

# Numerical Simulation Methods in Geophysics, Part 7: 2D Poisson+Helmholtz

## 1. MGPy+MGIN

*thomas.guenther@geophysik.tu-freiberg.de*



**TUBAF**  
Die Ressourcenuniversität.  
Seit 1765.

# Recap

- finite differences approximate partial derivatives
- finite elements approximate solution  $\Rightarrow$  preferred
- spatial discretization determines accuracy of solution
- implicit & mixed time-stepping schemes accurate & stable
- basic elements in 1D/2D/3D & higher order shape functions
- numerical integration , e.g. by Gaussian quadrature

# Inner vs. outer nodes

distinguish dofs into inner and outer  $[\mathbf{u}_i, \mathbf{u}_o]^T$

$$\mathbf{A} = \begin{pmatrix} \mathbf{A}_{ii} & \mathbf{A}_{io} \\ \mathbf{A}_{oi} & \mathbf{A}_{oo} \end{pmatrix}$$

$$\mathbf{A} \begin{bmatrix} \mathbf{u}_i \\ \mathbf{u}_o \end{bmatrix} = \begin{bmatrix} \mathbf{f}_i \\ \mathbf{f}_o \end{bmatrix}$$

$$\Rightarrow \mathbf{A}_{ii}\mathbf{u}_i = \mathbf{f}_i - \mathbf{A}_{oi}\mathbf{u}_o$$

# Green's functions

The Green's function  $G$  is the solution for a Dirac source  $\delta$

$$\mathcal{L}G = \delta(\mathbf{r})$$

The solution can then be obtained by convolution

$$u(\mathbf{r}) = G(\mathbf{r} - \mathbf{r}') * f(\mathbf{r}') =$$

# Secondary field approach

- singular or highly nonlinear source field
  - i.e.  $1/r$  behaviour of potential fields from point sources
- boundary conditions require *far-away* boundaries
- well-known field (*primary field*) for background

# Solution of Poisson's equation in 1D/2D/3D

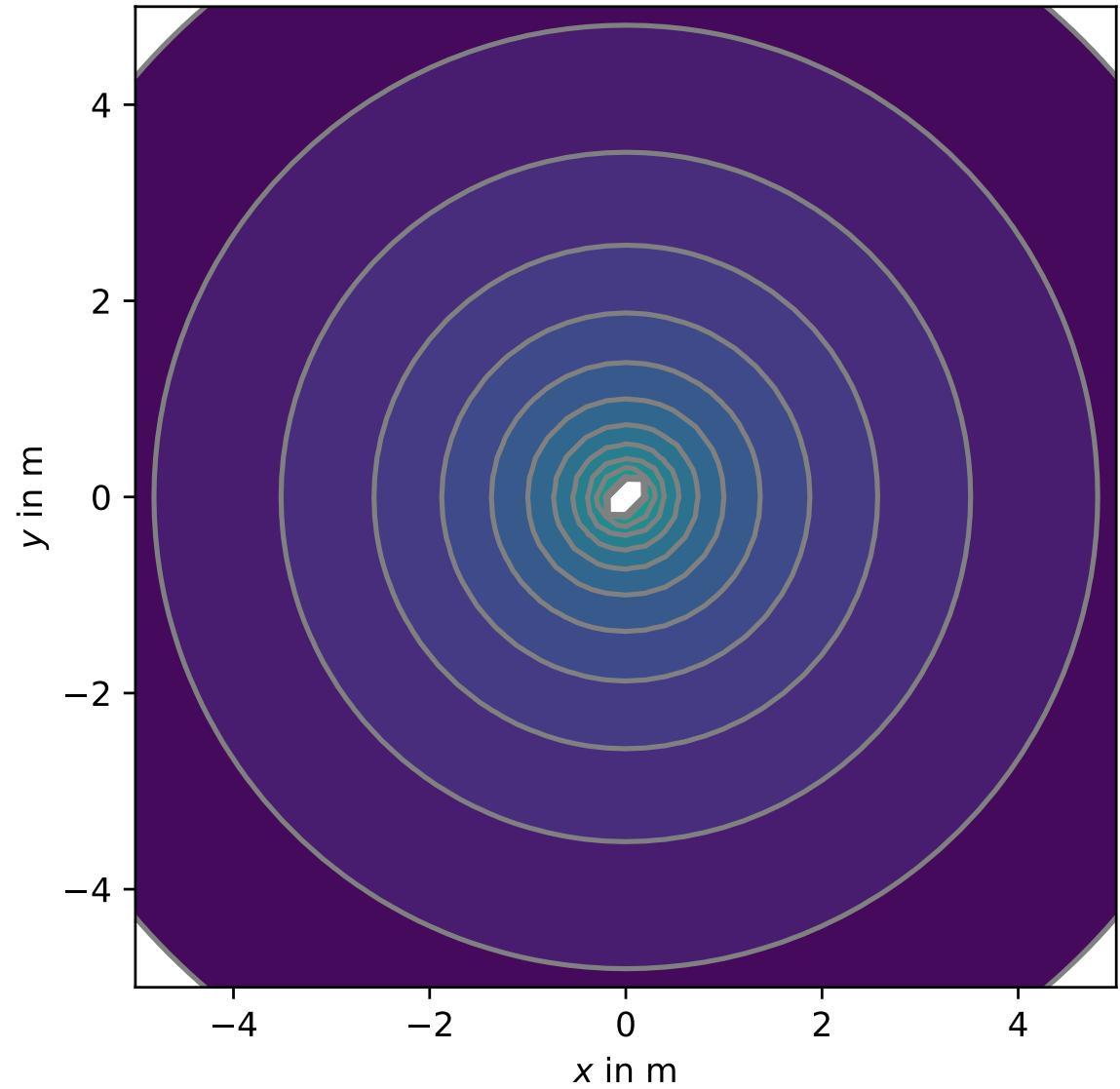
- point source (fundamental solution = Greens function)
- 1D: flow is constant (cannot spread)
- 2D: flow distributes on circle circumference (constant  $a$ )

$$q(r) = \frac{Q}{2\pi r} \quad \Rightarrow \quad u = -\frac{Q\rho}{2\pi} \ln r$$

