

# Simultaneous Transmission of Information and Power

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## Readings

[Simultaneous Transmission of Information and Power.pdf](#)

[Transporting Information and Energy Simultaneously.pdf](#)

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## Signal and Channel Model

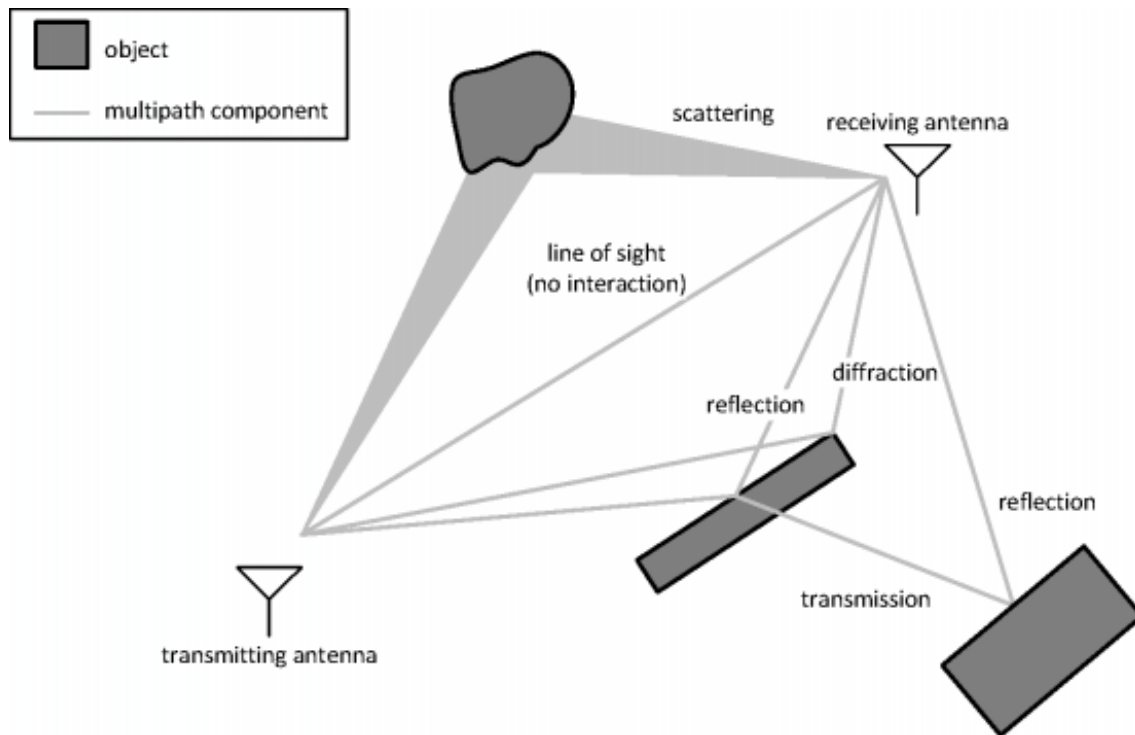
**Nyquist sampling theorem:** a signal with bandwidth  $B$  can be completely reconstructed by a sampling rate of  $2B$  samples per second.

A baseband equivalent signal of bandwidth  $W/2$  can be represented as

$$x(t) = \sum_n x\left(\frac{n}{W}\right) \text{sinc}(Wt - n) = \sum_n x[n] \text{sinc}(Wt - n)$$

- $x[n]$ : signal value at the  $n$ -th sample
- $\{\text{sinc}(Wt - n)\}_n$ : sampling kernels

**Multipath channel:** the receive signal is the superposition of the transmit signal propagating through different paths.



Multipath channel can be modelled as tapped delay line (TDL) [src]

$$y(t) = \sum_i a_i(t) x(t - \tau_i(t)) e^{-j2\pi f_c \tau_i(t)} + w(t)$$

function of t
AWGN

- $i$ : path index
- $a_i(t)$ : time-varying attenuation of path  $i$  channel response
- $\tau_i(t)$ : time-varying relative delay of path  $i$

