Applications

- High Frequency Synchronous Buck Converters for Computer Processor Power
- High Frequency Isolated DC-DC Converters with Synchronous Rectification for Telecom and Industrial Use
- Lead-Free

Benefits

- Very Low R_{DS(on)} at 4.5V V_{GS}
- Ultra-Low Gate Impedance
- Fully Characterized Avalanche Voltage and Current

IRFR3709ZPbFIRFU3709ZPbF

HEXFET® Power MOSFET

V _{DSS}	R _{DS(on)} max	Qg
30V	6.5m Ω	17nC



Absolute Maximum Ratings

	Parameter	Max.	Units	
V _{DS}	Drain-to-Source Voltage	30	V	
V _{GS}	Gate-to-Source Voltage	± 20		
I _D @ T _C = 25°C	Continuous Drain Current, V _{GS} @ 10V	86⊕	А	
I _D @ T _C = 100°C	Continuous Drain Current, V _{GS} @ 10V	61⊕		
I _{DM}	Pulsed Drain Current ①	340		
P _D @T _C = 25°C	Maximum Power Dissipation	79	W	
P _D @T _C = 100°C	Maximum Power Dissipation	39		
	Linear Derating Factor	0.53	W/°C	
TJ	Operating Junction and	-55 to + 175	°C	
T _{STG}	Storage Temperature Range			
_	Soldering Temperature, for 10 seconds	300 (1.6mm from case)		

Thermal Resistance

	Parameter	Тур.	Max.	Units
$R_{\theta JC}$	Junction-to-Case		1.9	°C/W
$R_{\theta JA}$	Junction-to-Ambient (PCB Mount) ^⑤		50	
$R_{\theta JA}$	Junction-to-Ambient		110	

Notes ① through ⑤ are on page 11

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Static @ T_J = 25°C (unless otherwise specified)

	Parameter	Min.	Тур.	Max.	Units	Conditions
BV _{DSS}	Drain-to-Source Breakdown Voltage	30			V	$V_{GS} = 0V, I_D = 250\mu A$
$\Delta BV_{DSS}/\Delta T_{J}$	Breakdown Voltage Temp. Coefficient		22		mV/°C	Reference to 25°C, I _D = 1mA
R _{DS(on)}	Static Drain-to-Source On-Resistance		5.2	6.5	mΩ	V _{GS} = 10V, I _D = 15A ③
			6.5	8.2	Ī	V _{GS} = 4.5V, I _D = 12A ③
V _{GS(th)}	Gate Threshold Voltage	1.35	1.80	2.25	V	$V_{DS} = V_{GS}$, $I_D = 250\mu A$
$\Delta V_{GS(th)}/\Delta T_{J}$	Gate Threshold Voltage Coefficient		-5.6		mV/°C	
I _{DSS}	Drain-to-Source Leakage Current			1.0	μA	$V_{DS} = 24V$, $V_{GS} = 0V$
				150	Ì	$V_{DS} = 24V, V_{GS} = 0V, T_{J} = 150^{\circ}C$
I _{GSS}	Gate-to-Source Forward Leakage			100	nA	V _{GS} = 20V
	Gate-to-Source Reverse Leakage			-100	Ī	V _{GS} = -20V
gfs	Forward Transconductance	51			S	$V_{DS} = 15V, I_D = 12A$
$\overline{Q_g}$	Total Gate Charge		17	26		
Q _{gs1}	Pre-Vth Gate-to-Source Charge		4.7		Ī	$V_{DS} = 15V$
Q _{gs2}	Post-Vth Gate-to-Source Charge		1.6		nC	$V_{GS} = 4.5V$
Q_{gd}	Gate-to-Drain Charge		5.7		Ī	I _D = 12A
Q_{godr}	Gate Charge Overdrive		5.0		Ī	See Fig. 16
Q _{sw}	Switch Charge (Q _{gs2} + Q _{gd})		7.3		Ī	
Q _{oss}	Output Charge		10		nC	V _{DS} = 16V, V _{GS} = 0V
t _{d(on)}	Turn-On Delay Time		12			V _{DD} = 16V, V _{GS} = 4.5V ③
t _r	Rise Time		12		Ī	I _D = 12A
t _{d(off)}	Turn-Off Delay Time		15		ns	Clamped Inductive Load
t _f	Fall Time		3.9			
C _{iss}	Input Capacitance		2330			V _{GS} = 0V
C _{oss}	Output Capacitance		460		рF	V _{DS} = 15V
C _{rss}	Reverse Transfer Capacitance		230		1	f = 1.0MHz

