

$$y_1[n] = x_1[n/L] \quad (n = r \cdot L)$$

$$x[n] = x_1[n - n_0]$$

$$\Rightarrow y[n] = x[n/L] = x_1[n/L - n_0]$$

$$= x_1 \left[\frac{n - L \cdot n_0}{L} \right] = y_1[n - L \cdot n_0] \neq y_1[n - n_0]$$

Not shift invariant

Another counterexample

$$y[n] = n \cdot x[n] \quad \text{scaling by time index}$$

- Hence $y_1[n - n_0] = (n - n_0) \cdot x_1[n - n_0]$

- If $x[n] = x_1[n - n_0]$
then $y[n] = n \cdot x_1[n - n_0] \neq$

- Not shift invariant
 - parameters depend on n

Linear Shift Invariant (LSI)

- Systems which are both **linear** and **shift invariant** are easily manipulated mathematically
- This is still a wide and useful class of systems
- If discrete index corresponds to time, called **Linear Time Invariant** (LTI)

Causality

- If **output** depends only on **past and current inputs** (not future), system is called **causal**
- Formally, if $x_1[n] \rightarrow y_1[n]$ & $x_2[n] \rightarrow y_2[n]$

$$\begin{aligned} \text{Causal} &\rightarrow x_1[n] = x_2[n] \quad \forall n < N \\ &\Leftrightarrow y_1[n] = y_2[n] \quad \forall n < N \end{aligned}$$

Causality example

- Moving average: $y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k]$

$y[n]$ depends on $x[n-k]$, $k \geq 0 \rightarrow$ **causal**

- ‘Centered’ moving average

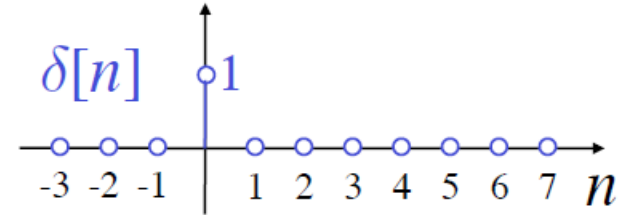
$$y_c[n] = y[n + (M-1)/2]$$

$$= \frac{1}{M} \left(x[n] + \sum_{k=1}^{(M-1)/2} x[n-k] + x[n+k] \right)$$

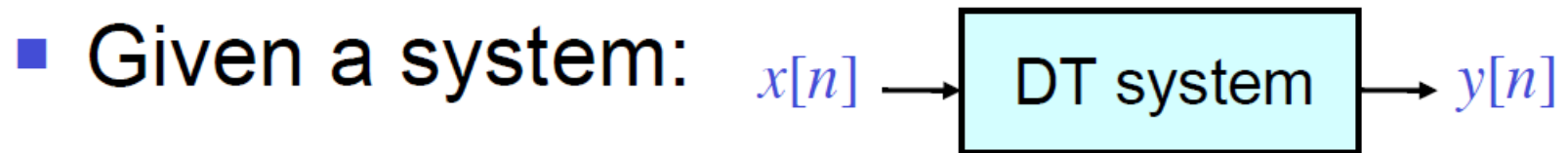
- .. looks **forward** in time \rightarrow **noncausal**
- .. Can make causal by **delaying**

Impulse response (IR)

- Impulse $\delta[n] = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$



(unit sample sequence)