# channel model :

May 2004

doc.: IEEE 802.11-03/940r4

## Appendix C – Model D

		Tap index	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18
		Excess delay [ns]	0	10	20	30	40	50	60	70	80	90	110	140	170	200	240	290	340	390
Cluster 1	Power [dB]		0	-0.9	-1.7	-2.6	-3.5	-4.3	-5.2	-6.1	-6.9	-7.8	-9.0	-11.1	-13.7	-16.3	-19.3	-23.2		
	AoA [°]		158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9	158.9		
	AS (receiver) AS [°]		27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7	27.7		
	AoD [°]		332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1	332.1		
	AS (transmitter) AS [°]		27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4	27.4		
Cluster 2	Power [dB]												-6.6	-9.5	-12.1	-14.7	-17.4	-21.9	-25.5	
	AoA [°]												320.2	320.2	320.2	320.2	320.2	320.2	320.2	
	AS [°]												31.4	31.4	31.4	31.4	31.4	31.4	31.4	
	AoD [°]												49.3	49.3	49.3	49.3	49.3	49.3	49.3	
	AS [°]												32.1	32.1	32.1	32.1	32.1	32.1	32.1	
Cluster 3	Power [dB]																-18.8	-23.2	-25.2	-26.7
	AoA [°]																276.1	276.1	276.1	276.1
	AS [°]																37.4	37.4	37.4	37.4
	AoD [°]																275.9	275.9	275.9	275.9
	AS [°]																36.8	36.8	36.8	36.8