Avalanche Characteristics

	Parameter	Тур.	Max.	Units
E _{AS}	Single Pulse Avalanche Energy®		100	mJ
I _{AR}	Avalanche Current ①		12	Α
E _{AR}	Repetitive Avalanche Energy ①		7.9	mJ

Diode Characteristics

	Parameter	Min.	Тур.	Max.	Units	Conditions
Is	Continuous Source Current			86@		MOSFET symbol
	(Body Diode)				Α	showing the
I _{SM}	Pulsed Source Current			340		integral reverse
	(Body Diode) ①					p-n junction diode.
V_{SD}	Diode Forward Voltage			1.0	V	$T_J = 25$ °C, $I_S = 12A$, $V_{GS} = 0V$ ③
t _{rr}	Reverse Recovery Time		29	44	ns	$T_J = 25$ °C, $I_F = 12A$, $V_{DD} = 15V$
Q _{rr}	Reverse Recovery Charge		25	37	nC	di/dt = 100A/µs ③
t _{on}	Forward Turn-On Time	Intrinsio	Intrinsic turn-on time is negligible (turn-on is dominated by LS+LD)			

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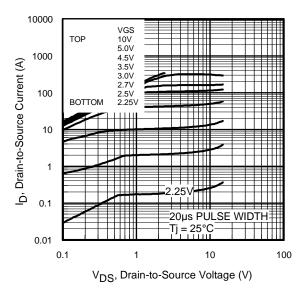