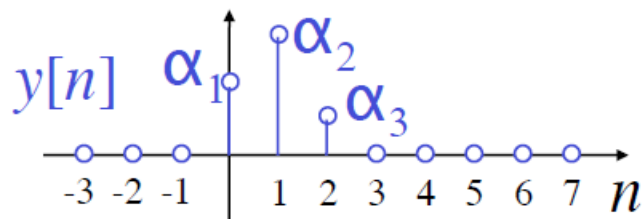
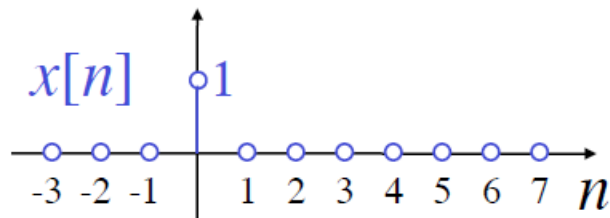
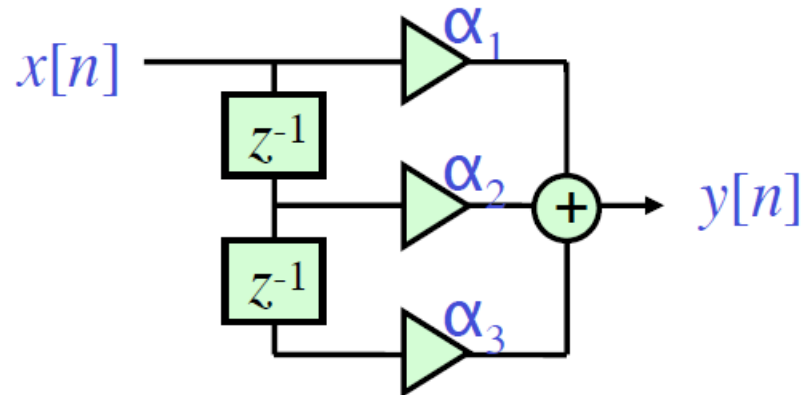
if $x[n] = \delta[n]$ then $y[n] \triangleq h[n]$

“impulse response”

- LSI system **completely specified** by $h[n]$

Impulse response example

- Simple system:



$x[n] = \delta[n]$ impulse



$y[n] = h[n]$ impulse response

2. Convolution

- Impulse response: $\delta[n] \rightarrow \boxed{\text{LSI}} \rightarrow h[n]$
- Shift invariance: $\delta[n-n_0] \rightarrow \boxed{\text{LSI}} \rightarrow h[n-n_0]$
- + Linearity: $\alpha \cdot \delta[n-k] + \beta \cdot \delta[n-l] \rightarrow \boxed{\text{LSI}} \rightarrow \alpha \cdot h[n-k] + \beta \cdot h[n-l]$
- Can express any sequence with δ s:
$$x[n] = x[0]\delta[n] + x[1]\delta[n-1] + x[2]\delta[n-2]..$$

Convolution sum

- Hence, since $x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$

- For LSI, $y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$ **Convolution sum**

written as $y[n] = x[n] \circledast h[n]$

- Summation is **symmetric** in x and h

i.e. $l = n - k \rightarrow$

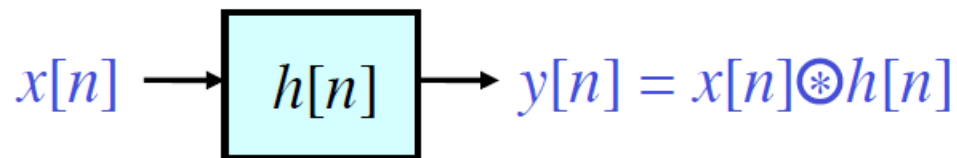
$$x[n] \circledast h[n] = \sum_{l=-\infty}^{\infty} x[n-l]h[l] = h[n] \circledast x[n]$$