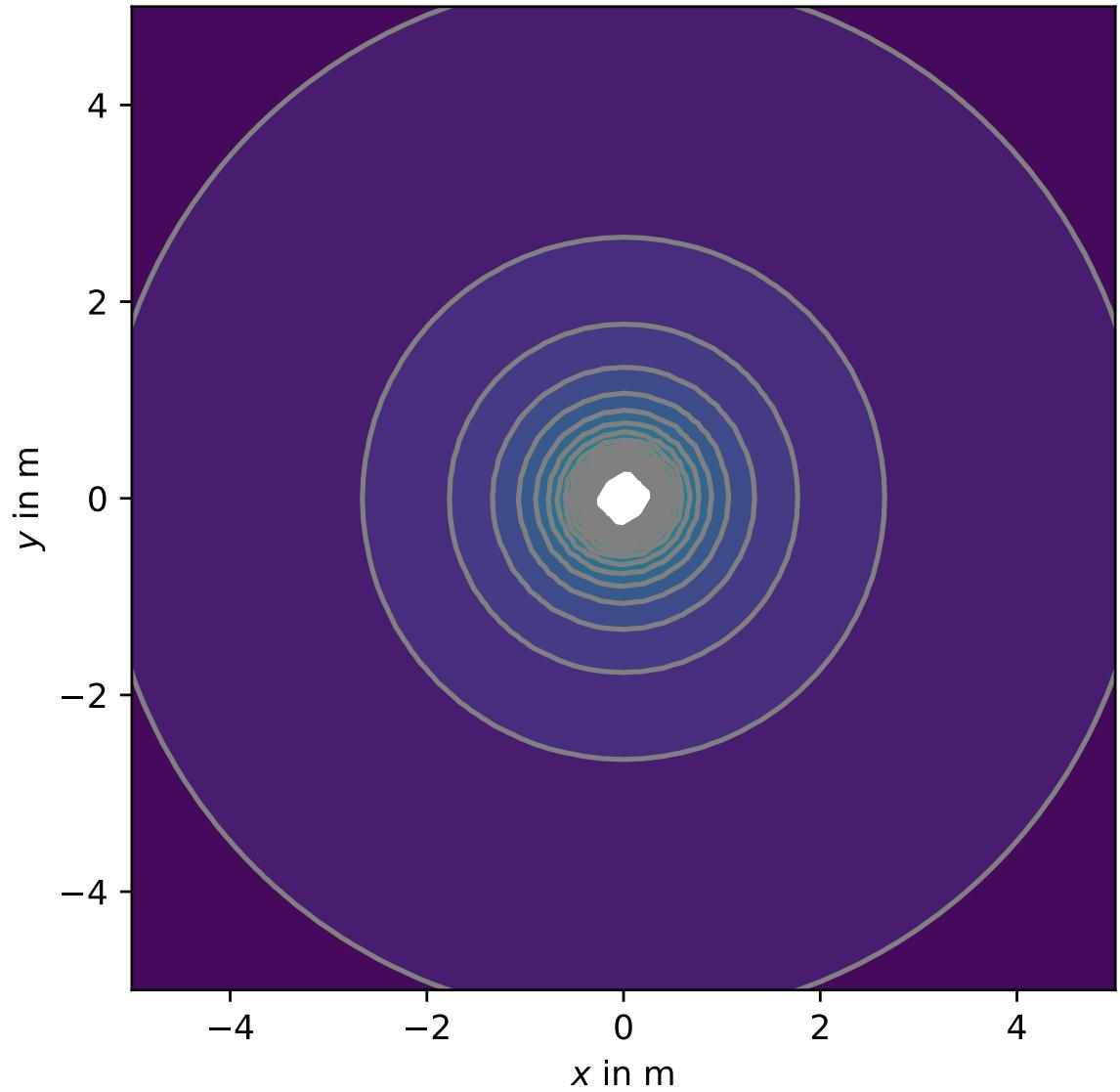
- 3D: flow distributes on sphere surface

