

MECH366 : Modeling of Mechatronic Systems

L3 : State-space models

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Review and today's topic

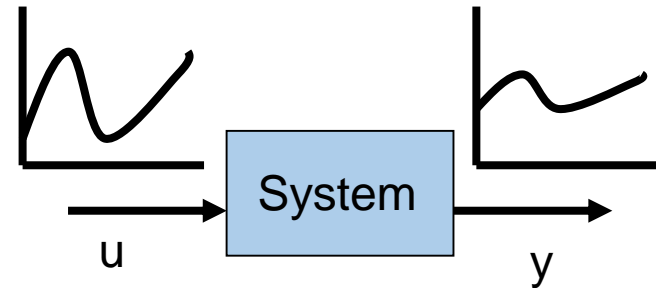
- Last lecture was about:
 - Dynamic model
 - Modeling procedure (next slide)
 - Analogies among different domains (incomplete)
- Today, we will study the **state-space model**.
 - Basic elements are used to generate a linear graph (in later classes, not today), from which we derive the state-space model.
 - State-space models are used for prediction, simulation, and controller design.



Modeling procedure (Review)

1. Identify a **physical system** to be modeled, and associated input and output variables.
2. Simplify the **physical system** with **basic elements**, based on your assumptions.
- ⇒ 3. By applying physical laws (Newton's second law, Kirchhoff's law etc.) to the **basic elements**, obtain **differential equations**.
4. Identify (estimate) **parameter values** in the **differential equations**.
5. Validate the obtained model (**differential equations** with estimated **parameter values**) experimentally. If the model turns out to be invalid (the model is invalidated), go back to Step 2 with modifications.

State-space model (Linear)



- Matrix-vector form

$$\begin{cases} \dot{x} &= Ax + Bu \\ y &= Cx + Du \end{cases}$$

x : state vector
 u : input vector
 y : output vector

- Element-wise representation

$$\begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix} \\ \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_m \end{bmatrix} \end{cases} = \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \cdots & a_{nn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} b_{11} & b_{12} & \cdots & b_{1r} \\ b_{21} & b_{22} & \cdots & b_{2r} \\ \vdots & \vdots & \ddots & \vdots \\ b_{n1} & b_{n2} & \cdots & b_{nr} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ \vdots \\ u_r \end{bmatrix}$$

$$= \begin{bmatrix} c_{11} & c_{12} & \cdots & c_{1n} \\ c_{21} & c_{22} & \cdots & c_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ c_{m1} & c_{m2} & \cdots & c_{mn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} d_{11} & d_{12} & \cdots & d_{1r} \\ d_{21} & d_{22} & \cdots & d_{2r} \\ \vdots & \vdots & \ddots & \vdots \\ d_{m1} & d_{m2} & \cdots & d_{mr} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ \vdots \\ u_r \end{bmatrix}$$



Remarks

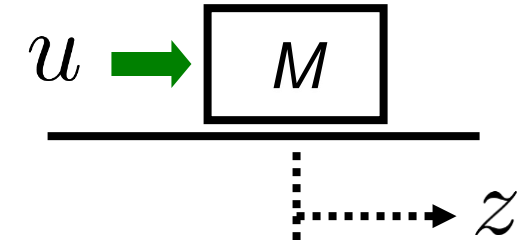
- The first equation, called *state equation*, is a first-order ordinary differential.
- The second equation, called *output equation*, is an algebraic equation.
- Two equations are called *state-space model*.
- Pay attention to *sizes of matrices and vectors*. They must be always compatible!
- “How to select the state vector?” is explained through examples.



Examples

- Mass with a driving force
- Mass-spring-damper system
 - 1-DOF (degree-of-freedom)
 - 2-DOF
- RLC circuit

Mass with a driving force

- By Newton's second law, we have 

$$M\ddot{z}(t) = u(t)$$

where input: force u , output: position z

- Define state variables $x_1 := z, x_2 := \dot{z}$

- Then,

$$\begin{cases} \dot{x}_1 = \dot{z} = x_2 \\ \dot{x}_2 = \ddot{z} = \frac{1}{M}u \\ y = x_1 \end{cases} \rightarrow \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix}}_B u(t) \\ y(t) = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_C \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underbrace{0}_D u(t) \end{cases}$$

Mass with a driving force

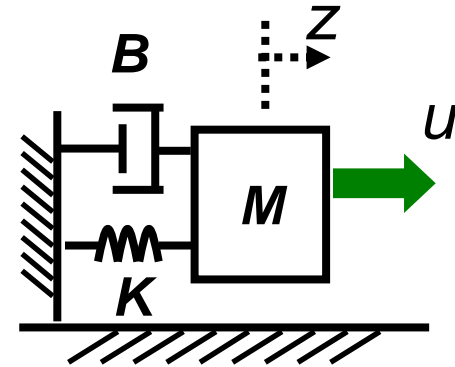
Another SS model

- Define state variables $x_1 := \dot{z}$, $x_2 := z$
- Then,
$$\begin{cases} \frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 1/M \\ 0 \end{bmatrix} u(t) \\ y(t) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} \end{cases}$$
- Derivation 1: Directly calculate.
- Derivation 2: Using the SS model in the previous slide, exchange rows/columns of A , B , C matrices.
- *SS model is not unique; it depends on the selection of states.*

