

Numerical Simulation Methods in Geophysics, Part 8: 2D Electromagnetics

1. MGPy+MGIN

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Recap & Motivation

- use *pyGMLi* for generating meshes, matrices & applying BC
- spatial discretization determines accuracy of solution
- source fields ($1/r$ etc.) require fine mesh or higher order elements
- boundaries for decaying fields need to be far away
- secondary field approach addresses both problems

Remaining lectures & topics

8. 11.12. 2D electromagnetics
 9. 18.12. 2D/3D EM and other element types
 10. 08.01. 3D vectorial electromagnetics (dies academicus on 15.1.)
 11. 22.01.
 12. 29.01. Advection problems and Finite Volumes
 13. 05.02. (online) special simulation methods
 14. 12.02. (not available) shift to later
-
- 7 excercises

Electromagnetics in 2D

- move from Poisson to Helmholtz type of PDE
- real-valued to complex-valued problem
- move from 1D to 2D (and eventually 3D)
- scalar to vectorial solution
- secondary field approach

Maxwells equations

- Faraday's law: currents & varying electric fields \Rightarrow magnetic field

$$\nabla \times \mathbf{H} = \frac{\partial \mathbf{D}}{\partial t} + \mathbf{j}$$

- Ampere's law: time-varying magnetic fields induce electric field

$$\nabla \times \mathbf{E} = - \frac{\partial \mathbf{B}}{\partial t}$$

- $\nabla \cdot \mathbf{D} = \varrho$, $\nabla \cdot \mathbf{B} = 0$, $\mathbf{D} = \epsilon \mathbf{E}$ & $\mathbf{B} = \mu \mathbf{H}$

Maxwell in frequency domain

Assumption: AC fields with angular frequency ω (decomposition)

$$\mathbf{E} = \mathbf{E}_0 e^{i\omega t} \quad \text{or} \quad \mathbf{H} = \mathbf{H}_0 e^{i\omega t}$$

$$\nabla \times \mathbf{H} = i\omega\epsilon\mathbf{E} + \sigma\mathbf{E} = (\sigma + i\omega\epsilon)\mathbf{E}$$

$$\nabla \times \mathbf{E} = -i\omega\mu\mathbf{H}$$

take curl of one of the equations and insert in the other

Helmholtz equation

see also [Theory EM](#)

take curl of one of the equations and insert in the other

$$\nabla \times \mu^{-1} \nabla \times \mathbf{E} + i\omega\sigma\mathbf{E} - \omega^2\epsilon\mathbf{E} = \nabla \times \mathbf{j}_s$$

$$\nabla \times \sigma^{-1} \nabla \times \mathbf{H} + i\omega\mu\mathbf{H} - \omega^2\mu\epsilon\mathbf{H} = 0$$

- Helmholtz type of PDE, vectorial

Quasi-static approximation

Assume: $\omega^2 \mu \epsilon < \omega \mu \sigma$, no sources ($\nabla \cdot \mathbf{j}_s = 0$), + vector identity

$$\nabla \times \nabla \times \mathbf{F} = \nabla \nabla \cdot \mathbf{F} - \nabla^2 \mathbf{F}$$

leads with $\nabla \cdot \mathbf{E} = 0 = \nabla \cdot \mathbf{B}$ to the vector Helmholtz PDE

$$-\nabla^2 \mathbf{E} + \imath \omega \mu \sigma \mathbf{E} = 0$$

$$-\nabla \cdot \sigma^{-1} \nabla \mathbf{H} + \imath \omega \mu \mathbf{H} = 0$$

Helmholtz equation $-\nabla^2 \mathbf{H} - k^2 \mathbf{H} = 0$ with $k^2 = -\imath \omega \mu \sigma$

TM polarization

Transverse magnetic (TM) mode

Assume the source field is oscillating perpendicular to the modelling plane, i.e.

$$\mathbf{H} = [H_x, 0, 0]^T e^{i\omega t}.$$

Then the PDE holds for the scalar H_x (now only H)

$$-\nabla \cdot \sigma^{-1} \nabla H_x(y, z) + i\omega \mu H_x(y, z) = 0$$

Halfspace solution

$$\frac{\partial^2 H}{\partial z^2} + k^2 H = 0 \quad \text{with} \quad k^2 = -\omega\mu\sigma$$