$$j(r) = \frac{I}{4\pi r^2} \quad \Rightarrow \quad u = \frac{I\rho}{4\pi r}$$

# Spatial 2D and 3D solutions

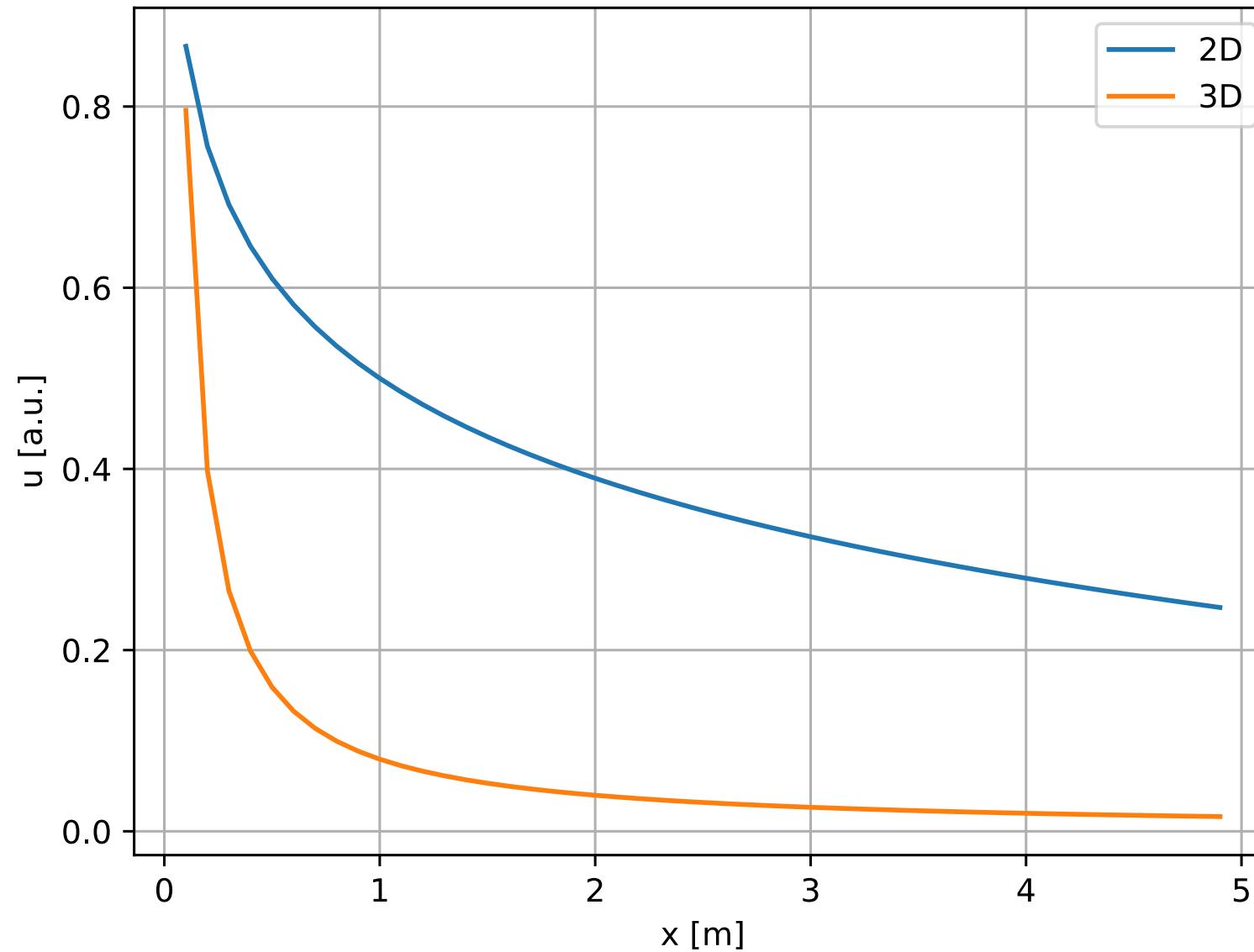


2D solution ( $-\ln r$ )



3D solution ( $1/r$ )

# Comparison of 2D and 3D solutions



# Direct current resistivity

$$\nabla \cdot \sigma \nabla u = -\nabla \cdot \mathbf{j}_s$$

See notebook for problems

# Secondary field approach

$$-\nabla \cdot \sigma \nabla u = \nabla \cdot \mathbf{j}_s$$

$u$  for  $\sigma=\text{const}$ :  $u_0 = u(\sigma_0) = \frac{I}{2\pi\sigma_0}$  fulfills eq.

Idea: split  $\sigma$  solution in  $u = u_0 + u_a$

$$-\nabla \cdot \sigma \nabla u_0 - \nabla \cdot \sigma \nabla u_a = \nabla \cdot \mathbf{j}_s = -\nabla \cdot \sigma_0 \nabla u_0$$

$$\Rightarrow -\nabla \cdot \sigma \nabla u_a = \nabla \cdot (\sigma - \sigma_0) \nabla u_0$$

Anomalies secondary sources:  $\mathbf{A}^\sigma \mathbf{u}_a = -\mathbf{A}^{\delta\sigma} \mathbf{u}_0 = (\mathbf{A}^{\sigma_0} - \mathbf{A}^\sigma) \mathbf{u}_0$

# Helmholtz equation in 2D

- move to another type of PDE
- move from 1D to 2D (and eventually 3D)
- scalar to vectorial solution
- complex-valued system
- secondary field approach

# Maxwells equations

- Faraday's law: currents & varying electric fields  $\Rightarrow$  magnetic field

$$\nabla \times \mathbf{H} = \frac{\partial \mathbf{D}}{\partial t} + \mathbf{j}$$

- Ampere's law: time-varying magnetic fields induce electric field

$$\nabla \times \mathbf{E} = - \frac{\partial \mathbf{B}}{\partial t}$$

- $\nabla \cdot \mathbf{D} = \varrho$ ,  $\nabla \cdot \mathbf{B} = 0$ ,  $\mathbf{D} = \epsilon \mathbf{E}$  &  $\mathbf{B} = \mu \mathbf{H}$

# Maxwell in frequency domain

$$\mathbf{E} = \mathbf{E}_0 e^{j\omega t} \quad \text{or} \quad \mathbf{H} = \mathbf{H}_0 e^{j\omega t}$$

$$\nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \sigma\mathbf{E} = (\sigma + j\omega\epsilon)\mathbf{E}$$

$$\nabla \times \mathbf{E} = -j\omega\mu\mathbf{H}$$

take curl of one of the equations and insert in the other

# Helmholtz equation

see also [Theory EM](#)

take curl of one of the equations and insert in the other

$$\nabla \times \mu^{-1} \nabla \times \mathbf{E} + j\omega\sigma\mathbf{E} - \omega^2\epsilon\mathbf{E} = \nabla \times \mathbf{j}_s$$

$$\nabla \times \sigma^{-1} \nabla \times \mathbf{H} + j\omega\mu\mathbf{H} - \omega^2\mu\epsilon\mathbf{H} = 0$$

# Quasi-static approximation

Assume:  $\omega^2 \mu \epsilon < \omega \mu \sigma$ , no sources ( $\nabla \cdot \mathbf{j}_s = 0$ ), + vector identity

$$\nabla \times \nabla \times \mathbf{F} = \nabla \nabla \cdot \mathbf{F} - \nabla^2 \mathbf{F}$$

leads with  $\nabla \cdot \mathbf{E} = 0 = \nabla \cdot \mathbf{B}$  to the vector Helmholtz PDE

$$-\nabla^2 \mathbf{E} + j\omega \mu \sigma \mathbf{E} = 0$$

$$-\nabla \cdot \sigma^{-1} \nabla \mathbf{H} + j\omega \mu \mathbf{H} = 0$$

# Variational form

$$-\nabla^2 u + \gamma\omega\mu\sigma u = f$$

$$-\int_{\Omega} w \nabla^2 u d\Omega + \int_{\Omega} w \gamma\omega\mu\sigma u d\Omega = \int_{\Omega} w f d\Omega$$

