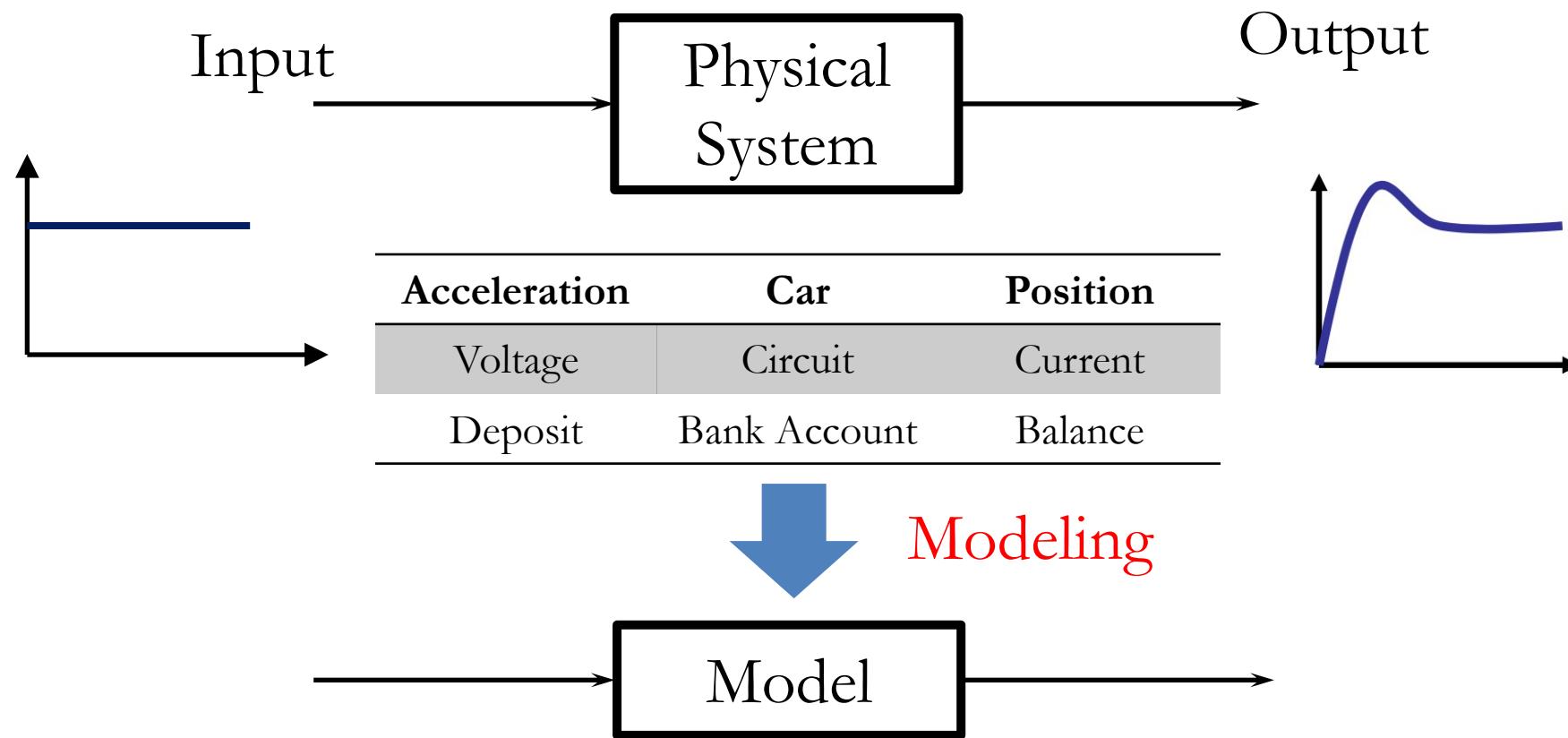




# Mechatronic Modeling and Design with Applications in Robotics

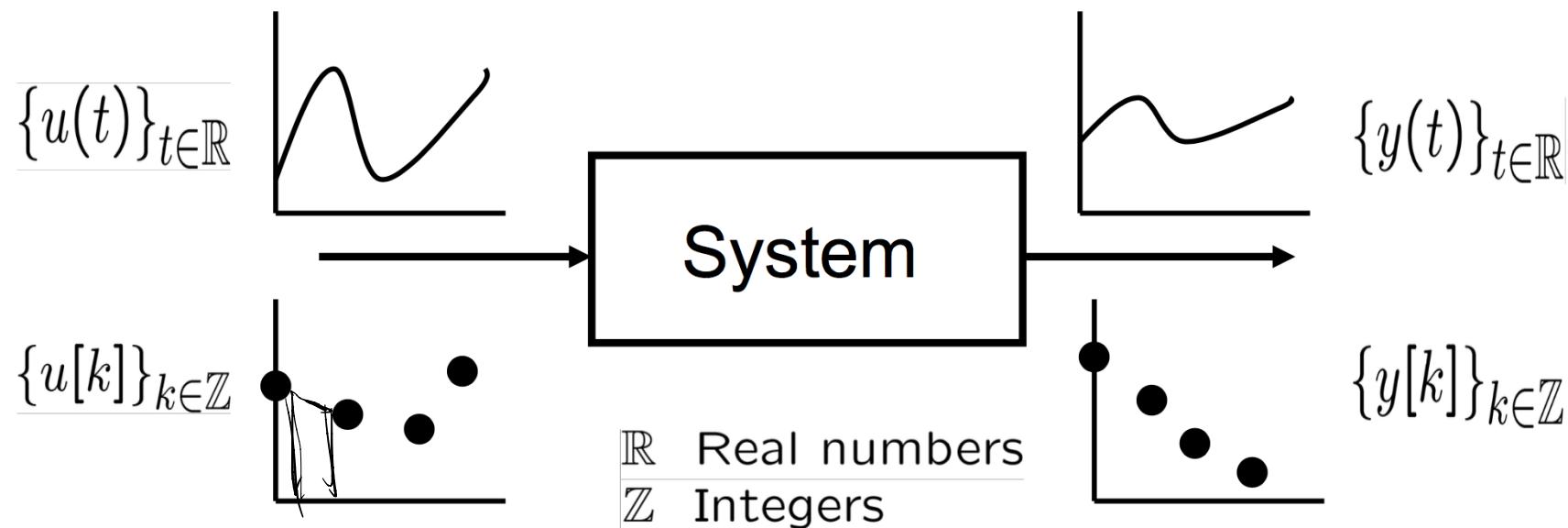
**Analytical Modeling (Part 1)**

Representation of the input-output relationship of a physical system.



- Continuous-time and discrete-time
- Memoryless, causal and noncausal
- Lumped and distributed
- Time-invariant and time-varying
- Linear and nonlinear