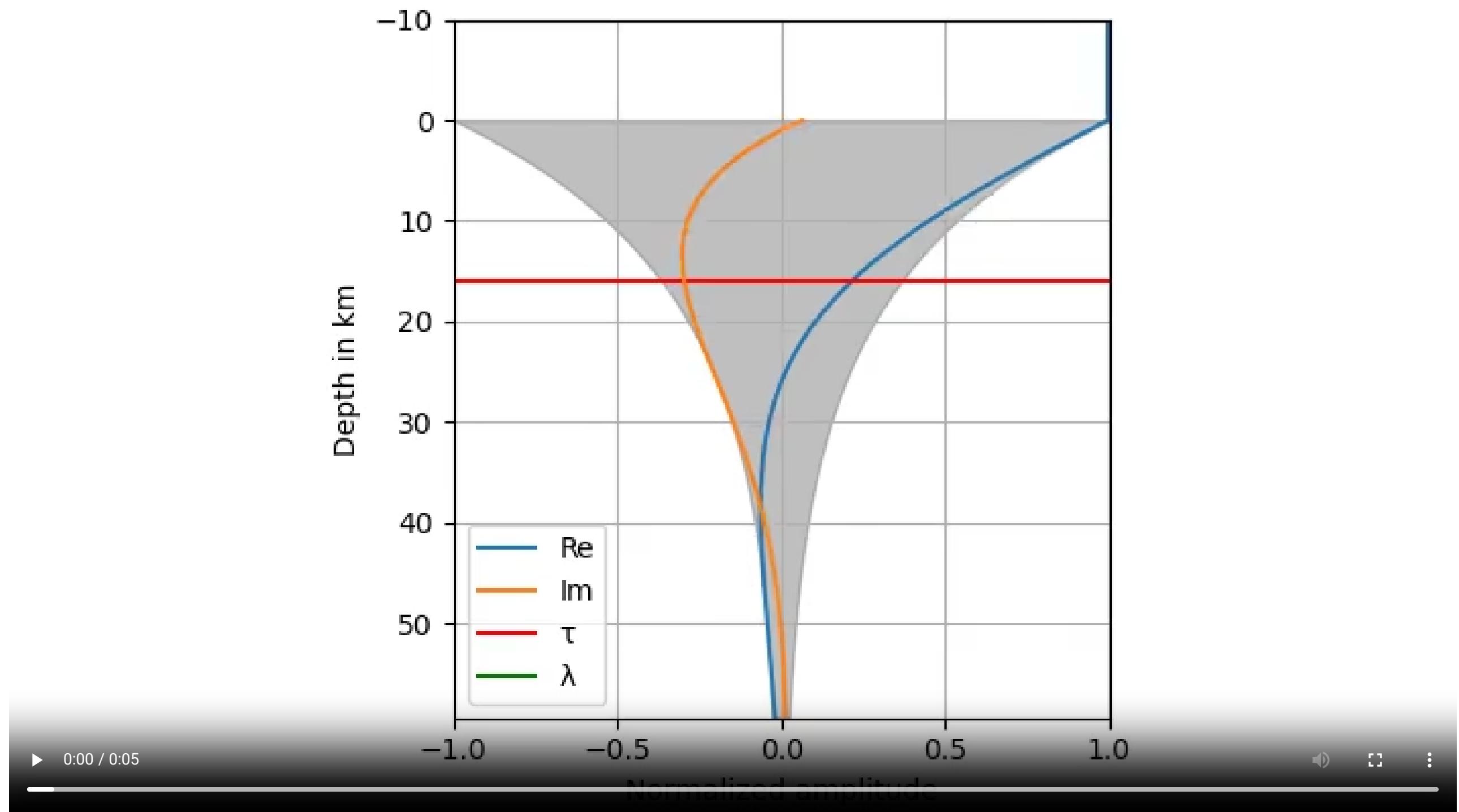
In the Earth, the solution is $H(z) = H_0 e^{-ikz}$ $\partial_{zz}H = -k^2 H$

$$k = \sqrt{-\omega\mu\sigma} = \sqrt{\omega\mu\sigma/2}(1 - i) = 1/d(1 - i)$$

with skin depth (1/e decay) $d = \sqrt{2/\omega\mu\sigma} = \sqrt{T/\pi\mu\sigma}$ ($\sim 500\sqrt{\rho/f}$)

$$-ik = -(1 + i)/d \Rightarrow H = H_0 e^{-z/d} e^{-iz/d} \equiv \cos(\omega t - z/d)$$

