

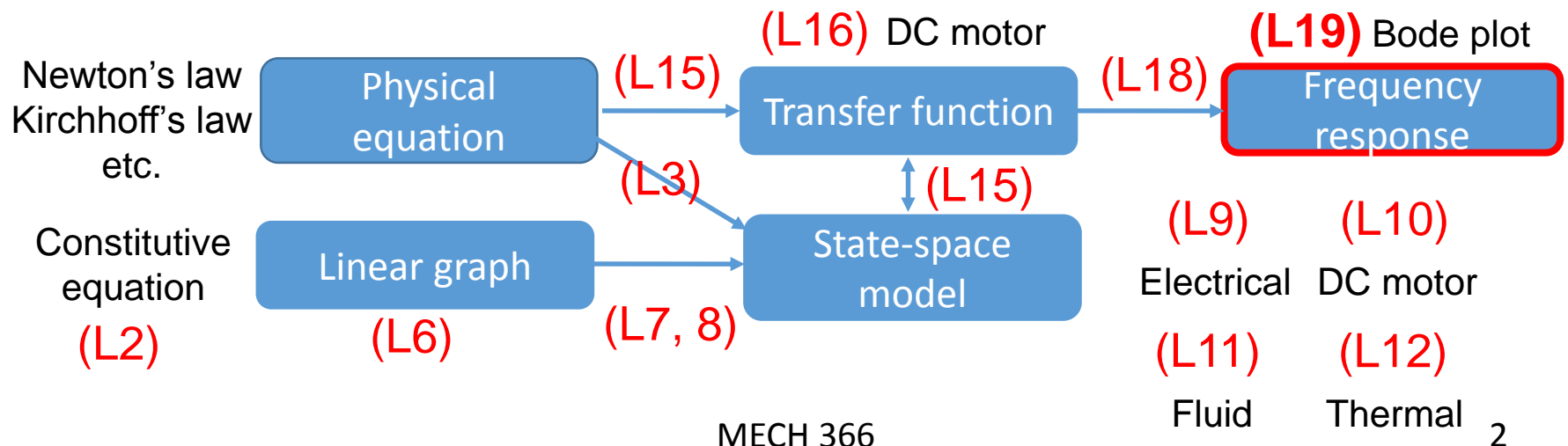
MECH366 : Modeling of Mechatronic Systems

L19 : Bode diagram of first-order and second-order systems


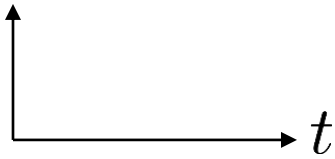
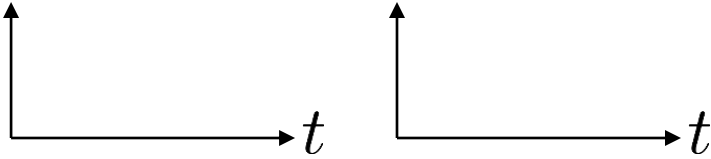
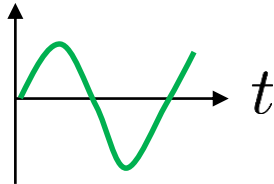
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Today's topic & class schedule

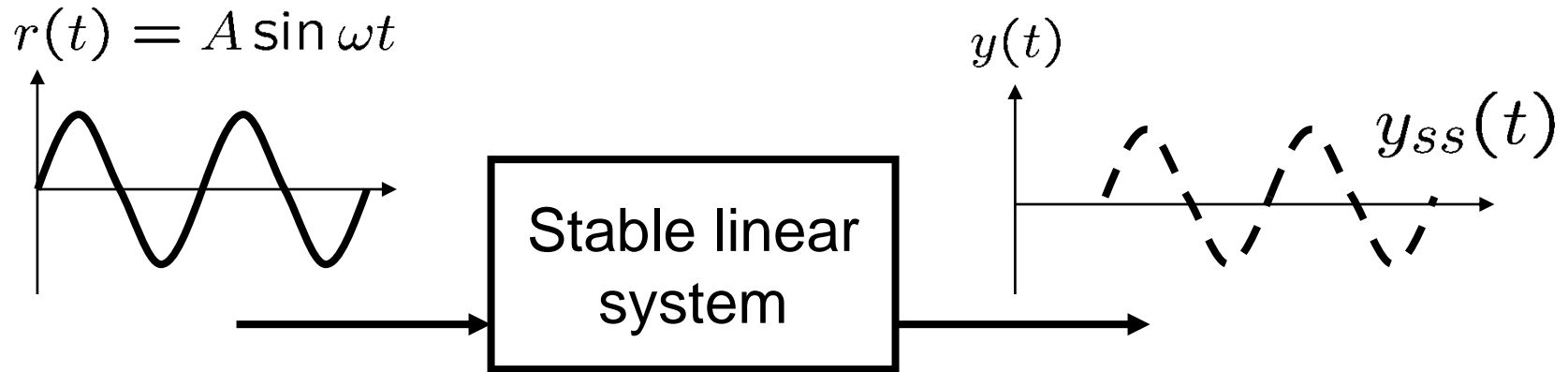
- L18: Nov 15 (Fri): Frequency response
- **L19**: Nov 18 (Mon): Bode diagram (Lab 4 report content, report due Nov 25, 6pm)
- L20: Nov 22 (Fri): Simulink, overdamped system
- L21: Nov 25 (Mon): Stability, course summary



Response analyses (useful for modeling and controller design)

$G(s)$	$\frac{K}{Ts + 1}$	$\frac{K\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$
Step response 	(L16) 	(L17) underdamped (L20) overdamped 
Frequency response (L18) 	(L19) Slide 13	(L19) Slide 16

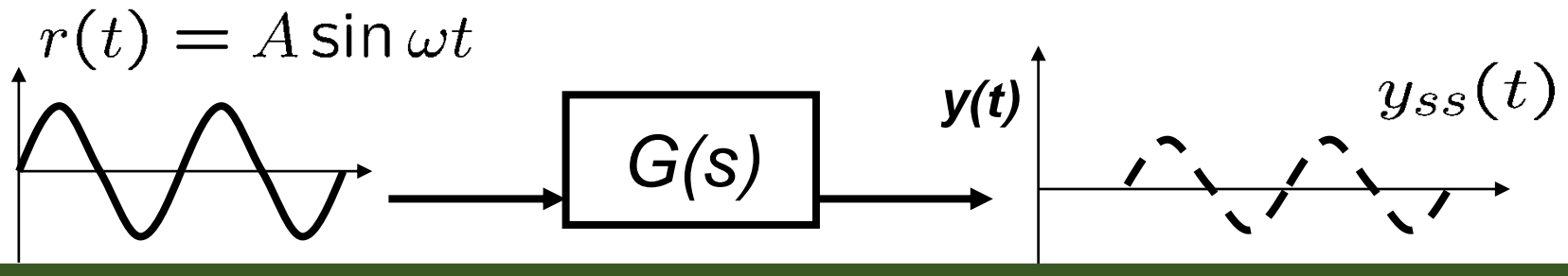
Frequency response (review)



- We would like to analyze a system property by applying a **sinusoidal input** $r(t)$ and observing a response $y(t)$.
- Steady state response $y_{ss}(t)$ (after transient dies out) of a system to sinusoidal inputs is called **frequency response**.