Input/output vectors are continuous-time signals



$\text{MHz}$

- Discrete-time system
- Input/output vectors are discrete-time signals

- Continuous-time system

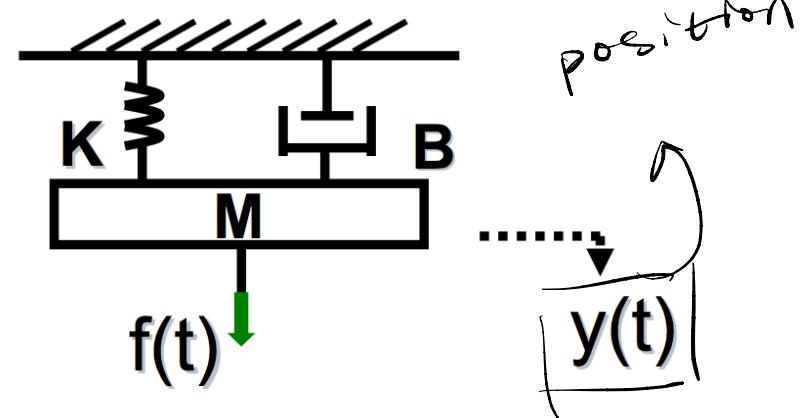
- Mass-spring-damper system

$$My''(t) = f(t) - By'(t) - Ky(t)$$

RLC circuit

$$v(t) = Ri(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt$$

*Differential equation*



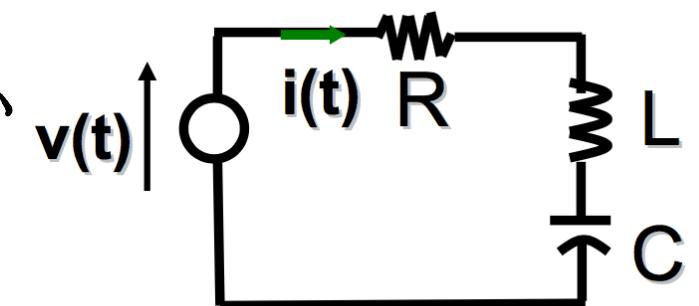
- Discrete-time System

- Digital computer
- Daily balance of a bank account

$$y[k+1] = (1+a)y[k] + u[k]$$

\*      step

*Difference equation*



**y[k] : balance at k-th day**  
**u[k] : deposit/withdrawal**  
**a : interest rate**

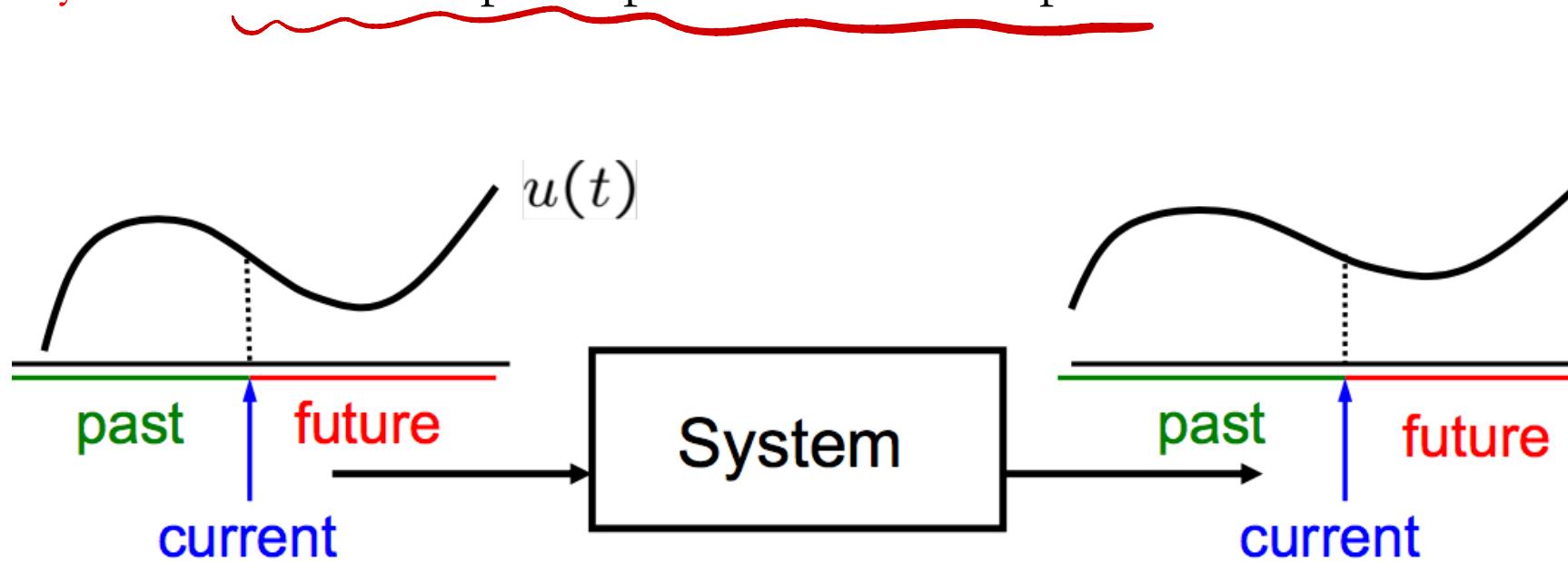
K :      day 1  
day 2  
day 3.

- Continuous-time and discrete-time
- Memoryless, causal and noncausal
- Lumped and distributed
- Time-invariant and time-varying
- Linear and nonlinear