Fig 1. Typical Output Characteristics

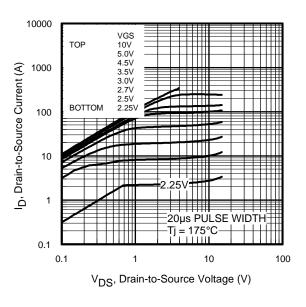


Fig 2. Typical Output Characteristics

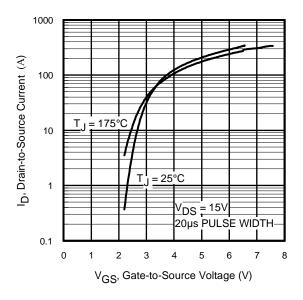


Fig 3. Typical Transfer Characteristics

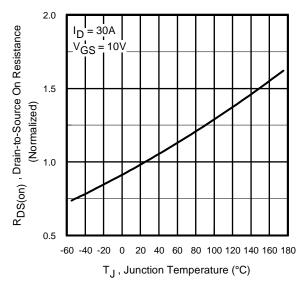


Fig 4. Normalized On-Resistance vs. Temperature

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TOR Rectifier

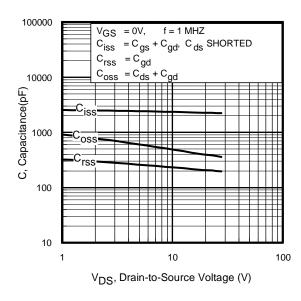


Fig 5. Typical Capacitance vs. Drain-to-Source Voltage

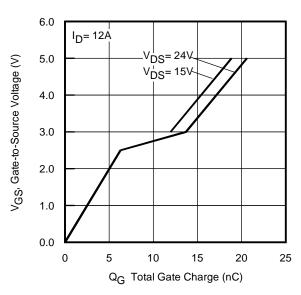


Fig 6. Typical Gate Charge vs. Gate-to-Source Voltage

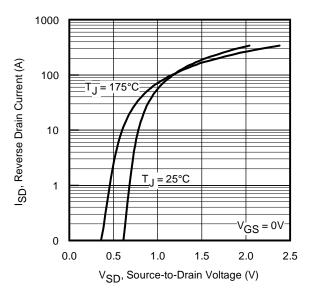


Fig 7. Typical Source-Drain Diode Forward Voltage

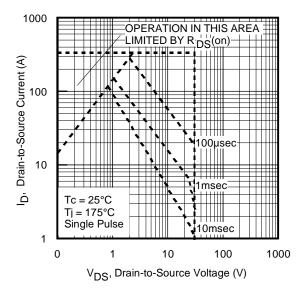
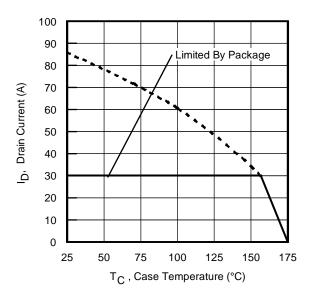


Fig 8. Maximum Safe Operating Area

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2.5 (A) a 2.0 (D) = $250\mu\text{A}$ 1.0 (D) = $250\mu\text{A}$ 1.0 (D) = $250\mu\text{A}$ 1.0 (D) = $250\mu\text{A}$ 1.75 (D) = 250

Fig 9. Maximum Drain Current vs. Case Temperature

Fig 10. Threshold Voltage vs. Temperature

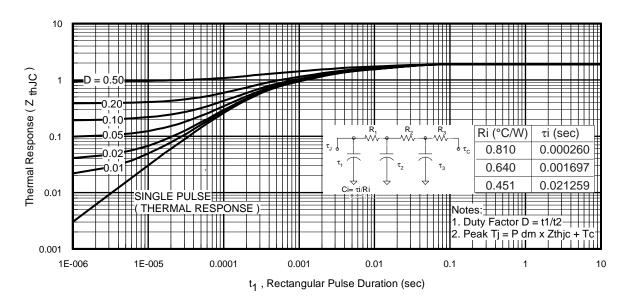


Fig 11. Maximum Effective Transient Thermal Impedance, Junction-to-Case

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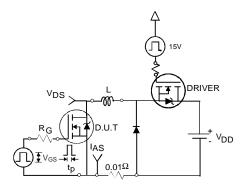


Fig 12a. Unclamped Inductive Test Circuit

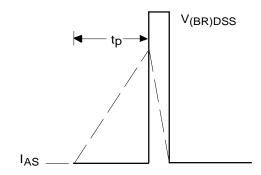


Fig 12b. Unclamped Inductive Waveforms

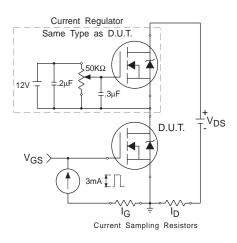


Fig 13. Gate Charge Test Circuit

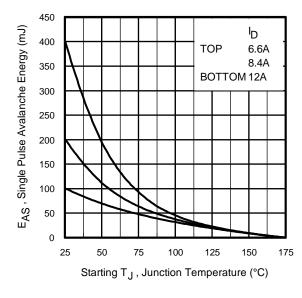


Fig 12c. Maximum Avalanche Energy vs. Drain Current

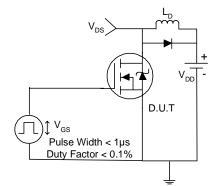


Fig 14a. Switching Time Test Circuit

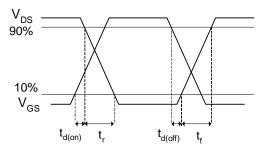


Fig 14b. Switching Time Waveforms www.irf.com

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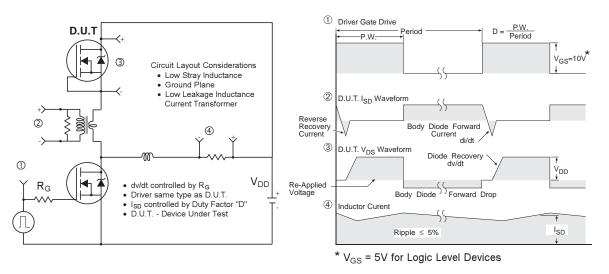