Submission

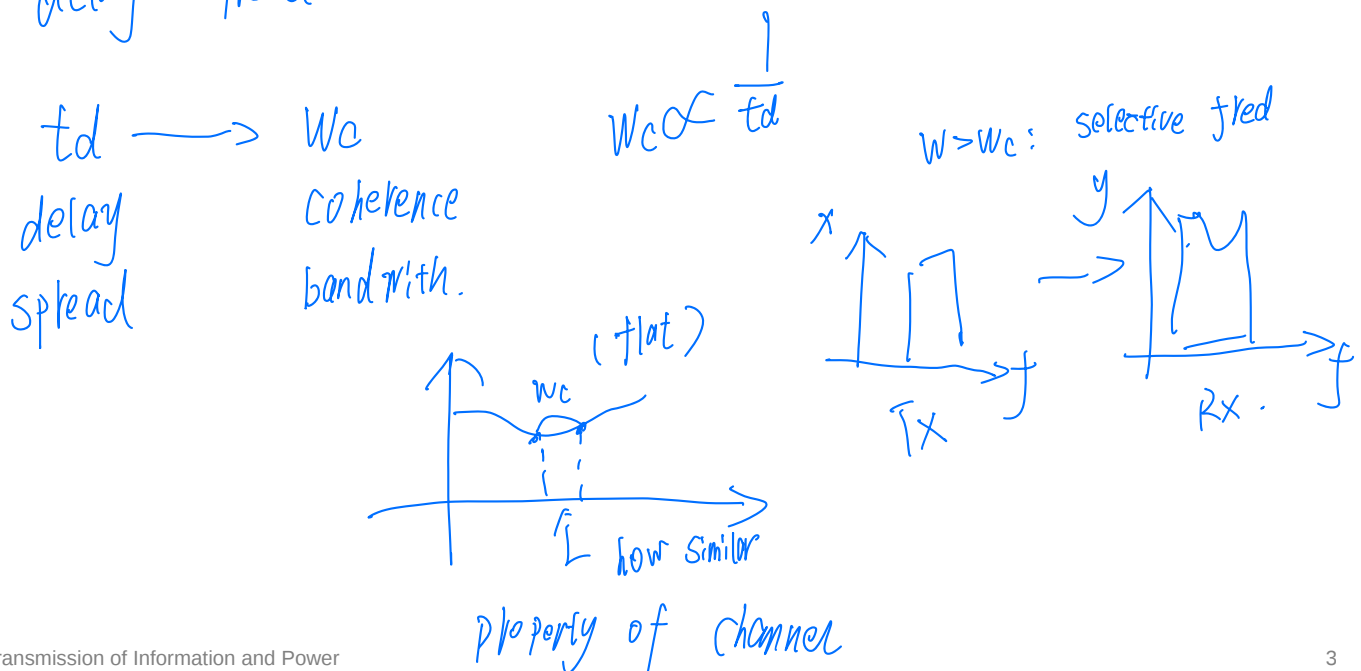
page 37

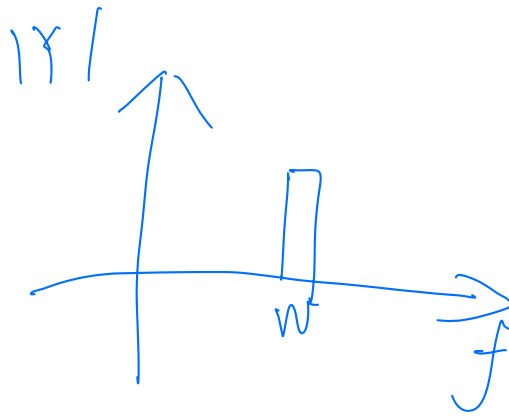
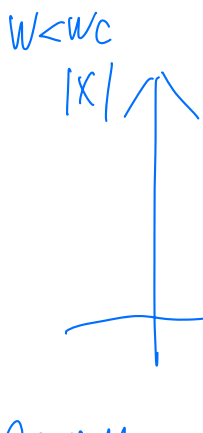
Vinko Erceg, Zyray Wireless; et al.

IEEE TGn channel model D, attenuation (power) and delay are time-invariant [\[src\]](#)

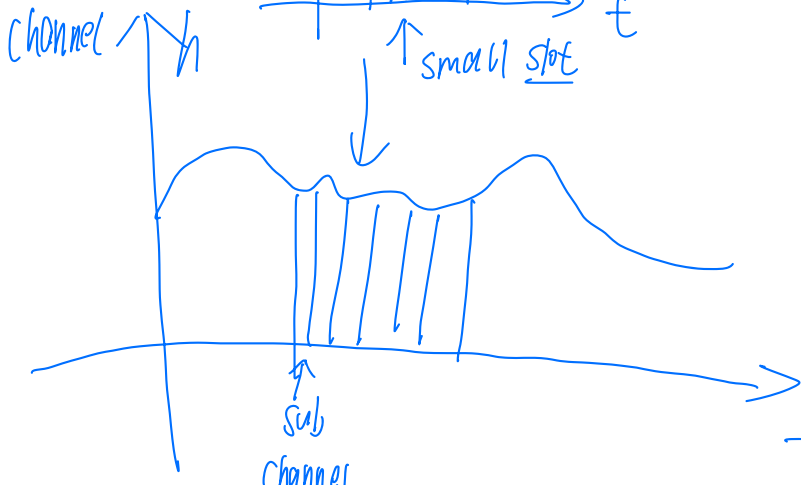
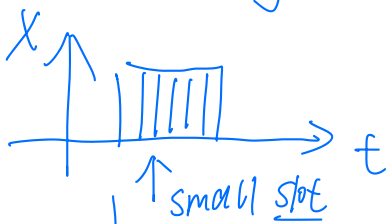
**Delay spread:** the span of signal arrival. (*definition not unique*)

$|t_n - t_1| = t_d$   
delay spread.





OFDM



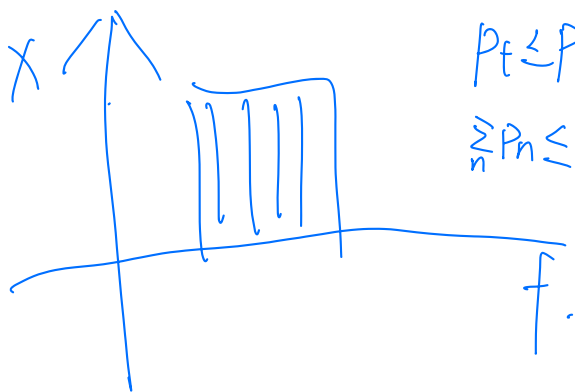
$y_1 \rightarrow h_1, x_1$

different noise.