**Memoryless system:** Current output depends on ONLY current input.

**Causal System:** Current output depends on current and past input.

**Noncausal system:** Current output depends on future input.



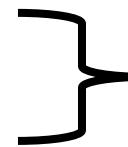
- Memoryless system
  - Spring: input  $f(t)$ , output  $x(t) \rightarrow f(t) = k\underline{x(t)}$
  - Resistor: input  $v(t)$ , output  $i(t) \rightarrow v(t) = Ri(t)$
- Causal System
  - Input: acceleration; output: position of a car  
  
Current position depends on not only current acceleration, but also all the past accelerations.
- Noncausal System does not exist in real world; it exists only mathematically. (We only consider causal systems)

- Continuous-time and discrete-time
- Memoryless, causal and noncausal
- **Lumped and distributed**
- Time-invariant and time-varying
- Linear and nonlinear

For a causal system,

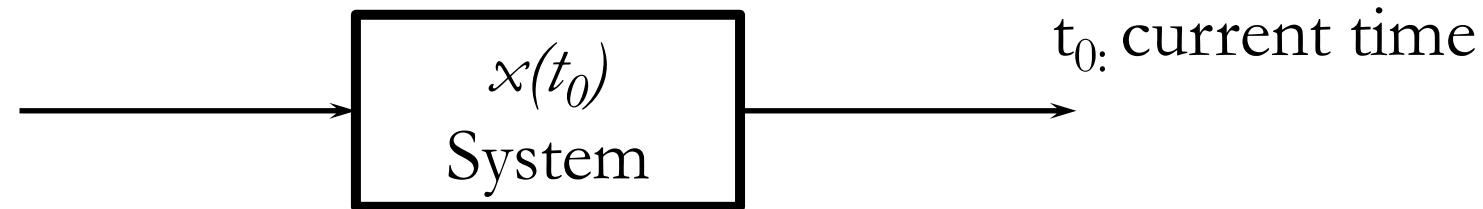
(Current/future input)

(past input)



Current/Future output

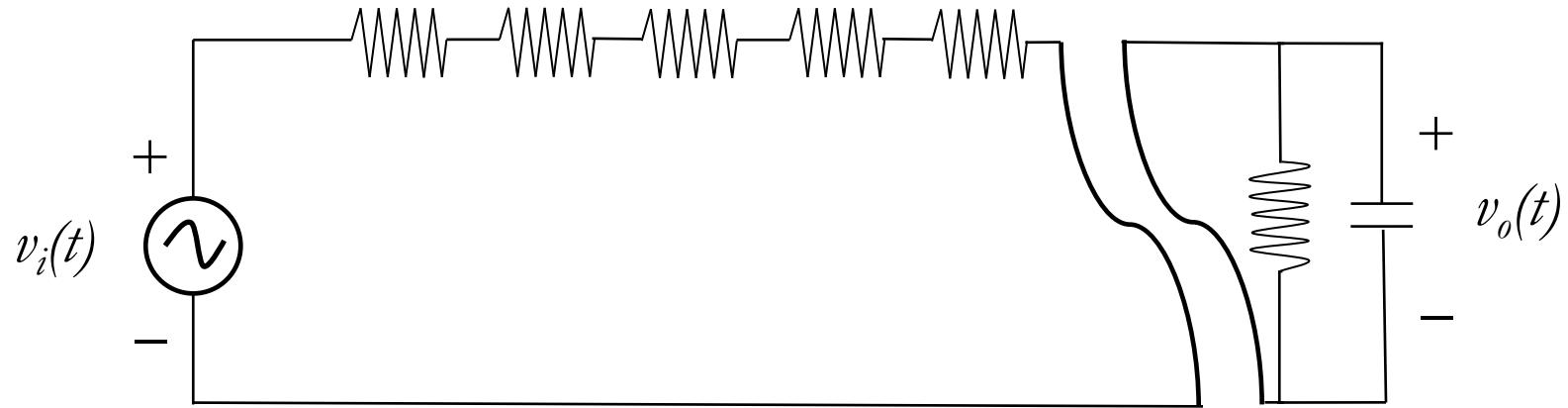
To Memorize this info, we use a state vector  $x(t_0)$