Convolution properties

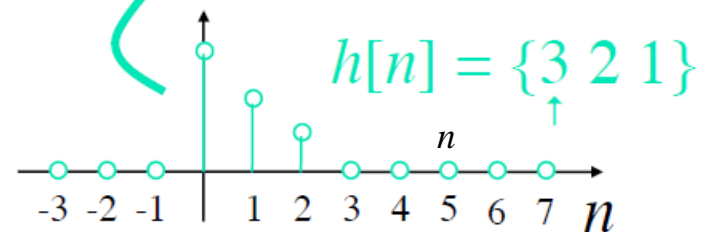
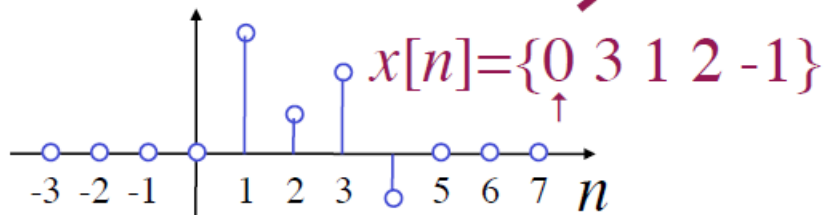
- **LSI System output** $y[n]$ **= input** $x[n]$ **convolved with impulse response** $h[n]$
→ $h[n]$ **completely describes system**
- **Commutative:** $x[n] \circledast h[n] = h[n] \circledast x[n]$
- **Associative:**
$$(x[n] \circledast h[n]) \circledast y[n] = x[n] \circledast (h[n] \circledast y[n])$$
- **Distributive:**
$$h[n] \circledast (x[n] + y[n]) = h[n] \circledast x[n] + h[n] \circledast y[n]$$

Interpreting convolution

- Passing a signal through a (LSI) system is equivalent to **convolving** it with the system's impulse response