Gauss's integral law

$$\int_{\Omega} \nabla w \cdot \nabla u d\Omega + \omega \int_{\Omega} \mu\sigma w u d\Omega = \int_{\Omega} w f d\Omega$$

# Weak formulation

$u = \sum_i u_i \mathbf{v}_i$  and  $w_i \in \{v_i\}$  leads to

$$\int_{\Omega} \nabla v_i \cdot \nabla v_j d\Omega + j\omega\mu \int_{\Omega} \sigma v_i v_j d\Omega = \int_{\Omega} v_i f d\Omega$$

$$\langle \nabla v_i | \nabla v_j \rangle + j\omega\mu \langle v_i | \sigma v_j \rangle = \langle v_i | f \rangle \quad \text{inner products}$$

representation by matrix-vector product  $(\mathbf{A} + j\omega\mathbf{M}^\sigma)\mathbf{u} = \mathbf{b}$

with  $A_{ij} = \langle \nabla v_i | \nabla v_j \rangle$ ,  $M_{ij}^\sigma = \langle v_i | v_j \rangle$  and  $b_i = \langle v_i | f \rangle$

# The finite element mass matrix

The mass matrix

$$M_{i,j} = \int_{\Omega} \mu \sigma v_i v_j d\Omega = \sum_c \int_{\Omega_c} \mu_c \sigma_c v_i v_j d\Omega$$

can be written for element-wise conductivity and permittivity

$$M_{i,j} = \sum_c \mu_c \sigma_c \int_{\Omega_c} v_i v_j$$

# Complex or real-valued?

The complex valued system

$$(\mathbf{A} + \imath\omega\mathbf{M})\mathbf{u} = (\mathbf{A} + \imath\omega\mathbf{M})(\mathbf{u}_r + \imath\mathbf{u}_i) = \mathbf{b}_r + \imath\mathbf{b}_i$$

can be transferred into a doubled real-valued system

$$\mathbf{A}\mathbf{u}_r + \imath\mathbf{A}\mathbf{u}_i + \imath\omega\mathbf{M}\mathbf{u}_r - \omega\mathbf{M}\mathbf{u}_i = \mathbf{b}_r + \imath\mathbf{b}_i$$

$$\begin{pmatrix} A & -\omega M \\ \omega M & A \end{pmatrix} \begin{pmatrix} u_r \\ u_i \end{pmatrix} = \begin{pmatrix} b_r \\ b_i \end{pmatrix}$$

# Secondary field approach

Consider the field to consist of a primary (background) and an secondary (anomalous) field  $F = F_0 + F_a$

solution for  $F_0$  known, e.g. analytically or 1D (semi-analytically)

⇒ form equations for  $F_a$ , because

- $F_a$  is weaker or smoother (e.g.  $F_0 \propto 1/r$  at sources)
- boundary conditions easier to set (e.g. homogeneous Dirichlet)

# Secondary field Helmholtz equation

The equation  $-\nabla^2 F - k^2 F = 0$  is solved by the primary field for  $k_0$ :

$-\nabla^2 F_0 - k_0^2 F_0 = 0$  and the total field for  $k_0 + \delta k$ :

$$-\nabla^2(F_0 + F_a) - (k_0^2 + \delta k^2)(F_0 + F_a) = 0$$

$$-\nabla^2 F_a - k^2 F_a = \delta k^2 F_0$$

 **Note**

Source terms only arise at anomalous terms, weighted by the primary field.

# Secondary field for EM

Maxwells equations  $k^2 = -\omega\mu\sigma$

$$-\nabla^2 \mathbf{E}_0 + \omega\mu\sigma \mathbf{E}_0 = 0$$

leads to

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$

 Note

Source terms only arise at anomalous conductivities and increase with primary field

# Secondary field for EM

Maxwells equations  $k^2 = -\omega\mu\sigma$

$$-\nabla^2 \mathbf{E}_0 + \omega\mu\sigma \mathbf{E}_0 = 0$$

leads to

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$

 Note

Source terms only arise at anomalous conductivities and increase with primary field

# Secondary field for EM

$$-\nabla^2 \mathbf{E}_a + \omega\mu\sigma \mathbf{E}_a = -\omega\mu\delta\sigma \mathbf{E}_0$$

leads to the discretized form (**A**-stiffness, **M**-mass)

$$\mathbf{A}\mathbf{E}_a + \omega\mathbf{M}_\sigma E_a = -\omega\mathbf{M}_{\delta\sigma} \mathbf{E}_0$$

```
1 A = stiffnessMatrix1DFE(x=z)
2 M = massMatrix1DFE(x=z, a=w*mu*sigma)
3 dM = massMatrix1DFE(x=z, a=w*mu*(sigma-sigma0))
4 u = uAna + solve(A+M*w*1j, dM@uAna * w*1j)
```

