#### 1

# Digital Signal Processing

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Abstract—This manual provides a simple introduction to digital signal processing.

## 1 SOFTWARE INSTALLATION

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

#### 2 DIGITAL FILTER

2.1 Download the sound file from

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/Sound\_Noise.wav

- 2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? Solution: By observing spectrogram, it clearly shows that tonal frequency is under 4kHz. And above 4kHz only noise is present.
- 2.3 Write the python code for removal of out of band noise and execute the code.

**Solution:** 

import soundfile as sf from scipy import signal # read .wav file input signal, fs = sf.read("Sound Noise.way # sampling frequency of Input signal  $sample\_freq = fs$ print(input signal, fs) # order of the filter order = 4# cutoff frquency 4kHz  $cutoff_freq = 4000.0$ # digital frequency Wn = 2\*cutoff freq/sample freq # b and a are numerator and denominator # polynomials respectively b, a = signal.butter(order, Wn, "low") # filter the input signal with butterworth filter output signal = signal.filtfilt(b, a, input signal # output signal = signal.lfilter(b, a, input\_signal) # write the output signal into .wav file sf.write("Sound\_With\_ReducedNoise.wav", output\_signal, fs)

2.4 The output of the python script 2.3 Problem is the audio file Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

**Solution:** The audio is subdued and the higher frequenices are just blank.

#### 3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ \frac{1}{2}, 2, 3, 4, 2, 1 \right\}$$
 (3.1)

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$
  
$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

**Solution:** The following code yields Fig. 3.2.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/xnyn.py

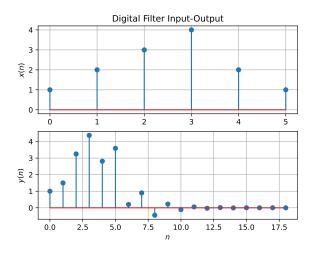


Fig. 3.2. figure

3.3 Repeat the above exercise using a C code. **Solution:** Download and run the C code for generating y and Python code for plotting y

wget https://github.com/gunjitmittal/EE3900 /blob/main/Assignment-1/codes/xnyn.c wget https://github.com/gunjitmittal/EE3900 /blob/main/Assignment-1/codes/xnyn (1).py

#### 4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

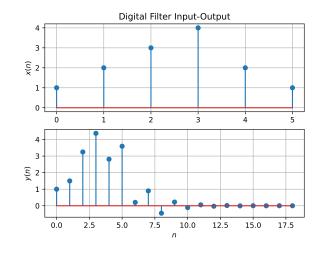


Fig. 3.3. figure

**Solution:** From (4.1),

$$\mathcal{Z}\{x(n-1)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$
 (4.4)

$$=\sum_{n=-\infty}^{\infty}x(n)z^{-n-1} \qquad (4.5)$$

$$= z^{-1} \sum_{n=-\infty}^{\infty} x(n) z^{-n} \qquad (4.6)$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\lbrace x(n-k)\rbrace = \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \quad (4.7)$$

$$=\sum_{n=-\infty}^{\infty}x(n)z^{-n-k} \qquad (4.8)$$

$$= z^{-k} \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$
 (4.9)

4.2 Obtain X(z) for x(n) defined in problem 3.1. **Solution:** 

$$X(z) = \sum_{n=0}^{5} x(n)z^{-n}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + 1z^{-5}$$
(4.10)
$$(4.11)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.12}$$

from (3.2) assuming that the Z-transform is a linear operation.

**Solution:** We have

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2)$$
 (4.13)

Finding the Z-transform

$$\mathcal{Z}\{y(n) + \frac{1}{2}y(n-1)\} = \mathcal{Z}\{x(n) + x(n-2)\}$$
(4.14)

Because Z-transform is a linear function

$$\mathcal{Z}y(n) + \frac{1}{2}\mathcal{Z}y(n-1) = \mathcal{Z}x(n) + \mathcal{Z}x(n-2)$$
(4.15)

Using (4.9)

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
(4.16)

$$\implies \frac{Y(z)}{X(z)} = \frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{4.17}$$

#### 4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.18)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.19)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1$$
 (4.20)

**Solution:** 

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \sum_{n=-\infty}^{\infty} \delta(n) z^{-n}$$
 (4.21)

$$=\sum_{n=0}^{\infty} z^{-n} = 1 \tag{4.22}$$

and from (4.19),

$$u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} U(z) = \sum_{n=-\infty}^{\infty} u(n)z^{-n}$$
 (4.23)

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.24)

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.25}$$

using the fomula for the sum of an infinite geometric progression.

