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# I Preliminaries

### 1.1 Review of Elementary Math

 $^*$ Sequence  $s_1, s_2, \ldots, s_n, \ldots$ 

Series is a sequence  $S_n$  where  $S_n = s_1 + s_2 + ... + s_n = \sum_{i=1}^n s_i$ 

### **Arithmetic Progression**

A sequence  $\{s_i\}$  is an arithmetic progression if there is d such that  $s_{n+1} - s_n = d \ \forall n$ 

Thus, we obtain

$$s_2 = s_1 + d$$

$$s_3 = s_2 + d = s_1 + 2d$$

. . .

$$s_n = s_{n-1} + d = s_1 + (n-1)d$$

In order to calculate the series  $S_n = \sum_{i=1}^n s_i$  we proceed as follows:

$$S_n = s_1 + s_2 + \ldots + s_n$$

Reversing it,

$$S_n = S_n + S_{n-1} + \ldots + S_1$$

By adding the above two expressions, we get

$$2S_n = (s_1 + s_n) + (s_2 + s_{n-1}) + \dots + (s_n + s_1)$$

$$= [s_1 + s_1 + (n-1)d] + [(s_1+d) + s_1 + (n-2)d] + \dots + [s_1 + (n-1)d + s_1]$$

$$= n[2s_1 + (n-1)d]$$

or

$$S_n = \frac{n(s_1 + s_n)}{2}$$

#### **Geometric Progression**

A geometric progression is a sequence of numbers  $s_1, s_2, ..., s_n, ...$  such that  $s_n = r s_{n-1}$ ; r is called the *common ratio*. Thus, we obtain

$$s_2 = rs_1$$

$$s_3 = rs_2 = r^2s_1$$

...

$$S_n = r^{n-1} S_1$$

In order to derive a closed formula for series  $S_n = \sum_{i=1}^n s_i$  we proceed as follows:

$$S_n = s_1 + s_2 + \ldots + s_n$$

$$rS_n = rs_1 + rs_2 + \dots + rs_n = s_2 + s_3 + \dots + s_{n+1}$$

Taking the difference, we get

$$S_n - rS_n = S_1 - S_{n+1}$$

$$S_n = \frac{s_1 - s_{n+1}}{1 - r} = \frac{s_1(1 - r^n)}{1 - r}$$

For |r| < 1, as  $n \to \infty$ , the sum converges to  $S_{\infty} = \frac{s_1}{1 - r}$ 

Also assuming 
$$s_1 = 1$$
, we get  $S_{\infty} = \frac{s_1}{1 - r} \Rightarrow S_{\infty} = \frac{1}{1 - r}$ 

#### **Power and Binomial Series**

#### Taylor series

A Taylor series is a series expansion of a function about a point  $x = x_0$  (sometimes written instead  $x = \alpha$ ).

$$f(x) = f(\alpha) + f'(\alpha)(x - \alpha) + \frac{f''(\alpha)(x - \alpha)^{2}}{2!} + \dots + \frac{f^{(n-1)}(\alpha)(x - \alpha)^{n-1}}{(n-1)!} + R_{n}$$

where  $R_n$  is a remainder term known as the Lagrange remainder, which is given by

$$R_n = \frac{f^{(n)}(x^*)(x-\alpha)^n}{n!} \text{ for some } x^* \in (\alpha, x).$$

If  $\alpha = 0$ , the expansion is known as a *Maclaurin series*.

#### Example 1.1

 $f(x) = \frac{1}{1-x}$ ; using the Taylor series above, we get

$$\frac{1}{1-x} = 1 + x + x^2 + \dots + x^n + \dots$$
 (geometric series)

For |x| < 1, the series converges as  $n \to \infty$ .

For |x| > 1, the series diverges.

Similarly, we can derive:

$$e^{x} = 1 + x + \frac{x^{2}}{2!} + \frac{x^{3}}{3!} + \dots = \sum_{i=0}^{\infty} \frac{x^{i}}{i!}$$

$$\sin x = x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + \dots$$

$$\cos x = 1 - \frac{x^2}{2!} + \frac{x^4}{4!} - \frac{x^6}{6!} + \dots$$

The last three series  $(e^x, \sin x, \cos x)$  converge for all values of x.

#### **Binomial Theorem**

$$(\alpha + b)^p = \alpha^p + \frac{p}{1!}\alpha^{p-1}b + \frac{p(p-1)}{2!}\alpha^{p-2}b^2 + \dots + b^p$$

$$= \sum_{n=0}^p \frac{p!}{(p-n)!n!}\alpha^{p-n}b^n \qquad \text{for every positive integer } p$$

where the numbers  $\binom{p}{n} = \frac{p!}{n!(p-n)!}$ , n = 0,1,...,p are referred to as binomial

coefficients.

#### Example 1.2 Binomial series

$$(1+x)^{p} = 1 + px + \frac{p(p-1)}{2!}x^{2} + \frac{p(p-1)(p-2)}{3!}x^{3} + \dots + x^{p}$$
$$= \sum_{n=0}^{p} \frac{p!}{(p-n)!n!}x^{n} = \sum_{n=0}^{p} \binom{p}{n}x^{n}$$

### **Multinomial Theorem**

$$(x_1 + x_2 + \dots + x_r)^n = \sum_{\substack{(n_1, n_2, \dots, n_r): \\ n_1 + n_2 + \dots + n_r = n}} \binom{n}{n_1 n_2 \dots n_r} x_1^{n_1} x_2^{n_2} \dots x_r^{n_r}$$

where the numbers  $\binom{n}{n_1 n_2 \dots n_r} = \frac{n!}{n_1! n_2! \dots n_r!}$  are referred to as multinomial coefficients.

#### Leibnitz's Formula

$$\frac{\partial^n}{\partial x^n} f(x)g(x) = \sum_{k=0}^n \binom{n}{k} f^{(k)}(x)g^{(n-k)}(x)$$

# 1.2 Permutations and Combinations

The number of ways to arrange N objects in N positions, is

$$N \times N - 1 \times N - 2 \times ... \times 2 \times 1 \equiv N!$$

The number of ways of arranging  $K \leq N$  objects chosen (from N objects) in K positions, without replacement, is

$$N \times N - 1 \times N - 2 \times ... \times (N - (K-1))$$

$$= N(N-1)(N-2)...(N-K+1) = \frac{N!}{(N-K)!}$$

With replacement allowed, the number of permutations is  $N^k$ 

Combination is the number of ways of selecting k objects out of N objects.