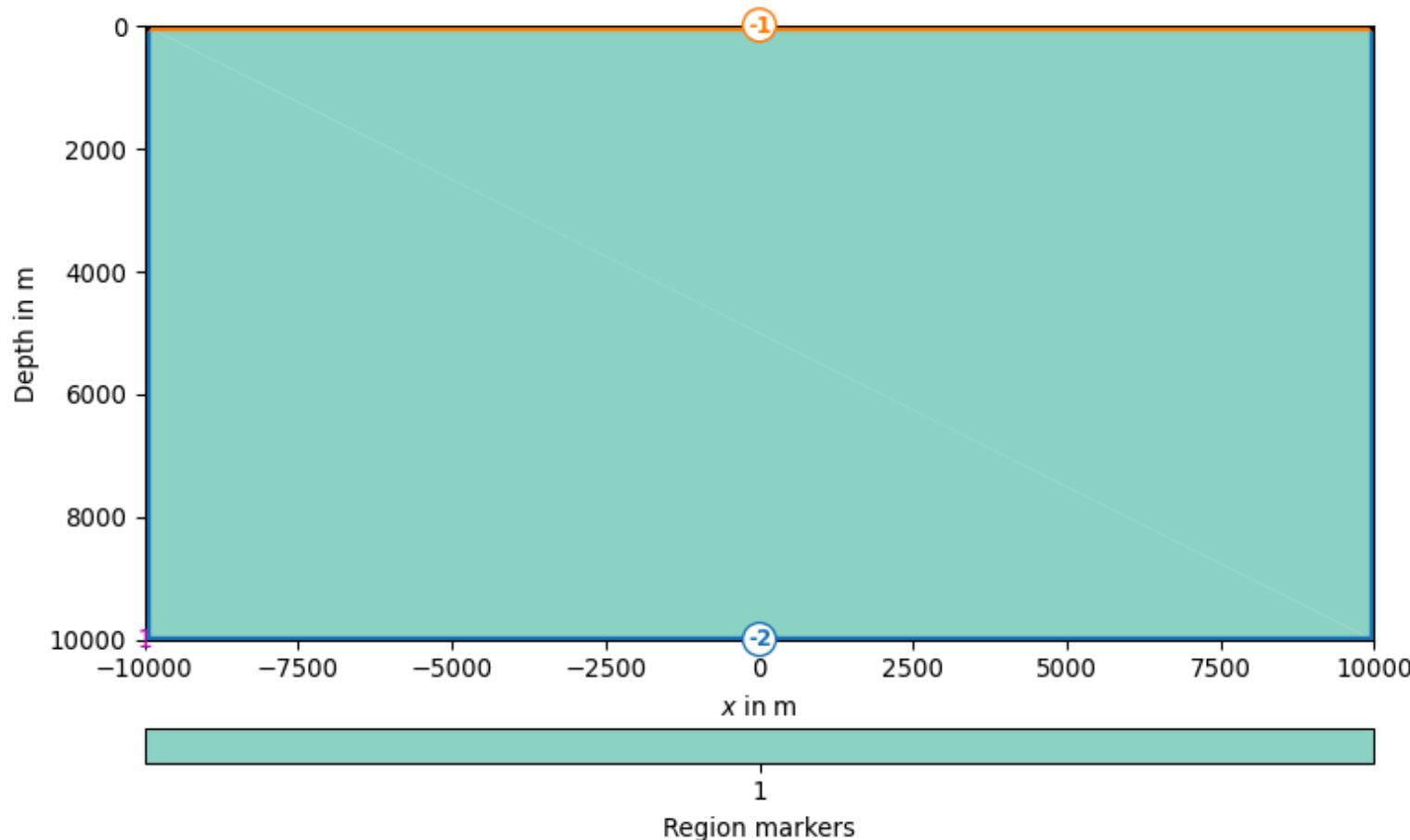
# 2D/3D problems

Make use of pyGIMLi

See documentation on [pyGIMLi.org](http://pyGIMLi.org)

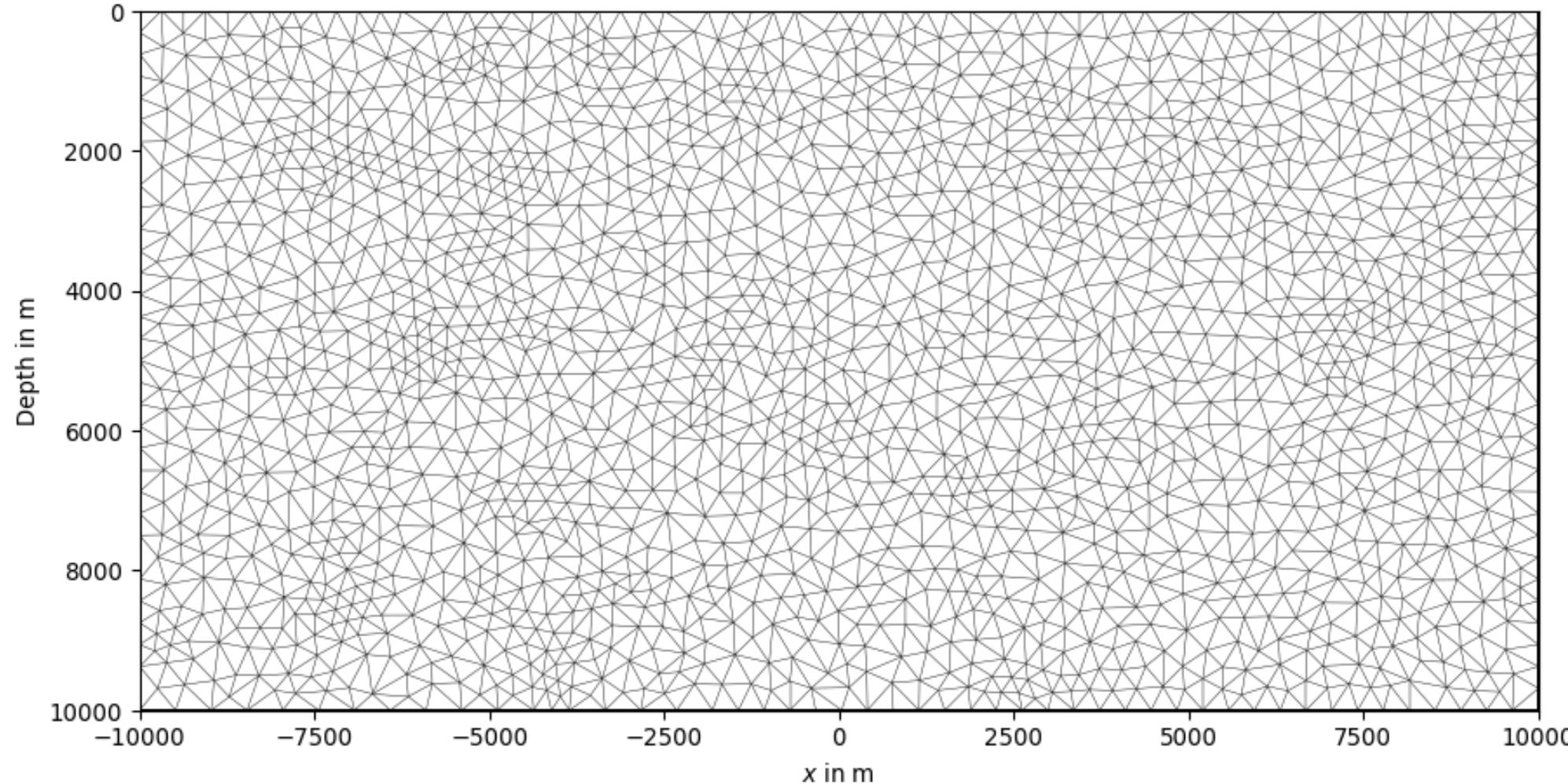
# The meshtools module

```
1 import pygimli as pg
2 import pygimli.meshtools as mt
3 world = mt.createWorld(start=[-10000, -10000], end=[10000, 0])
4 pg.show(world, boundaryMarkers=True)
```