$y_1 = h_1 x_1 + n_1$     $y_2 = h_2 x_2 + n_2$

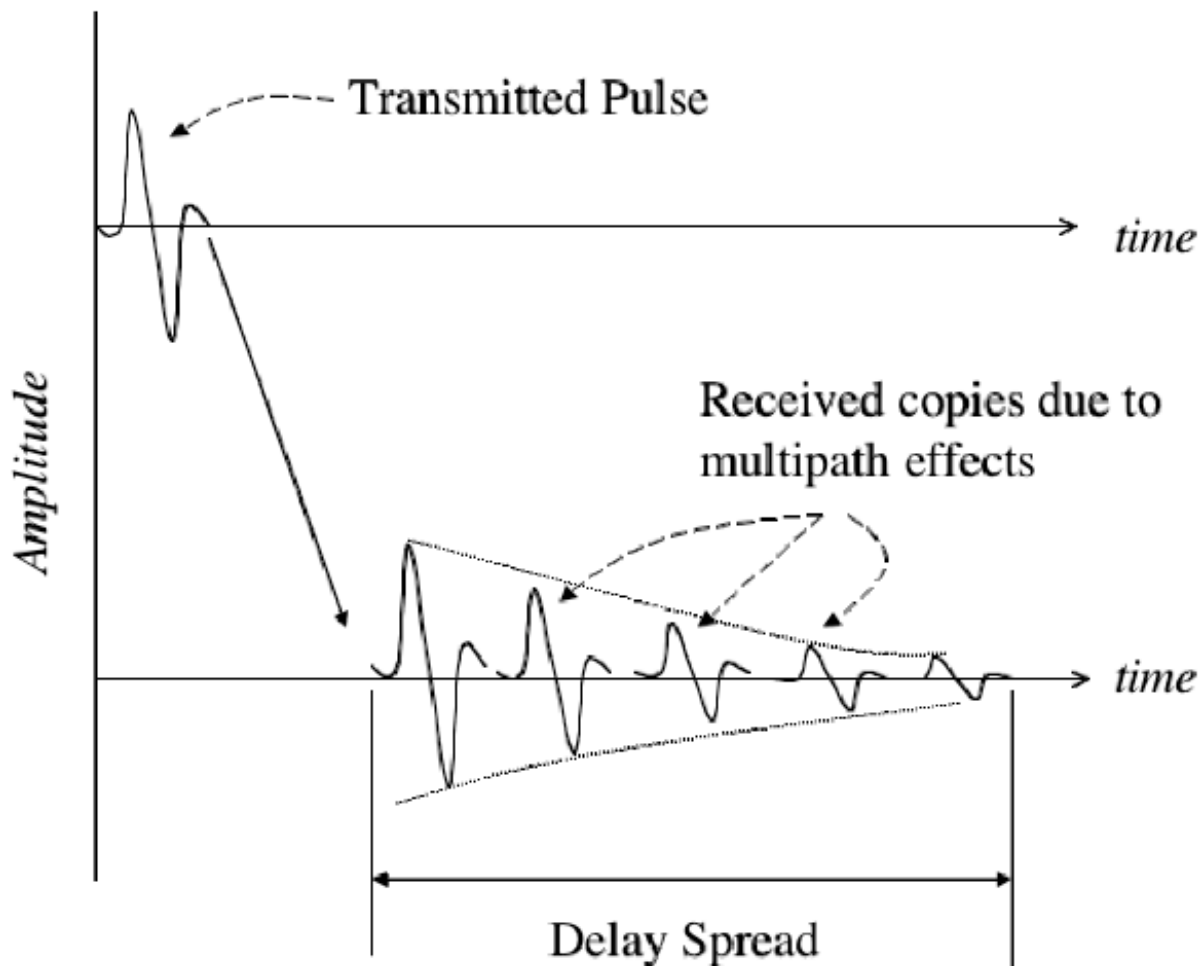
① flat fading for each component.

② subchannel are parallel.



$P_t \leq P$

$\sum_n P_n \leq P$     $P_n \geq 0$



Delay spread and multipath effect [src]

Delay spread  $T_d \rightarrow$  coherence bandwidth  $W_c$  (channel characteristics)

- $W < W_c$ : **flat fading** (different frequency components see the same channel)
- $W > W_c$ : **frequency-selective fading** (different frequency components see different channels)

**OFDM**: divide frequency-selective channel into multiple orthogonal subchannels, with independent AWGN.

$$\log_2(1 + \text{SNR}) \quad (\text{bps})$$

- power allocation on subcarrier  $n$ :  $\sum_n P_n = P, P_n \geq 0, \forall n$
- capacity of subband  $n$ :  $\log_2 \left( 1 + \frac{P_n |h_n|^2}{N_0} \right)$ 
  - $|h_n|$ : amplitude response at subchannel  $n$
  - $N_0$ : average noise power