Fields in homogeneous half-space



Electromagnetic fields in the Earth

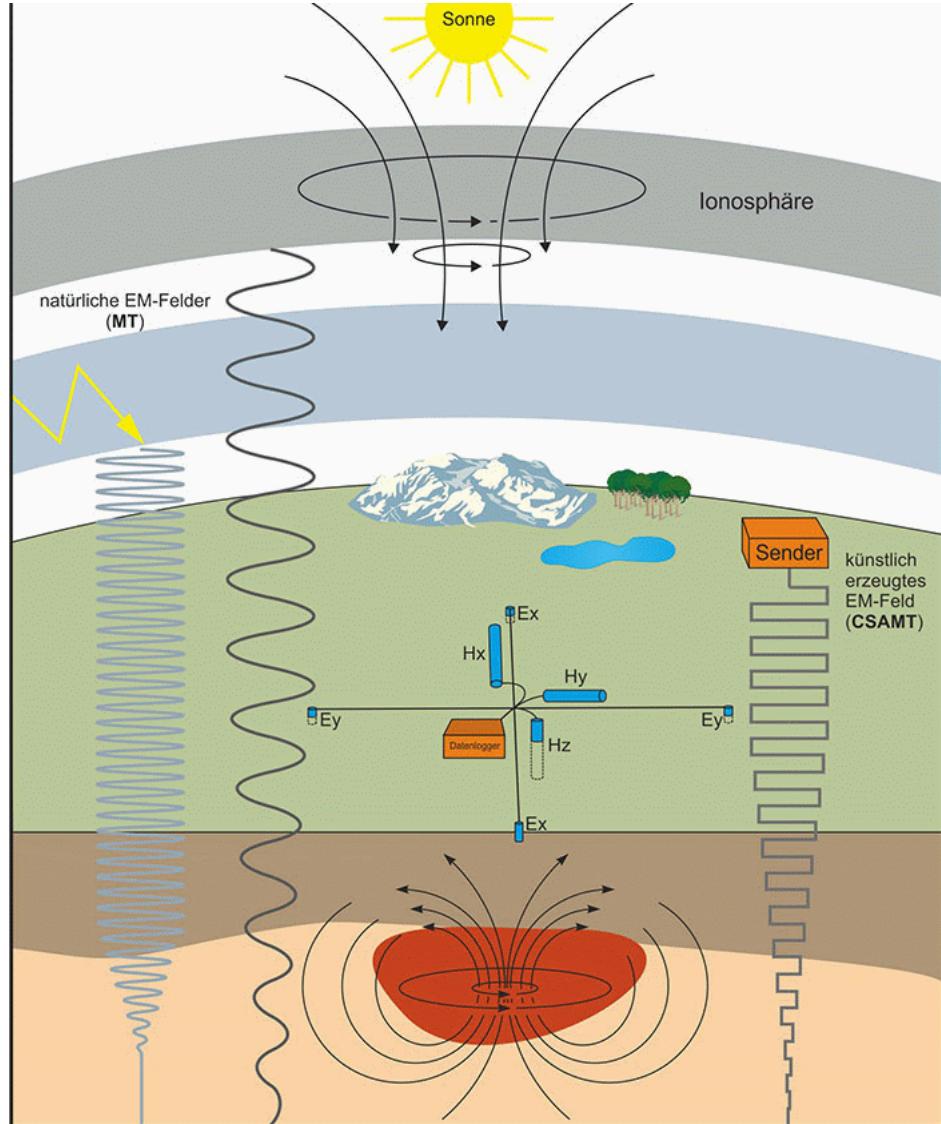
Assume E field in x direction

$$E_x = -\frac{1}{\mu_0 \sigma} \frac{\partial B_y}{\partial z} = \frac{B_0}{\mu_0 \sigma d} e^{-z/d} \sqrt{2} \cos(\omega t - z/d + \pi/4)$$