Fig 15. Peak Diode Recovery dv/dt Test Circuit for N-Channel HEXFET® Power MOSFETs

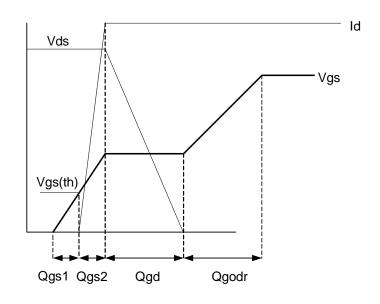


Fig 16. Gate Charge Waveform

Power MOSFET Selection for Non-Isolated DC/DC Converters

Control FET

Special attention has been given to the power losses in the switching elements of the circuit - Q1 and Q2. Power losses in the high side switch Q1, also called the Control FET, are impacted by the $R_{\text{ds(on)}}$ of the MOSFET, but these conduction losses are only about one half of the total losses.

Power losses in the control switch Q1 are given by;

$$P_{loss} = P_{conduction} + P_{switching} + P_{drive} + P_{output}$$

This can be expanded and approximated by;

$$\begin{split} P_{loss} &= \left(I_{rms}^{2} \times R_{ds(on)}\right) \\ &+ \left(I \times \frac{Q_{gd}}{i_{g}} \times V_{in} \times f\right) + \left(I \times \frac{Q_{gs2}}{i_{g}} \times V_{in} \times f\right) \\ &+ \left(Q_{g} \times V_{g} \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) \end{split}$$

This simplified loss equation includes the terms ${\rm Q_{gs2}}$ and ${\rm Q_{oss}}$ which are new to Power MOSFET data sheets.

 Q_{gs2} is a sub element of traditional gate-source charge that is included in all MOSFET data sheets. The importance of splitting this gate-source charge into two sub elements, Q_{gs1} and Q_{gs2} , can be seen from Fig 16.

 Q_{gs2} indicates the charge that must be supplied by the gate driver between the time that the threshold voltage has been reached and the time the drain current rises to I_{dmax} at which time the drain voltage begins to change. Minimizing Q_{gs2} is a critical factor in reducing switching losses in Q1.

 $\rm Q_{oss}$ is the charge that must be supplied to the output capacitance of the MOSFET during every switching cycle. Figure A shows how $\rm Q_{oss}$ is formed by the parallel combination of the voltage dependant (nonlinear) capacitance's $\rm C_{ds}$ and $\rm C_{dg}$ when multiplied by the power supply input buss voltage.

Synchronous FET

The power loss equation for Q2 is approximated by;

$$\begin{split} P_{loss} &= P_{conduction} + P_{drive} + P_{output}^* \\ P_{loss} &= \left(I_{rms}^2 \times R_{ds(on)}\right) \\ &+ \left(Q_g \times V_g \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) + \left(Q_{rr} \times V_{in} \times f\right) \end{split}$$

*dissipated primarily in Q1.

For the synchronous MOSFET Q2, $R_{ds(on)}$ is an important characteristic; however, once again the importance of gate charge must not be overlooked since it impacts three critical areas. Under light load the MOSFET must still be turned on and off by the control IC so the gate drive losses become much more significant. Secondly, the output charge Q_{oss} and reverse recovery charge Q_{rr} both generate losses that are transfered to Q1 and increase the dissipation in that device. Thirdly, gate charge will impact the MOSFETs' susceptibility to Cdv/dt turn on.

The drain of Q2 is connected to the switching node of the converter and therefore sees transitions between ground and $V_{\rm in}$. As Q1 turns on and off there is a rate of change of drain voltage dV/dt which is capacitively coupled to the gate of Q2 and can induce a voltage spike on the gate that is sufficient to turn the MOSFET on, resulting in shoot-through current . The ratio of $Q_{\rm gd}/Q_{\rm gs1}$ must be minimized to reduce the potential for Cdv/dt turn on.

