

Effective area

$$A_e = \frac{P_{\text{int}}}{S_i} \Rightarrow A_e = \frac{\lambda^2 D}{4\pi} \quad (\text{matched impedance condition})$$

Short dipole:  $A_e = \frac{3\lambda^2}{8\pi}$

halfwave:  $D = 1.64$

Friis Transmission Formula

$$\frac{P_{\text{rec}}}{P_t} = G_t G_r \left( \frac{\lambda}{4\pi R} \right)^2 \quad [\text{power transfer ratio}]$$

Signal-to-noise ratio:

$$P_n = K T_{\text{sys}} B \quad \text{where } K = 1.38 \times 10^{-23} \text{ J/K} \quad B \text{ is the receiver bandwidth in Hz}$$

$$\text{SNR} = \frac{P_{\text{rec}}}{P_n} \quad (\text{dimensionless})$$

electric and magnetic field vectors

$$\vec{E} = \hat{x} \tilde{E}_x e^{-jkz} + \hat{y} \tilde{E}_y e^{-jkz} \quad \text{wave propagate in } +z \text{ direction}$$

$$\vec{H} = \frac{1}{\eta} \hat{k} \times \vec{E} \quad (\hat{k} \text{ is the wave propagation direction})$$

$$k = \omega \sqrt{\mu\epsilon} = \frac{\omega}{u_p} \quad \text{wavenumber}$$

$$\Rightarrow \vec{E} = (\hat{x} a_x e^{j\phi_x} + \hat{y} a_y e^{j\phi_y}) e^{-jkz} \quad +z \text{ direction}$$

If  $\phi_y - \phi_x = 90^\circ$

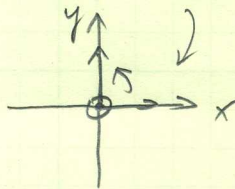
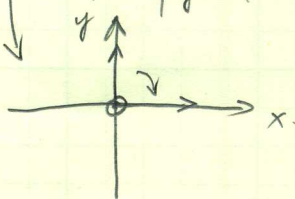
If  $\phi_y - \phi_x = -90^\circ$

Right Hand

Left Hand

y leads x

x leads y





Maxwell's Equations:

	time-domain	phasor
Gauss's Law	$\nabla \cdot \vec{D} = \rho_v$	$\nabla \cdot \vec{D} = \tilde{\rho}_v$
Faraday's Law	$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$	$\nabla \times \vec{E} = -j\omega \vec{B}$
"Magnetic Gauss's" Law	$\nabla \cdot \vec{B} = 0$	$\nabla \cdot \vec{B} = 0$
Ampere's Law:	$\nabla \times \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$	$\nabla \times \vec{H} = \vec{J} + j\omega \vec{D}$
Constitutive relations:	$\vec{D} = \epsilon \vec{E}, \quad \vec{B} = \mu \vec{H}, \quad \vec{J} = \sigma \vec{E}$	

Antenna:

far fields  $d\vec{E} = \hat{\theta} \frac{jk\eta \tilde{I}_0 dz}{4\pi R} e^{-jkr} \sin\theta, \quad d\vec{H} = \hat{\phi} \frac{jk\tilde{I}_0 dz}{4\pi R} e^{-jkr} \sin\theta$

Hertzian Dipole:

Exact:  $\vec{E} = \hat{R} \frac{\tilde{I}_0 k^2 l}{4\pi} \eta e^{-jkr} \left[ \frac{e}{(kr)^2} - \frac{j^2}{(kr)^3} \right] \cos\theta + \hat{\theta} \frac{\tilde{I}_0 k^2 l}{4\pi} \eta e^{-jkr} \left[ \frac{j}{kr} + \frac{1}{(kr)^2} - \frac{j}{(kr)^3} \right] \sin\theta$

$\vec{H} = \hat{\phi} \frac{\tilde{I}_0 k^2 l}{4\pi} e^{-jkr} \left[ \frac{j}{kr} + \frac{1}{(kr)^2} \right] \sin\theta$

Far field:  $kr \gg 1$

$\vec{E} = \hat{\theta} \frac{jk\eta \tilde{I}_0 l}{4\pi R} e^{-jkr} \sin\theta, \quad \vec{H} = \hat{\phi} \frac{jk\tilde{I}_0 l}{4\pi R} e^{-jkr} \sin\theta$