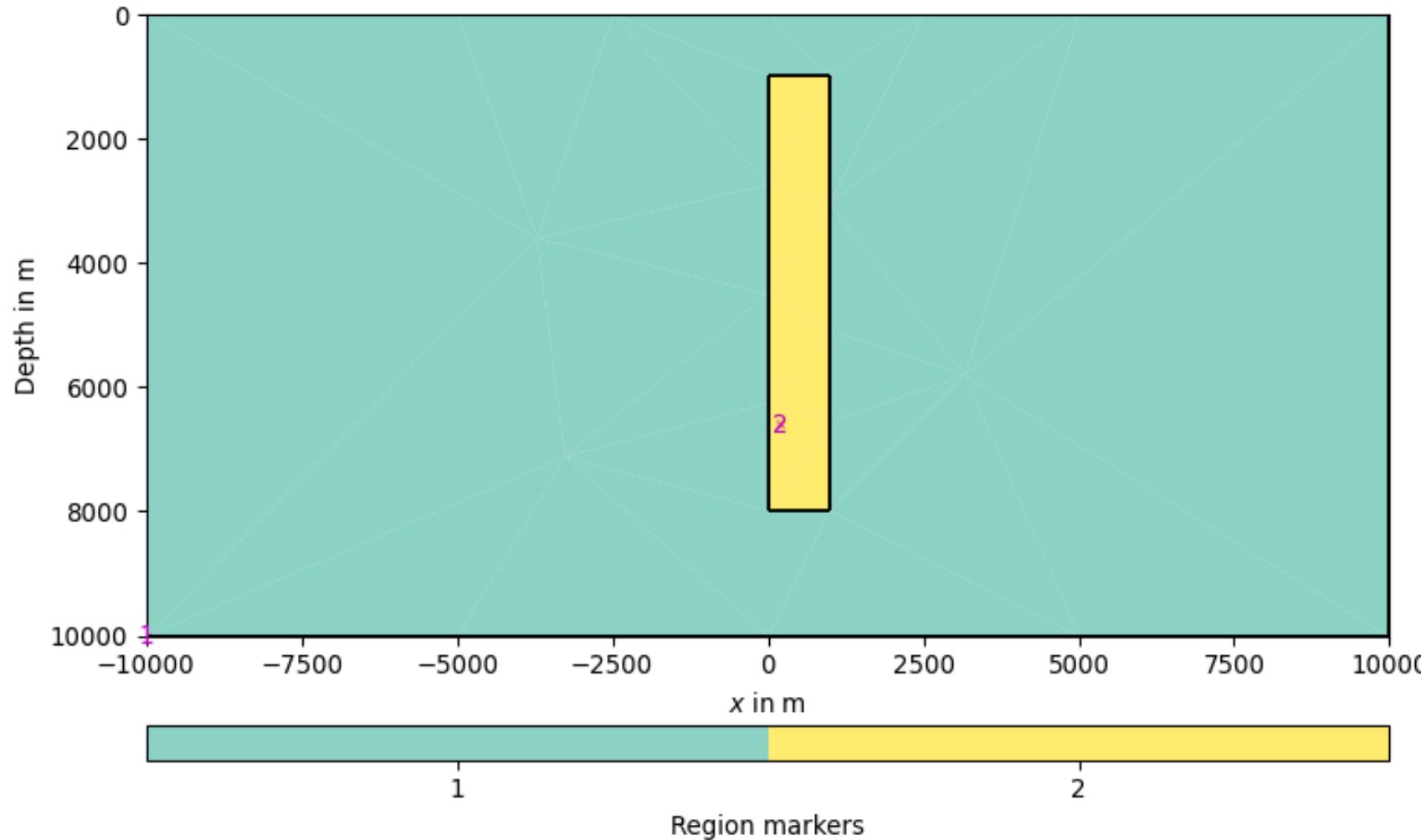
# The meshtools module

```
1 mesh = mt.createMesh(world, quality=34, area=1e5)
2 pg.show(mesh)
```