$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

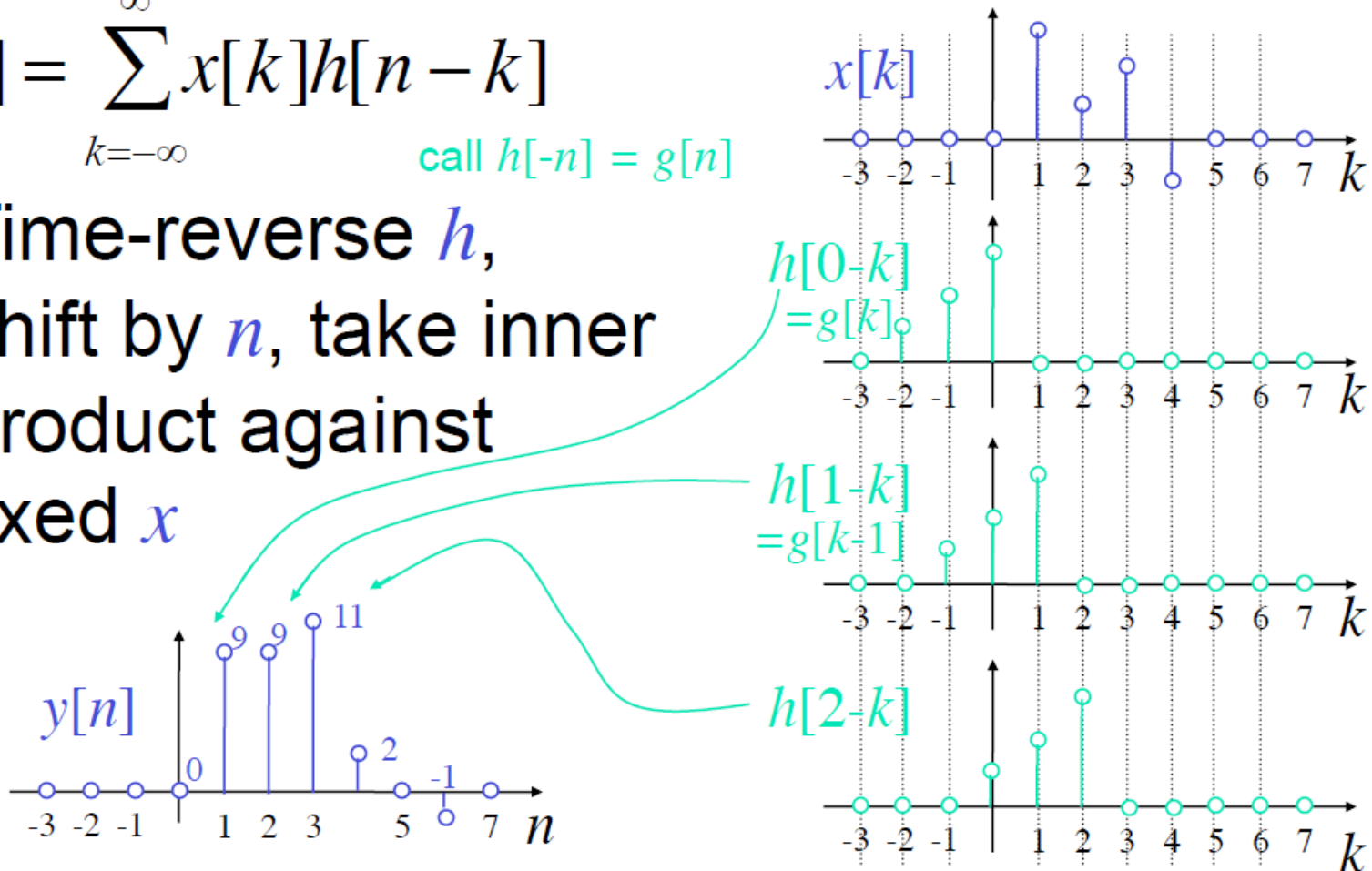


Convolution interpretation 1

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

call $h[-n] = g[n]$

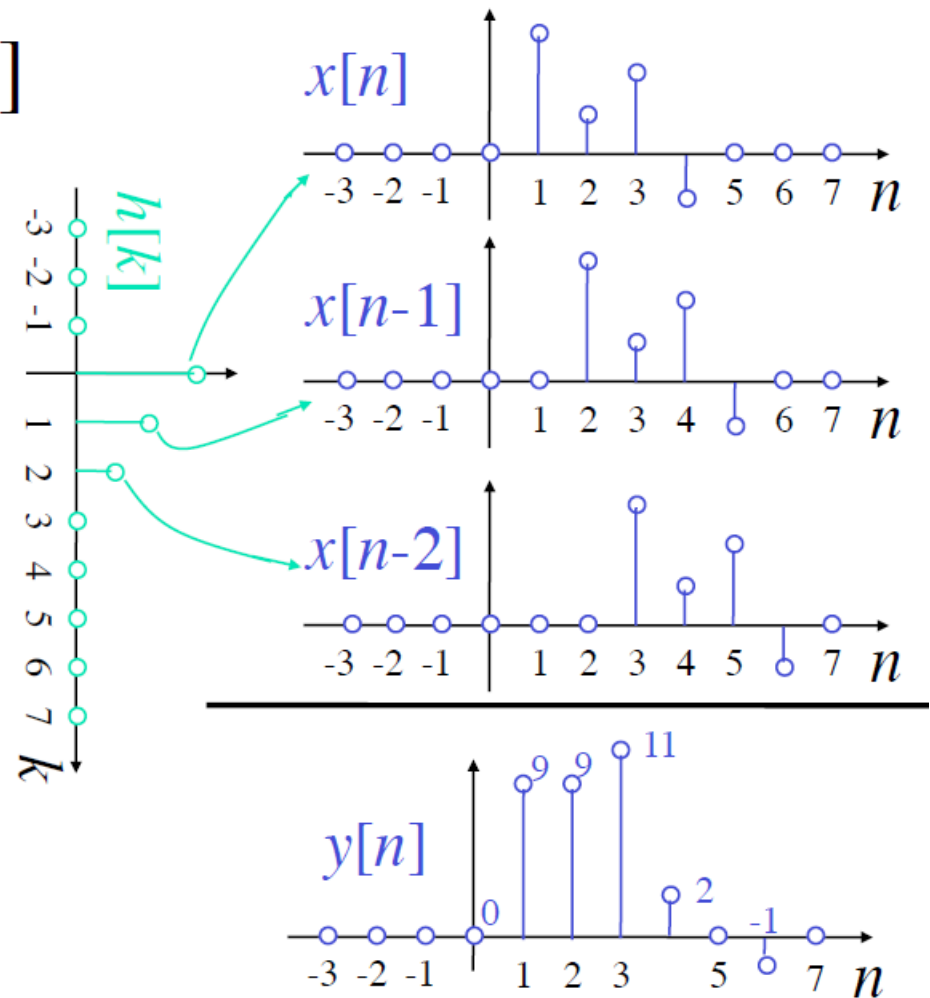
- Time-reverse h ,
shift by n , take inner
product against
fixed x