Mass-spring-damper system (1-DOF)

- By Newton's second law,

$$M\ddot{z} = u - Kz - B\dot{z}$$



- Input: force (u), output: position (z)
- Define state variables: $x_1 := z$, $x_2 := \dot{z}$

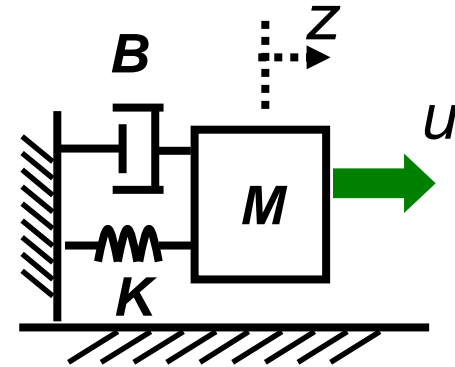
$$\rightarrow \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -K/M & -B/M \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1/M \end{bmatrix} u \\ y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \end{cases}$$

Second order model
(because it has 2 states)

Mass-spring-damper system (1-DOF)

- By Newton's second law,

$$M\ddot{z} = u - \underbrace{Kz}_{=:f_K} - B\dot{z}$$

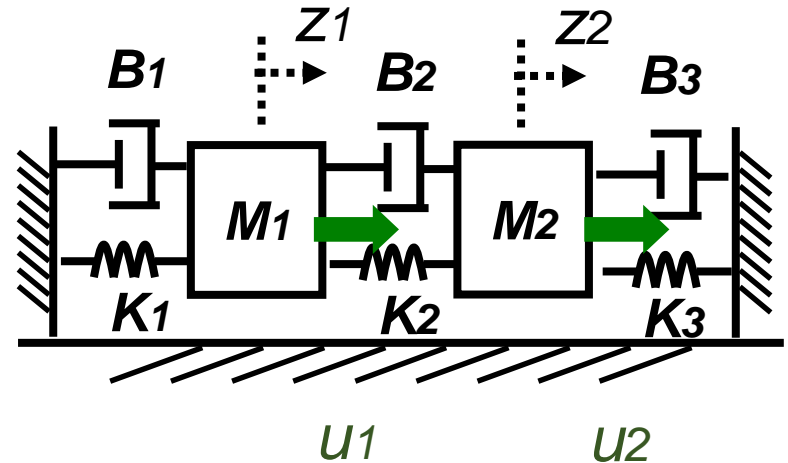


- Define state variables:** $x_1 := \dot{z}$, $x_2 := f_K$
 - Across variable (velocity) for A-type element (Mass)
 - Through variable (force) for T-type element (Spring)

$$\rightarrow \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -B/M & -1/M \\ K & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1/M \\ 0 \end{bmatrix} u \\ y = \begin{bmatrix} 0 & 1/K \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \end{cases}$$

Mass-spring-damper system (2-DOF)

- Inputs: forces (u_1 & u_2)
- Outputs: position (z_1 & z_2)
- Newton's second law



$$\begin{cases} M_1 \ddot{z}_1 &= u_1 - K_1 z_1 - B_1 \dot{z}_1 - K_2(z_1 - z_2) - B_2(\dot{z}_1 - \dot{z}_2) \\ M_2 \ddot{z}_2 &= u_2 - K_2(z_2 - z_1) - B_2(\dot{z}_2 - \dot{z}_1) - K_3 z_2 - B_3 \dot{z}_2 \end{cases}$$



Mass-spring-damper system (2-DOF)

$$\begin{cases} M_1 \ddot{z}_1 &= u_1 - K_1 z_1 - B_1 \dot{z}_1 - K_2(z_1 - z_2) - B_2(\dot{z}_1 - \dot{z}_2) \\ M_2 \ddot{z}_2 &= u_2 - K_2(z_2 - z_1) - B_2(\dot{z}_2 - \dot{z}_1) - K_3 z_2 - B_3 \dot{z}_2 \end{cases}$$

- State variables: $x_1 := z_1$, $x_2 := \dot{z}_1$, $x_3 := z_2$, $x_4 := \dot{z}_2$

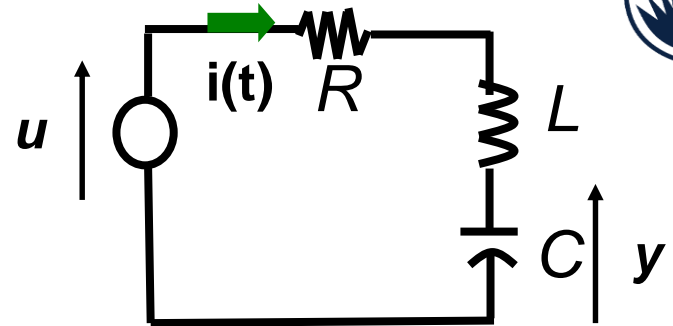
$$\begin{aligned} \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} &= \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{K_1+K_2}{M_1} & -\frac{B_1+B_2}{M_1} & \frac{K_2}{M_1} & \frac{B_2}{M_1} \\ 0 & 0 & 0 & 1 \\ \frac{K_2}{M_2} & \frac{B_2}{M_2} & -\frac{K_2+K_3}{M_2} & -\frac{B_2+B_3}{M_2} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ \frac{1}{M_1} & 0 \\ 0 & 0 \\ 0 & \frac{1}{M_2} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} \\ \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} &= \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} \end{cases} \end{aligned}$$

Fourth order model
(because it has 4 states)

RLC circuit (one-output case)

- By Kirchhoff's voltage law

$$u(t) = Ri(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(\tau) d\tau$$



- Define state variables:** $x_1(t) := i(t)$, $x_2(t) := \frac{1}{C} \int i(\tau) d\tau$
 - Current (through variable) for inductor (T-type element)
 - Voltage (across variable) for capacitor (A-type element)

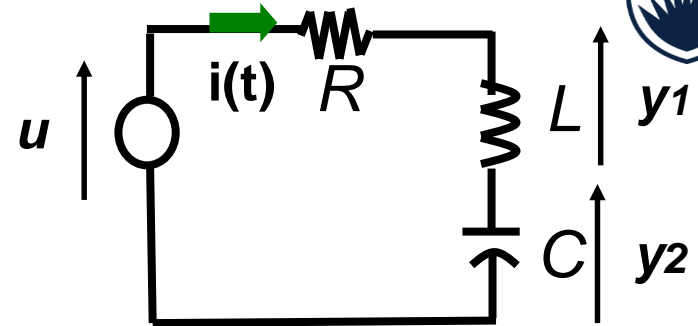
$$\Rightarrow \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -R/L & -1/L \\ 1/C & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1/L \\ 0 \end{bmatrix} u \\ y = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \end{cases}$$



RLC circuit (two-output case)

- By Kirchhoff's voltage law

$$u(t) = Ri(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(\tau) d\tau$$



- Define state variables:** $x_1(t) := i(t)$, $x_2(t) := \frac{1}{C} \int i(\tau) d\tau$

$$\begin{aligned} \begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} &= \begin{bmatrix} -R/L & -1/L \\ 1/C & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1/L \\ 0 \end{bmatrix} u \\ \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} &= \begin{bmatrix} -R & -1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \end{cases} \end{aligned}$$



Summary

- Today's topics
 - Linear state-space model
 - Examples for deriving state-space models
 - Mechanical system
 - Electrical system
- Next, linearization of nonlinear models
- **HW1**: given on Monday, and due in one week.



O : State variable

System Type	Energy Storage Elements		Energy Dissipating Elements
	A-type (Across) Element	T-type (Through) Element	D-type (Dissipative) Element
<i>Translatory-Mechanical</i> v = velocity f = force	<i>Mass</i> $m \frac{dv}{dt} = f$ (Newton's second law) m = mass	<i>Spring</i> $\frac{df}{dt} = kv$ (Hooke's law) k = stiffness	<i>Viscous Damper</i> $f = bv$ b = damping constant
<i>Electrical</i> v = voltage i = current	<i>Capacitor</i> $C \frac{dv}{dt} = i$ C = capacitance	<i>Inductor</i> $L \frac{di}{dt} = v$ L = inductance	<i>Resistor</i> $Ri = v$ R = resistance
<i>Thermal</i> T = temperature difference Q = heat transfer rate	<i>Thermal Capacitor</i> $C_t \frac{dT}{dt} = Q$ C_t = thermal capacitance	<i>None</i>	<i>Thermal Resistor</i> $R_t Q = T$ R_t = thermal resistance
<i>Fluid</i> P = pressure difference Q = volume flow rate	<i>Fluid Capacitor</i> $C_f \frac{dP}{dt} = Q$ C_f = fluid capacitance	<i>Fluid Inertor</i> $I_f \frac{dQ}{dt} = P$ I_f = inertance	<i>Fluid Resistor</i> $R_f Q = P$ R_f = fluid resistance

Review of linear algebra

- Product of a matrix and a vector

$$\begin{aligned}
 Ax &= \begin{matrix} (n \times n) & (n \times 1) \\ \begin{bmatrix} a_{11} & \cdots & a_{1n} \\ \vdots & \ddots & \vdots \\ a_{n1} & \cdots & a_{nn} \end{bmatrix} & \begin{bmatrix} x_1 \\ \vdots \\ x_n \end{bmatrix} \end{matrix} \\
 &= \begin{matrix} \begin{bmatrix} a_{11}x_1 + \cdots + a_{1n}x_n \\ \vdots \\ a_{n1}x_1 + \cdots + a_{nn}x_n \end{bmatrix} \\ (n \times 1) \end{matrix}
 \end{aligned}$$