

EE-568

HW-2

**Motor Winding Design and
Analysis**



Prepared by: Hakan POLAT
Submitted to: Dr. Ozan KEYSAN

**Electrical and Electronics Engineering
Department
ANKARA
31.03.2020**

Contents

1	Introduction	2
2	Integral-Slot Winding Design	2
3	Fractional-Slot Winding Design	3
4	2D FEA of 20 Pole 24 Slot PMSM	6
5	Conclusion	8

1 Introduction

This report consists of 4 parts. In the first section an analysis on aq 20-pole 120 slot 3-phase permanent magnet machine will be made. In the next section a fractional-slot winding design will be presented where a pole number 20 or 22 will be selected. Then a slot number between 20-30 will be selected usioing Emetor Winding Design. Then various analysis will be delivered. In the last section, a fractional slot PMSM will be analyzed usign ANSYS Maxwell finite element analysis(FEA) where a general 2D drawing and winding diagram will be presented. Then the airgap flux distribution and the induced voltage waveforms (phase and line-line) at rated speed will be given. Finally the cogging torque of the proposed machine will be investigated.

2 Integral-Slot Winding Design

In this section it is assumed that the PMSM has 20-poles, 120-slot and the number of phases is three. As for λ a full pitch design is selected and hence $\lambda = 1$. Then slots per pole por phase ,denoted as q, can be calculated as in Eq. 1 .

$$q = \frac{120}{20 \times 3} = 2 \quad (1)$$

Moreover, it is clear that there are in total of 6 slots per pole and hence the winding diagram of a sinle pole-pair is given in Table 1 .

Table 1: Winding Diagram of a Single Pole-Pair

1	2	3	4	5	6	7	8	9	10	11	12
A1	A2	-C1	-C2	B1	B2	-A1	-A2	C1	C2	-B1	-B2

Then the distribution factor, pitch factor and the winding factor for the fundamantal component can be calculated as in Eq., Eq. and Eq. respectively.

$$k_d = \frac{\sin(q\frac{\alpha}{2})}{q\sin(\frac{\alpha}{2})} = 0.965 \quad (2)$$

$$k_p = \sin(\frac{\lambda}{2}) = \sin(\frac{180}{2}) = 1 \quad (3)$$

$$k_w = k_p k_d = k_d = 0.965 \quad (4)$$

The same procedure can be generalized for the n-th harmonic of the input voltage waveform. However, in three phase machines the 3-rd and 5-th 7-th harmonics are often under investigation since the rest of the harmonic have relatively low magnitudes. In this question the 3-rd and 5-th harmonics are calculated.

The k_p , k_d and k_w for 3-rd harmonic are given in Eq. 6, Eq.5 and Eq. 7 respectively.

$$k_p = \sin\left(\frac{3\lambda}{2}\right) = \sin\left(\frac{3 \times 180}{2}\right) = -1 \quad (5)$$

$$k_d = \frac{\sin(3q\frac{\alpha}{2})}{q\sin(\frac{3\alpha}{2})} = 0.707 \quad (6)$$

$$k_w = k_p x k_d = k_d = -0.707 \quad (7)$$

Similarly the k_p , k_d and k_w for 5-th harmonic are given in Eq. 9, Eq.8 and Eq.10 respectively.

$$k_p = \sin\left(\frac{5\lambda}{2}\right) = \sin\left(\frac{5 \times 180}{2}\right) = 1 \quad (8)$$

$$k_d = \frac{\sin(5q\frac{\alpha}{2})}{q\sin(\frac{5\alpha}{2})} = -0.44 \quad (9)$$

$$k_w = k_p x k_d = k_d = -0.44 \quad (10)$$

Compared to the fundamental component 3-rd and 5-th harmonics have high magnitude. This is due to selection of full-pitch winding. If a 144 degrees of coil pitch angle were to be selected, the $k_p = 0$ meaning there are no 5-th harmonic flowing in the system. Although the magnitude of the 5-th harmonic is significantly lower than 3-rd harmonic pitch-factor is usually selected to reduce the 5-th harmonic. This is due to elimination the 3-rd harmonic when the three phases are connected as a Wye connection.

3 Fractional-Slot Winding Design

In this part it is asked to choose a pole number 20 or 22. As for the number of slot any slot number between 20 to 30 can be selected. In order to achieve high winding factor a pole number of 22 was selected. As for the number of slots 24 is selected. During the selection Emotor Winding Design was used.

Firstly the phase angle of the induced voltage in each slot is presented in Table 2.

Table 2: Angle of Induced Voltage in each Slot

1	2	3	4	5	6	7	8	9	10	11	12
0	165	330	135	300	105	270	75	240	45	210	15
13	14	15	16	17	18	19	20	21	22	23	24
180	345	150	315	120	285	90	255	60	225	30	195

In Fig. 1 the vectors for phase-a are present. There is 15 degrees between each voltage vector and hence slots 1, 2, 12, 13, 14, 23, 24 can be used.

