

### EECE 376, Home Work Assignment # 1:

Follow through Example 1A, and do Study Problem, SP1.2-2, SP1.2-4

Follow through Example 1B, and do Study Problems SP1.3-1, SP1.3-3

Follow through Example 1D, and do Study Problems SP1.5-2 and SP1.5-3.

Textbook Chapter Problem(s): 3, 8, 9, 15, and 16.

Solutions:

#### SP1.2-2

From SP1.2-1, we have phasors of voltage and current:

$$\tilde{V} = 1 \angle 0^\circ, \quad \tilde{I} = 1 \angle 180^\circ$$

Then we have  $\omega = 2\pi \cdot 60 \approx 377 \text{ rad/s}$

$$v = \sqrt{2} \cdot \cos(\omega t) = \sqrt{2} \cdot \cos(377 \cdot t)$$

$$i = \sqrt{2} \cdot \cos(\omega t + \pi) = \sqrt{2} \cdot \cos(377 \cdot t + \pi)$$

$$\begin{aligned} p &= v \cdot i = \sqrt{2} \cdot \cos(377 \cdot t) \cdot \sqrt{2} \cdot \cos(377 \cdot t + \pi) \\ &= 2 \left( \frac{1}{2} \cos(754 \cdot t + \pi) + \frac{1}{2} \cos(-\pi) \right) = \\ &= -1 + \cos(754 \cdot t + \pi) \end{aligned}$$

SP 1.2-4 For simplicity, and without lack of generality, neglect resistance and assume the voltage is along the real axis. In this case the relationship between voltage and current is given by:

$$V \angle 0 = j(X_L - X_C) \tilde{I} + \tilde{E}$$

One can observe that the magnitude and phase of  $\tilde{E}$  controls the magnitude and direction of  $\tilde{I}$ . For example, if  $\tilde{E} = 2 \angle 0^\circ$  and  $X_L > X_C$ ,  $\tilde{I}$  will lead  $\tilde{V}$ . In contrast, if  $\tilde{E} = 0.5 \angle 0^\circ$  and  $X_L > X_C$ ,  $\tilde{I}$  will lag  $\tilde{V}$ .

$$\text{SP1.3-1. } Ni = \mathcal{R}_{ab}(\Phi_1 + \Phi_2) + \mathcal{R}_{bcda} \Phi_1$$

$$\Phi_1 = \frac{1}{\mathcal{R}_{bcda}} [Ni - \mathcal{R}_{ab}(\Phi_1 + \Phi_2)]$$

$$\Phi_1 = \frac{1}{358,099} \left[ -(109,419)(2.547 \times 10^{-3}) \right] = 2.014 \times 10^{-3} \text{ Wb}$$

SP1.3-3. With windings as shown in Fig. 1B-1 and with the center leg removed, the total mmf is

$$\begin{aligned} \text{mmf}_t &= \text{mmf}_1 + \text{mmf}_2 = N_1 I_1 + N_2 I_2 \\ &= (150)(9) + (90)(-15) = 0 \end{aligned}$$

**SP1.5-2.** During steady-state conditions, the time rate-of-change of  $i_1$  is zero; therefore, a voltage is not induced in the 2-winding. Hence, for the 2-winding open or short circuited  $I_1 = \frac{V}{r_1} = \frac{12}{6} = 2 \text{ A}$  and  $I_2 = 0$ .

**SP1.5-3.**  $Z = r_1 + j\omega_e(L_{l1} + L_{m1}) = 6 + j(100)(13.5 + 263.9) \times 10^{-3}$   
 $= 6 + j27.74 = 28.38 \angle 77.8^\circ$

$$\tilde{I}_1 = \frac{\tilde{V}_1}{Z} = \frac{10 \angle 0^\circ}{28.38 \angle 77.8^\circ} = 0.352 \angle -77.8^\circ \text{ A}$$

**3.** The reluctance of the iron is

$$\mathcal{R}_m = \frac{l}{\mu_1 A} = \frac{(4)(0.25)}{(4000)(4\pi \times 10^{-7})(0.05)^2} = 79,577 \text{ H}^{-1}$$

$$L_{12} = \frac{N_1 N_2}{\mathcal{R}_m} = \frac{(50)(100)}{79,577} = 0.0628 \text{ H}$$

$$L_{m1} = \frac{N_1^2}{\mathcal{R}_m} = \frac{50^2}{79,577} = 0.0314 \text{ H}$$

$$L_{m2} = \frac{N_2^2}{\mathcal{R}_m} = \frac{100^2}{79,577} = 0.1257 \text{ H}$$

**8.**  $\tilde{V}_1 = \frac{10}{2\sqrt{2}} \angle 0^\circ = 3.54 \angle 0^\circ \text{ V}$

$$Z = r_1 + r_2' + j\omega_e(L_{l1} + L_{l2}') = 10 + 10 + j(2\pi)(30)(30 + 30) \times 10^{-3}$$

$$= 20 + j11.31 \Omega$$

$$\tilde{I}_1 = \frac{\tilde{V}_1}{Z} = \frac{3.54 \angle 0^\circ}{20 + j11.31} = 0.154 \angle -29.5^\circ \text{ A}$$

**9.** Since  $\omega_e = 400$ ,  $X_{m1} = 400 \Omega$ . Neglecting the magnetizing current  $i_1 = -i_2'$ .

(a)  $\tilde{V}_1 = \frac{2}{\sqrt{2}} \angle 0^\circ = \sqrt{2} \angle 0^\circ$

$$\tilde{I}_1 = \frac{\tilde{V}_1}{(r_1 + r_2 + R_L) + j\omega_e(L_{l1} + L_{l2})} = \frac{\sqrt{2} \angle 0^\circ}{4 + j(400)(0.02)} = 0.158 \angle -63.4^\circ \text{ A}$$

(b)  $I_1 = \sqrt{2} 0.158 \cos(400t - 63.4^\circ)$

$$15. \quad L_m(x) = \frac{k}{k_0 + x}$$

$$k = \frac{N^2 \mu_0 A_i}{2} = \frac{(500)^2 (4\pi \times 10^{-7}) (4 \times 10^{-4})}{2} = 2\pi \times 10^{-5}$$

$$k_0 = \frac{l_i}{2\mu_{ti}} = \frac{20 \times 10^{-2}}{(2)(1000)} = 10^{-4}$$

$$L_m(x) = \frac{2\pi \times 10^{-5}}{10^{-4} + x} \text{ H}$$

The approximation for  $x > 0$  is

$$L_m(x) = \frac{2\pi \times 10^{-5}}{x} \text{ H}$$

Now to find minimum value of  $x$

$$\frac{2\pi \times 10^{-5}}{x} = 1.1 \frac{2\pi \times 10^{-5}}{10^{-4} + x}$$

Solving for  $x$  yields  $x = 1 \text{ mm}$ . Thus, the approximate expression is 10% in error at  $x = 1 \text{ mm}$  and less than this for  $x > 1 \text{ mm}$ .

$$16. \quad L_m(x) = \frac{k}{k_0 + x}$$

$$\frac{\partial L_m(x)}{\partial x} = \frac{-k}{(k_0 + x)^2}$$

$$v = r i + \left( L_1 + \frac{k}{k_0 + x} \right) \frac{di}{dt} - i \frac{k}{(k_0 + x)^2} \frac{dx}{dt}$$

$$v = r \sqrt{2} I_s \cos \omega_e t - \left( L_1 + \frac{k}{k_0 + t} \right) \omega_e \sqrt{2} I_s \sin \omega_e t - \sqrt{2} I_s \cos \omega_e t \left[ \frac{k}{(k_0 + t)^2} \right]$$

Gathering terms

$$v = \left[ r - \frac{k}{(k_0 + t)^2} \right] \sqrt{2} I_s \cos \omega_e t - \left( L_1 + \frac{k}{k_0 + t} \right) \omega_e \sqrt{2} I_s \sin \omega_e t$$

Taking the limit as  $t \rightarrow \infty$

$$v = r \sqrt{2} I_s \cos \omega_e t - L_1 \omega_e \sqrt{2} I_s \sin \omega_e t$$

which is the voltage equation for a linear r-L circuit. In phasor form,

$$\tilde{V} = (r + j \omega_e L_1) \tilde{I}$$