#### 4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a|$$
 (4.26)

#### **Solution:**

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$
 (4.27)

$$=\sum_{n=0}^{\infty} (az^{-1})^{-n} \tag{4.28}$$

$$= \frac{1}{1 - az^{-1}}, \quad |z| > |a| \qquad (4.29)$$

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \tag{4.30}$$

Plot  $|H(e^{j\omega})|$ . Comment.  $H(e^{j\omega})$  is known as the *Discret Time Fourier Transform* (DTFT) of h(n).

#### **Solution:**

$$H(e^{J\omega}) = \frac{1 + e^{-2J\omega}}{1 + \frac{1}{2}e^{-J\omega}}$$
 (4.31)

$$\implies |H(e^{j\omega})| = \frac{|1 + \cos 2\omega - j\sin 2\omega|}{|1 + \frac{1}{2}\cos \omega - \frac{1}{2}\sin \omega|}$$

$$= \sqrt{\frac{(1 + \cos 2\omega)^2 + (\sin 2\omega)^2}{(1 + \frac{1}{2}\cos \omega)^2 + (\frac{1}{2}\sin \omega)^2}}$$

$$(4.33)$$

$$=\sqrt{\frac{2+2\cos 2\omega}{\frac{5}{4}+\cos \omega}}\tag{4.34}$$

$$= \sqrt{\frac{2(2\cos^2\omega)4}{5 + 4\cos\omega}}$$
 (4.35)

$$=\frac{4\left|\cos\omega\right|}{\sqrt{5+4\cos\omega}}\tag{4.36}$$

The following code plots Fig. 4.6.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/dtft.py

The plot is even and has a period of  $2\pi$ .

(5.9)

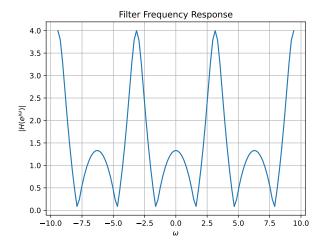


Fig. 4.6.  $|H(e^{j\omega})|$ 

4.7 Express h(n) in terms of  $H(e^{j\omega})$ .

**Solution:** Since  $H(e^{j\omega})$  is the DTFT of h(n)

$$\int_{-\pi}^{\pi} H\left(e^{j\omega}\right) e^{jn\omega} d\omega \tag{4.37}$$

$$= \int_{-\pi}^{\pi} \left( \sum_{k=-\infty}^{\infty} h(k) e^{-jk\omega} \right) e^{jn\omega} d\omega \quad (4.38)$$

$$= \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{j(n-k)\omega} d\omega$$
 (4.39)

$$=2\pi\sum_{k=-\infty}^{\infty}h(k)\delta(n-k) \tag{4.40}$$

$$=2\pi h(n) \tag{4.41}$$

$$\therefore h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H\left(e^{j\omega}\right) e^{jn\omega} d\omega \qquad (4.42)$$

#### 5 IMPULSE RESPONSE

## 5.1 Using long division, find

$$h(n), \quad n < 5 \tag{5.1}$$

for H(z) in (4.17)

**Solution:** 

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.2)

Substitute 
$$z^{-1} = x$$

$$2x - 4$$

$$\frac{1}{2}x + 1$$

$$-x^2 - 2x$$

$$-2x + 1$$

$$2x + 4$$

$$\implies 1 + z^{-2} = \left(1 + \frac{1}{2}z^{-1}\right)\left(-4 + 2z^{-1}\right) + 5$$
(5.3)

$$\implies H(z) = -4 + 2z^{-1} + \frac{5}{1 + \frac{1}{2}z^{-1}}$$
 (5.4)

On applying the inverse Z-transform on both sides of the equation

$$H(z) \stackrel{\mathcal{Z}}{\rightleftharpoons} h(n)$$
 (5.5)

$$-4 \stackrel{\mathcal{Z}}{\rightleftharpoons} -4\delta(n) \tag{5.6}$$

$$2z^{-1} \stackrel{\mathcal{Z}}{\rightleftharpoons} 2\delta(n-1) \tag{5.7}$$

$$\frac{5}{1 + \frac{1}{2}z^{-1}} \stackrel{\mathcal{Z}}{\rightleftharpoons} 5\left(-\frac{1}{2}\right)^n u(n) \tag{5.8}$$