Convolution interpretation 2

$$y[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$$

- Shifted x 's weighted by points in h
- Conversely, weighted, delayed versions of h ...



Matrix interpretation

$$y[n] = \sum_{k=-\infty}^{\infty} x[n-k]h[k]$$

$$\begin{bmatrix} y[0] \\ y[1] \\ y[2] \\ \dots \end{bmatrix} = \begin{bmatrix} x[0] & x[-1] & x[-2] \\ x[1] & x[0] & x[-1] \\ x[2] & x[1] & x[0] \\ \dots & \dots & \dots \end{bmatrix} \begin{bmatrix} h[0] \\ h[1] \\ h[2] \end{bmatrix}$$

- **Diagonals** in **X** matrix are equal

Convolution notes

- Total nonzero length of convolving N and M point sequences is $N+M-1$
- Adding the indices of the terms within the summation gives n :

$$y[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k] \quad k + (n-k) = n$$

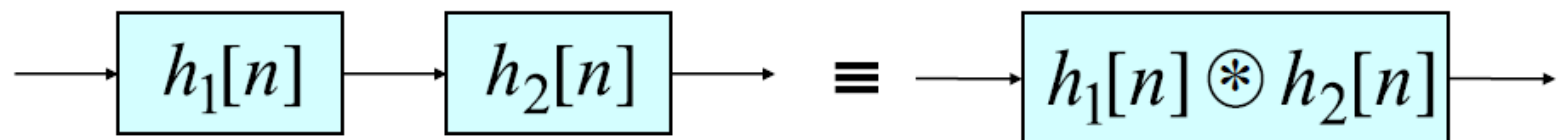
i.e. summation indices move in opposite senses

Convolution in MATLAB

- The M-file `conv` implements the convolution sum of two finite-length sequences
- If $a = [0 \ 3 \ 1 \ 2 \ -1]$
 $b = [3 \ 2 \ 1]$
then `conv(a,b)` yields
 $[0 \ 9 \ 9 \ 11 \ 2 \ 0 \ -1]$

Connected systems

- **Cascade** connection:



Impulse response $h[n]$ of the **cascade** of two systems with impulse responses $h_1[n]$ and $h_2[n]$ is $h[n] = h_1[n] \circledast h_2[n]$

- By commutativity,

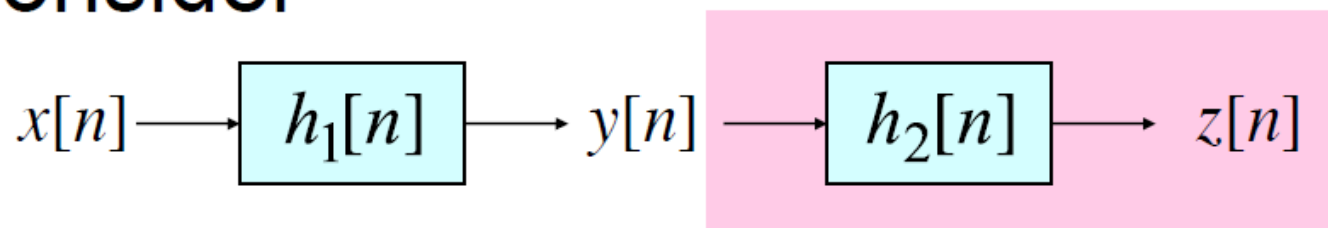


Inverse systems

- $\delta[n]$ is **identity** for convolution

i.e. $x[n] \circledast \delta[n] = x[n]$

- Consider



$$\begin{aligned} z[n] &= h_2[n] \circledast y[n] = h_2[n] \circledast h_1[n] \circledast x[n] \\ &= x[n] \quad \text{if} \quad h_2[n] \circledast h_1[n] = \delta[n] \end{aligned}$$

- $h_2[n]$ is the **inverse system** of $h_1[n]$

Inverse systems

- Use inverse system to **recover** input $x[n]$ from output $y[n]$ (e.g. to undo effects of transmission channel)
- Only sometimes possible - e.g. cannot 'invert' $h_1[n] = 0$
- In general, attempt to solve $h_2[n] \otimes h_1[n] = \delta[n]$

Inverse system example

- Accumulator:

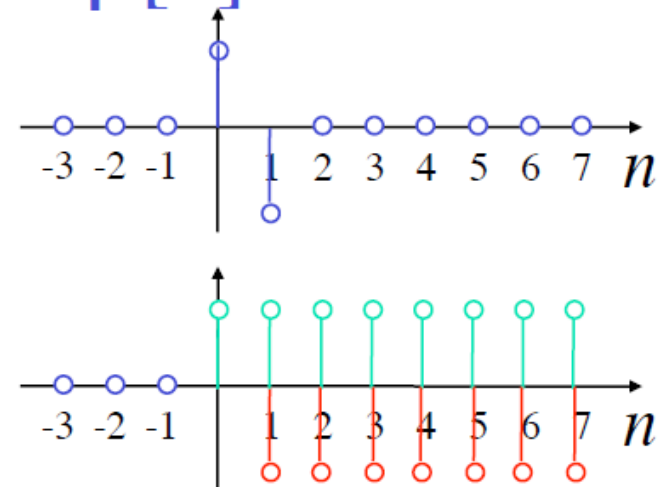
Impulse response $h_1[n] = \mu[n]$

- ‘Backwards difference’

$$h_2[n] = \delta[n] - \delta[n-1]$$

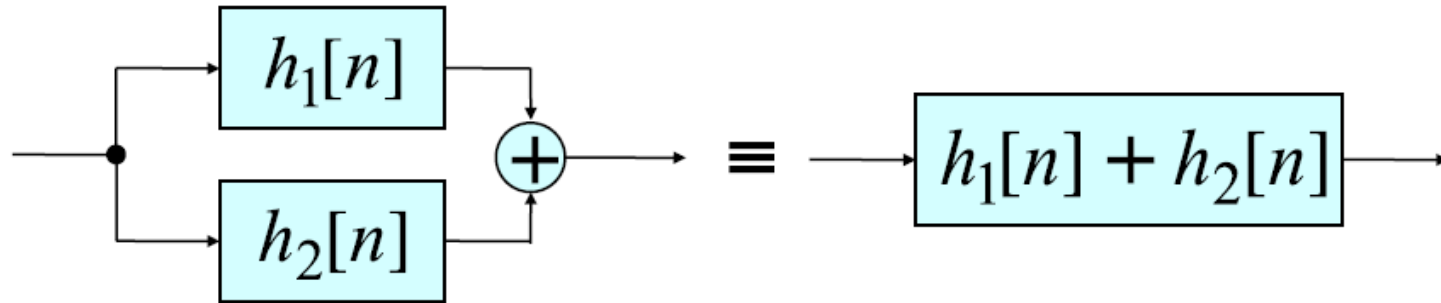
.. has desired property:

$$\mu[n] - \mu[n-1] = \delta[n]$$



- Thus, ‘backwards difference’ is inverse system of accumulator.

Parallel connection



- Impulse response of two parallel systems added together is:

$$h[n] = h_1[n] + h_2[n]$$