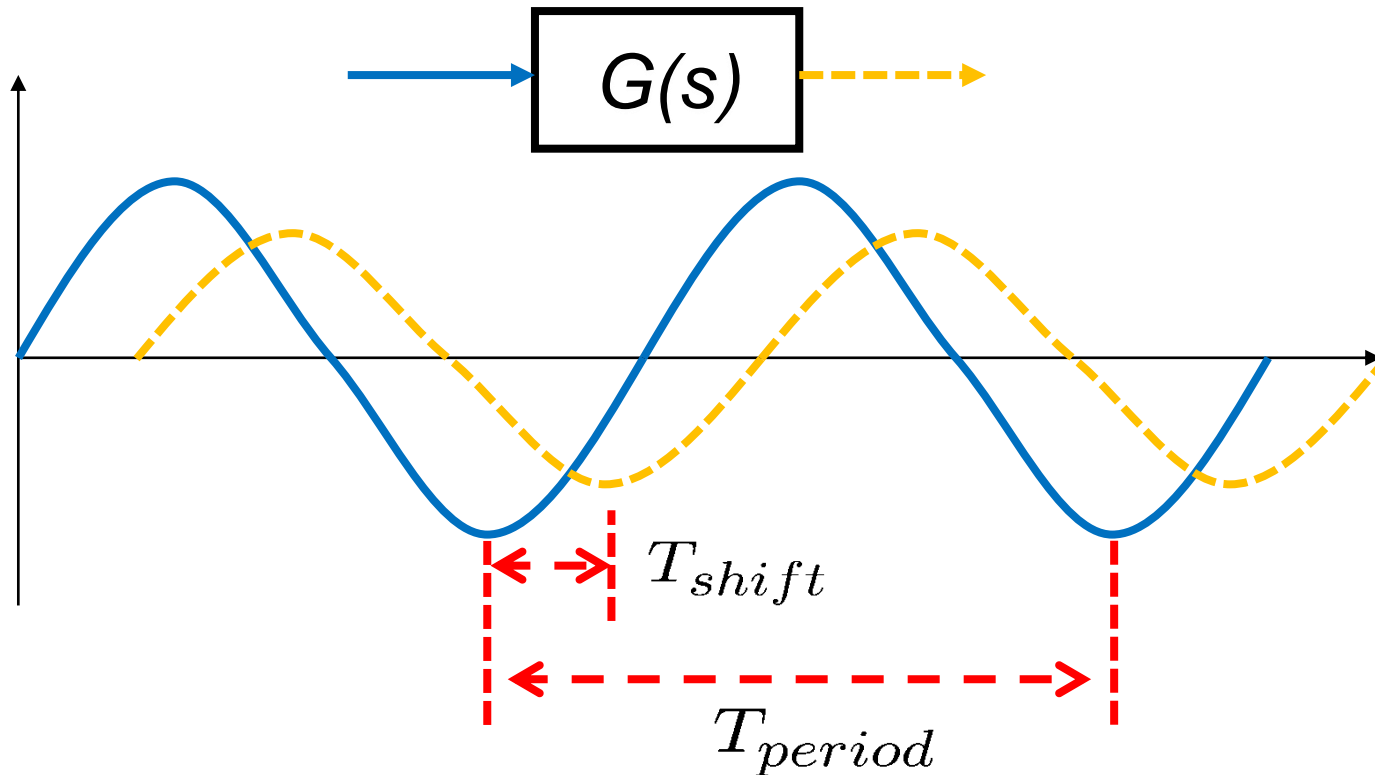
Response to sinusoidal input (review)

- What is the steady state output of a stable linear system when the input is sinusoidal?



- Steady state** output $y_{ss}(t) = A |G(j\omega)| \sin(\omega t + \angle G(j\omega))$
 - Frequency** is same as the input frequency ω
 - Amplitude** is that of input (A) multiplied by $|G(j\omega)|$
 - Phase** shifts $\angle G(j\omega)$ **Gain**

Phase shift (review)