It is the same as number of ways of placing k out of N in k positions, divided by the number of ways k objects can be arranged in k positions, thus:

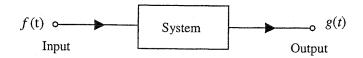
$$\binom{N}{k} = \frac{\frac{N!}{(N-k)!}}{k!} = \frac{N!}{(N-k)!k!}$$

#### 1.3 Transforms

#### Why

- 1) Naturally arise in formulation
- 2) Simplify calculations

## 1.3.1 Linear Time-Invariant Systems (LTI)



Without loss of generality, let f and g be functions of time, that is, f(t) and g(t)Notation:  $f(t) \rightarrow g(t)$  represents an input/output relationship.

#### Linearity

If  $f_1(t) \to g_1(t)$ , and  $f_2(t) \to g_2(t)$ , and  $a f_1(t) + b f_2(t) \to a g_1(t) + b g_2(t)$ , for a, b constants with respect to time, the system is said to be linear.

#### **Time Invariance**

If  $f(t) \to g(t)$  and  $f(t+\tau) \to g(t+\tau)$ , the system is said to be time-invariant.

#### Question

Which forms of f(t) passes through with no change in form except for scalar (time-invariant) multiplier, i.e.  $f(t) \rightarrow H f(t)$ 

Functions satisfying the above relation are called **characteristic functions**, **invariants**, or **eigenfunctions** for the LTI systems.

Let  $f_e(t)$  be the eigenfunction of a linear, time-invariant system. We can prove that  $f_e(t) = e^{st}$ , where s, in general, is a complex variable.

#### **Proof**

Let 
$$f_e(t) = e^{st} \rightarrow g_e(t)$$

By the linearity property,  $\alpha f_{\rm e}(t) \rightarrow \alpha g_{\rm e}(t)$ 

Let  $\tau$  be a constant time interval; thus  $\alpha = e^{s\tau}$  is also a constant. Then,

$$e^{s\tau}f_e(t) \to e^{s\tau}g_e(t)$$

$$e^{s\tau}e^{st} \rightarrow e^{s\tau}g_e(t)$$

$$e^{s(\tau+t)} \to e^{s\tau} g_e(t) \tag{1.3.1}$$

By the time-invariant property,

$$f_e(t+\tau) \to g_e(t+\tau)$$
 
$$e^{s(t+\tau)} \to g_e(t+\tau)$$
 (1.3.2)

The (unique) solution to (1.3.1) and (1.3.2) is

$$g_e(t) = He^{st}$$

where H is independent of t.

Thus, we have proved that the eigenfunction of a linear, time-invariant system is of the form  $f_{\epsilon}(t) = e^{st}$ . Thus,

$$e^{st} \to H(s)e^{st} \tag{1.3.3}$$

H(s) is called the system or transfer function.

# 1.3.2 Transform Method of Analysis

If we were able to decompose a function f(t) into complex exponentials whose sum or integral contributes to g(t) as above, we have a (simpler) method of calculating g(t) given f(t):

- 1. Decompose input into sum of exponentials
- 2. Compute response of each exponential as in (1.3.3)
- 3. Reconstitute output from sum of exponentials

#### 1.3.3 Discrete Functions of Time

 $f(t) \equiv f(t = nT)$ , where n is in  $\{..., -2, -1, 0, 1, 2, ...\}$  and T is a constant interval.

Notation:

$$f(nT) \equiv f_n$$

The input/output relationship can be expressed as  $f_n \rightarrow g_n$ 

### Linearity

$$\alpha f_n^{(1)} + b f_n^{(2)} \to \alpha g_n^{(1)} + b g_n^{(2)}$$

#### Time Invariance

$$f_{n+m} \to g_{n+m}$$

#### **Eigenfunctions**

$$f_n^{(e)} \equiv e^{st} \equiv e^{snT}$$

Let  $z \equiv e^{-sT}$ , where z is a complex variable, then  $f_n^{(e)} = z^{-n}$ 

#### Exercise

Show that  $z^{-n} \to H(z)z^{-n}$  (where H(z) is independent of n), thus proving that  $z^{-n}$  are indeed eigenfunctions for discrete, linear, time-invariant systems.

Again H(z) is called a *system* or *transfer function* of the LTI system. It gives the output we get when the input is one unit of the exponential. It defines the impact of the system on the exponential function.

**Unit Function (Kronecker Delta Function)** 

$$u_n = \begin{cases} 1, & n = 0 \\ 0, & \text{elsewhere} \end{cases}$$

• 1

$$n = -1$$
  $n = 0$   $n = 1$ 

The input/output relationship is  $u_n \to h_n$ , where  $h_n$  is called the *unit response*.

Using time invariance:

$$u_{n+m} \rightarrow h_{n+m}$$

And using linearity:

$$z^m u_{n+m} \to z^m h_{n+m}$$

Multiplying both sides of the relation above by  $z^{-n}z^n$  we get:

$$z^{-n}z^nz^mu_{n+m}\to z^{-n}z^nz^mh_{n+m}$$

$$z^{-n} \sum_{m} z^{n+m} u_{n+m} \longrightarrow z^{-n} \sum_{m} z^{n+m} h_{n+m}$$

But,  $\sum_{m} z^{n+m} u_{n+m} = 1$  by definition of  $u_n$ , so

$$z^{-n} \to z^{-n} \sum_{m} z^{n+m} h_{n+m}$$

or

$$z^{-n} \to z^{-n} \sum_{k} z^{k} h_{k}$$

Thus, the transfer function is

$$H(z) = \sum_{k} z^{k} h_{k}$$

H(z) is called the **z-transform** of  $h_k$ .

As we see from the above derivation, the system or transfer function can be determined from the unit response function  $h_n$ .

It is worth noting that only a single experiment (applying the unit function as input and measuring the resulting output  $h_n$ ) is sufficient to compute all responses of the system.

### 1.3.4 z-Transforms

It is also called **characteristic function** or **generating function**.

The z-transform maps a series  $f_n$  indexed by an integer value n into a function of a complex variable z.

Consider a sequence  $f_n$ , n = 0, 1, 2,...