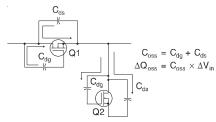
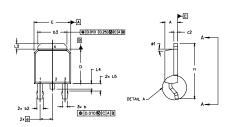


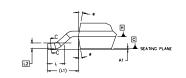
Figure A: Qoss Characteristic

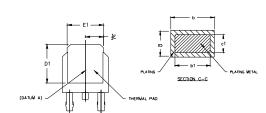
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D-Pak (TO-252AA) Package Outline

Dimensions are shown in millimeters (inches)



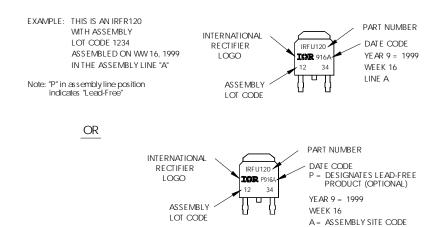




NOTE	'S:
1,0	DIMENSIONING AND TOLERANCING PER ASME Y14.5 M- 1994.
2.0	DIMENSIONS ARE SHOWN IN INCHES [MILLIMETERS]
3.0	LEAD DIMENSION UNCONTROLLED IN L5
4.0	DIMENSION DI AND EI ESTABLISH A MINIMUM MOUNTING SURFACE FOR THERMAL PAD.
5.0	SECTION C-C DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN .005 [0.127] AND
	.010 [0.2540 FROM THE LEAD TIP.
6.0	DIMENSION D & E DO NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED
	.005" (0.127) PER SIDE, THESE DIMENSIONS ARE MEASURED AT THE OUTERMOST
	EXTREMES OF THE PLASTIC BODY,
7.0	OUTLINE CONFORMS TO JEDEC OUTLINE TO 25244

		DIMEN	ISIONS				
SYMBOL	MILLIMETERS		INCHES		1		
	VIN.	MAX.	MINL	WAX.	NOTES		
A	2.18	2.39	.086	.094			
A1		0.13		.005			
b	0,64	0.89	.025	.035	5	LEAD ASSIGNMENTS	
ь1	0.64	0.79	.025	0.031	5		
b2	0.76	1,14	.030	.045		HEXFET	
b3	4.95	5.46	.195	.215			
c	0.46	0,61	.018	.024	5	1 GATE	
c1	0.41	0.56	.016	.022	5	2 DRAIN	
c2	.046	0.89	.018	.035	5	3 SOURCE	
D	5.97	6.22	.235	.245	6	4 DRAIN	
D1	5.21	-	.205	-	4		
Ε	6.35	6.73	.250	.265	6	IGBTs, CoPACK	
E1	4.32	-	.170		4	10010, 00111011	
e	2.	29	.090	BSC	1	1,- GATE	
н	9.40	10:41	.370	.410	1	2 COLLECTOR	
L	1.40	1.78	.055	.070		3 EMITTER	
L1	2.74	REF.	.108	REF.]	4 COLLECTOR	
L2	0,05	1 BSC	.020	BSC]		
L3	0.89	1,27	.035	.050	1		
L4		1.02		.040			
L5	1,14	1.52	.045	.060	3		
	o.	10"	o.	10*			
81	0"	15"	o.	15"			
		1			1		

D-Pak (TO-252AA) Part Marking Information

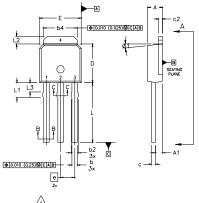


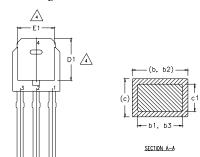
International

TOR Rectifier

I-Pak (TO-251AA) Package Outline

Dimensions are shown in millimeters (inches)





VIEW A-A

NOTES:

- 1 DIMENSIONING AND TOLERANCING PER ASME Y14.5 M- 1994.
- DIMENSIONS ARE SHOWN IN MILLIMETERS [INCHES],
- JIMENSION D & E DO NOT INCLUDE MOLD FLASH, MOLD FLASH SHALL NOT EXCEED 0.005" (0.127) PER SIDE. THESE DIMENSIONS ARE MEASURED AT THE OUTERMOST EXTREMES OF THE PLASTIC BODY.
- 4 THERMAL PAD CONTOUR OPTION WITHIN DIMENSION 64, L2, E1 & D1.
- 5 LEAD DIMENSION UNCONTROLLED IN L3.