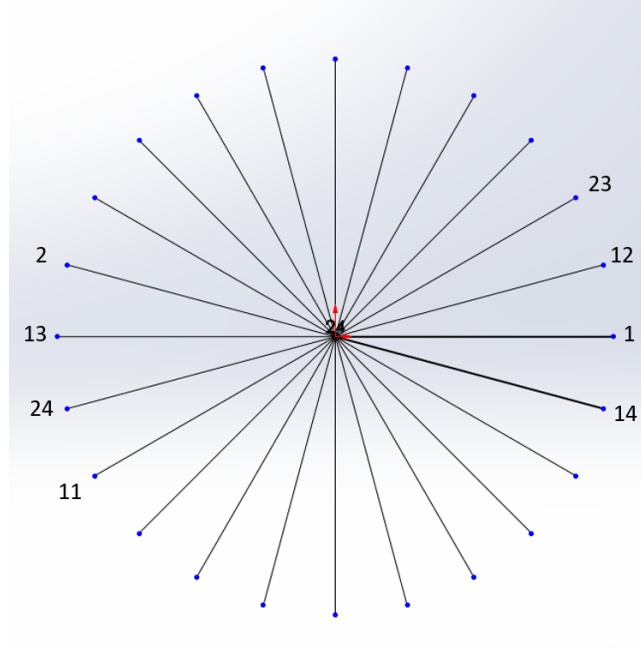


Figure 1: Voltage vector in the selected fractional-slot winding.

Distribution factor can be calculated as the vectoral sum of voltage vectors 23, 12, 1 and 14. Then the k_d can be calculated as in Eq. 11

$$k_d = \frac{3.82}{4} = 0.956 \quad (11)$$

The pitch factor can be calculated as in Eq. 12.

$$k_p = \sin(165/2) = 0.991 \quad (12)$$

The winding factor can be calculated as Eq. 13.

$$k_w = k_p k_d = 0.991 \times 0.956 = 0.947 \quad (13)$$

The third and the fifth harmonic can be also be calculated.

$$k_p = \sin(3 * 165/2) = 0.924 \quad (14)$$

$$k_p = \sin(5 * 165/2) = 0.793 \quad (15)$$

$$k_d = \frac{\sin(3 \times 0.36 \times 82.5)}{0.36 \sin(3 \times 82.5)} = \quad (16)$$

$$k_d = \frac{\sin(3 \times 0.36 \times 82.5)}{0.36 \sin(3 \times 82.5)} = \quad (17)$$

Then another fractional slot design is made for 20 pole. As for slot number 24 is selected. The electrical angle between two neighbour slots is given in Eq. 18.

$$Angle = \frac{360}{24} \frac{20}{2} = 150 \quad (18)$$

Table 3: Angle of Induced Voltage in each Slot

1	2	3	4	5	6	7	8	9	10	11	12
0	150	300	90	240	30	180	330	120	270	60	210
13	14	15	16	17	18	19	20	21	22	23	24
0	150	300	90	240	30	180	330	120	270	60	210

The voltage angle between each slot is given in Table 3.
The voltage vectors are given in Fig. 2.

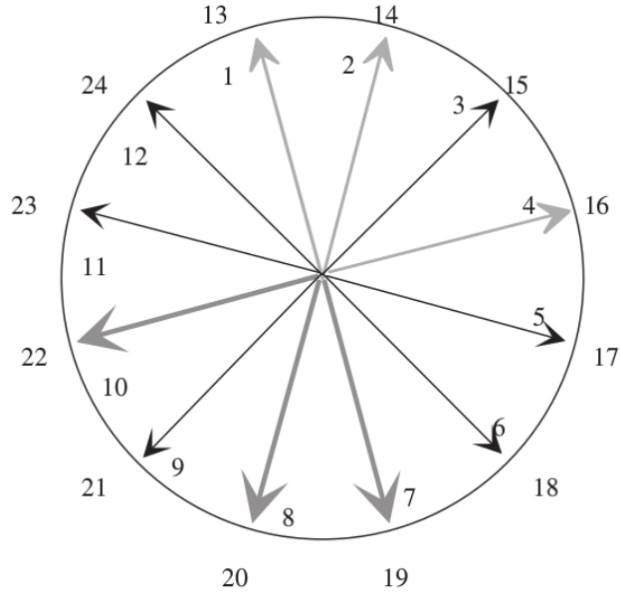


Figure 2: Voltage vector in the selected fractional-slot winding.

For a single phase slots 1, 13, 8 and 20 forms the A phase. Hence the distribution factor can be calculated as in Eq. 19.

$$k_d = \frac{2 * 2\cos(15)}{4} = 0.965 \quad (19)$$

If A+ is wound on slot number 1, A- can be connected to slot 2. Hence the pitch factor is given in Eq.20

$$k_p = \sin(150/2) = 0.965 \quad (20)$$

Which results in a winding factor given in Eq. 21

$$k_w = k_p k_d = 0.965 * 0.965 = 0.933 \quad (21)$$

For the third harmonic the winding factor can be calculated as in Eq.22.

$$k_w = k_p k_d = \frac{\sin(3q\alpha/2)}{q\sin(3\alpha/2)} \sin(3\lambda/2) = 0.707x - 0.707 = -0.5 \quad (22)$$

For the fifth harmonic the winding factor can be calculated as in Eq. 23.

$$k_w = k_p k_d = \frac{\sin(5q\alpha/2)}{q\sin(5\alpha/2)} \sin(3\lambda/2) = -0.707x0.258 = 0.183 \quad (23)$$

I couldnt do the necessary calculations for the 22pole 24 slot case.