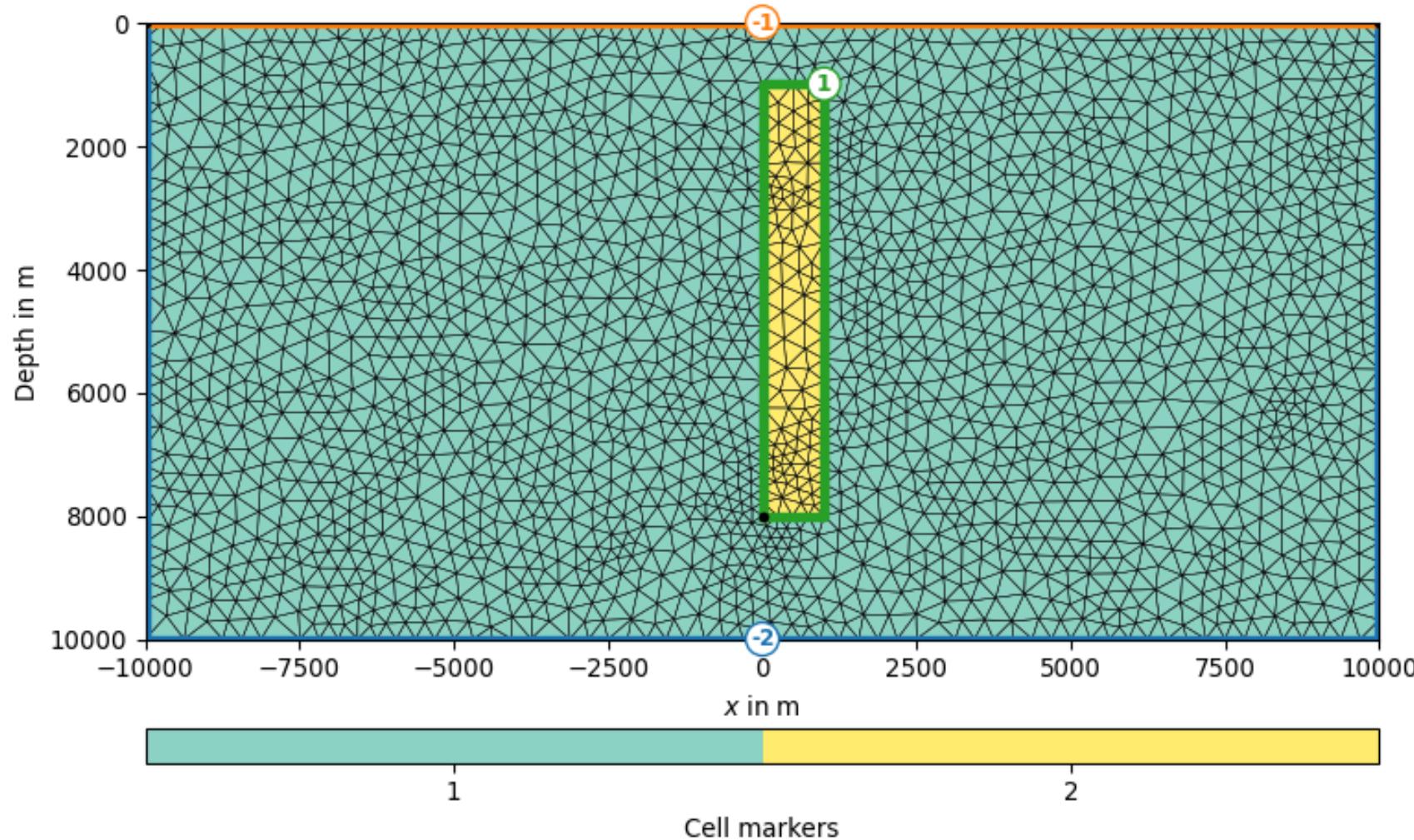
# Creating a 2D geometry

```
1 anomaly = mt.createRectangle(start=[0, -8000], end=[1000, -1000], marker=2)  
2 pg.show(world+anomaly)
```



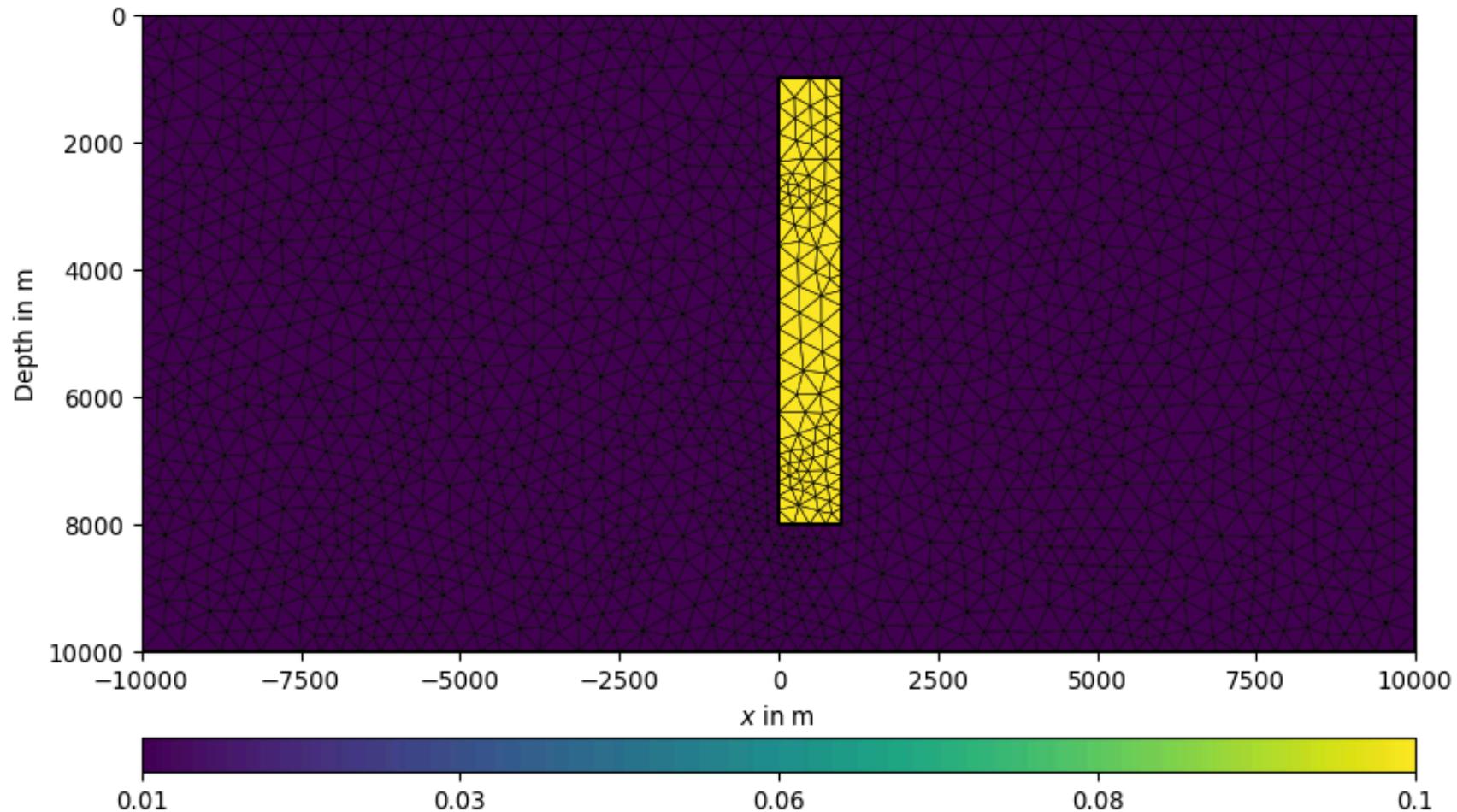
# Creating a 2D mesh

```
1 mesh = mt.createMesh(world+anomaly, quality=34, smooth=True, area=1e5)  
2 pg.show(mesh, markers=True, showMesh=True);
```