- if use  $\log$  instead of  $\log_2$ , the unit is nats/s per symbol rather than bits/s per symbol

## Optimization

Consider a sum-rate maximization problem subject to average transmit power constraint:

$$\begin{aligned} \max_{P_0, \dots, P_{N_c-1}} \quad & \sum_{n=0}^{N_c-1} \log_2 \left( 1 + \frac{P_n |h_n|}{N_0} \right) \\ \text{s.t.} \quad & \begin{cases} \sum_{n=0}^{N_c-1} P_n = P \\ P_n \geq 0, \quad \forall n \end{cases} \end{aligned}$$

① local = global optim  
② KKT conditions  
y = x<sup>2</sup>  
Global optim

This is a convex problem (rest assured), which enables optimization tools such as KKT optimality conditions.

1. objective is convex.  
ii. constraints is convex.

The **standard form** of an optimization problem is

$$\begin{aligned} \min_x \quad & f_0(x) \\ \text{s.t.} \quad & \begin{cases} f_i(x) \leq 0, & i = 1, \dots, m \\ h_i(x) = 0, & i = 1, \dots, p \end{cases} \end{aligned}$$

min convex  
max. concave

(FYI only, to be a convex problem,  $f$ 's should be convex while  $h$ 's should be linear)

It's **Lagrangian** is defined as

just define make KKT

$$L(x, \lambda, \nu) = f_0(x) + \sum_{i=1}^m \lambda_i f_i(x) + \sum_{i=1}^p \nu_i h_i(x)$$

Lagrangian multiplier

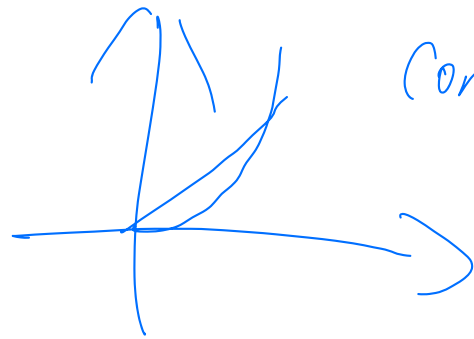
easy to present.

The **KKT conditions** are

- $f_i(x) \leq 0, \quad i = 1, \dots, m$
- $h_i(x) = 0, \quad i = 1, \dots, p$
- $\lambda_i \geq 0, \quad i = 1, \dots, m$



concave.



Convex.

$$\sum p_n$$

↓  
linear. (both concave & convex)

standard convex.

$$\min f_0(x)$$

$$\text{s.t. } f_i(x) \leq 0.$$

$$h_j(x) = 0 \quad j = 1, \dots, n$$

$$\begin{cases} f_0: \text{convex.} \\ f_i(x): \text{convex} \end{cases} \quad h_j(x) = \text{linear}$$

- $\lambda_i f_i(x) = 0, \quad i = 1, \dots, m$
- $\nabla f_0(x) + \sum_{i=1}^m \lambda_i \nabla f_i(x) + \sum_{i=1}^p \nu_i \nabla h_i(x) = 0$

Why KKT conditions?

- it is necessary for local optimality, for solvable problem
- it is sufficient for global optimality, for convex problem

Hence, if the problem is convex, we can solve KKT conditions for the optimal solution.

Tips for Exercise 2, 3

- write the optimization problem in the standard form
- use log for simplicity
- use `bar([carrierNoise; carrierPower], 'stacked')` for a stacked view

## Comments

Now it's your turn! Few suggestions:

- for  $P_d = 0$  (wireless communication only), the solution is the famous "water-filling" power allocation, could you explain the intuition behind?
- ~~how do  $\lambda$  and  $\mu$  balance the information and power transfer?~~
- ask questions and discuss together
- don't worry if you find it hard, you don't need to understand everything - this lab is just an introduction, more details in module *Wireless Communications* and *Topics in Large Dimensional Data Processing*

## Q&A

☐ Post any public question in Teams channel! You can also email Yang ([yang.zhao18@imperial.ac.uk](mailto:yang.zhao18@imperial.ac.uk)) or Prof Bruno Clerckx ([b.clerckx@imperial.ac.uk](mailto:b.clerckx@imperial.ac.uk)) for questions.



*(Last updated by @Yang Zhao at @September 30, 2020)*