4 2D FEA of 20 Pole 24 Slot PMSM

The geometry of a 20 pole 224 slot machine is given in Fig. 3.

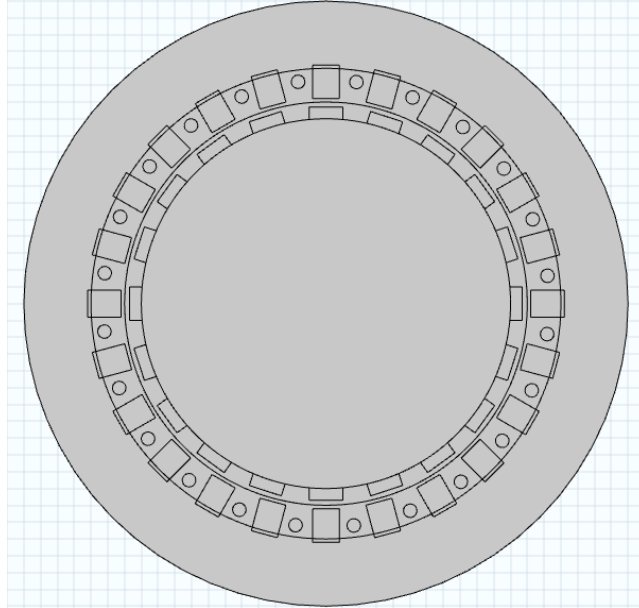


Figure 3: Motor Geometry

During operation the magnetic field is given in Fig. 4

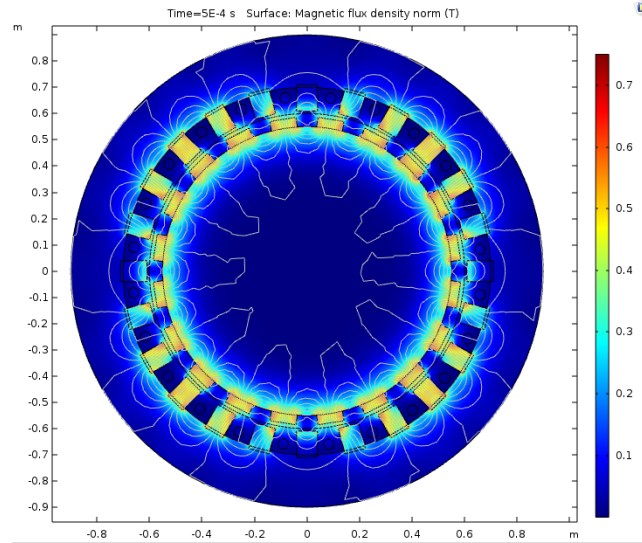


Figure 4: Airgap flux density distribution

The induced voltages are given in Fig. 5

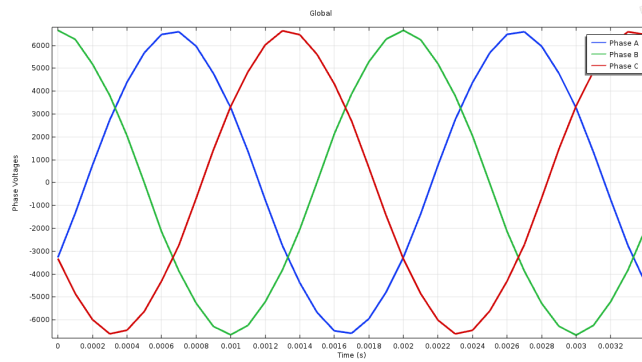


Figure 5: Induced voltages per phase

The cogging torque is given in Fig. 6.

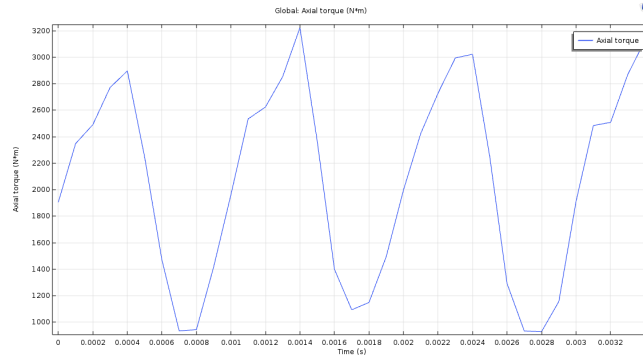


Figure 6: Cogging Torque

Also a 3d moving gif of the motor is in the Github repository.

5 Conclusion

In this homework, it was asked to desing various PMSM machine windings where concepts like pitch factor, distribution factor and winding factor was under investigation. The effect of winding design on the harmonic content was also discussed.