⇒ phase shift of 45° ($\pi/4$) between E_x und B_y

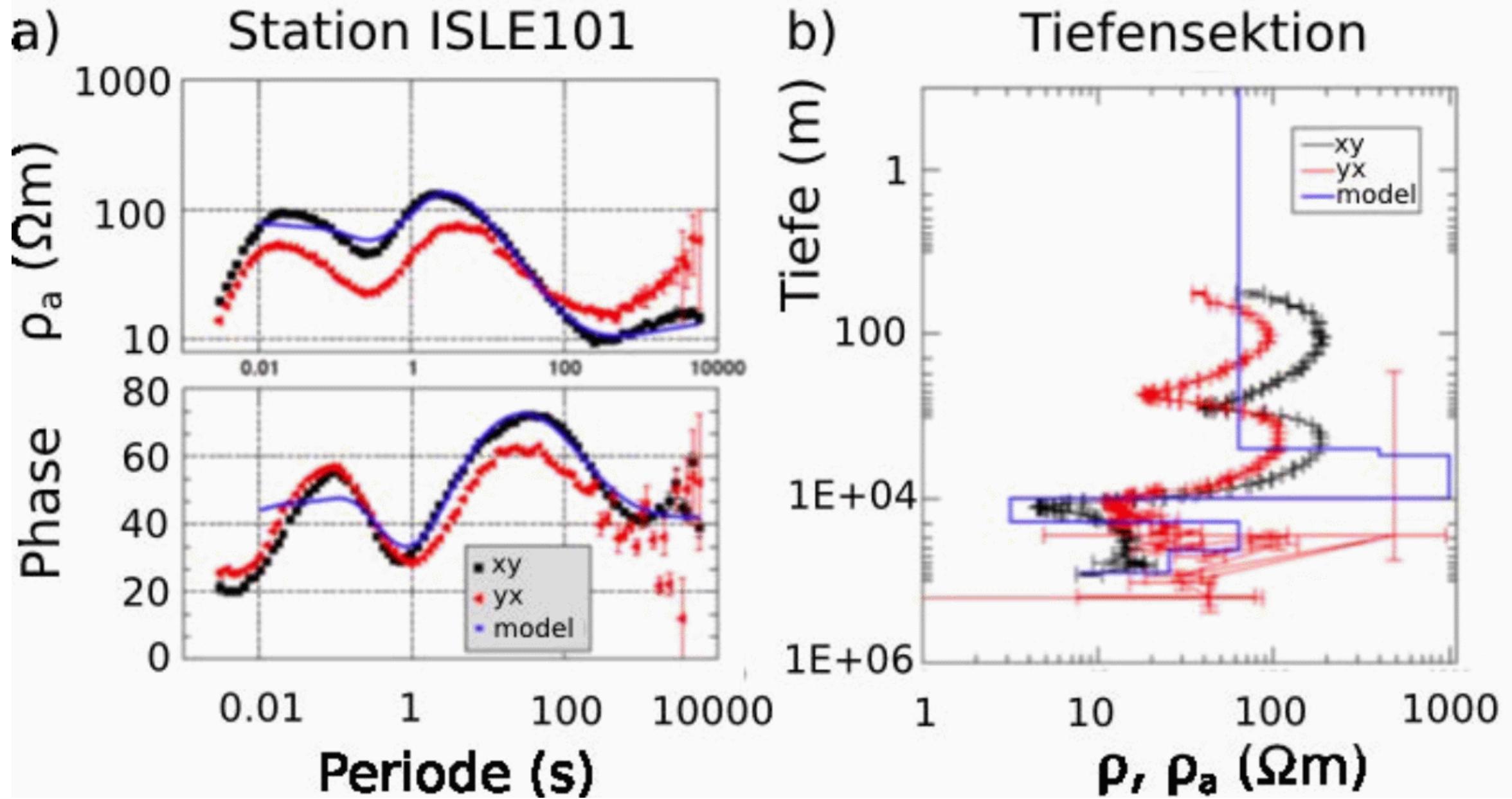
$$\frac{|E_x|}{|B_y|} = \frac{\sqrt{2}}{\mu_0 \sigma d} \Rightarrow \rho = \frac{\mu_0}{\omega} \left| \frac{E_x}{B_y} \right|^2$$

Magnetotellurics (MT)