The z-transform of  $f_n$  is defined by:

$$F(z) = \sum_{n=0}^{\infty} f_n z^n$$

For z = 1, we get

$$F(1) = \sum_{n=0}^{\infty} f_n$$

Notation: We denote the relationship between the transform pair as  $f_n \Leftrightarrow F(z)$ .

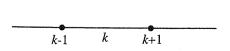
### Example 1.3

Let us consider the unit function  $u_n = \begin{cases} 1 & n=0 \\ 0 & n \neq 0 \end{cases}$ 

Then, the z-transform of  $u_n$  is  $F(z) = \sum_{n=0}^{\infty} u_n z^n = 1$ 

Using the notation we defined earlier, we can write  $u_n \Leftrightarrow 1$ 

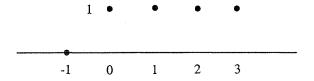
We can show that for a unit function shifted to the right k units, that is,  $u_{n-k} = \begin{cases} 1 & n = k \\ 0 & n \neq k \end{cases}$ 



we get  $u_{n-k} \Leftrightarrow z^k$ 

**Example 1.4** The (discrete) unit step function  $\delta_n$ 

$$\delta_n' = \begin{cases} 1 & n \ge 0 \\ 0 & n < 0 \end{cases}$$



Then, 
$$\delta_n \Leftrightarrow \sum_{n=0}^{\infty} z^n \delta_n$$

$$\delta_n \Leftrightarrow \sum_{n=0}^{\infty} z^n = 1 + z + z^2 + \dots$$

$$\delta_n \Leftrightarrow \frac{1}{1-z}$$

### Example 1.5 Geometric sequence

$$f_n = A \alpha^n, n = 0,1,2,...$$

Then, 
$$A\alpha^n \Leftrightarrow A\sum_{n=0}^{\infty} \alpha^n z^n$$

$$A\alpha^n \Leftrightarrow \frac{A}{1-\alpha z}$$

### Convolution: Important z - Transform Property

Consider the sequences  $f_n$  and  $g_n$  for n = 0, 1, 2,...

Denoting the convolution operator by  $\otimes$  the convolution is defined as

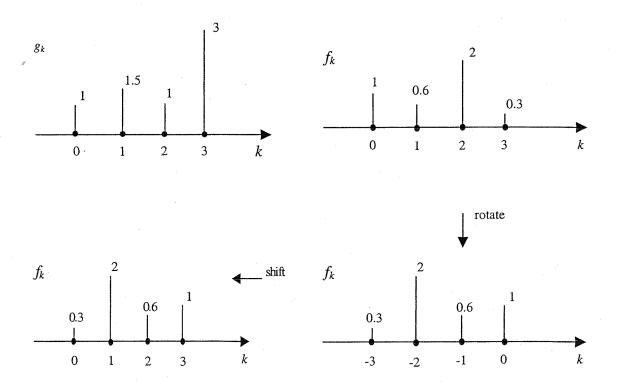
$$f_n \otimes g_n = \sum_{k=0}^n f_{n-k} g_k$$

z-transforms are useful in simplifying convolution calculations. Procedure:

- 1. Calculate F(z), G(z)
- 2. Perform inverse transformation of the product F(z)G(z)

### Example 1.6

$$f_3 \otimes g_3 = [(0.3 \times 1) + (2 \times 1.5) + (0.6 \times 1) + (1 \times 3)] = 0.3 + 3.0 + 0.6 + 3.0 = 6.9$$

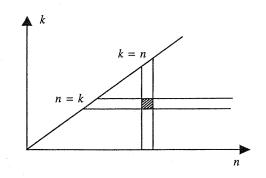


We can show that convolution operation corresponds to a product operation in the transfer domain.

$$f_n \otimes g_n \iff \sum_{n=0}^{\infty} (f_n \otimes g_n) z^n$$

$$= \sum_{n=0}^{\infty} \sum_{k=0}^{n} f_{n-k} g_k z^n = \sum_{k=0}^{\infty} \sum_{n=k}^{\infty} f_{n-k} g_k z^n = \sum_{k=0}^{\infty} g_k z^k \sum_{n=k}^{\infty} f_{n-k} z^{n-k}$$

$$f_n \otimes g_n \iff G(z) F(z)$$



### Other z-Transform Properties

$$F(z) = \sum_{k=0}^{\infty} f_k z^k = f_0 + f_1 z + f_2 z^2 + f_3 z^3 + \dots$$

1. 
$$F(0) = f_0$$

2. 
$$F'(1) = \sum_{k=0}^{\infty} k f_k$$

3. To show that  $f_n = \frac{1}{n!} \frac{\partial^n F(z)}{\partial z^n} \Big|_{z=0}$ , we proceed as follows:

$$\frac{\partial F(z)}{\partial z} = f_1 + 2f_2z + 3f_3z^2 + \dots$$

$$\left. \frac{\partial F(z)}{\partial z} \right|_{z=0} = f_1$$

$$\frac{\partial^2 F(z)}{\partial z^2} = 2f_2 + 6f_3 z + \dots$$

$$\frac{1}{2} \frac{\partial^2 F(z)}{\partial z^2} \bigg|_{z=0} = f_2$$

$$\frac{\partial^3 F(z)}{\partial z^3} = 6f_3 + \dots$$

$$\left. \frac{1}{6} \frac{\partial^3 F(z)}{\partial z^3} \right|_{z=0} = f_3$$

4. 
$$nf_n \Leftrightarrow \sum_{n=0}^{\infty} nf_n z^n = z \sum_{n=0}^{\infty} nf_n z^{n-1}$$

$$= z \frac{\partial}{\partial z} \sum_{n=0}^{\infty} f_n z^n = z \frac{\partial F(z)}{\partial z}, \text{ where } f_n \Leftrightarrow F(z)$$

5. 
$$\frac{1}{n!} \Leftrightarrow \sum_{n=0}^{\infty} \frac{1}{n!} z^n = e^z$$

6. 
$$n \Leftrightarrow \sum_{n=1}^{\infty} nz^n = z \sum_{n=1}^{\infty} nz^{n-1}$$

$$= z \frac{\partial}{\partial z} \sum_{n=1}^{\infty} z^n = z \frac{\partial}{\partial z} \frac{1}{1-z}$$

$$n \Leftrightarrow \frac{z}{(1-z)^2}$$

#### 1.3.5 Continuous, Linear, Time-Invariant Systems

$$f(t) \circ$$
 LTI Output

For continuous time, we have shown that the eigenfunctions are of the form  $e^{st}$  and  $e^{st} \to H(s)e^{st}$ 

Consider now 
$$u_0(t) = \begin{cases} 0 & t \neq 0 \\ \infty & t = 0 \end{cases}$$
 where  $\int_{-\infty}^{\infty} u_0(t) dt = 1$ 

This is the **Dirac delta function** or the **unit impulse function** (we shall see it in more detail later).