Short Dipole:

far field:  $\vec{E} = \hat{\theta} \frac{jk\eta \tilde{I}_0 l}{8\pi R} e^{-jkr} \sin\theta, \quad \vec{H} = \hat{\phi} \frac{jk\tilde{I}_0 l}{4\pi R} e^{-jkr} \sin\theta$

1) Triangular current distribution

→ Poynting vector:  $\vec{S}_{av} = \frac{1}{2} \text{Re} \{ \vec{E} \times \vec{H}^* \} = \hat{R} S_{max} F(\theta, \phi)$   
max magnitude normalized Radiation Pattern Max = 1

## → Directivity and Gain:

Radiated Power:  $P_{rad} = S_{max} R^2 \int_0^{2\pi} \int_0^\pi F(\theta, \phi) \sin\theta d\theta d\phi$

$D = 4\pi R^2 \frac{S_{max}}{P_{rad}} = \frac{S_{max}}{|S_{iso}|} = \frac{4\pi}{\iint F(\theta, \phi) \sin\theta d\theta d\phi} \quad (S_{iso} = \hat{R} \frac{P_{in}}{4\pi R^2})$

$\epsilon = \frac{P_{rad}}{P_{rad} + P_{loss}} \Rightarrow G = \epsilon D, \quad P_{rad} = \epsilon P_{in} \quad \boxed{S_{max} = \frac{P_{in} G}{4\pi R^2}} \star$

→ Radiation resistance:  $R_{rad} = 2P_{rad}/|\tilde{I}_0|^2$

→ Loss resistance:

$R_{loss} = \frac{2P_{loss}}{|\tilde{I}_0|^2}, \quad \text{Hertzian: } R_{loss} = \frac{1}{2\pi a} \sqrt{\frac{\pi f \mu_0}{\sigma_c}}; \quad \text{short: } R_{loss} = \frac{1}{6\pi a} \sqrt{\frac{\pi f \mu_0}{\sigma_c}}$

$\lambda/2$  dipole:  $R_{loss} = \frac{\lambda}{8\pi a} \sqrt{\frac{\pi f \mu_0}{\sigma_c}}; \quad \text{arbitrary length: } R_{loss} = \frac{1}{2\pi a |\tilde{I}_0|^2} \sqrt{\frac{\pi f \mu_0}{\sigma_c}} \int_{-\lambda/2}^{\lambda/2} |I(z)|^2 dz$

Center-fed Arbitrary length:

$I(z) = \begin{cases} \tilde{I}_0 \sin[k(\frac{\lambda}{2} - |z|)] & -\frac{\lambda}{2} \leq z \leq \frac{\lambda}{2} \\ 0 & \text{elsewhere} \end{cases}$   
 $\vec{E} = \hat{\theta} j \frac{\eta}{2\pi} \tilde{I}_0 \frac{e^{-jkr}}{R} \left[ \frac{\cos(0.5\pi \cos\theta) - \cos(0.5\pi k l)}{\sin\theta} \right] \quad |\vec{H}| = \frac{|\vec{E}|}{\eta}$   
 $F(\theta) = \frac{1}{[1 - \cos(0.5\pi k l)]^2} \left| \frac{\cos[0.5\pi k l \cos\theta] - \cos(0.5\pi k l)}{\sin\theta} \right|^2$

Half-wave dipole:

$D = \begin{pmatrix} 1.64 \\ 2.15 \text{ dBi} \end{pmatrix} \vec{E} = \hat{\theta} j \frac{\eta}{2\pi} \tilde{I}_0 \frac{e^{-jkr}}{R} \left[ \frac{\cos(0.5\pi \cos\theta)}{\sin\theta} \right] \quad |\vec{H}| = \frac{|\vec{E}|}{\eta}$

Phase approximation:

$e^{-jk|R-R'|} \approx e^{-jkr} e^{jk\tilde{r}' \cos\theta}$



Input Impedance

$$Z_{in} = Z_0 \left( \frac{Z_L + jZ_0 \tan \beta l}{Z_0 + jZ_L \tan \beta l} \right)$$

lossy line:

$$Z_{in} = Z_0 \left( \frac{Z_L + Z_0 \tanh \gamma l}{Z_0 + Z_L \tanh \gamma l} \right)$$

$$Y_{in} = Y_0 \left( \frac{Y_L + jY_0 \tan \beta l}{Y_0 + jY_L \tan \beta l} \right) \quad \text{lossless}$$

$$\tanh x = \frac{e^x - e^{-x}}{e^x + e^{-x}}$$

Special Cases:

$$1) Z_L = Z_0 \Rightarrow \Gamma = 0 \Rightarrow \boxed{Z_{in} = Z_0}$$

$$2) l \ll \lambda \rightarrow \beta l \ll 1$$

$$\Rightarrow Z_0 \tan \beta l \ll Z_L \Rightarrow Z_{in} = Z_L$$

$$3) Z_L = 0 \Rightarrow \boxed{Z_{in} = jZ_0 \tan \beta l} = \text{closed end.}$$

$$4) Z_L \rightarrow \infty \Rightarrow \boxed{Z_{in} = -jZ_0 \cot \beta l} = \text{open end}$$

$$5) l = \lambda/2 \Rightarrow \boxed{Z_{in} = Z_L}$$

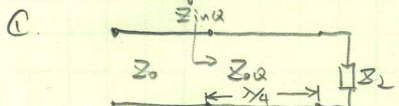
$$6) l = \lambda/4 \Rightarrow \beta l = 2\pi \times \lambda/4 = \pi/2 \Rightarrow \boxed{\frac{Z_0^2}{Z_L}} \leftarrow \text{important}$$

When load is a reactance  $Z_L = jX$ 

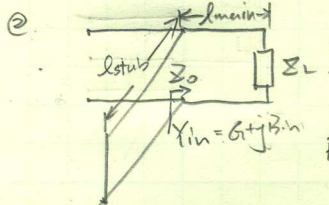
$$Z_{in} = jZ_0 \frac{X + Z_0 \tan \beta l}{Z_0 - X \tan \beta l} \quad |\Gamma| = 1$$

Wave Trap $Z_{in} = 0$  @ Designed frequency.

$$\Rightarrow X + Z_0 \tan \beta l = 0 \Rightarrow X = -Z_0 \tan \left( \frac{2\pi l}{\lambda} \right)$$

Match Impedance ① Maximum power transfer ② prevent reflection ③ specified  $Z_L$ make  $Z_{in} = Z_0$ If  $Z_L \rightarrow$  Pure Real  $Z_0 = \sqrt{Z_{in} Z_L}$ Otherwise, add at  $V_{max}$  or  $V_{min}$ .  $2\beta z + \theta_r = 2\pi n$  or  $(2n+1)\pi$ 

$$Z(-d_{max}) = Z_0 \frac{1+|\Gamma|}{1-|\Gamma|} \quad Z(-d_{min}) = Z_0 \frac{1-|\Gamma|}{1+|\Gamma|}$$

Let  $Y_{in} = Y_0 = 1/Z_0$ 

$$l_{shunt, min} = \frac{\lambda}{4\pi} [\theta_r \pm \cos^{-1}(-|\Gamma|)]$$

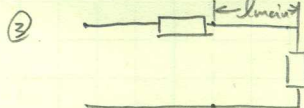
$$Z_{stub} = -B_{in}$$

$$X_{stub} = -X_{in}$$

$$B_{shunt, in} = \pm \frac{2|\Gamma|}{\sqrt{1-|\Gamma|^2}} Y_0$$

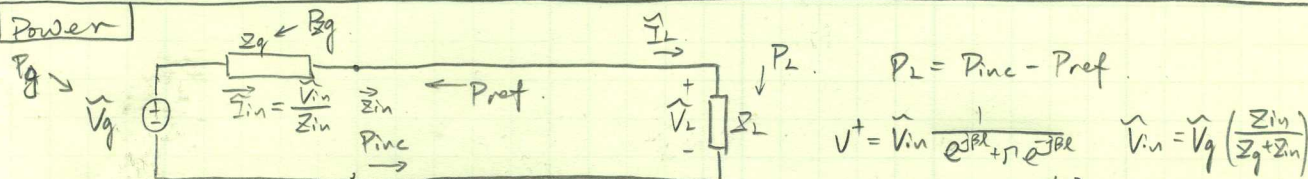
$$l_{stub, sc} = \frac{\lambda}{2\pi} \tan^{-1} \left( \frac{-Y_0}{B_{stub, sc}} \right)$$