6 DIMENSION 61, 63 APPLY TO BASE METAL ONLY.

7 OUTLINE CONFORMS TO JEDEC OUTLINE TO-251AA.

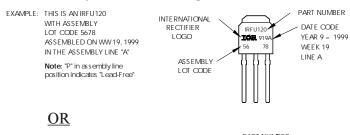
8 CONTROLLING DIMENSION: INCHES.

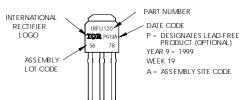
LEAD ASSIGNMENTS

<u>HE</u> >	HEXFET								
2 3	GATE DRAIN SOURCE DRAIN								

	DIMENSIONS						
SYMBOL	MBOL MILLIMETERS INCHES		HES				
	MIN.	MAX.	MIN.	MAX.	NOTES		
A	2,18	2,39	0,086	,094			
A1	0,89	1.14	0,035	0,045			
b	0,64	0,89	0.025	0,035			
ь1	0,64	0.79	0.025	0.031	4		
b2	0.76	1.14	0.030	0.045			
b3	0.76	1.04	0.030	0.041			
ь4	5.00	5.46	0.195	0.215	4		
c	0.46	0.61	0.018	0.024			
c1	0,41	0,56	0.016	0.022			
c2	.046	0.86	0.018	0.035			
D	5.97	6.22	0.235	0.245	3, 4		
D1	5.21	-	0.205	-	4		
E	6.35	6.73	0.250	0.265	3, 4		
E1	4.32	-	0.170	-	4		
e	2.	29	0.090	BSC			
L	8.89	9.60	0.350	0.380			
L1	1.91	2.29	0.075	0.090			
L2	0.89	1.27	0.035	0.050	4		
L3	1,14	1,52	0.045	0.060	5		
ø1	0*	15*	0*	15*			
1							

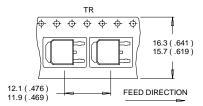
I-Pak (TO-251AA) Part Marking Information

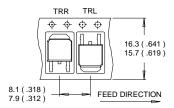




D-Pak (TO-252AA) Tape & Reel Information

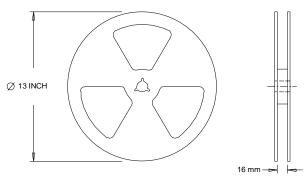
Dimensions are shown in millimeters (inches)





NOTES :

- 1. CONTROLLING DIMENSION : MILLIMETER.
- 2. ALL DIMENSIONS ARE SHOWN IN MILLIMETERS (INCHES).
 3. OUTLINE CONFORMS TO EIA-481 & EIA-541.



1. OUTLINE CONFORMS TO EIA-481.

Notes:

- ① Repetitive rating; pulse width limited by max. junction temperature.
- Starting $T_J = 25$ °C, L = 1.4mH, $R_G = 25\Omega$, $I_{AS} = 12A$.
- 3 Pulse width \leq 400 μ s; duty cycle \leq 2%.
- 4 Calculated continuous current based on maximum allowable junction temperature. Package limitation current is 30A.
- ⑤ When mounted on 1" square PCB (FR-4 or G-10 Material). For recommended footprint and soldering techniques refer to application note #AN-994.

Data and specifications subject to change without notice. This product has been designed and qualified for the Industrial market. Qualification Standards can be found on IR's Web site.



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Note: For the most current drawings please refer to the IR website at: http://www.irf.com/package/