Consider applying a unit impulse function as input to the linear, time-invariant system. Let the output be the function h(t) usually called the **unit impulse response function**. Thus,  $u_0(t) \to h(t)$ .

Using time invariance, we can write  $u_0(t+\tau) \rightarrow h(t+\tau)$ 

And using linearity, we get

$$e^{-s\tau}u_0(t+\tau) \rightarrow e^{-s\tau}h(t+\tau)$$

also,

$$e^{-st}e^{st}e^{-s\tau}u_0(t+\tau) \rightarrow e^{-st}e^{st}e^{-s\tau}h(t+\tau)$$

$$e^{st}\int_{-\infty}^{\infty}e^{-s(t+\tau)}u_0(t+\tau)d\tau \to e^{st}\int_{-\infty}^{\infty}e^{-s(t+\tau)}h(t+\tau)d\tau$$

by 
$$\int_{-\infty}^{\infty} e^{-s(t+\tau)} u_0(t+\tau) d\tau \to e^{-s(t+\tau)} \Big|_{t+\tau=0} = 1$$

Thus,

$$e^{st} \rightarrow e^{st} \int_{-\infty}^{\infty} e^{-s(t+\tau)} h(t+\tau) d\tau$$

or by changing variables in the integral

$$e^{st} \to e^{st} \int_{-\infty}^{\infty} e^{-s\tau} h(\tau) d\tau$$

Thus

$$H(s) = \int_{-\infty}^{\infty} e^{-s\tau} h(\tau) d\tau$$

This relates the system transfer function to the unit impulse response function.

The above is also known as the Laplace transform. It transforms the time domain function  $h(\tau)$  to the complex domain function H(s).

### 1.3.6 Laplace Transforms

The Laplace transform maps a continuous function f(t), where f(t) = 0 for t < 0, into a function of a complex variable s. It is defined as follows:

$$F^*(s) \equiv \int_{0^-}^{\infty} f(t)e^{-st}dt$$

0 means any accumulation at 0 will be included in the integration.

$$F*(0) = \int_{0^{-}}^{\infty} f(t)dt$$

Many properties of z-transforms at z = 1 are equivalent to those of Laplace transforms at s = 0.

Notation:  $f(t) \Leftrightarrow F^*(s)$ 

Laplace transforms have a number of useful properties that make them a very practical tool for calculations that would otherwise be difficult in the time domain!

#### Example 1.7

Consider the function  $f(t) = \begin{cases} Ae^{-\alpha t} & t \ge 0 \\ 0 & t < 0 \end{cases}$ 

Then,

$$F^*(s) = A \int_0^\infty e^{-\alpha t} e^{-st} dt = A \int_0^\infty e^{-(\alpha+s)t} dt = A \left[ \frac{e^{-(\alpha+s)t}}{-(\alpha+s)} \right]_0^\infty = \frac{A}{\alpha+s}$$

Thus,

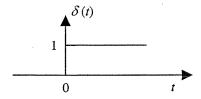
$$Ae^{-at}\delta(t) \Leftrightarrow \frac{A}{\alpha+s}$$

where  $\delta(t)$  is the unit step function.

#### Example 1.8 Continuous unit step function

Consider the function  $f(t) = \delta(t) = \begin{cases} 1 & t \ge 0 \\ 0 & t < 0 \end{cases}$ 

The Laplace transform of the unit step function is given by  $F^*(s) = \int_0^\infty \delta(t)e^{-st} dt$ 



Using the function of the previous example we view the  $\delta(t)$  as  $\delta(t) = Ae^{-\alpha t}$  with A = 1 and  $\alpha = 0$ . Then the Laplace transform is  $F*(s) = \frac{1}{s}$ 

Thus, 
$$\delta(t) \Leftrightarrow \frac{1}{s}$$

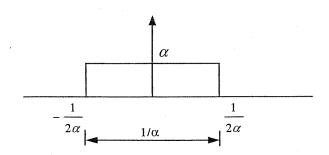
#### Example 1.9 Unit impulse function

The Dirac delta function provides a way for handling discontinuities

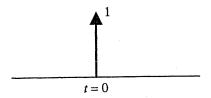
Consider 
$$f_{\alpha}(t) = \begin{cases} \alpha & |t| \le \frac{1}{2\alpha} \\ 0 & |t| > \frac{1}{2\alpha} \end{cases}$$
 and  $\int_{-\infty}^{\infty} f_{\alpha}(t)dt = 1$ 

The unit impulse function  $u_0(t)$  is obtained from  $f_{\alpha}(t)$  through  $\lim_{\alpha \to \infty} f_{\alpha}(t)$ . Note that,

$$\int_{-\infty}^{\infty} u_0(t)dt = 1.$$

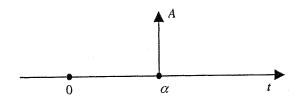


A graphical notation for  $u_0(t)$  is shown below:



# Time shift impact and amplitude change impact

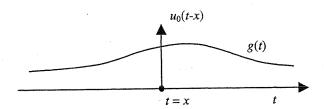
$$f(t) = Au_0(t-\alpha)$$



### Sampling or sifting property

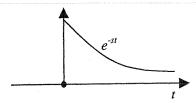
For an arbitrary differentiable function g(t)

$$\int_{-\infty}^{\infty} u_0(t-x)g(t)dt = g(x)$$



We can use the sifting property to derive the Laplace transform for  $u_0(t)$  as follows:

$$u_0(t) \Leftrightarrow \int_{0^-}^{\infty} u_0(t) e^{-st} dt = 1$$



Thus,  $u_0(t) \Leftrightarrow 1$ 

### 1.3.6.1 Properties of Laplace Transforms

We can show

1. 
$$f(t-\alpha) \Leftrightarrow e^{-\alpha s} F^*(s)$$

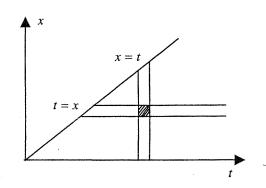
2. 
$$tf(t) \Leftrightarrow -\frac{dF^*(s)}{ds}$$