Therefore,

$$h(n) = -4\delta(n) + 2\delta(n-1) + 5\left(-\frac{1}{2}\right)^n u(n)$$
(5.10)

$$h(0) = -4 + 5 = 1 \tag{5.11}$$

$$h(1) = 2 - 2.5 = -0.5 (5.12)$$

$$h(2) = 1.25 \tag{5.13}$$

$$h(3) = -0.625 (5.14)$$

$$h(4) = 0.3125 \tag{5.15}$$

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z)$$
 (5.16)

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

**Solution:** From (4.17),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.17)

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.18)

using (4.26) and (4.9).

5.3 Sketch h(n). Is it bounded? Justify theoretically.

**Solution:** The following code plots Fig. 5.3.

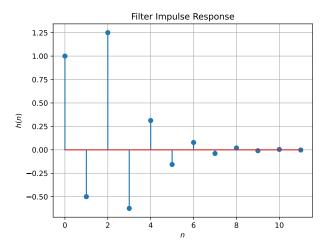


Fig. 5.3. h(n) as the inverse of H(z)

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/hn.py

As we can see from the plot h(n) is bounded. Theoretically,

$$|u(n)| \leq 1 (5.19)$$

$$\left| \left( -\frac{1}{2} \right)^n \right| \leq 1 (5.20)$$

$$\left| \left( -\frac{1}{2} \right) \right| \leq 1 \qquad (5.20)$$

$$\implies \left| \left( -\frac{1}{2} \right)^n u(n) \right| \leq 1 \qquad (5.21)$$

Similarly,

$$\left| \left( -\frac{1}{2} \right)^{n-2} u(n-2) \right| \le 1 \quad (5.22)$$

$$\implies h(n) \qquad \leq 2 \quad (5.23)$$

Therefore h(n) is bounded.

5.4 Is it convergetn? Justify using ratio test.

Solution: Using the ratio test for convergence

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \to \infty} \left| \frac{\left(-\frac{1}{2}\right)^{n-1} \left(\frac{1}{4} + 1\right)}{\left(-\frac{1}{2}\right)^{n-2} \left(\frac{1}{4} + 1\right)} \right|$$
(5.24)

$$=\lim_{n\to\infty} \left| -\frac{1}{2} \right| \tag{5.25}$$

$$=\frac{1}{2}<1$$
 (5.26)

Therefore, h(n) is convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.27}$$

Is the system defined by (3.2) stable for the impulse response in (5.16)?

#### **Solution:**

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^n u(n)$$

$$+ \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.28)$$

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n + \sum_{n=2}^{\infty} \left(-\frac{1}{2}\right)^{n-2}$$
(5.29)

These are both sums of infinite geometric progressions with first terms 1 and common ratios  $-\frac{1}{2}$ 

$$\sum_{n=-\infty}^{\infty} h(n) = \frac{1}{1 - \left(-\frac{1}{2}\right)} + \frac{1}{1 - \left(-\frac{1}{2}\right)} \quad (5.30)$$
$$= \frac{4}{3} < \infty \qquad (5.31)$$

Therefore, the system is stable.

5.6 Verify the above result using a Python code.
Solution: The stability has been verified in the following code

wget https://github.com/gunjitmittal/ EE3900/blob/main/Assignment-1/ codes/hndef.py

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2),$$
 (5.32)

This is the definition of h(n).

**Solution:** 

$$h(0) = 1 (5.33)$$

Now, for n=1,

$$h(1) + \frac{1}{2}h(0) = \delta(1) + \delta(-1) = 0$$
 (5.34)

$$\implies h(1) = -\frac{1}{2}h(0) = -\frac{1}{2} \tag{5.35}$$

For n=2,

$$h(2) + \frac{1}{2}h(1) = \delta(2) + \delta(0) = 1$$
 (5.36)

$$\implies h(2) = 1 - \frac{1}{2}h(1) = \frac{5}{4}$$
 (5.37)