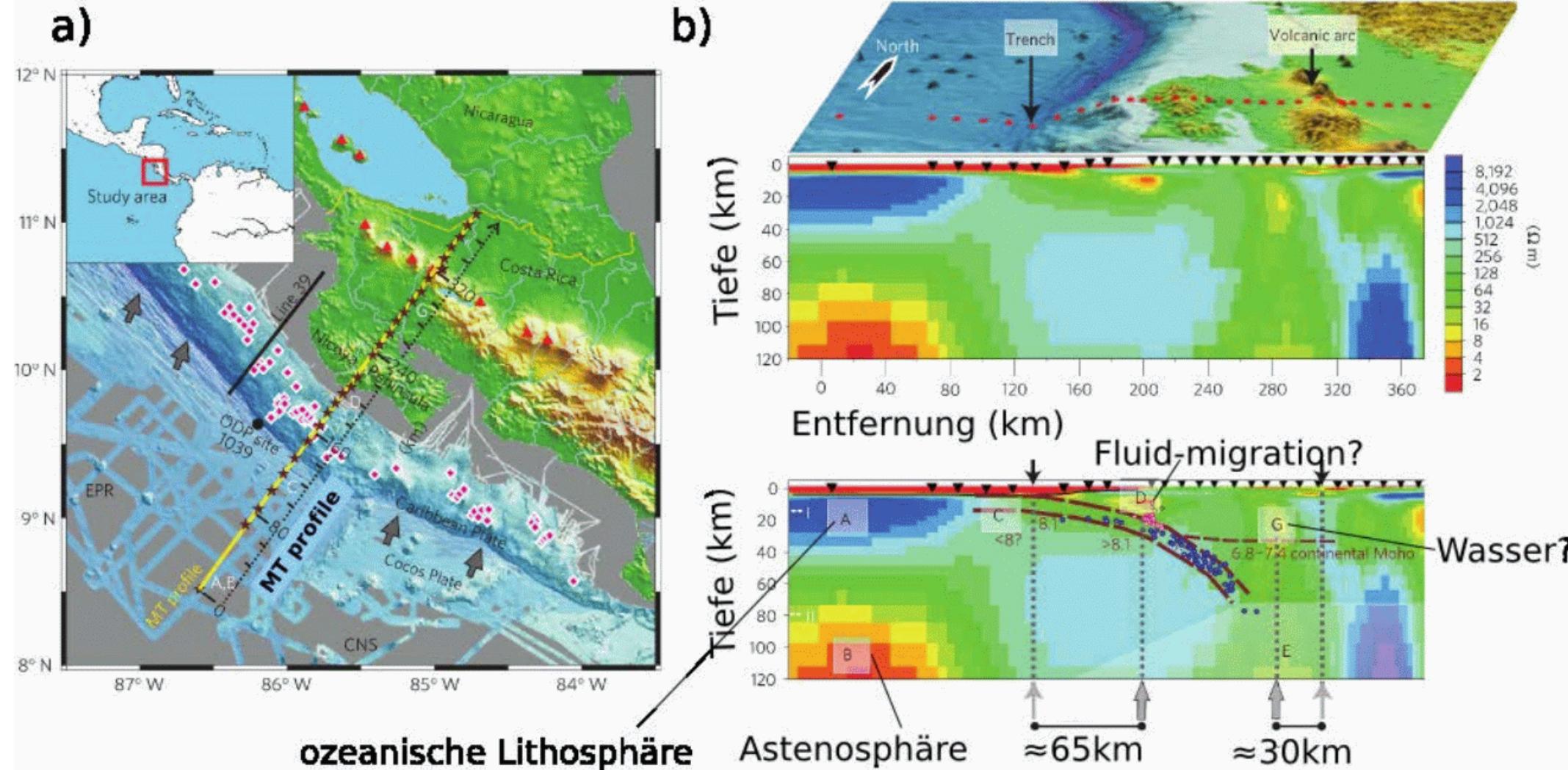


- depth sounding with $T=0.01\text{-}1000\text{s}$ ($f=0.001\text{-}100,\text{Hz}$)
- source in ionosphere (natural source) or on ground (controlled source)
- measure magnetic and electric fields
- analyse (complex) ratio in frequ. domain
- depth sounding (ρ_a & ϕ) with T
- MT course of Anna Marti (U Barcelona)

Magnetotelluric depth sounding



2D/3D Magnetotellurics



Imaging of a subduction zone (Worszewski et al., 2011)

EM with finite elements

$$-\nabla^2 u + \omega\mu\sigma u = f$$

$$-\int_{\Omega} w \nabla^2 u d\Omega + \int_{\Omega} w \omega \mu \sigma u d\Omega = \int_{\Omega} w f d\Omega$$

Gauss's integral law

$$\int_{\Omega} \nabla w \cdot \nabla u d\Omega + \omega \int_{\Omega} \mu \sigma w u d\Omega = \int_{\Omega} w f d\Omega$$

Weak formulation

$u = \sum_i u_i \mathbf{v}_i$ and $w_i \in \{v_i\}$ leads to

$$\int_{\Omega} \nabla v_i \cdot \nabla v_j d\Omega + \omega \mu \int_{\Omega} \sigma v_i v_j d\Omega = \int_{\Omega} v_i f d\Omega$$

$$\langle \nabla v_i | \nabla v_j \rangle + \omega \mu \langle v_i | \sigma v_j \rangle = \langle v_i | f \rangle \quad \text{inner products}$$

representation by matrix-vector product $(\mathbf{A} + \omega \mathbf{M}^\sigma) \mathbf{u} = \mathbf{b}$

with $A_{ij} = \langle \nabla v_i | \nabla v_j \rangle$, $M_{ij}^\sigma = \langle v_i | v_j \rangle$ and $b_i = \langle v_i | f \rangle$

The finite element mass matrix

The mass matrix

$$M_{i,j} = \int_{\Omega} \mu \sigma v_i v_j d\Omega = \sum_c \int_{\Omega_c} \mu_c \sigma_c v_i v_j d\Omega$$

can be written for element-wise conductivity and permittivity

$$M_{i,j} = \sum_c \mu_c \sigma_c \int_{\Omega_c} v_i v_j$$

Complex or real-valued?

The complex-valued system

$$(\mathbf{A} + \imath\omega\mathbf{M})\mathbf{u} = (\mathbf{A} + \imath\omega\mathbf{M})(\mathbf{u}_r + \imath\mathbf{u}_i) = \mathbf{b}_r + \imath\mathbf{b}_i$$

can be transferred into a doubled real-valued system

$$\mathbf{A}\mathbf{u}_r + \imath\mathbf{A}\mathbf{u}_i + \imath\omega\mathbf{M}\mathbf{u}_r - \omega\mathbf{M}\mathbf{u}_i = \mathbf{b}_r + \imath\mathbf{b}_i$$

$$\begin{pmatrix} A & -\omega M \\ \omega M & A \end{pmatrix} \begin{pmatrix} u_r \\ u_i \end{pmatrix} = \begin{pmatrix} b_r \\ b_i \end{pmatrix}$$

Secondary field approach

Consider the field to consist of a primary (background) and an secondary (anomalous) field $F = F_0 + F_a$

solution for F_0 known, e.g. analytically or 1D (semi-analytically)

⇒ form equations for F_a , because

- F_a is weaker or smoother (e.g. $F_0 \propto 1/r$ at sources)
- boundary conditions easier to set (e.g. homogeneous Dirichlet)

Secondary field Helmholtz equation

The equation $-\nabla^2 F - k^2 F = 0$ is solved by the primary field for k_0 :

$-\nabla^2 F_0 - k_0^2 F_0 = 0$ and the total field for $k_0 + \delta k$:

$$-\nabla^2(F_0 + F_a) - (k_0^2 + \delta k^2)(F_0 + F_a) = 0$$

$$-\nabla^2 F_a - k^2 F_a = \delta k^2 F_0$$

 **Note**

Source terms only arise at anomalous terms, weighted by the primary field.