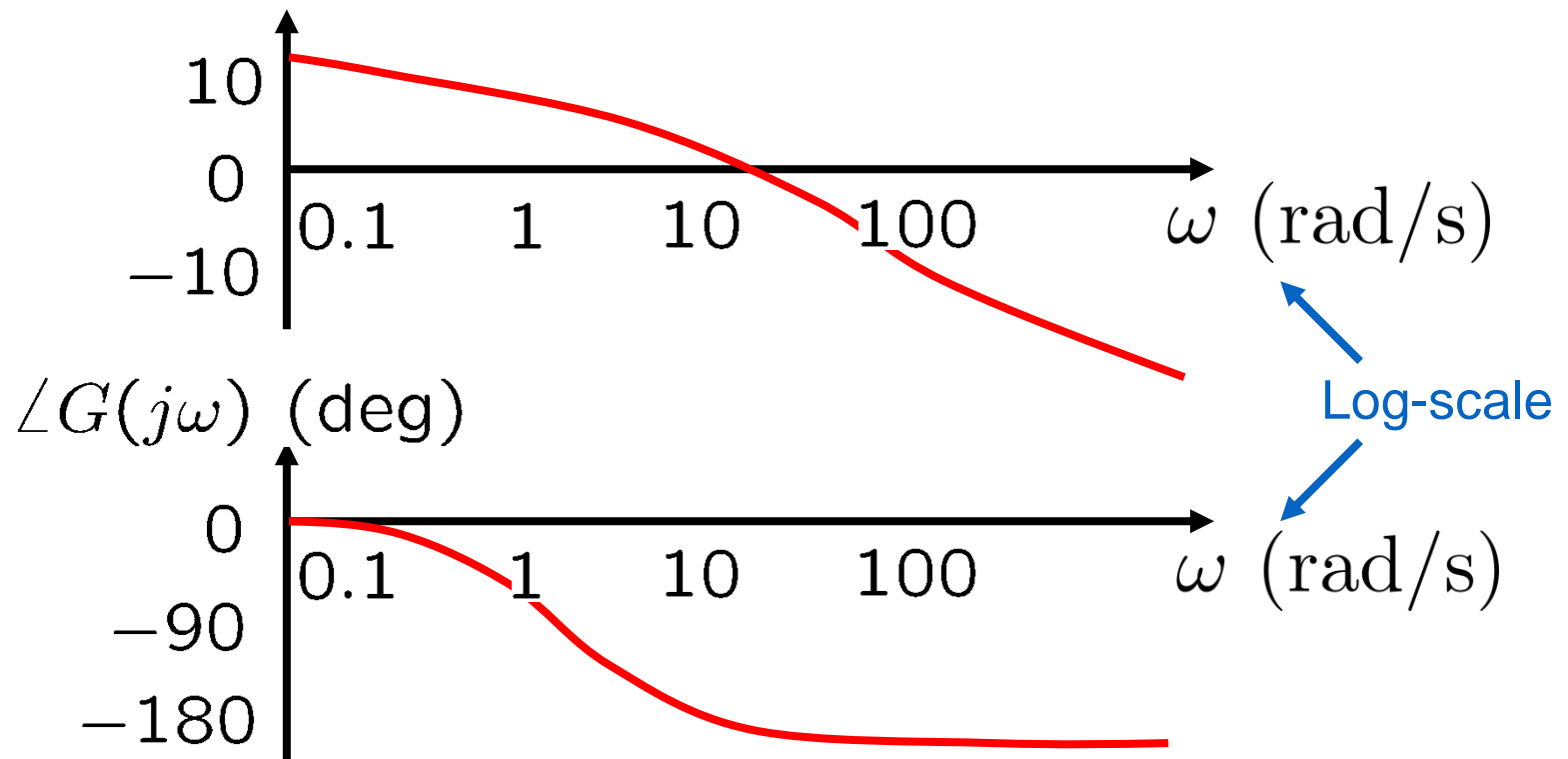


$$\frac{T_{shift}}{T_{period}} = \frac{-\angle G(j\omega)}{360^\circ} \quad \longrightarrow \quad \angle G(j\omega) = -\frac{T_{shift}}{T_{period}} \times 360^\circ$$

Bode plot (Bode diagram) of $G(j\omega)$

- Bode diagram consists of **gain plot** & **phase plot**

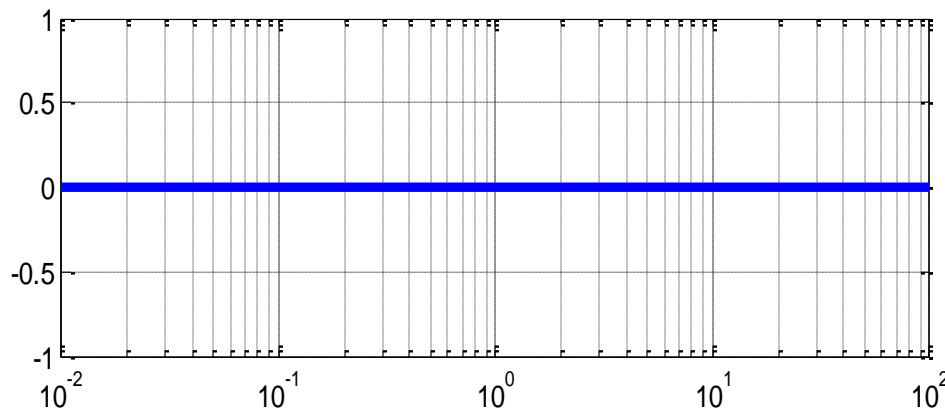
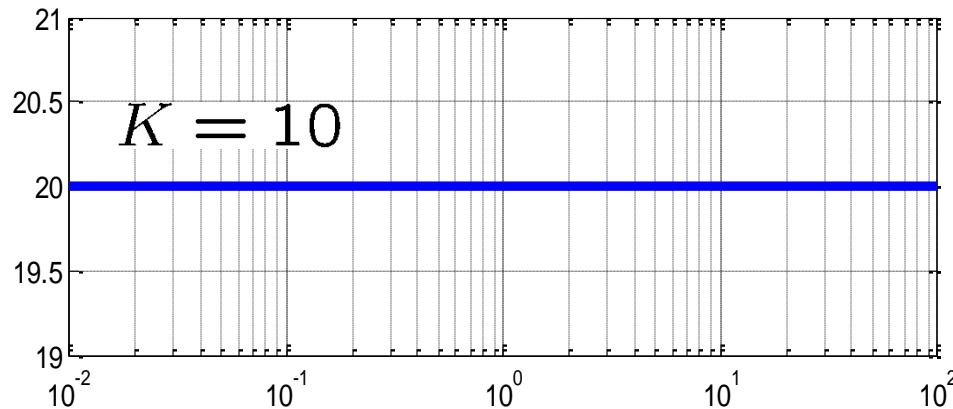
$$20 \log_{10} |G(j\omega)| \text{ (dB)}$$



Bode plot of a constant gain

$$G(s) = K \Rightarrow |G(j\omega)| = K, \angle G(j\omega) = 0^\circ, \forall \omega$$

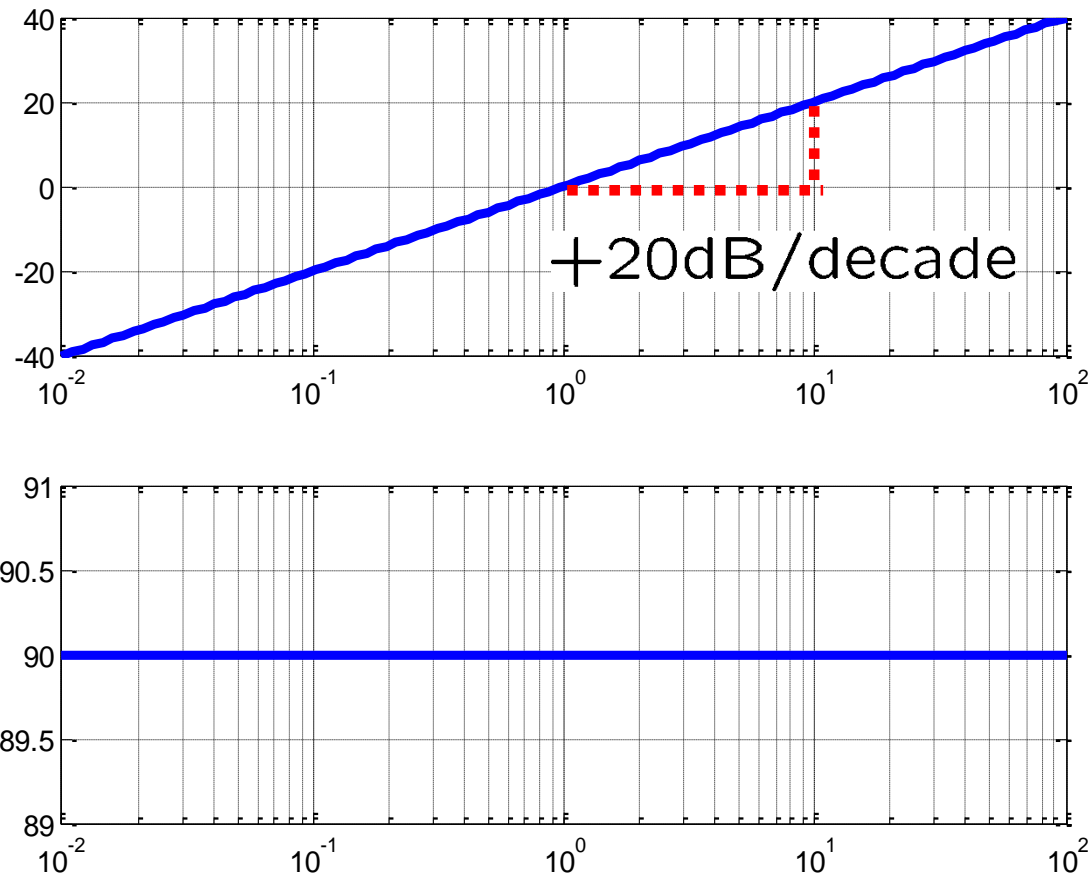
↑
(for all)