For n > 2, the right hand side of the equation is always zero. Thus,

$$h(n) = -\frac{1}{2}h(n-1)$$
  $n > 2$  (5.38)

$$h(3) = \frac{5}{4} \left( -\frac{1}{2} \right) \tag{5.39}$$

$$h(4) = \frac{5}{4} \left( -\frac{1}{2} \right)^2 \tag{5.40}$$

$$\vdots (5.41)$$

$$h(n) = \frac{5}{4} \left( -\frac{1}{2} \right)^{n-2} \tag{5.42}$$

Therefore,

$$h(n) = \begin{cases} 1 & n = 0 \\ -\frac{1}{2} & n = 1 \\ \frac{5}{4} \left( -\frac{1}{2} \right)^{n-2} & n \ge 2 \end{cases}$$
 (5.43)

Thus, it is bounded and convergent to 0

$$\lim_{n \to \infty} h(n) = 0 \tag{5.44}$$

The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/hndef.py

#### 5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.45)

Comment. The operation in (5.45) is known as *convolution*.

**Solution:** The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/ynconv.py

5.9 Express the above convolution using a Toeplitz matrix.

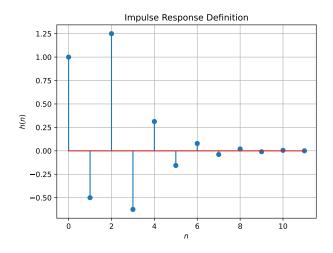


Fig. 5.7. h(n) from the definition

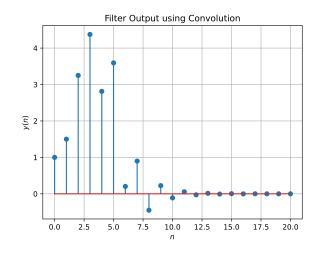


Fig. 5.8. y(n) from the definition of convolution

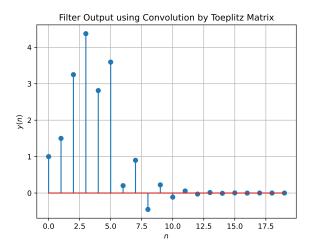
#### Solution: Let

$$\mathbf{x} = \begin{pmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \end{pmatrix} \qquad \mathbf{h} = \begin{pmatrix} 1 \\ -0.5 \\ 1.25 \\ -0.62 \\ 0.31 \\ -0.16 \end{pmatrix} \tag{5.46}$$

Their convolution is given by the product of

the following Toeplitz matrix T

$$\begin{pmatrix} 1 & 0 & 0 & 0 & 0 & 0 \\ -0.5 & 1 & 0 & 0 & 0 & 0 \\ 1.25 & -0.5 & 1 & 0 & 0 & 0 \\ -0.62 & 1.25 & -0.5 & 1 & 0 & 0 \\ 0.31 & -0.62 & 1.25 & -0.5 & 1 & 0 \\ -0.16 & 0.31 & -0.62 & 1.25 & -0.5 & 1 \\ 0 & -0.16 & 0.31 & -0.62 & 1.25 & -0.5 \\ 0 & 0 & -0.16 & 0.31 & -0.62 & 1.25 \\ 0 & 0 & 0 & -0.16 & 0.31 & -0.62 \\ 0 & 0 & 0 & 0 & -0.16 & 0.31 \\ 0 & 0 & 0 & 0 & 0 & -0.16 \\ 0 & 0 & 0 & 0 & 0 & -0.16 \end{pmatrix}$$



and x

Fig. 5.9. Plot of the convolution of x(n) and h(n)

$$\mathbf{y} = \mathbf{x} \circledast \mathbf{h} = \mathbf{T}\mathbf{x} = \begin{pmatrix} 1\\ 1.5\\ 3.25\\ 4.38\\ 2.81\\ 3.59\\ 0.12\\ 0.78\\ -0.62\\ 0\\ -0.16 \end{pmatrix}$$
 (5.48)

Download the following Python code for computing the convolution by using a Toeplitz matrix and plotting Fig. 5.9

> wget https://github.com/ Ankit-Saha-2003/ EE3900/raw/main/ Assignment\_1/codes/5.9.

Run the Python code by executing

5.10 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.49)

**Solution:** From (5.45)

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.50)

Substituting k with n-k

$$=\sum_{n-k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.51)

as n remains constant, we can rewrite it as

$$= \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.52)

# 6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

**Solution:** The following code plots Fig. 6.1.

wget https://github.com/gunjitmittal/EE3900/ blob/main/Assignment-1/codes/6\_1.py

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

**Solution:** The following code plots Fig. 6.2.

wget https://github.com/gunjitmittal/EE3900/ blob/main/Assignment-1/codes/6\_2.py

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

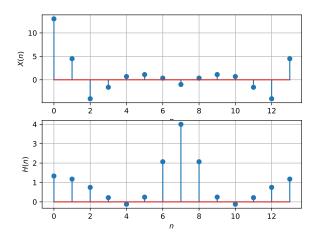


Fig. 6.1. X(n) and H(n)

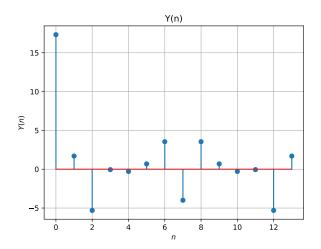


Fig. 6.2. Y(n)

**Solution:** The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/yndft.py

6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT. **Solution:** The following code plots Fig. 6.4.

wget https://github.com/gunjitmittal/EE3900/blob/main/Assignment-1/codes/6\_4.py

6.5 Wherever possible, express all the above equations as matrix equations.

Solution: We use the DFT Matrix, where

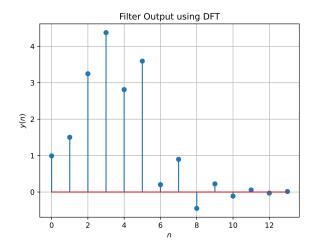


Fig. 6.3. y(n) from the DFT

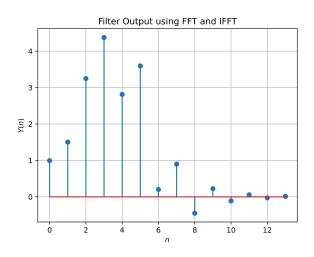


Fig. 6.4. y(n) using FFT and IFFT

 $\omega = e^{-\frac{j2k\pi}{N}}$ , which is given by

$$\mathbf{W} = \begin{pmatrix} \omega^0 & \omega^0 & \dots & \omega^0 \\ \omega^0 & \omega^1 & \dots & \omega^{N-1} \\ \vdots & \vdots & \ddots & \vdots \\ \omega^0 & \omega^{N-1} & \dots & \omega^{(N-1)(N-1)} \end{pmatrix}$$
(6.4)

i.e.  $W_{jk} = \omega^{jk}, \ 0 \le j, k < N.$  Hence, we can write any DFT equation as

$$X = Wx = xW \tag{6.5}$$

where

$$\mathbf{x} = \begin{pmatrix} x(0) \\ x(1) \\ \vdots \\ x(n-1) \end{pmatrix} \tag{6.6}$$

Using (6.3), the inverse Fourier Transform is given by

$$\mathbf{x} = \mathcal{F}^{-1}(\mathbf{X}) = \mathbf{W}^{-1}\mathbf{X} = \frac{1}{N}\mathbf{W}^{\mathbf{H}}\mathbf{X} = \frac{1}{N}\mathbf{X}\mathbf{W}^{\mathbf{H}}$$

$$(6.7)$$

$$\Longrightarrow \mathbf{W}^{-1} = \frac{1}{N}\mathbf{W}^{\mathbf{H}}$$

$$(6.8)$$

where H denotes hermitian operator. We can rewrite (6.2) using the element-wise multiplication operator as

$$\mathbf{Y} = \mathbf{H} \cdot \mathbf{X} = (\mathbf{W}\mathbf{h}) \cdot (\mathbf{W}\mathbf{x}) \tag{6.9}$$

#### 7 EXERCISES

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m) y(n-m) = \sum_{k=0}^{N} b(k) x(n-k)$$
(7.1)

where the input signal is x(n) and the output signal is y(n) with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

- 7.2 Repeat all the exercises in the previous sections for the above a and b.
- 7.3 What is the sampling frequency of the input signal?

**Solution:** Sampling frequency(fs)=44.1kHZ.

7.4 What is type, order and cutoff-frequency of the above butterworth filter

**Solution:** The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

7.5 Modifying the code with different input parameters and to get the best possible output.