Secondary field for EM

Maxwells equations $k^2 = -\omega\mu\sigma$

$$-\nabla^2 \mathbf{E}_0 + \omega\mu\sigma \mathbf{E}_0 = 0$$

leads to

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$



Note

Source terms only arise at anomalous conductivities and increase with primary field

Secondary field for EM

Maxwells equations $k^2 = -\omega\mu\sigma$

$$-\nabla^2 \mathbf{E}_0 + \omega\mu\sigma \mathbf{E}_0 = 0$$

leads to

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$

 Note

Source terms only arise at anomalous conductivities and increase with primary field

Secondary field for EM

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$

leads to the discretized form (**A**-stiffness, **M**-mass)

$$\mathbf{A}\mathbf{E}_a + \omega\mathbf{M}_\sigma E_a = -\omega\mathbf{M}_{\delta\sigma} \mathbf{E}_0$$

```
1 A = stiffnessMatrix1DFE(x=z)
2 M = massMatrix1DFE(x=z, a=w*mu*sigma)
3 dM = massMatrix1DFE(x=z, a=w*mu*(sigma-sigma0))
4 u = uAna + solve(A+M*w*1j, dM@uAna * w*1j)
```

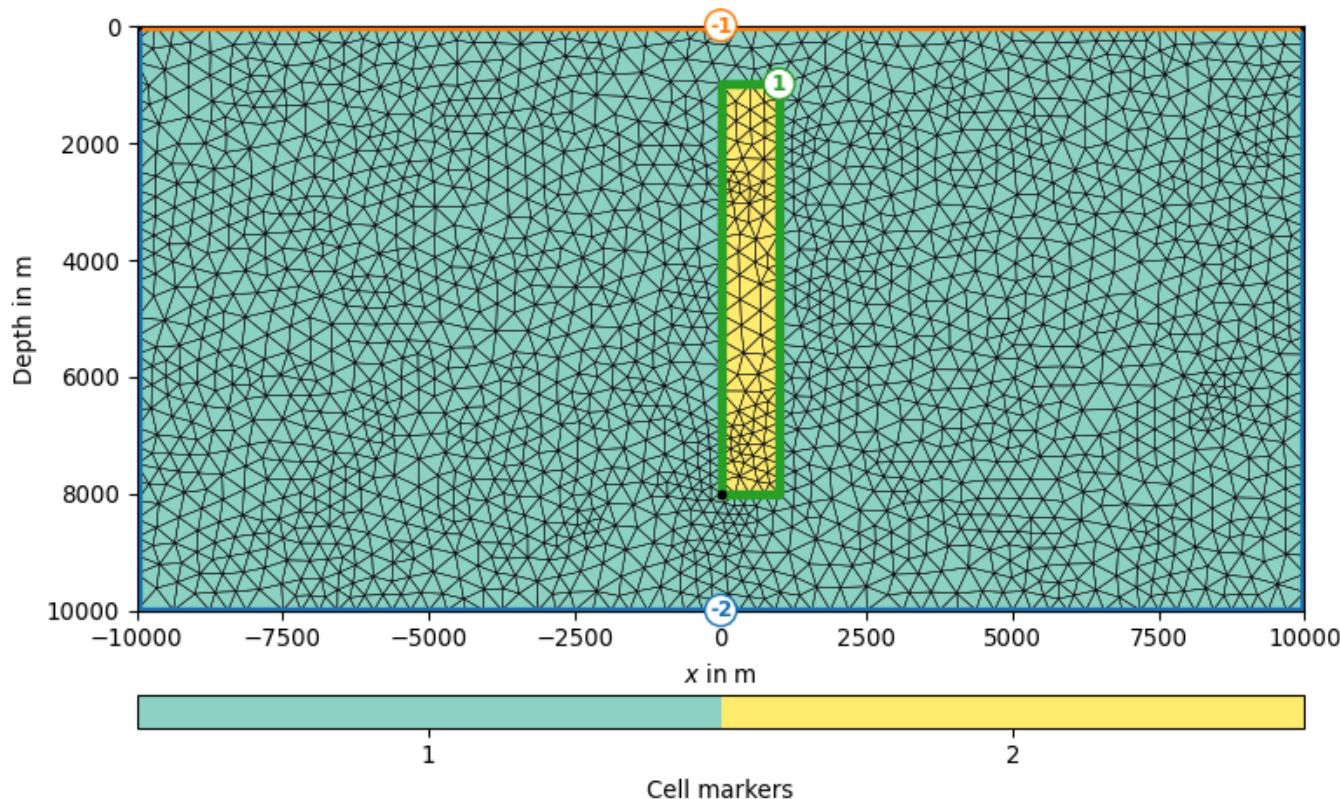
2D/3D problems

Make use of pyGIMLi

See documentation on pyGIMLi.org

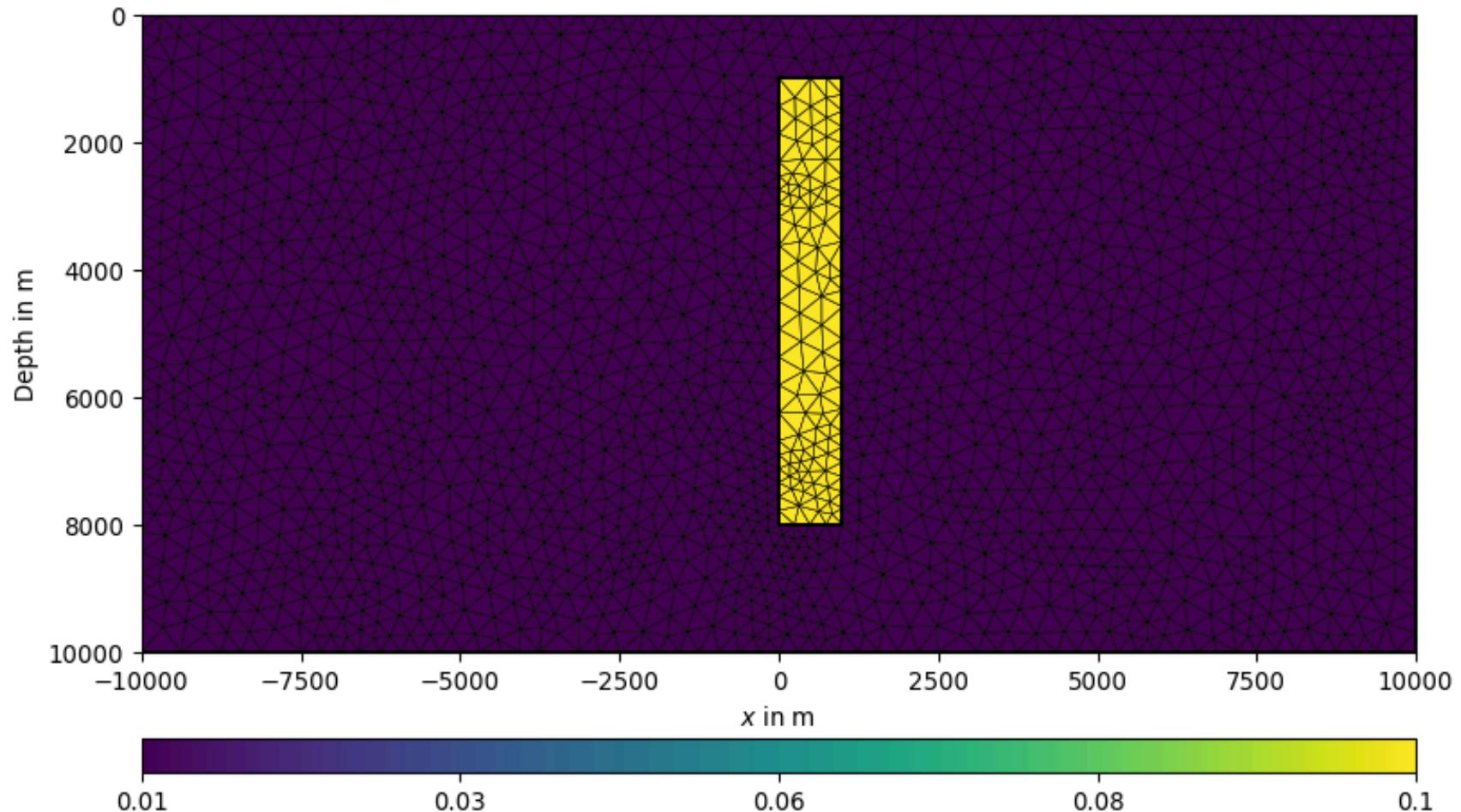
Generating a 2D model

```
1 import pygimli as pg
2 import pygimli.meshTools as mt
3 world = mt.createWorld(start=[-10000, -10000], end=[10000, 0])
4 anomaly = mt.createRectangle(start=[0, -8000], end=[1000, -1000], marker=2)
5 mesh = mt.createMesh(world+anomaly, quality=34, smooth=True, area=1e5)
6 pg.show(mesh, markers=True, showMesh=True);
```