$$l_{stub, oc} = \frac{\lambda}{2\pi} \tan^{-1} \left( \frac{-Z_0}{X_{stub, oc}} \right)$$



$$l_{series} = \frac{\lambda}{4\pi} [\theta_r \pm \cos^{-1}(|\Gamma|)]$$

$$X_{in}^{series} = \text{Im} \{ Z_{in}(-l_{series}) \} = \pm \frac{2|\Gamma|}{\sqrt{1-|\Gamma|^2}} Z_0$$

Power

$$P_2 = P_{inc} - P_{ref}$$

$$V^+ = \tilde{V}_{in} \frac{1}{e^{j\beta l} + \Gamma} e^{-j\beta l} \quad \tilde{V}_{in} = \tilde{V}_g \left( \frac{Z_{in}}{Z_g + Z_{in}} \right)$$

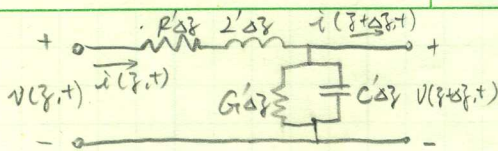
$$\tilde{V}_r = \tilde{V}_0^+ (1 + \Gamma) \quad \tilde{I}_L = \frac{\tilde{V}_0^+}{Z_0} (1 - \Gamma) \Rightarrow P_2 = \frac{1}{2} \text{Re} \{ \tilde{V}_L \tilde{I}_L^* \} = \frac{|\tilde{V}_0^+|^2}{2Z_0} (1 - |\Gamma|^2)$$

$$P_{inc} = \frac{|\tilde{V}_0^+|^2}{2Z_0} \quad P_{ref} = |\Gamma|^2 \frac{|\tilde{V}_0^+|^2}{2Z_0}$$

$$P_g = -\frac{1}{2} \text{Re} \{ \tilde{V}_g \tilde{I}_{in}^* \}$$

$$P_{zg} = \frac{1}{2} \text{Re} \{ (\tilde{I}_{in} Z_g) \tilde{I}_{in}^* \} = \frac{1}{2} |\tilde{I}_{in}|^2 Z_g$$





1 Xmn line equations

$$① - \frac{\partial v(z,t)}{\partial z} = R' i(z,t) + L' \frac{\partial i(z,t)}{\partial t} \Rightarrow - \frac{d\tilde{v}(z)}{dz} = (R' + j\omega L') \tilde{i}(z)$$

$$② - \frac{\partial i(z,t)}{\partial z} = G' v(z,t) + C' \frac{\partial v(z,t)}{\partial t} \Rightarrow - \frac{d\tilde{i}(z)}{dz} = (G' + j\omega C') \tilde{v}(z)$$

Wave Equations:

 $\frac{dQ}{dz}$ 

$$\Rightarrow \frac{d^2 \tilde{v}(z)}{dz^2} - \gamma^2 \tilde{v}(z) = 0$$

 $\frac{dQ}{dz}$ 

$$\Rightarrow \frac{d^2 \tilde{i}(z)}{dz^2} - \gamma^2 \tilde{i}(z) = 0$$

complex propagation constant  $\rightarrow \gamma^2 = (R' + j\omega L')(G' + j\omega C') = \alpha + j\beta$  ← phase constant  
attenuation constant

$$\Rightarrow \tilde{v}(z) = V_0^+ e^{-\gamma z} + V_0^- e^{+\gamma z} \quad \tilde{i}(z) = I_0^+ e^{-\gamma z} + I_0^- e^{+\gamma z}$$

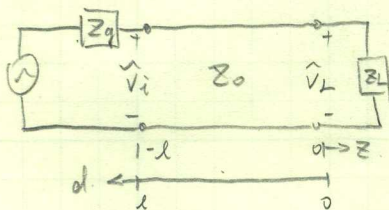
$$\frac{V_0^+}{I_0^+} = Z_0 = \frac{-V_0^-}{I_0^-} = \frac{R' + j\omega L'}{\gamma} = \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}} \quad \boxed{\text{Characteristic impedance}}$$

$$v(z,t) = \text{Re}(\tilde{v}(z) e^{j\omega t}) = |V_0^+| e^{-\alpha z} \cos(\omega t - \beta z + \phi^+) + |V_0^-| e^{+\alpha z} \cos(\omega t + \beta z + \phi^-)$$

$$\mu_p = p\lambda = \frac{\omega}{\beta} \quad \lambda_g = \text{guide wavelength}$$

Lossless XMN LINE  $R' \ll \omega L' \quad G' \ll \omega C' \Rightarrow \gamma = j\omega \sqrt{L'C'} = j\beta \quad Z_0 = \sqrt{\frac{L'}{C'}}$   
 $\lambda f = \mu_p = \frac{1}{\sqrt{\mu_0 \epsilon_r \epsilon_0}} = \frac{c}{\sqrt{\epsilon_r}} \quad \mu_0 = 4\pi \times 10^{-7} \text{ H/m} \quad \epsilon_0 = 8.854 \times 10^{-12} \text{ F/m} \quad \mu_0 = \frac{1}{\lambda f c}$

$$\tilde{v} = \underbrace{V_0^+ e^{-j\beta z}}_{\text{incident wave}} + \underbrace{V_0^- e^{+j\beta z}}_{\text{reflected wave}} \quad \tilde{i} = I_0^+ e^{-j\beta z} + I_0^- e^{+j\beta z}$$



$$\tilde{v}_L = \tilde{v}(z=0) = V_0^+ + V_0^- \quad \tilde{i}_L = \tilde{i}(z=0) = \frac{V_0^+}{Z_0} - \frac{V_0^-}{Z_0}$$

$$Z_L = \left( \frac{V_0^+ + V_0^-}{V_0^+ - V_0^-} \right) Z_0 \Rightarrow V_0^- = \left( \frac{Z_L - Z_0}{Z_L + Z_0} \right) V_0^+$$

$$-\frac{I_0^-}{I_0^+} = \Gamma = \frac{V_0^-}{V_0^+} = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{Z_L - Z_0}{Z_L + Z_0} \quad \boxed{\text{Voltage reflection coefficient}}$$

$$\Gamma = |\Gamma| e^{j\theta_r} \quad (|\Gamma| \leq 1) \quad ① \Gamma = 0 (Z_L = Z_0) \quad ② \Gamma = 1 (Z_L = \infty) \quad ③ \Gamma = -1 (Z_L = 0)$$

$$|\tilde{v}(z)| = [\tilde{v}(z) \tilde{v}^*(z)]^{1/2} = \left\{ [V_0^+ (e^{-j\beta z} + |\Gamma| e^{j\theta_r} e^{j\beta z})] \cdot [V_0^+ (e^{+j\beta z} + |\Gamma| e^{-j\theta_r} e^{-j\beta z})] \right\}^{1/2}$$

$$= |V_0^+| [1 + |\Gamma|^2 + 2|\Gamma| \cos(2\beta z + \theta_r)]^{1/2}$$

$$|\tilde{v}(z)|_{\max} @ (2\beta z + \theta_r) = 2n\pi \Rightarrow z = \frac{2n\pi - \theta_r}{2\beta} \quad |\tilde{v}|_{\max} = |V_0^+| [1 + |\Gamma|]$$

$$|\tilde{v}(z)|_{\min} @ (2\beta z + \theta_r) = (2n+1)\pi \Rightarrow z = \frac{(2n+1)\pi - \theta_r}{2\beta} \quad |\tilde{v}|_{\min} = |V_0^+| [1 - |\Gamma|]$$

Voltage Standing Wave Ratio  $S = \frac{|\tilde{v}|_{\max}}{|\tilde{v}|_{\min}} = \frac{1 + |\Gamma|}{1 - |\Gamma|} \quad (\text{SWR})$

$$|\Gamma| = \frac{S-1}{S+1}$$

$$R_s = \sqrt{\frac{\mu_0 \mu_c}{\epsilon_0 \epsilon_c}} \quad L'C' = \mu \epsilon \quad \frac{G'}{C'} = \frac{\sigma}{\epsilon}$$