K	$20 \log_{10} K$
100	40 dB
10	20 dB
2	≈ 6 dB
1	0 dB
0.1	-20 dB
0.01	-40 dB

Bode plot of a differentiator

$$G(s) = s \Rightarrow |G(j\omega)| = \omega, \angle G(j\omega) = \angle j\omega = 90^\circ, \forall \omega$$

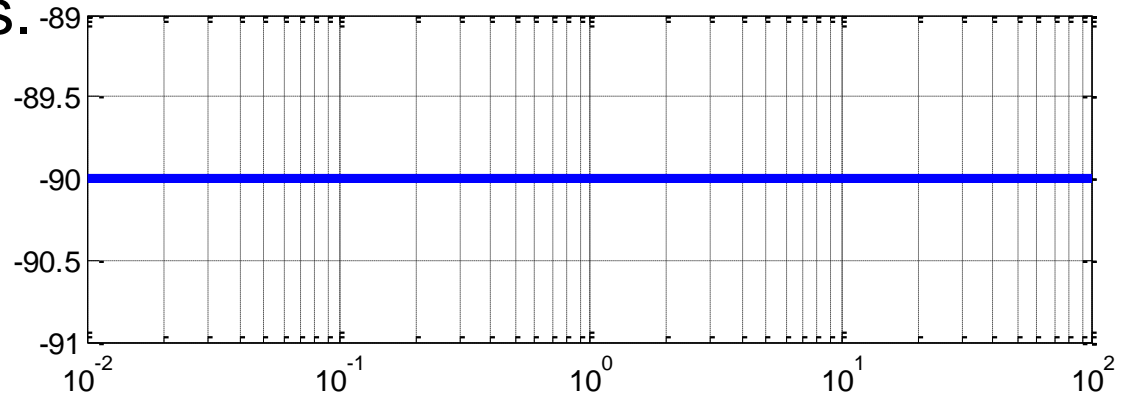
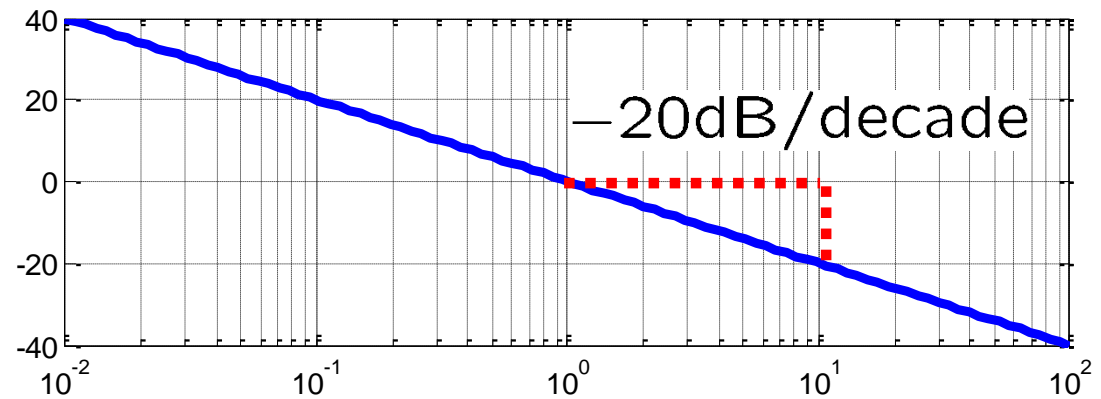




Bode plot of an integrator

$$G(s) = \frac{1}{s} \Rightarrow |G(j\omega)| = \frac{1}{\omega}, \quad \angle G(j\omega) = \angle \frac{1}{j\omega} = -90^\circ, \quad \forall \omega$$

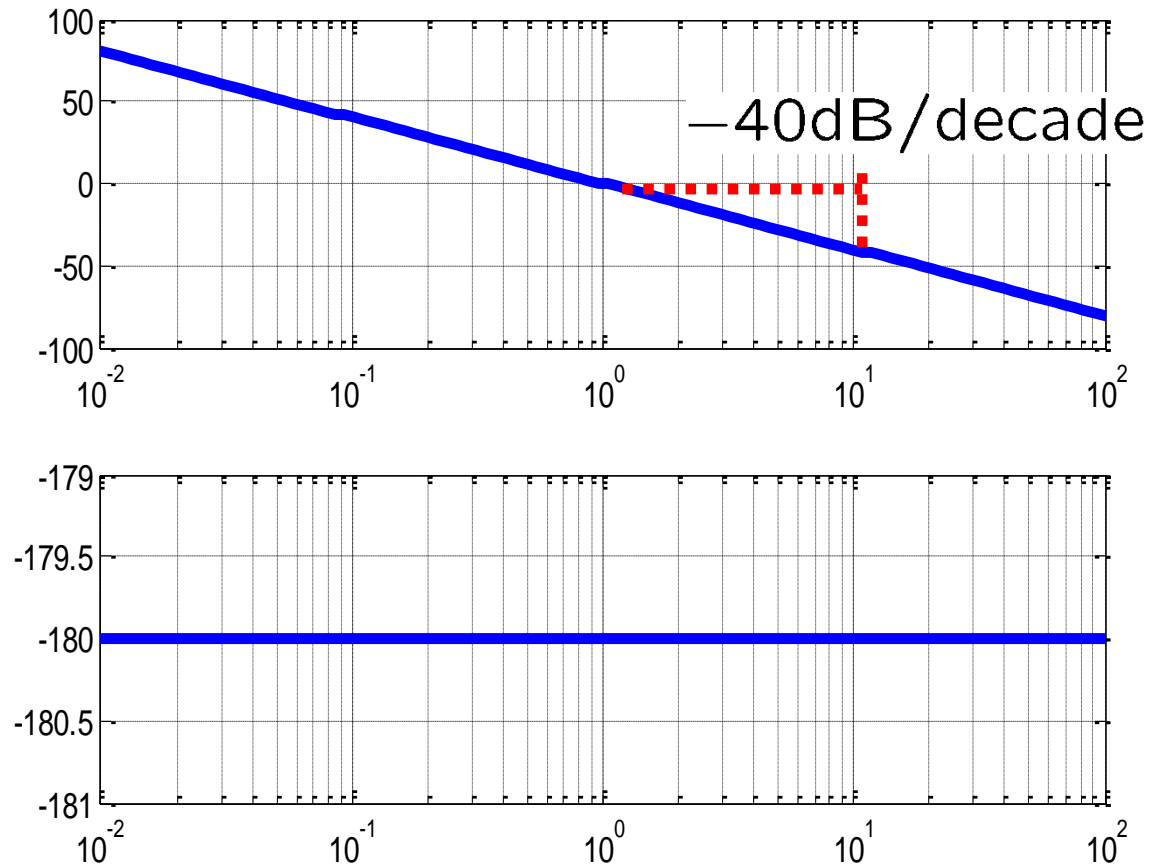
Mirror image of the
Bode plot of $G(s)=s$
with respect to ω -axis.





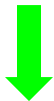
Bode plot of a double integrator

$$G(s) = \frac{1}{s^2} \Rightarrow |G(j\omega)| = \frac{1}{\omega^2}, \angle G(j\omega) = \angle \frac{1}{(j\omega)^2} = -180^\circ, \forall \omega$$



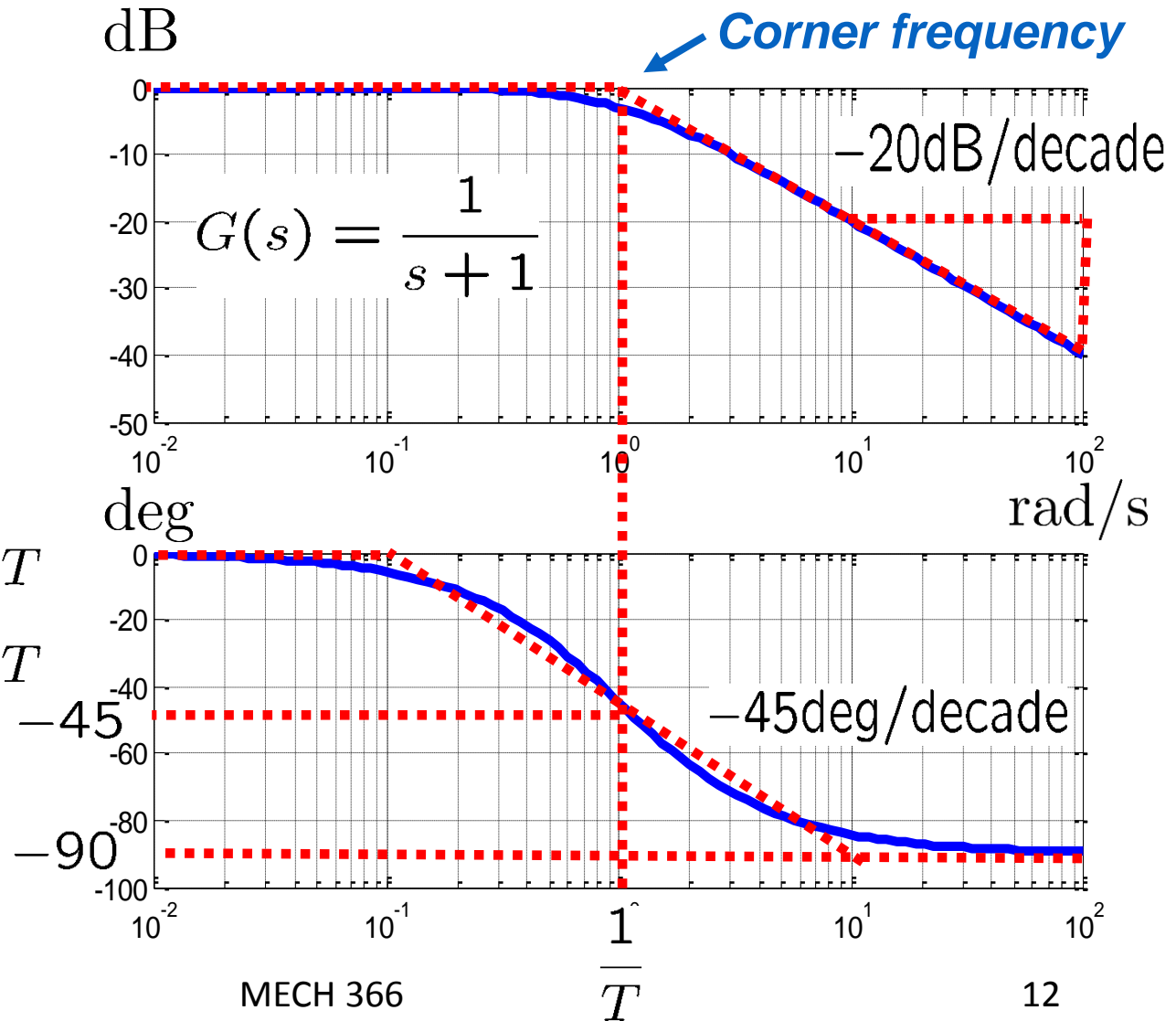
Bode plot of a 1st order system

$$G(s) = \frac{1}{Ts + 1}$$



$$G(j\omega) = \frac{1}{j\omega T + 1}$$

$$\approx \begin{cases} 1 & \text{if } 1 \gg \omega T \\ \frac{1}{j\omega T} & \text{if } 1 \ll \omega T \end{cases}$$





Exercises of sketching Bode plot

$$G(s) = \frac{1}{s + 1}$$

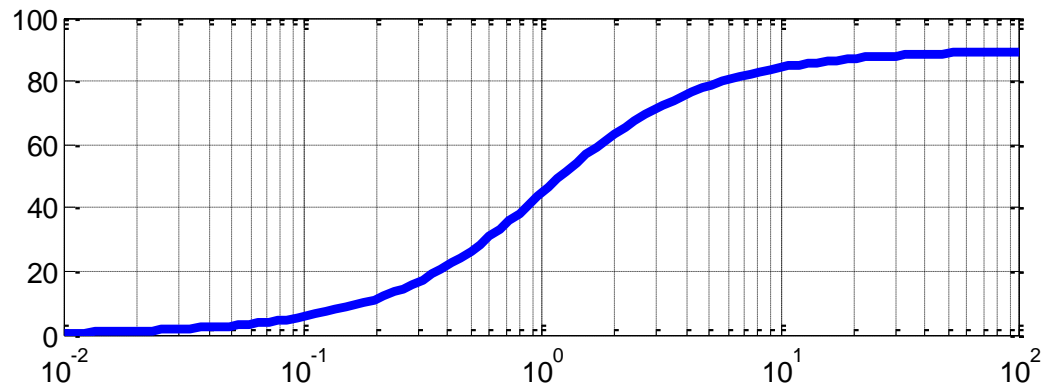
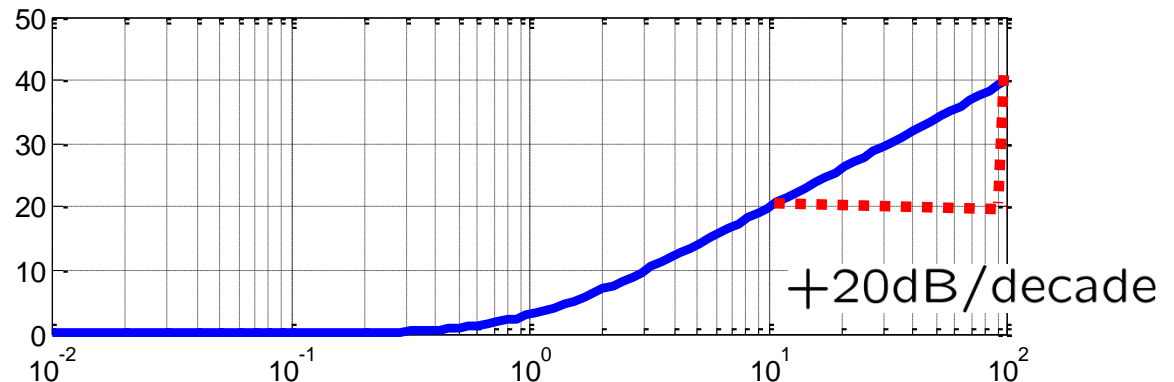
$$G(s) = \frac{1}{0.1s + 1}$$

$$G(s) = \frac{1}{10s + 1}$$

Bode plot of an inverse system

$$G(s) = Ts + 1 = \left(\frac{1}{Ts + 1} \right)^{-1}$$

Mirror image of the original Bode plot with respect to ω -axis.



Bode plot of a 2nd order system

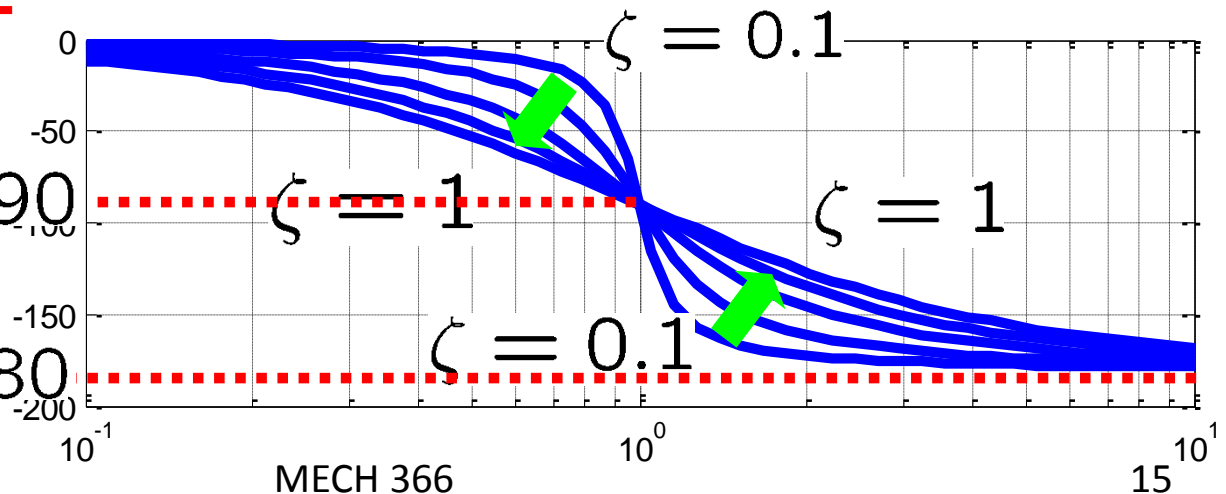
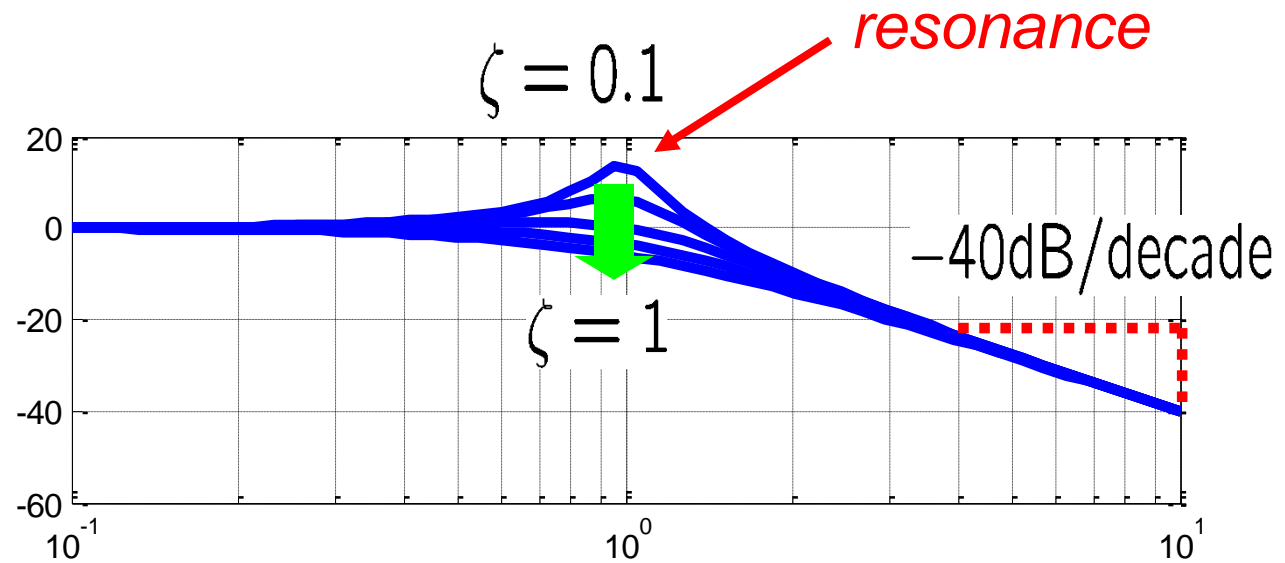
$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

Resonant freq.

$$\omega_n \sqrt{1 - 2\zeta^2} \approx \omega_n$$

Peak gain

$$\frac{1}{2\zeta\sqrt{1 - \zeta^2}} \approx \frac{1}{2\zeta}$$



An advantage of Bode plot

- Bode plot of a series connection $G_1(s)G_2(s)$ is the addition of each Bode plot of G_1 and G_2 .

- Gain

$$20 \log_{10} |G_1(j\omega)G_2(j\omega)| = 20 \log_{10} |G_1(j\omega)| + 20 \log_{10} |G_2(j\omega)|$$

- Phase

$$\angle G_1(j\omega)G_2(j\omega) = \angle G_1(j\omega) + \angle G_2(j\omega)$$

- Later, we use this property to design $C(s)$ so that $G(s)C(s)$ has a “desired” shape of Bode plot. (in Controls course MECH467)

Short proofs (optional)

- Use polar representation

$$G_1(j\omega) = |G_1(j\omega)|e^{j\angle G_1(j\omega)} \quad G_2(j\omega) = |G_2(j\omega)|e^{j\angle G_2(j\omega)}$$

$$\begin{aligned} \text{Then, } G_1(j\omega)G_2(j\omega) &= |G_1(j\omega)||G_2(j\omega)|e^{j\angle G_1(j\omega)}e^{j\angle G_2(j\omega)} \\ &= |G_1(j\omega)||G_2(j\omega)|e^{j\{\angle G_1(j\omega)+\angle G_2(j\omega)\}} \end{aligned}$$

Therefore,

$$20\log_{10}|G_1(j\omega)G_2(j\omega)| = 20\log_{10}|G_1(j\omega)| + 20\log_{10}|G_2(j\omega)|$$

$$\angle G_1(j\omega)G_2(j\omega) = \angle G_1(j\omega) + \angle G_2(j\omega)$$

Example 1

- Sketch the Bode plot of a transfer function

$$G(s) = \frac{10}{s}$$

1. Decompose $G(s)$ into a product form:

$$G(s) = 10 \cdot \frac{1}{s}$$

2. Sketch a Bode plot for each component on the same graph.
3. Add them all on both gain and phase plots.

Example 1 (cont'd)

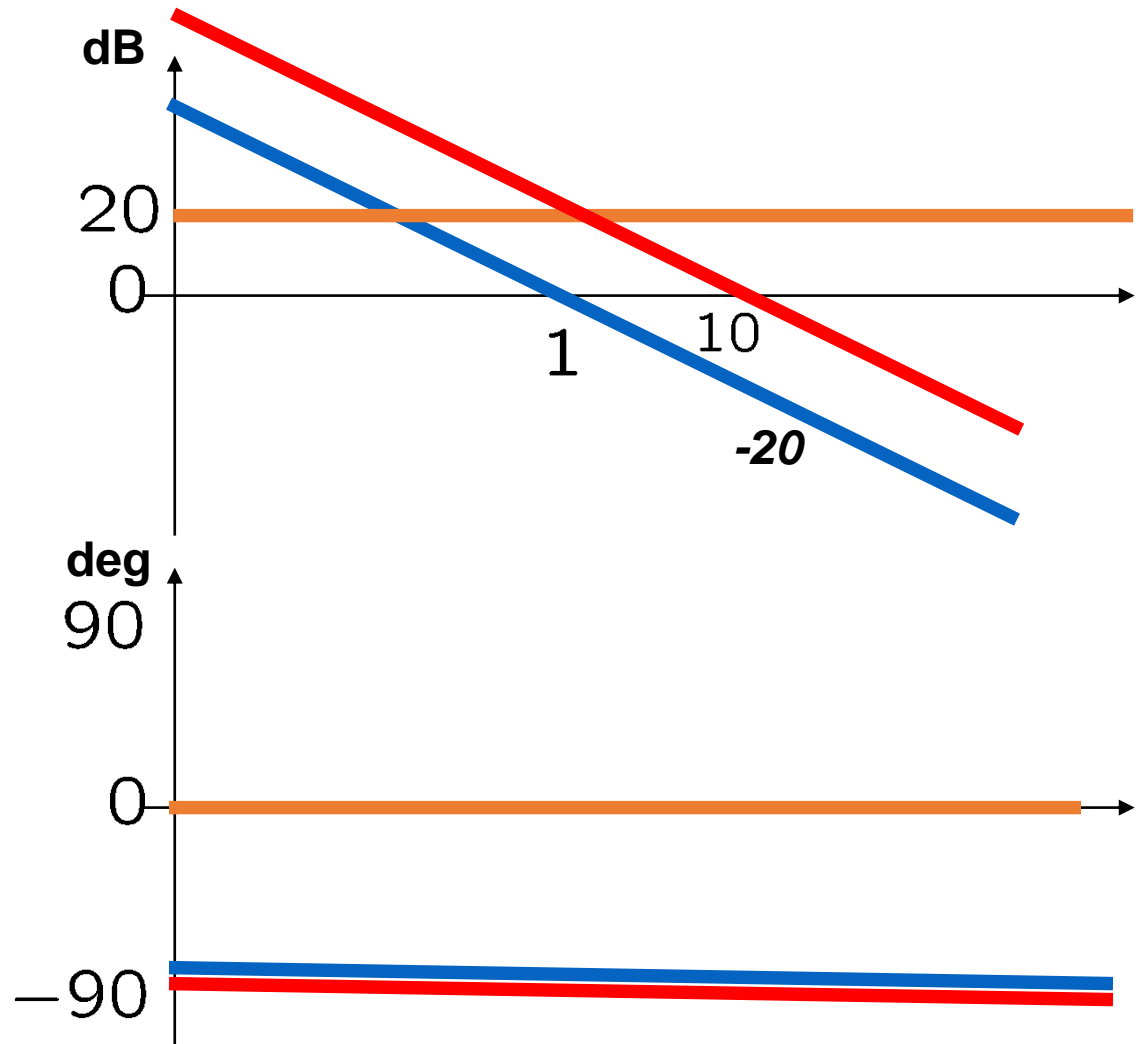
$$G(s) = 10$$

×

$$G(s) = \frac{1}{s}$$



$$G(s) = \frac{10}{s}$$



Example 2

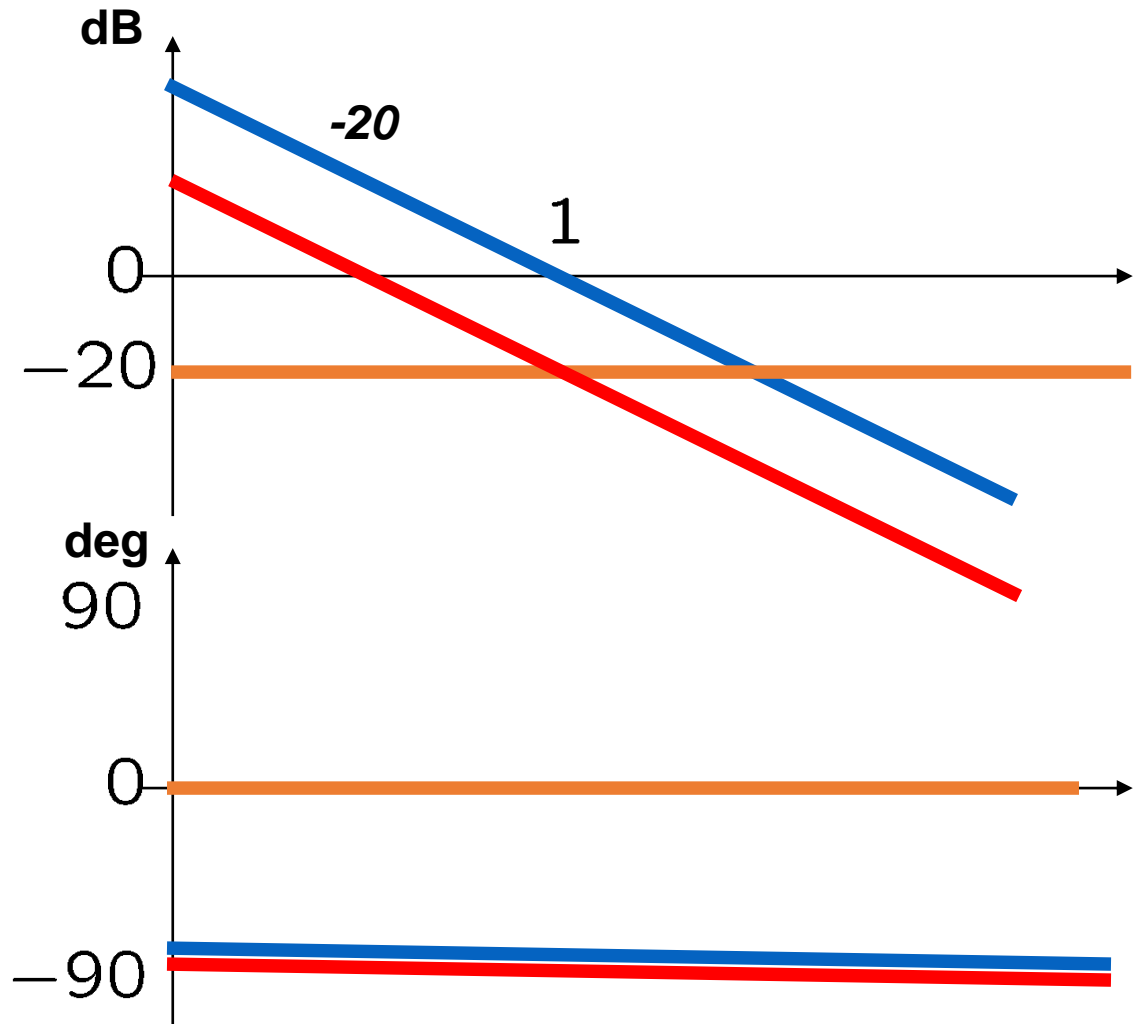
$$G(s) = 0.1$$

×

$$G(s) = \frac{1}{s}$$



$$G(s) = \frac{0.1}{s}$$



Example 3

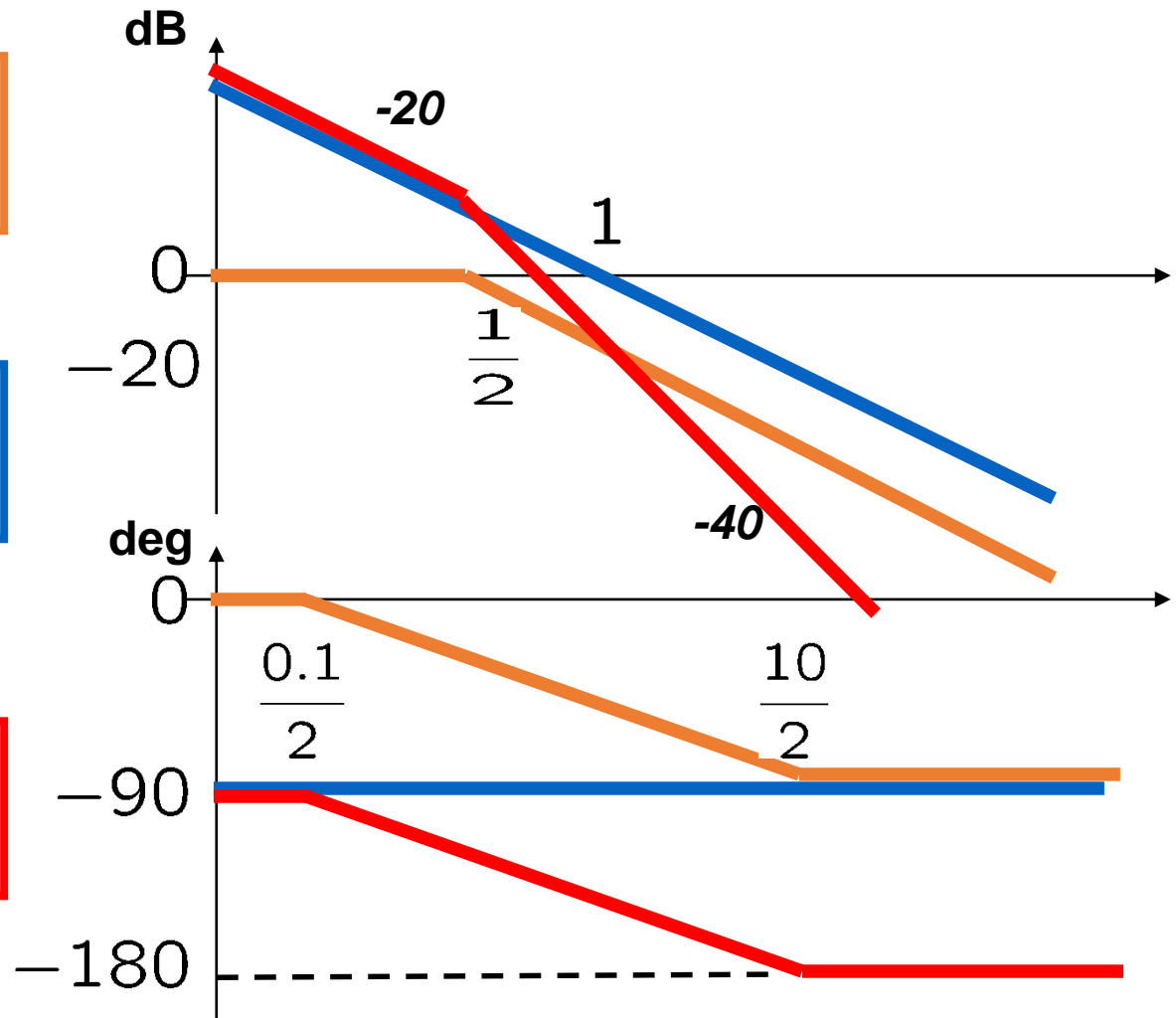
$$G(s) = \frac{1}{2s + 1}$$

×

$$G(s) = \frac{1}{s}$$



$$G(s) = \frac{1}{s(2s + 1)}$$





Summary

- How to sketch/draw Bode diagram (bode.m)
- Next,
 - Simulink
 - Step response of overdamped systems
- **Project:** Fridays Nov 22, 29 (presentation)
- **Lab 4 report:** Due Nov 25 (Monday), 6pm