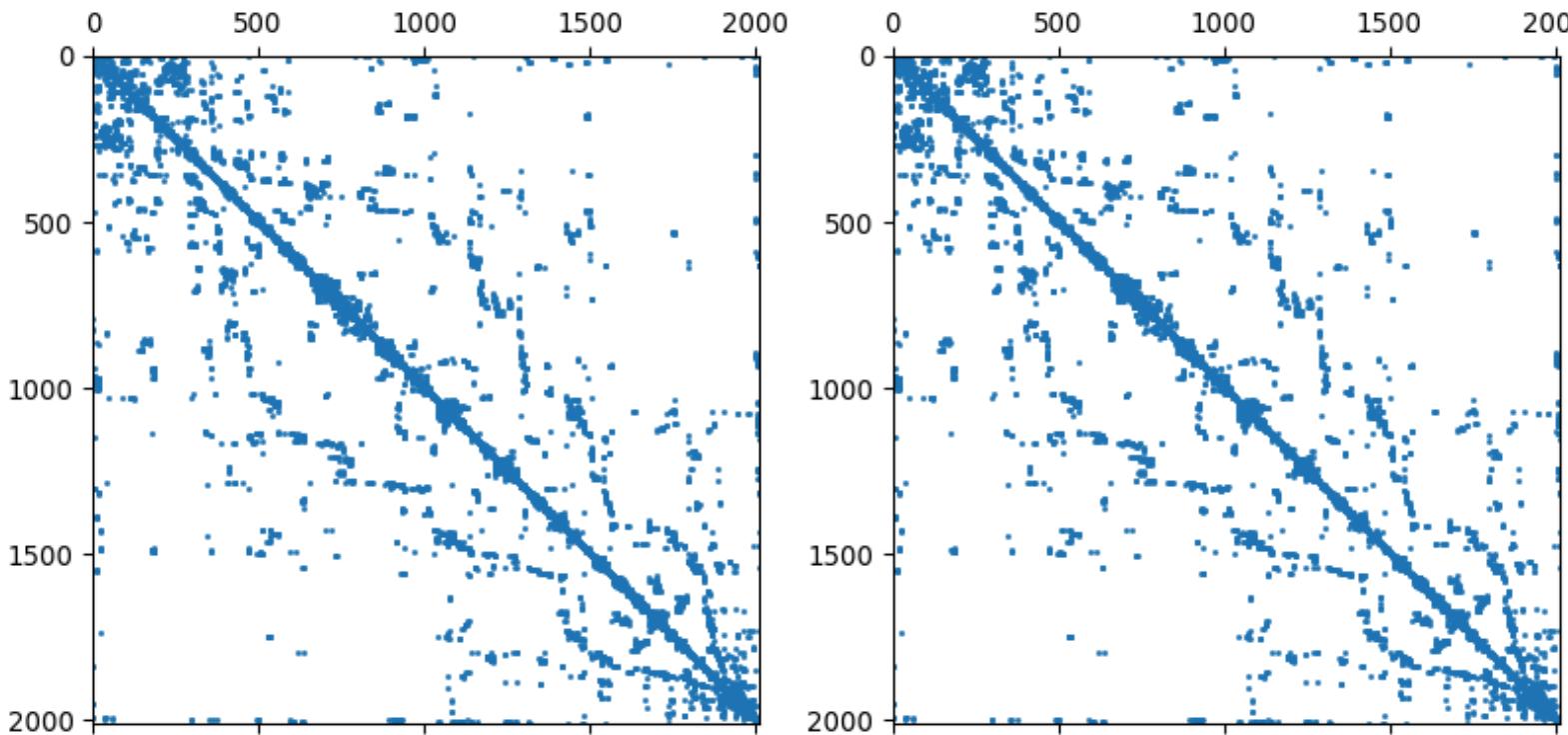
Creating a 2D conductivity model

```
1 sigma0 = 1 / 100 # 100 Ohmm  
2 sigma = mesh.populate("sigma", {1: sigma0, 2: sigma0*10})  
3 pg.show(mesh, "sigma", showMesh=True);
```



The solver module

```
1 import pygimli.solver as ps
2 mesh["my"] = 4 * np.pi * 1e-7
3 A = ps.createStiffnessMatrix(mesh, a=1/mesh["my"])
4 M = ps.createMassMatrix(mesh, mesh["sigma"])
5 fig, ax = plt.subplots(ncols=2)
6 pg.show(A, ax=ax[0], markersize=1)
7 pg.show(M, ax=ax[1], markersize=1)
```



The complex problem matrix

$$\mathbf{B} = \begin{pmatrix} \mathbf{A} & -\omega\mathbf{M} \\ \omega\mathbf{M} & \mathbf{A} \end{pmatrix}$$

```
1 w = 0.1
2 nd = mesh.nodeCount()
3 B = pg.BlockMatrix()
4 B.Aid = B.addMatrix(A)
5 B.Mid = B.addMatrix(M)
6 B.addMatrixEntry(B.Aid, 0, 0)
7 B.addMatrixEntry(B.Aid, nd, nd)
8 B.addMatrixEntry(B.Mid, 0, nd, scale=-w)
9 B.addMatrixEntry(B.Mid, nd, 0, scale=w)
10 pg.show(B)
```