# Creating a 2D conductivity model

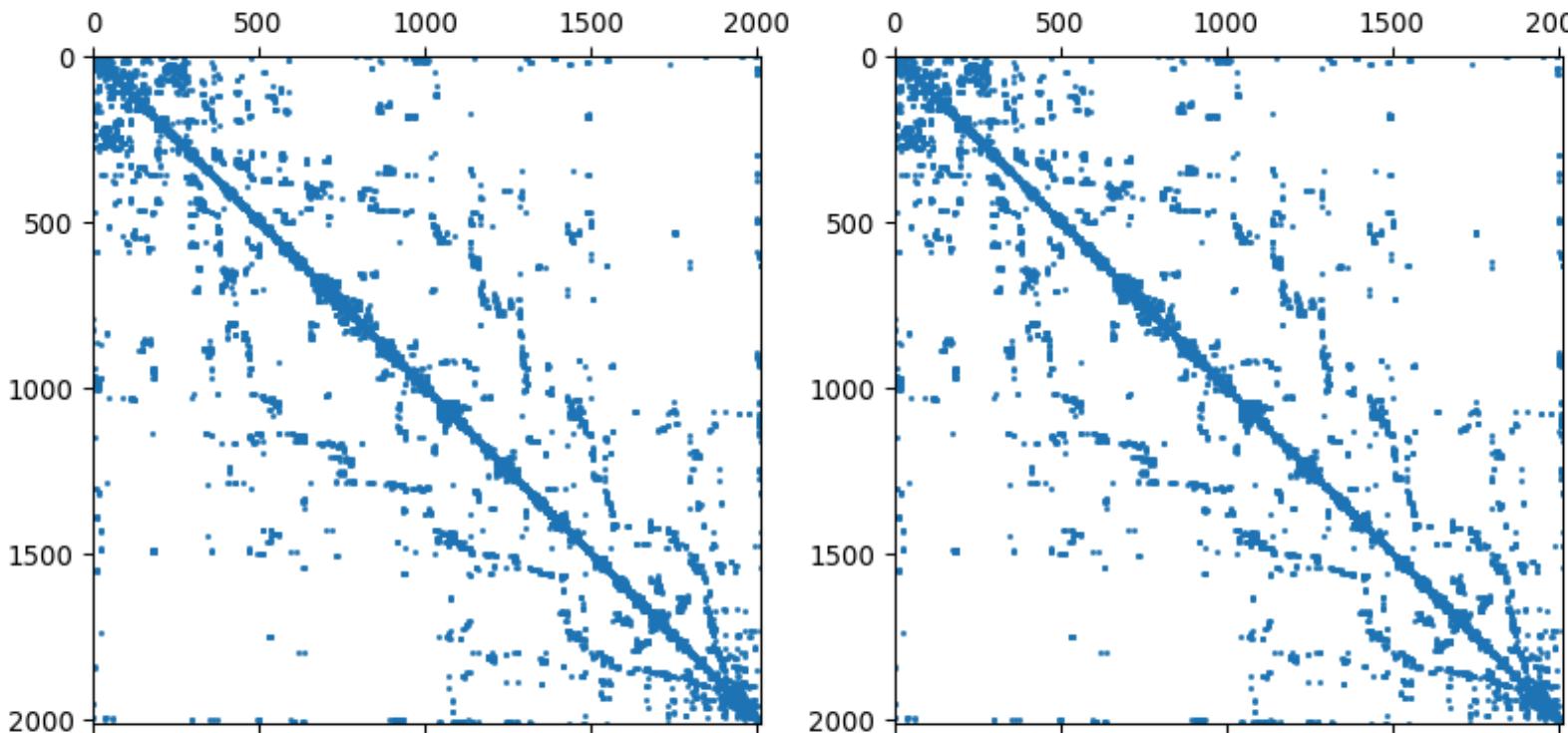
```
1 sigma0 = 1 / 100 # 100 Ohmm  
2 sigma = mesh.populate("sigma", {1: sigma0, 2: sigma0*10})  
3 pg.show(mesh, "sigma", showMesh=True);
```



# The solver module

# The solver module

```
1 import pygimli.solver as ps
2 mesh["my"] = 4 * np.pi * 1e-7
3 A = ps.createStiffnessMatrix(mesh, a=1/mesh["my"])
4 M = ps.createMassMatrix(mesh, mesh["sigma"])
5 fig, ax = plt.subplots(ncols=2)
6 ax[0].spy(pg.utils.toCSR(A), markersize=1)
7 ax[1].spy(pg.utils.toCSR(M).todense(), markersize=1)
```



# The complex problem matrix

$$\mathbf{B} = \begin{pmatrix} \mathbf{A} & -\omega\mathbf{M} \\ \omega\mathbf{M} & \mathbf{A} \end{pmatrix}$$

```
1 w = 0.1
2 nd = mesh.nodeCount()
3 B = pg.BlockMatrix()
4 B.Aid = B.addMatrix(A)
5 B.Mid = B.addMatrix(M)
6 B.addMatrixEntry(B.Aid, 0, 0)
7 B.addMatrixEntry(B.Aid, nd, nd)
8 B.addMatrixEntry(B.Mid, 0, nd, scale=-w)
9 B.addMatrixEntry(B.Mid, nd, 0, scale=w)
10 pg.show(B)
```