3. 
$$t^n f(t) \Leftrightarrow (-1)^n \frac{d^n F^*(s)}{ds^n}$$

4. 
$$f(t) \otimes g(t) \equiv \int_{0^{-}}^{t} f(t-x)g(x)dx \Leftrightarrow F^{*}(s)G^{*}(s)$$
 (Convolution)

**Proof** 

$$f(t) \otimes g(t) \Leftrightarrow \int_{t=0^{-}}^{\infty} \int_{x=0^{-}}^{t} f(t-x)g(x)dx e^{-st} dt$$



By changing variables

$$f(t) \otimes g(t) \Leftrightarrow \int_{x=0^{-}}^{\infty} \int_{t=x}^{\infty} e^{-st} f(t-x)g(x) dt dx$$

$$= \int_{x=0^{-}}^{\infty} \int_{t=x}^{\infty} e^{-s(t-x)} f(t-x) dt \ g(x) e^{-sx} dx$$

$$= \int_{x=0^{-}}^{\infty} g(x) e^{-sx} dx \int_{t=x}^{\infty} e^{-s(t-x)} f(t-x) dt$$

$$= G * (s) \int_{\tau=0^{-}}^{\infty} e^{-s\tau} f(\tau) d\tau$$

$$= G * (s) F * (s)$$

# 1.3.7 Using Transforms to Solve Difference and Differential Equations

#### Example 1.10

Consider the difference equation

$$2f_n = 3f_{n-1} - f_{n-2}$$
 for  $n = 2, 3, ...$ 

with initial conditions:  $f_0 = 1$  and  $f_1 = 2$ . We want to find  $f_n$  that satisfies the above difference equation. We will use z-transform to determine  $f_n$ .

Solution

Multiplying both sides by  $z^n$  and summing we get

$$2\sum_{n=2}^{\infty} f_n z^n = 3\sum_{n=2}^{\infty} f_{n-1} z^n - \sum_{n=2}^{\infty} f_{n-2} z^n$$

$$2 [F(z) - f_0 - f_1 z] = 3z [F(z) - f_0] - z^2 F(z)$$

$$F(z) (2 - 3z + z^2) = f_0 (2 - 3z) + 2f_1 z$$

Using initial conditions we get

$$F(z) = \frac{2 - 3z + 4z}{z^2 - 3z + 2} = \frac{2 + z}{(2 - z)(1 - z)}$$

But 
$$F(z) = \frac{2+z}{(2-z)(1-z)} = \frac{A}{2-z} + \frac{B}{1-z}$$

$$A = \left[ (2-z)F(z) \right]_{z=2} = \frac{2+z}{1-z} \bigg|_{z=2} = -4$$

$$B = [(1-z)F(z)]_{z=1} = \frac{2+z}{2-z}\Big|_{z=1} = 3$$

Thus

$$F(z) = \frac{3}{1-z} - \frac{4}{2-z}$$

By inspection we can do the inverse z-transform to obtain:

$$f_n = 3(1)^n - 2\left(\frac{1}{2}\right)^n$$
 for  $n = 0, 1, 2, ...$ 

or

$$f_n = 3 - 2\left(\frac{1}{2}\right)^n$$

# **II** Introduction to Probability Theory

Provides mathematical models to analyze random phenomena and random behavior.

Random: Unpredictable or irregular

#### Example 2.1

Tossing a coin once is a random event because we cannot predict the outcome of a trial.

**Statistical Regularity**: Applies to behavior about which one can make accurate statements of prediction for a large collection of trials.

#### Example 2.2

Tossing a coin a large number of times will yield about one-half of the times heads and the other half tails, if the coin is fair. When we say a coin is fair, we mean that when it is flipped, it is equally likely to come up heads or tails.

# 2.1 Experiments and Their Corresponding Probability Model

Experiment	Model	Example	
Set of Outcomes	Sample Space	Toss a die	
	Let $S = \{ \omega : \omega \text{ corresponds to an outcome} \}$	1 2	
		3 4	
		5 6	
Set of Results	Set of <b>Events</b> , $E = \{A_1, A_2,\}$ where		
	$A_i \subseteq S$ subset of $S$	$1 \sqrt{2}$	
		3 4	
		5 6	
		A = [even number]	
Relative frequency	Probability Measure	$P[A] = \frac{1}{2}$	
	Mapping from $E$ into real line, with the	2	
	following properties:	3	
	$(1) P[A] \ge 0$		
	(2) P[S] = 1		
	(3) $P[A \cup B] = P[A] + P[B] - P[A \cap B]$		

# 2.2 Sample Spaces

### Example 2.3

Toss a coin twice;

 $S = \{(H, H), (H, T), (T, H), (T, T)\}$ 

#### Example 2.4

Observe 2 devices at t = 120 minutes

Two possible sample spaces:

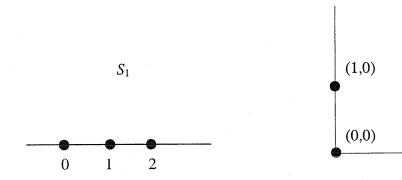
 $S_1 = \{0, 1, 2\}$  (where 0 = both down, 1 = one up, and 2 = both up)

 $S_2 = \{(0, 0), (0, 1), (1, 0), (1, 1)\}\ (0 \Rightarrow \text{device down}, 1 \Rightarrow \text{device up})$ 

 $S_2$ 

(1,1)

(0,1)



### Example 2.5

Observe *n* devices at  $t = \tau$ ;

$$S = \{(0, 0, ..., 0), (0, 0, ..., 1), (0, 0, ..., 1, 0), ...\}$$

This is an n-dimensional space with  $2^n$  sample points.

#### Example 2.6

Lifetime of a car;

 $S = [0, \infty)$ 

#### Example 2.7

Check devices coming out of an assembly line; continue until a defective device is found

$$S = \{1, 2, 3, ..., \infty\}$$

### **Sample Space Classification**

(a) Finite (Examples 2.3, 2.4, 2.5)

Outcomes are countable (one-to-one correspondence with natural numbers) and finite.

(b) Discrete (Example 2.7)

Outcomes are countable but infinite.