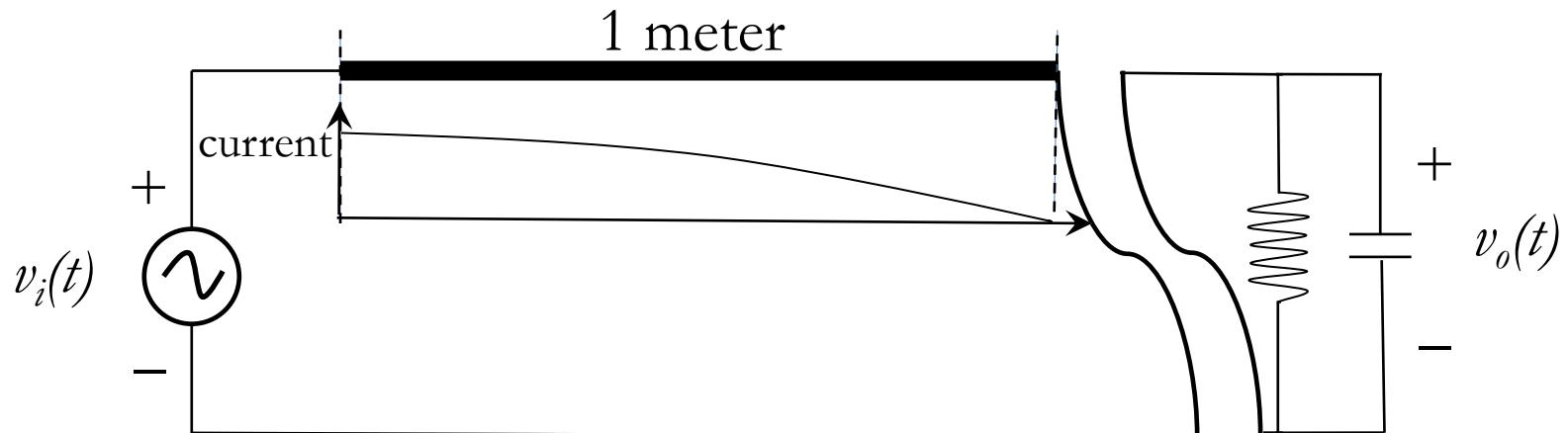
**Lumped system:** State vector is finite dimensional

**Distributed system:** State vector is infinite dimensional

- Lumped System



- Distributed System



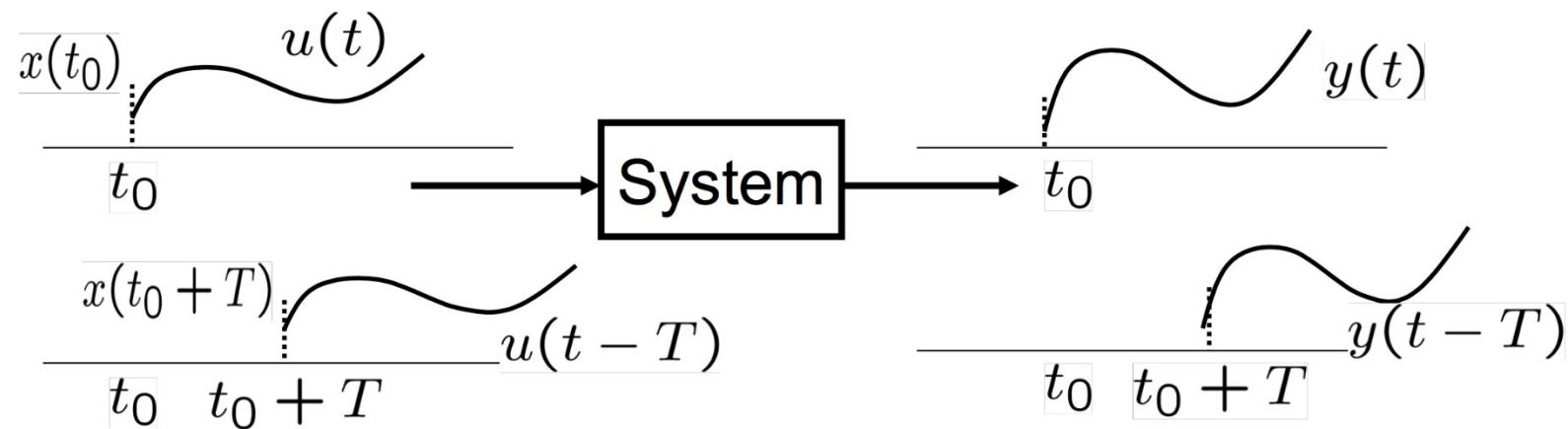
- Continuous-time and discrete-time
- Memoryless, causal and noncausal
- Lumped and distributed
- Time-invariant and time-varying
- Linear and nonlinear

For a causal system,  $\left. \begin{array}{l} x(t_0) \\ u(t), t \geq t_0 \end{array} \right\} \rightarrow y(t), t \geq t_0$

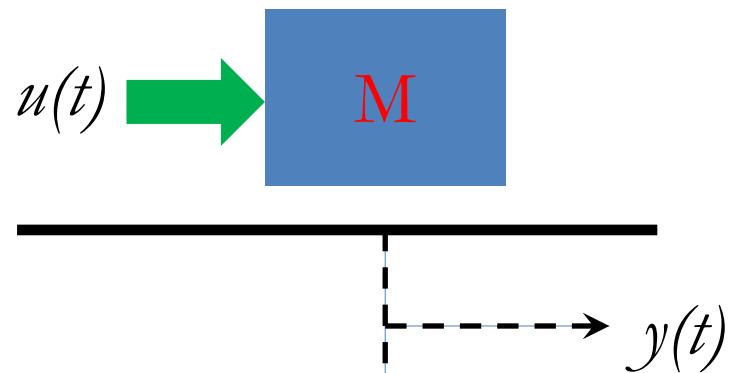
**Time-invariant system:** For any time shift  $T$ ,

$\left. \begin{array}{l} x(t_0 + T) \\ u(t - T), t \geq t_0 + T \end{array} \right\} \rightarrow y(t - T), t \geq t_0 + T$

Time-varying system: Not time-invariant



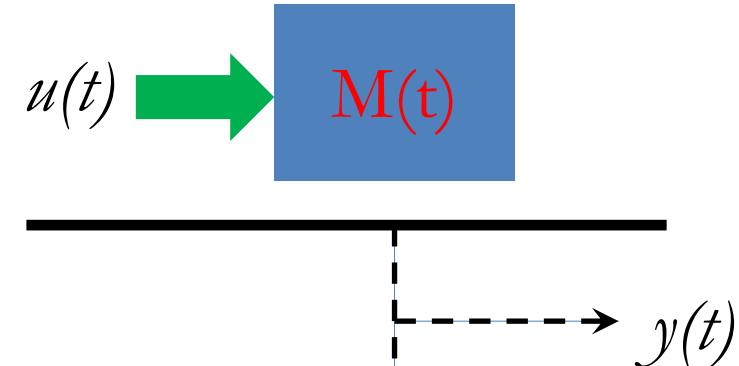
- Car, Rocket etc.



If we regard M to be **constant** (even though M changes very slowly), then this system is **time-invariant**.

$$My''(t) = u(t)$$

(Laplace applicable)



If we regard M to be **Changing** (due to **fuel consumption**), then this system is **time-varying**.

$$M(t)y''(t) = u(t)$$

(Laplace not applicable)

- Continuous-time and discrete-time
- Memoryless, causal and noncausal
- Lumped and distributed
- Time-invariant and time-varying
- Linear and nonlinear

For a causal system,

$$\left. \begin{array}{l} x_i(t_0) \\ u_i(t), t \geq t_0 \end{array} \right\} \rightarrow y_i(t), t \geq t_0, i = 1, 2$$

output  $y = f(x)$   $\pi$  input  $x = x_1$

$y_1 = f(x_1)$   $x = x_2$

**Linear system:** A system satisfying **superposition property**

$$\left. \begin{array}{l} \alpha_1 x_1(t_0) + \alpha_2 x_2(t_0) \\ \alpha_1 u_1 + \alpha_2 u_2(t), t \geq t_0 \\ t \geq t_0 \quad \forall \alpha_1, \alpha_2 \in \mathbb{R} \end{array} \right\} \rightarrow \alpha_1 y_1(t) + \alpha_2 y_2(t),$$

*Homogeneity*

$y = f(x)$   
 $\alpha y = f(\alpha x)$

$y_2 = f(x_2)$   
 $x = x_1 + x_2$   
 $f(x_1 + x_2) = y_1 + y_2.$

**Nonlinear system:** A system that does not satisfy superposition property.

$y = 5x$

$x = x_1$

$y_1 = 5 \cdot x_1$

$x = x_2$

$y_2 = 5 \cdot x_2$

$x = x_1 + x_2$

$y = 5(x_1 + x_2)$

$= 5x_1 + 5x_2 = y_1 + y_2$

$y = 5x + 1$

$x = x_1, y = 5x_1 + 1$

$x = x_2, y_2 = 5x_2 + 1$

$x = x_1 + x_2, y = 5(x_1 + x_2) + 1$

$= 5x_1 + 5x_2 + 1$

$\neq y_1 + y_2 = 5x_1 + 5x_2 + 2$

- All systems in real world are nonlinear.

$$f(t) = Ky(t) \rightarrow$$

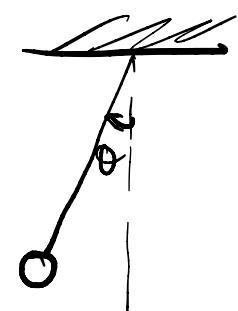
This linear relation holds only for small  $y(t)$  and  $f(t) = \text{Approximation}$

### Linearization

Model  $\neq$  perfect

- However, linear approximation is often good enough for control purposes

- Linearization:** approximation of a nonlinear system by linear system around some operating point



$$mL^2 \ddot{\theta}(t) = T(\theta) - mgL \sin(\theta(t))$$

$$\theta \rightarrow 0^\circ$$

$$\sin \theta \approx \theta$$

$$\sin \theta(t)$$

$$mgL\theta(t)$$

Taylor series expansion

Nonlinear term



Continuous-time

$$\begin{cases} \dot{x} = Ax + Bu \\ y = Cx + Du \end{cases}$$

$t \in \mathbb{R}$  (Real number)

① state equation

Discrete-time

② output equation

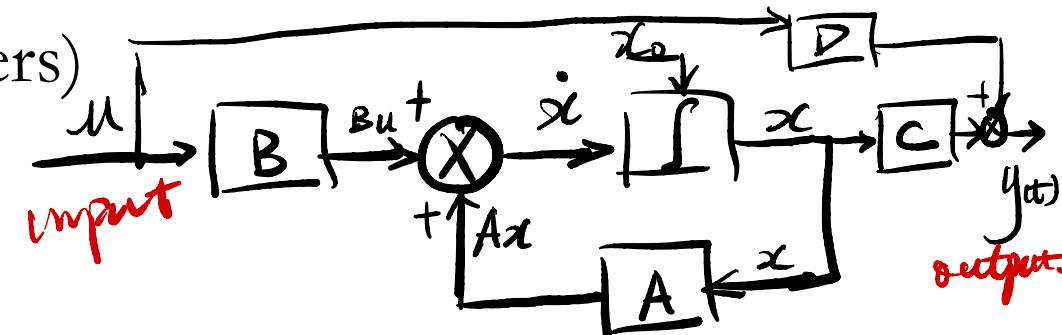
$$\begin{cases} x[k+1] = Ax[k] + Bu[k] \\ y[k] = Cx[k] + Du[k] \end{cases}$$

$$\dot{x} = Ax + Bu$$

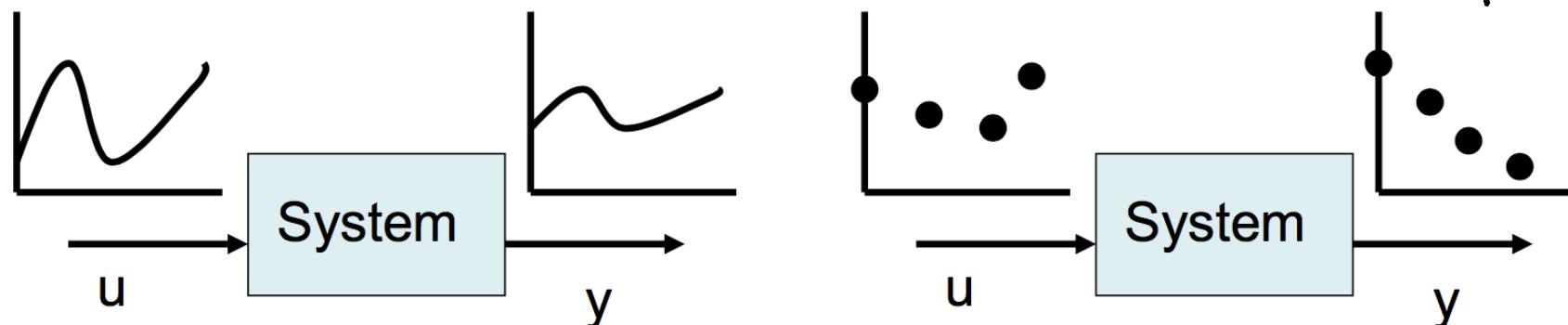
$$y = Cx + Du$$

$k \in \mathbb{Z}$  (Integers)

- x: state vector
- u: input vector
- y: output vector



Block diagram



- The first equation, called **state equation**, is a first order ordinary differential (CT case) and difference (DT case) equation.
- The second equation, called **output equation**, is an algebraic equation.
- Two equations are called **state-space model**.
- If a system is **time-invariant**, the matrices A, B, C, D are constant (independent of time).  
 $A, B, C, D,$
- Pay attention to sizes of matrices and vectors. They must be always compatible!

$$\dot{x} = Ax + Bu$$

$$y = Cx + Du$$

*n<sup>th</sup> order DE*

Consider a general  $n$ th-order model of a dynamic system:

$$\frac{d^n y(t)}{dt^n} + \underbrace{a_{n-1} \frac{d^{n-1}y(t)}{dt^{n-1}}}_{b_1 \frac{du(t)}{dt} + b_0 u(t)} + \cdots + a_1 \frac{dy(t)}{dt} + a_0 y(t) = b_n \frac{d^n u(t)}{dt^n} + b_{n-1} \frac{d^{n-1}u(t)}{dt^{n-1}} + \cdots +$$

*input*      *output*

Assuming all initial conditions are all **zeros**.

**Goal:** to derive a systematic procedure that transforms a differential equation of order  $n$  to a state space form representing a system of  $n$  first-order differential equations.

$$\dot{\underline{x}} = \underline{A}\underline{x} + \underline{B}u \quad \text{1st order DE}$$

*n<sup>th</sup> order differential equation  
to a 1st order differential equation*

*how to formulate the state equations*

→ 6th order differential equation.

Consider a dynamic system represented by the following differential equation:

$$y^{(6)} + 6y^{(5)} - 2y^{(4)} + y^{(2)} - 5y^{(1)} + 3y = 7u^{(3)} + u^{(1)} + 4u$$

where  $y^{(i)}$  stands for the  $i$ th derivative:  $y^{(i)} = d^i y / dt^i$ . Find the state space model of the above system.

$$\begin{aligned} A_{6 \times 6} &= \begin{bmatrix} 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{bmatrix} & B_{6 \times 1} &= \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} \\ C_{1 \times 6} &= \begin{bmatrix} 4 & 1 & 0 & 7 & 0 & 0 \end{bmatrix} & D &= [0] \end{aligned}$$

$$\begin{array}{l} \dot{x} = Ax + Bu \\ y = Cx + Du \end{array}$$

A  
 B  
 C  
 D  
 x is defined.  
 x  
 u - input  
 y output

State-space model

# Example: Mass with a Driving Force

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- By Newton's law, we have

$$M\ddot{y}(t) = u(t)$$

$u$ : input force

$y$ : output position

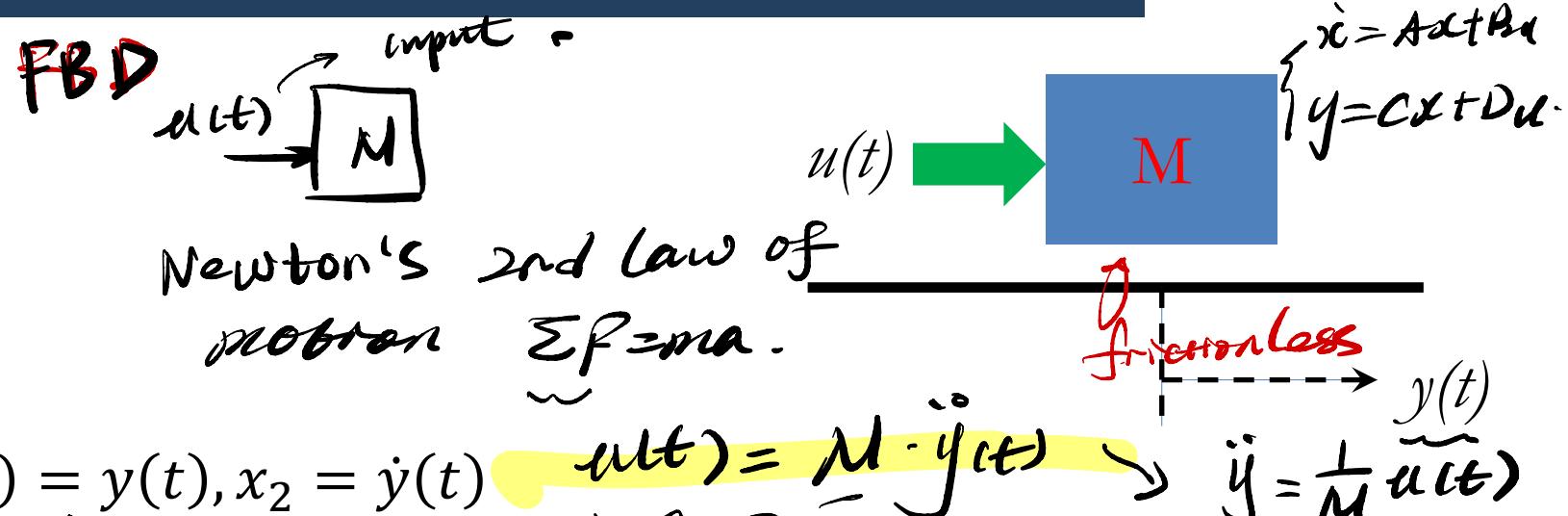
- Define state variables:  $x_1(t) = y(t), x_2 = \dot{y}(t)$

state

$$\begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} y(t) \\ \dot{y}(t) \end{bmatrix}$$

- Then,

$$\begin{cases} \dot{x}_1(t) = \dot{y}_1(t) = x_2(t) \\ \dot{x}_2(t) = \ddot{y}(t) = \frac{1}{M}u(t) \\ y(t) = x_1(t) \end{cases}$$



Newton's 2nd law of motion  $\sum F = ma$ .

$$\ddot{y} = \frac{1}{M}u(t)$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \dot{y}(t) \\ \ddot{y}(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} y(t) \\ \dot{y}(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} u(t)$$

$$y(t) = [1 \ 0] \begin{bmatrix} y(t) \\ \dot{y}(t) \end{bmatrix} + [0] u(t)$$

$$\begin{cases} \dot{x}_1(t) = x_2(t) \\ \dot{x}_2(t) = \frac{1}{M}u(t) \end{cases} \rightarrow \begin{cases} \frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} u(t) \\ y(t) = [1 \ 0] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} \end{cases}$$

$$B = \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix}$$

$$C = \begin{bmatrix} 1 & 0 \end{bmatrix} \quad D = \begin{bmatrix} 0 \end{bmatrix}$$

# Mass-Spring-Damper System

- By Newton's law

$$M\ddot{y}(t) = u(t) - B\dot{y}(t) - ky(t)$$

$$\ddot{y}(t) = \frac{1}{M}u(t) - \frac{B}{M}\dot{y}(t) - \frac{k}{M}y(t)$$

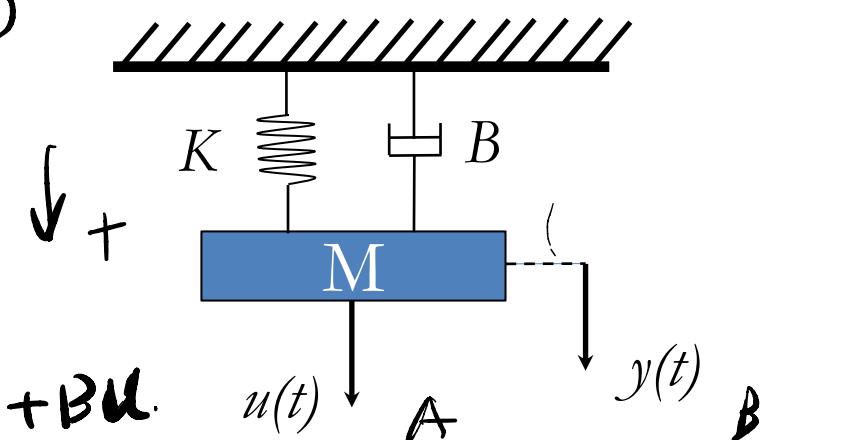
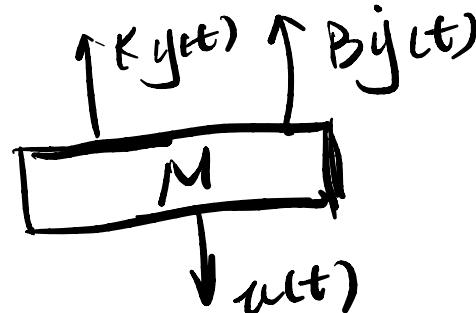
- Define state variables

$$x_1(t) = y(t), x_2(t) = \dot{y}(t)$$

$$\begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} y(t) \\ \dot{y}(t) \end{bmatrix} \quad \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \dot{y} \\ \ddot{y} \end{bmatrix}$$

$$\begin{cases} \frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -K/M & -B/M \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1/M \end{bmatrix} u(t) \\ y(t) = [1 \ 0] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} \end{cases}$$

FBD.



$$\dot{x} = Ax + Bu$$

$$y = Cx + Du$$

$$\begin{cases} \begin{bmatrix} \dot{y} \\ \ddot{y} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{K}{M} & -\frac{B}{M} \end{bmatrix} \begin{bmatrix} y \\ \dot{y} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} u(t) \\ y(t) = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} y \\ \dot{y} \end{bmatrix} + \begin{bmatrix} 0 \end{bmatrix} u(t) \end{cases}$$

$$\begin{aligned} A &= \begin{bmatrix} 0 & 1 \\ -\frac{K}{M} & -\frac{B}{M} \end{bmatrix} & B &= \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} \\ C &= \begin{bmatrix} 1 & 0 \end{bmatrix} & D &= \begin{bmatrix} 0 \end{bmatrix} \end{aligned}$$

- $u(t)$ : input voltage

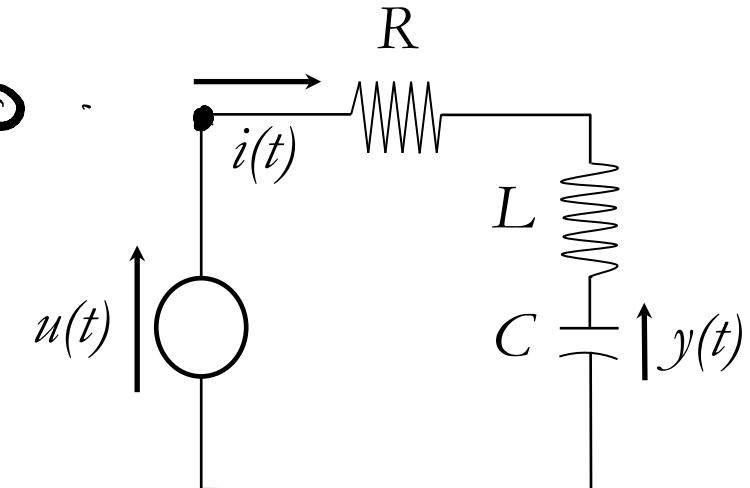
- $y(t)$ : output voltage

- By Kirchhoff's voltage law

$$u(t) = \underbrace{Ri(t)}_{\text{resistor}} + \underbrace{L \frac{di(t)}{dt}}_{\text{inductor}} + \underbrace{\frac{1}{C} \int i(\tau) d\tau}_{\text{capacitor}}$$

*current law (node)  
loop.)*

$\sum \text{Voltage} = 0$



\* Define State Variables (current for inductor, voltage for capacitor):

$$x_1(t) = i(t), x_2(t) = \frac{1}{C} \int i(\tau) d\tau$$

$$\begin{cases} \dot{x} = Ax + Bu \\ y = Cx + Du \end{cases}$$

$$\begin{cases} \frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} -R/L & -1/L \\ 1/C & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 1/L \\ 0 \end{bmatrix} u(t) \\ y(t) = [0 \quad 1] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} \end{cases}$$

2 additional  
questions in  
homework 1.