(c) Continuous (Example 2.6)

Uncountable or non-denumerable; sample space constitutes a continuum on some interval of the real line.

## 2.3 Events and Their Algebra

An **event** A is a subset of the sample space S,  $A \subseteq S$ 

If  $s \in S$ ,  $s \in A$ , and s is the outcome of the trial, we say event A has occurred. That is, it is sufficient that the outcome of the trial corresponds to one element of A to say that A has occurred.

### Example 2.8

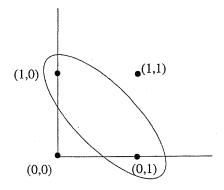
 $A \equiv [\text{outcome of tossing a die is a power of 2}]$ 



### Example 2.9

Observe two devices at  $t = \tau$ 

Let 0 means device is down and 1 means device is up.



 $S = \{(0, 0), (0, 1), (1, 0), (1, 1)\}$  is the sample space.

Let us define the event

A = [exactly one device is up]

Then,

 $A^C = [0 \text{ or both devices are up}]$ 

and

 $A \cup A^C = S$  (S which is also called the **universal** event)

Let

 $B = \{(0,0)\} \equiv [\text{no devices up}];$ 

 $\boldsymbol{B}$  is called an **elementary** event since it is made up of exactly one sample element.

 $S^C = \emptyset$  where  $\emptyset$  is called the **null** event

In this example, there are 16 subsets of S, including  $\emptyset$ . Each subset corresponds to an event. Note that, here, the cardinality of the set S and its power set is |S| = 4 and  $|E| = 2^4 = 16$  respectively.

### Example 2.10

Let  $S = [0, \infty)$  represents the lifetime of a car (Jaguar?)

Let  $A = [\text{car lasts at least 4 years}] = \{x: x \ge 4\} = [4, \infty)$ 

**Example 2.11** System with five tape drives is observed at time  $t = \tau$ 

Let 1 means device is available and 0 means device is in use (occupied).

 $S = \{(n_1, n_2, n_3, n_4, n_5): n_i = 0 \text{ or } 1 \text{ depending on the status of the drive}\}$ 

$$|S| = 2^5 = 32$$
; and  $|E| = 2^{32}$ .

Let  $E_1 \equiv [\text{at least 4 drives are available}] = \{(1, 1, 1, 1, 1), (1, 1, 1, 1, 0), (1, 1, 1, 0, 1), (1, 1, 0, 1, 1, 1), (1, 0, 1, 1, 1), (0, 1, 1, 1, 1)\}$ 

If we write each bit string (event) as binary integers between 0 and 31 (i.e. (0, 0, 0, 0, 0) is  $S_0$  (0, 0, 0, 0, 1) is  $S_1, \ldots (1, 1, 1, 1, 1)$  is  $S_{31}$ ), we can rewrite  $E_1$  as follows:

$$E_1 = \{S_{15}, S_{23}, S_{27}, S_{29}, S_{30}, S_{31}\}$$

 $E_1^C = [\text{at most 3 drives are available}] = \{S_0 \to S_{14}, S_{16} \to S_{22}, S_{24} \to S_{26}, S_{28}\}.$ 

### 2.3.1 Event Intersection

 $A_1 \cap A_2 = \{ \omega \colon \omega \in A_1 \text{ and } \omega \in A_2 \}$ 

$$\bigcap_{i=1}^n A_i \equiv A_1 \cap A_2 \cap ... \cap A_n$$

In example 2.11 above, set  $E_2 \equiv [\text{at most 4 drives are available}].$ 

Then,  $E_2 = \{S_0 \to S_{30}\}$  and  $E_2^C = \{S_{31}\}$ .

Recall,  $E_1 \equiv$  [at least 4 drives are available]. Then,  $E_1 \cap E_2 \equiv$  [exactly 4 drives are available]. We can also define a set  $E_3 = E_1 \cap E_2 = \{S_{15}, S_{23}, S_{27}, S_{29}, S_{30}\}.$ 

#### 2.3.2 Event Union

 $A_1 \cup A_2 = \{ \omega : \omega \in A_1 \text{ or } \omega \in A_2 \}$ 

$$\bigcup_{i=1}^{n} A_{i} \equiv A_{1} \cup A_{2} \cup ... \cup A_{n}$$

In example 2.11 above, set  $E_4 \equiv$  [tape drive 1 is available]. Then,  $E_4 = \{S_{16} \rightarrow S_{31}\}$ . Recall,  $E_1 = \{S_{15}, S_{23}, S_{27}, S_{29}, S_{30}, S_{31}\}$ . We can define set  $E_5 = E_1 \cup E_4 = \{S_{15} \rightarrow S_{31}\}$ . Then,  $E_5 \equiv$  [drive 1 or at least 4 drives are available].

### 2.3.3 Cardinality

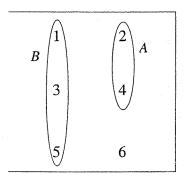
It is the number of elements in an event.

Note that  $|E_5| \le |E_1| + |E_4|$  in the above case.

#### 2.3.4 Mutual Exclusion

Events A and B are **mutually exclusive** if they cannot occur in the same trial. That is, for such events  $A \cap B = \emptyset$ .

#### Example 2.12 Toss a die



 $A \equiv [power of 2 > 1]$ 

 $B \equiv [odd]$ 

 $A \cap B = \emptyset$ .

#### **Extension of Definition**

Events  $A_1, A_2, ..., A_n$  are mutually exclusive if and only if (iff)  $A_i \cap A_j = \emptyset$ ,  $\forall i \neq j$ .

#### 2.3.5 Exhaustive Events

Events *A* and *B* are **exhaustive** iff  $A \cup B = S$ .

Also,  $\bigcup_{i=1}^{n} A_i = S \iff \{A_i\}$  is a set of exhaustive events.

# 2.3.6 Sample Space Partition

Let a set of events  $A_i$ , i = 1, 2, ..., n be a set of mutually exclusive and exhaustive (M.E events. We say that  $A_i$ , i = 1, 2, ..., n define a **partition** of the sample space.

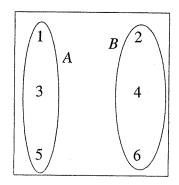
### Example 2.13

Consider once again tossing a die.

 $A \equiv [odd]$ 

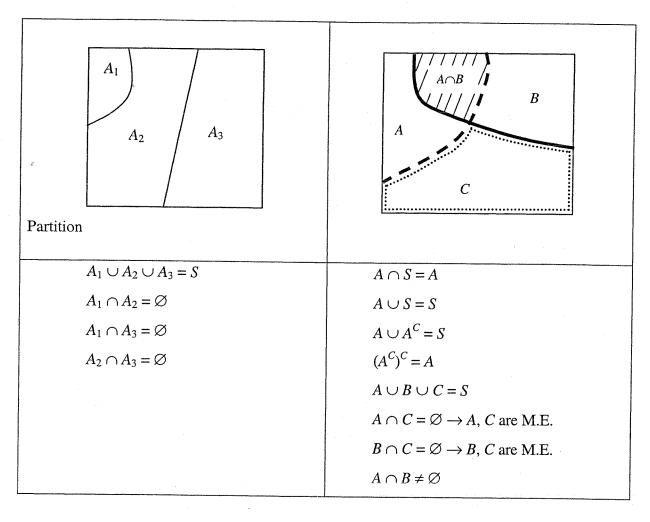
 $B \equiv [even]$ 

A, B partition the sample space S



## Venn Diagrams

A sometimes useful representation of (S, E), the sample space and the set of events defir on it.



### **Tree Representation**

Useful for sequential sample spaces.

### Example 2.14

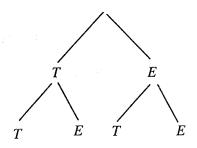
If C, then T else E

Let the trial be two executions of the if statement above.

The outcome of each trial is a pair of the following sample space:

 $S = \{(T,\,T),\,(T,\,E),\,(E,\,T),\,(E,\,E)\}$ 

Each event is a path from the root to some leaf in the tree below:



### **Coordinate System**

Useful in a 2-dimensional system representation

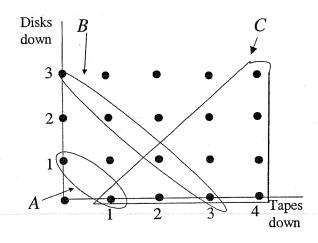
### Example 2.15 4 tape drives, 3 disk drives

At time  $t = \tau$ , how many of each type of drive are down?

 $A \equiv [$ one device is down]

 $B \equiv [3 \text{ devices are down}]$ 

 $C \equiv [\text{more tape drives are down than disk drives}].$ 



### 2.4 Probability Measure

Mapping from set of events E into [0,1].

P[A] represents the relative likelihood that A would occur. P[A] is assigned to A based on intuition, assumed, or based on experiments.

#### **Axioms (or Conditions)**

- (i)  $P[A] \ge 0$ : nonnegative real number
- (ii) P[S] = 1: normalization effect  $\Rightarrow P[A] \le 1$
- (iii)  $P[A \cup B] = P[A] + P[B] P[A \cap B]$  in general  $P[A \cup B] = P[A] + P[B]$  when A and B are mutually exclusive.

#### **Corollaries**

a) 
$$P\left[\bigcup_{i=1}^{n} A_i\right] = \sum_{i=1}^{n} P\left[A_i\right]$$
 for  $\{A_i\}$  mutually exclusive events.

b) 
$$P[S] = 1 = P[A \cup A^C] = P[A] + P[A^C]$$
  
 $\Rightarrow P[A] = 1 - P[A^C]$ 

Can similarly show that  $P[\emptyset] = 0$ .

#### Method to Calculate P [A]

 $S = \{s_i : s_i \text{ is an outcome}\}\$ 

Recall that  $s_i$  are called elementary events.  $P[s_i]$  is assigned by intuition or experimentation and  $\{s_i\}$  is a set of mutually exclusive events.

$$P[A] = \sum_{i=1}^{n} P[s_i]$$
 for  $s_i \in A$ .

In many cases  $P[s_i] = p$ , i.e. all outcomes have same probability p. The remaining task is to count the number of outcomes that belong to some event  $A_j$ . This leads to the need for combinatorics.

### Example 2.16

A box contains 75 good Integrated Circuits (IC) and 25 bad integrated circuits. Twelve ar pulled at random. The condition of any chip is independent from the condition of all othe chips. All chips in the box are equally likely to be pulled. What is *P* [at least one IC in the 1: pulled is defective]?

Solution

Sample Space  $S = \{\text{all possible combinations of } 12 \text{ chips} \}$ 

$$|S| = \binom{100}{12}$$

Let  $E \equiv [at least one IC of those pulled is defective]$ 

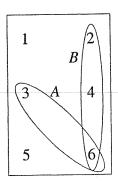
Then,  $E^C \equiv [\text{no IC pulled is defective}]$ 

$$|E^c| = \binom{75}{12}$$

$$P\left[E^{C}\right] = \frac{\binom{75}{12}}{\binom{100}{12}}$$

$$P\left[E\right] = 1 - P\left[E^C\right]$$

# Example 2.17 Event union and intersection



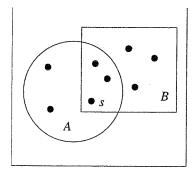
Toss a die.

Let  $A \equiv [divisible by 3]$ , and

 $B \equiv [even].$ 

 $P[A \cup B] = P[A] + P[B] - P[A \cap B] = 1/3 + 1/2 - 1/6 = 2/3.$ 

### 2.5 Conditional Probabilities



Let  $P[s \mid B] \equiv P$  [outcome corresponding to s occurs given that event B has occurred]

$$P[s \mid B] = \begin{cases} \frac{P[s]}{P[B]}, & s \in B \\ 0, & s \notin B \end{cases}$$

$$P[A \mid B] = \sum_{s \in A \cap B} P[s \mid B]$$

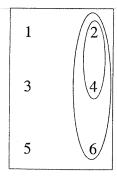
or

$$P[A \mid B] = \frac{P[A \cap B]}{P[B]}$$

Knowledge that B occurred changes the effective sample space.

Division by P[B] in  $P[s \mid B] = \frac{P[s]}{P[B]}$  is necessary for normalization, i.e.  $\sum_{s \in B} P[s \mid B] = 1$ .

### Example 2.18 Toss a die once



$$P[\text{power of two} > 1 | \text{even}] = \frac{P[\text{power of two} > 1 \text{ and even}]}{P[\text{even}]} = \frac{\frac{1}{3}}{\frac{1}{2}} = \frac{2}{3}$$

Example 2.19 Box of 5000 IC's

1000 from manufacturer X

4000 from manufacturer Y

10% of chips from X are defective

5% of chips from Y are defective

What is P [a chip picked at random was from X | defective]?

Solution

Let A be the event [picked chip was from X]

and B be the event [defective chip]. Then,

$$P[A \mid B] = \frac{P[A \cap B]}{P[B]} = \frac{\frac{100}{5000}}{\frac{300}{5000}} = \frac{1}{3}$$

# 2.6 Independent Events

Given that B occurred, P[A] may increase, decrease, or remain the same.

#### **Definition**

A and B are independent iff P[A|B] = P[A].

From the definition of conditional probability, it follows that:

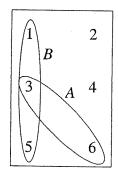
$$P[A \mid B] = \frac{P[A \cap B]}{P[B]} = P[A]$$

 $\Rightarrow$   $P[A \cap B] = P[A]P[B]$  for independent events A, B.

#### **Notes**

- 1. Independence is *not* transitive; i.e. if A and B are independent events and B and C are independent, it does *not* follow that A and C are independent. They may or may not be.
- 2.  $A_1, A_2, \ldots, A_n$  are mutually independent events **iff all subsets of all sizes** are mutually independent.

### Example 2.20 Toss a fair die



Let

 $A \equiv [\text{outcome is divisible by 3}], \text{ and }$ 

 $B \equiv [outcome is odd]$ 

$$P[\text{divisible by 3}|\text{odd}] = \frac{P[A \cap B]}{P[B]} = \frac{\frac{1}{6}}{\frac{1}{2}} = \frac{1}{3}.$$

By 
$$P[A] = \frac{1}{3} \Rightarrow A$$
,  $B$  are independent.

### Example 2.21

Toss 2 fair dice. Each outcome of the possible 36 outcomes is equally likely.

	1,1	2,1	3,1	4,1	5,1	6,1
	1,2	2,2	3,2	4,2	5,2	6,2
	1,3	2,3	3,3	4,3	5,3	6,3
	1,4	2,4	3,4	4,4	5,4	6,4
	1,5	2,5	3,5	4,5	5,5	6,5
0	1,6	2,6	3,6	4,6	5,6	6,6

#### Note

[first = 4] reduces the effective sample space to 6 possible outcomes that are all equally likely.

Consider the following events:

 $A \equiv [\text{sum of two dice results} = 6]$ 

 $B \equiv [first dice result = 4]$ 

 $C \equiv [\text{sum of two dice results} = 7]$ 

Determine whether A, B are independent events

Determine whether B, C are independent events

$$P[\text{sum} = 6 | \text{first} = 4] = \frac{P[\text{sum} = 6 \text{ and first} = 4]}{P[\text{first} = 4]} = \frac{\frac{1}{36}}{\frac{6}{36}} = \frac{1}{6}$$

P [sum = 6] = 5/36.

Thus, [sum = 6], [first = 4] are not independent.

$$P[\text{sum} = 7 \mid \text{first} = 4] = \frac{P[\text{sum} = 7 \text{ and first} = 4]}{P[\text{first} = 4]} = \frac{1}{6} \left( = \frac{\frac{1}{36}}{\frac{1}{6}} \right)$$

P[sum = 7] = 1/6.

Thus, [sum = 7], [first = 4] are independent events!

What is the difference? When [sum = 6] is the event of interest, the first die result has a impact. For example  $P[sum = 6 \mid first = 6] = 0$ .

This is not the case for [sum = 7]; P[sum = 7 | first outcome = x] is the same for all x.

### Example 2.22

Given an urn with 4 distinct balls. Pick 1 ball at random; balls are equally likely to be pickel Let  $E = \{1, 2\}, F = \{1, 3\}, G = \{1, 4\}.$ 

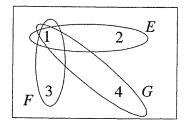
Consider the probabilities of the intersections below:

$$P[E \cap F] = P[\{1\}] = 1/4 = P[E]P[F]$$

$$P[F \cap G] = P[\{1\}] = 1/4 = P[F]P[G]$$

$$P\left[E\cap G\right]=P\left[\left\{1\right\}\right]=1/4=P\left[E\right]P\left[G\right]$$

$$P[E \cap F \cap G] = P[\{1\}] = 1/4 \neq P[E] P[F] P[G]$$



The events E, F, G are called **pairwise** independent. They are **not** a set of **mutually** independent events.

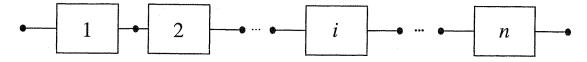
### Example 2.23 Reliability of series/parallel system structures

Series  $\rightarrow$  if any component fails, the system fails

 $Parallel \rightarrow all$  components must fail for the system to fail

Assume component failures are mutually independent events.

#### **Series System**

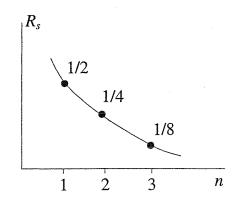


Let  $P[A_i] = P$  [component i is functioning]

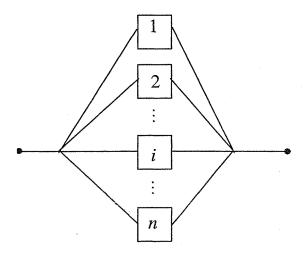
Define Series system Reliability as:  $R_s = \prod_{i=1}^n P[A_i] \equiv P$  [series system is functional at some

instance in time] since  $R_S = P[A_1 \cap A_2 \cap ... \cap A_n]$ .

 $R_s$  can reduce fast as n grows as shown:



### **Parallel System**



 $A_i \equiv [\text{component } i \text{ is functioning}]$ 

 $A_i^C \equiv [\text{component } i \text{ fails}]$ 

Define Parallel system Reliability as: P [parallel system is functioning at instance in time]  $R_p$ 

$$R_p = 1 - P$$
 [parallel system is not functioning] =  $1 - P\left[A_1^C \cap A_2^C \cap ... \cap A_n^C\right] = 1 - \prod_{i=1}^n P\left[A_i^C\right]$ 

As n increases,  $R_p$  approaches 1 as shown: