

## ASSIGNMENT 8 SOLUTIONS.

sd① Half wave symmetry  $x(t) = -x(t - T/2)$

FS coefficients

$$a_n = \frac{2}{T} \int_{-T/2}^{T/2} x(t) \cos(n\omega_0 t) dt \text{ where } \omega_0 = \frac{2\pi}{T}$$

$$= \frac{2}{T} \left[ \int_{-T/2}^0 x(t) \cos(n\omega_0 t) dt + \int_0^{T/2} x(t) \cos(n\omega_0 t) dt \right]$$

$$= \frac{2}{T} [I_1 + I_2]$$

making the substitution  $\tau = t + T/2$  in  $I_1$  gives

$$I_1 = \int_0^{T/2} x(\tau - T/2) \cos(n\omega_0 (\tau - T/2)) d\tau$$

$$= - \int_0^{T/2} x(\tau) \cos(n\omega_0 \tau - n\pi) d\tau$$

$$= - \int_0^{T/2} x(\tau) (-1)^n \cos n\omega_0 \tau d\tau$$

$$= -(-1)^n \int_0^{T/2} x(t) \cos(n\omega_0 t) dt$$

we can write

$$a_n = \frac{2}{T} (1 - (-1)^n) \int_0^{T/2} x(t) \cos(n\omega_0 t) dt$$

from this expression we find that  $a_n = 0$

whenever  $n$  is even. In fact, we have

$$a_n = \begin{cases} \frac{4}{T} \int_0^{T/2} x(t) \cos(n\omega_0 t) dt, & n \text{ is odd} \\ 0, & n \text{ is even} \end{cases}$$

\* similar derivation leads to

$$b_n = \begin{cases} \frac{4}{T} \int_0^{T/2} x(t) \sin(n\omega_0 t) dt, & n \text{ is odd} \\ 0, & n \text{ is even} \end{cases}$$

$$(2) \quad x(t) \leftrightarrow X(\omega), \quad y(t) \leftrightarrow Y(\omega)$$

$$r_{xy}(t) = \int_{-\infty}^{\infty} x(\tau) y^*(t+\tau) d\tau, \quad r_{xy}(t) \leftrightarrow R_{xy}(\omega)$$

$$\text{i.e., } X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt, \quad Y(\omega) = \int_{-\infty}^{\infty} y(t) e^{-j\omega t} dt.$$

$$\text{Similarly, } R_{xy}(\omega) = \int_{-\infty}^{\infty} r_{xy}(t) e^{-j\omega t} dt$$

$$\text{i.e., } R_{xy}(\omega) = \int_{-\infty}^{\infty} e^{-j\omega t} \left[ \int_{-\infty}^{\infty} x(\tau) y^*(t+\tau) d\tau \right] dt$$

$$= \int_{-\infty}^{\infty} e^{-j\omega t} \left[ \int_{-\infty}^{\infty} x(\tau) y^*(t+\tau) dt \right] d\tau$$

$$= \int_{-\infty}^{\infty} x(\tau) \left[ \int_{-\infty}^{\infty} y^*(t+\tau) e^{-j\omega t} dt \right] d\tau$$

Now, using the property of Fourier transform that

$$\text{if } \begin{cases} x(t) \leftrightarrow X(\omega) \\ x(t-t_0) \leftrightarrow X(\omega) e^{-j\omega t_0} \end{cases}$$

$$\text{we get, } R_{xy}(\omega) = \int_{-\infty}^{\infty} x(\tau) \cdot Y^*(-\omega) e^{-j\omega \tau} d\tau$$

$$= Y(\omega) \int_{-\infty}^{\infty} x(\tau) e^{-j\omega \tau} d\tau = Y(\omega) \cdot X(\omega)$$

$$\text{Hence, } R_{xy}(\omega) = \mathcal{F} \left[ \int_{-\infty}^{\infty} x(\tau) y^*(t+\tau) d\tau \right] = X(\omega) \cdot Y^*(-\omega)$$

Q3) ⇒

a) No, Not every LTI system have an inverse

Ex:  $y(t) = x(t-4)$  it is a LTI system.

Now, if we give different input it will produce different output.

i/p	$\delta(t)$	$-\delta(t)$	$u(t)$	$-u(t)$
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o/p	$\delta(t-4)$	$-\delta(t-4)$	$u(t-4)$	$-u(t-4)$
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Hence, the system is invertible. and the inverse system gives  $y(t+4)$ .

2)  $y(t) = \frac{d}{dt} x(t)$  is a LTI system.

different i/p gives

i/p	2	5	10
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o/p	0	0	0
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Hence, for different i/p it gives same output  
therefore, the system doesn't have inverse.

→ b) To have inverse of system, it must give different output for different input.

Hence, it should have one one onto mapping between input & output.

c) Given,  $h(t) * h'(t) = \delta(t)$ . Take F.T. on both side.

$$\Rightarrow H(\omega) \cdot H'(\omega) = 1.$$

$$\Rightarrow \boxed{H(\omega) = 1/H'(\omega)}$$

4. Let  $x(t)$  have a support of  $(a, b)$ . Then with no loss of generality we can write

$$x(t) = x(t)p(t) \quad \text{where}$$
$$p(t) = \begin{cases} 1: & |t| < \max(|a|, |b|) \\ 0: & \text{otherwise} \end{cases}$$

$$\text{Then } X(\omega) = \frac{1}{2\pi} X(\omega) * P(\omega)$$

$P(\omega)$ , the transform of a rectangular pulse, has an infinite support as it is of the form  $(\sin x)/x$ . Since  $P(\omega)$  is convolved with  $X(\omega)$ , to get  $X(\omega)$  the result,  $X(\omega)$  will have infinite support.

To prove the vice versa result, express  $X(\omega)$  similarly as a product of a symmetric rectangular freq domain pulse with  $X(\omega)$  and then apply the modulation property discussed in class.

$$H(\omega) = \frac{5j\omega + 7}{(j\omega + 5)(-3j^2\omega^2 - 2j\omega + 2)}$$

Inverse,  $H^{-1}(\omega)$ :

$$\text{Let } sw = s \quad \frac{(jw+5)(-3jw^2-2jw+2)}{5jw+7}$$

$$H(s) = \frac{5s+7}{(s+5)(3s^2+2s+2)}$$

$$H(s) = \left( \frac{5s+7}{s+5} \right) \cdot \frac{1}{(3s^2-2s+2)}$$

The diagram shows a control system with two feedback loops. The first loop has a forward path with a summing junction, a  $\frac{1}{s}$  block, and a feedback path with a  $-5$  block. The second loop has a forward path with a summing junction, a  $(5s+7)$  block, a  $\frac{1}{3s}$  block, and a feedback path with a summing junction, a  $\frac{1}{s}$  block, and a  $2$  block. There is also a direct path from the output of the first loop to the second summing junction.

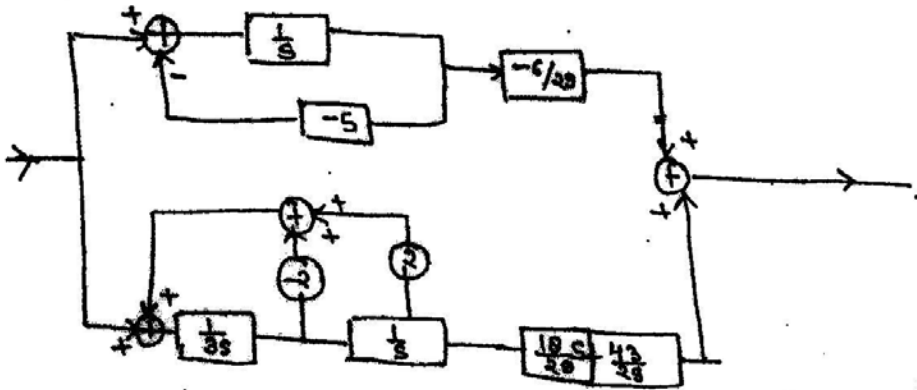
$$\boxed{\frac{1}{s}} \rightarrow \boxed{s}$$

$$\boxed{S} \rightarrow \boxed{\frac{d}{dt}}$$

$$H(s) = \frac{5s+7}{(s+5)(3s^2-2s+2)}$$

$$= \frac{A}{s+5} + \frac{Bs+C}{3s^2-2s+2}$$

$$\Rightarrow \frac{-6/29}{s+5} + \left(\frac{18}{29}\right)s + (43/29)$$



$$\frac{1}{s} \rightarrow \boxed{s}$$

$$s \rightarrow \boxed{\frac{d}{dt}}$$

$$G(\omega) = \frac{16 + (j\omega)^2}{(j\omega)(j\omega + 4)(j\omega - 4)}$$

Put  $j\omega = s$

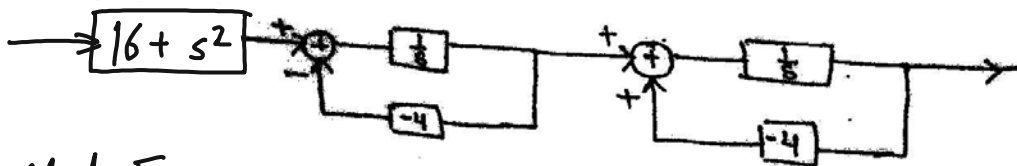
$$\text{Then } G(s) = \frac{16 + s^2}{s(s+4)(s-4)}$$

Cascade Form:

$$G(s) = \frac{(16 + s^2)}{(s)} \cdot \frac{1}{(s+4)} \cdot \frac{1}{(s-4)}$$

Inverse  $G^{-1}(\omega)$ :

$$\frac{j\omega(j\omega + 4)(j\omega - 4)}{(16 - \omega^2)}$$



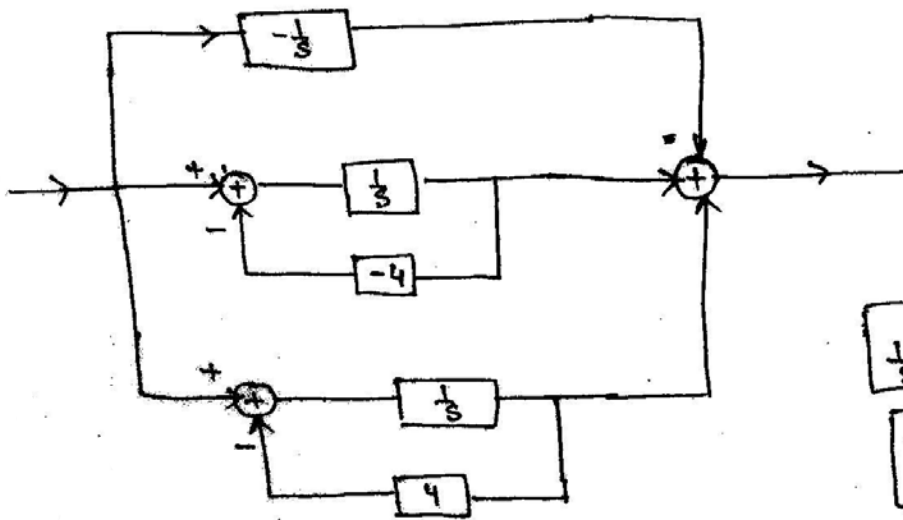
Parallel Form:

$$\frac{1}{s} \rightarrow \boxed{s}$$

$$s \rightarrow \boxed{\frac{d}{dt}}$$

$$G(s) = \frac{A}{s} + \frac{B}{s+4} + \frac{C}{s-4}$$

$$= \frac{-1}{s} + \frac{1}{s+4} + \frac{1}{s-4}$$



$\frac{1}{s}$	$\rightarrow$	$s$
$s$	$\rightarrow$	$\frac{d}{dt}$

Sol<sup>n</sup> ⑥

$$x(n) \xleftrightarrow{\text{DFT}} X(k)$$

$$y(n) = X(n) \xleftrightarrow{\text{DFT}} Y(k)$$

we know,

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi}{N} kn}$$

and

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j \frac{2\pi}{N} kn}$$

$$N x(n) = \sum_{k=0}^{N-1} X(k) e^{j \frac{2\pi}{N} kn}$$

$$N x(-n) = \sum_{k=0}^{N-1} X(k) e^{-j \frac{2\pi}{N} kn}$$

Exchanging the role of  $k$  and  $n$  we get

$$N x(-k) = \sum_{n=0}^{N-1} X(n) e^{-j \frac{2\pi}{N} kn}$$

$$N x(-k) = Y(k)$$

$$\boxed{Y(n) = N x(-n)}$$



$$\begin{aligned}
 \textcircled{7} \textcircled{a} \quad & x[n] = (-1)^n \quad \therefore X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}} \\
 \Rightarrow X[k] &= \sum_{n=0}^{N-1} (-1)^n e^{-j\frac{2\pi kn}{N}} = \sum_{n=0}^{N-1} \left( e^{-j\frac{2\pi k}{N}} \right)^n \\
 X[k] &= 1 - \left( e^{-j\frac{2\pi k}{N}} \right) + \left( e^{-j\frac{2\pi k}{N}} \right)^2 + \dots + \left( -e^{-j\frac{2\pi k}{N}} \right)^{N-1} \\
 \therefore X[k] &= \frac{1 - (-1)^N \exp(-j\frac{2\pi k}{N})}{1 + e^{-j\frac{2\pi k}{N}}}
 \end{aligned}$$

$$\boxed{X[k] = \frac{1 - (-1)^N \exp(-j2\pi k)}{1 + \exp(-j\frac{2\pi k}{N})} \quad \text{for } 0 \leq k < N}$$

$$\textcircled{b} \quad x[n] = 1 + (-1)^n = x_1[n] + x_2[n]$$

$$x[n] \leftrightarrow X[k], \quad \text{By linearity property,} \\
 X[k] = X_1[k] + X_2[k]$$

$$x_1[n] = 1 \quad 0 \leq n < N$$

$$x_1[k] = N \delta(k)$$

and,  $X_2[k]$  as found in the previous problem

$$\therefore \text{so, } x[n] = 1 + (-1)^n \leftrightarrow X[k]$$

$$\text{Where, } \boxed{X[k] = N \delta(k) + \frac{1 - (-1)^N \exp(-j2\pi k)}{1 + \exp(-j\frac{2\pi k}{N})} }$$

$$\text{for } 0 \leq k < N$$

(7) (c)  $x[n] = j^n \quad 0 \leq n < N$

$$\therefore X[k] = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi k n}{N}}$$

$$= \sum_{n=0}^{N-1} j^n e^{-j \left(\frac{2\pi k}{N}\right) n} = \sum_{n=0}^{N-1} \left( j e^{-j \frac{2\pi k}{N}} \right)^n$$

$$= \frac{1 - \left( j e^{-j \frac{2\pi k}{N}} \right)^N}{1 - j e^{-j \frac{2\pi k}{N}}}$$

$$\therefore X[k] = \frac{1 - \left( j e^{-j \frac{2\pi k}{N}} \right)^N}{1 - j e^{-j \frac{2\pi k}{N}}}$$

for  $0 \leq k < N$

(d)  $x[n] = 1, 0, \dots, 0$

or,  $x[n] = \delta[n] \quad 0 \leq n < N$

$$X[k] = \sum_{n=0}^{N-1} x[n] \exp\left(-j \frac{2\pi k n}{N}\right)$$

$$= \sum_{n=0}^{N-1} \delta[n] \exp\left(-j \frac{2\pi k n}{N}\right)$$

$$= 1$$

$$\therefore X[k] = 1 \quad 0 \leq k \leq N-1$$

Q8 → Given,  $x[n]$  is  $N$ -length sequence and  $x'[n]$  is  $2N$ -length sequence.

Then, DFTs of above signals are,

$$X[k] = \sum_{n=0}^{N-1} x[n] \cdot e^{-j2\pi nk/N}$$

$$X[k] = x[0] + x[1]e^{-j2\pi k/N} + \dots + x[N-1]e^{-j2\pi k(N-1)/N}$$

$$X[0] = x[0] + x[1] + \dots + x[N-1]$$

$$X[1] = x[0] + x[1]e^{-j2\pi/N} + \dots + x[N-1]e^{-j2\pi(N-1)/N}$$

⋮

$$X[N-1] = x[0] + x[1]e^{-j2\pi(N-1)/N} + \dots + x[N-1]e^{-j2\pi(N-1)^2/N}$$

And  $X[k]$  is periodic with period  $N$ .

Similarly,

$$X'[k] = x[0] + x[1]e^{-j2\pi k/2N} + \dots + x[N-1]e^{-j2\pi k(N-1)/2N}$$

$$X'[0] = x[0] + x[1] + \dots + x[N-1]$$

$$X'[1] = x[0] + x[1]e^{-j2\pi/2N} + \dots + x[N-1]e^{-j2\pi(N-1)/2N}$$

⋮

$$X'[2N-1] = x[0] + x[1]e^{-j2\pi(2N-1)/2N} + \dots + x[N-1]e^{-j2\pi(N-1)(2N-1)/2N}$$

Hence every member of  $X[k]$  is found to be on even

values of  $k$  in  $X'[k]$  and values at odd value of

$k$  in  $X'[k]$  are to be evaluated separately their no

such relation with old ' $N$  members'. (No shortcuts)

9.  $y[n] = x[n] - \alpha y[n-1]$ .

Assuming  $Y(\Omega)$  exists, we have

$$Y(\Omega) = X(\Omega) - \alpha Y(\Omega) e^{-j\Omega}$$

$$\text{So } H(\Omega) = \frac{Y(\Omega)}{X(\Omega)} = \frac{1}{1 - \alpha e^{-j\Omega}}$$

$Y(\Omega)$  will exist whenever  $|h[n]|$  is summable.

$h[n]$  is found as follows.

$$y[n] = x[n] - \alpha [x[n-1] - \alpha [x[n-2] - \dots]]$$

Setting  $x[n] = \delta[n]$ , we have

$$h[n] = \sum_{k=0}^{\infty} (-1)^k \alpha^n$$

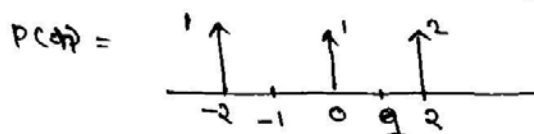
This geometric series is summable only for  $|\alpha| < 1$ . But this resulted because we solved the equation forward in time. Solving backward in time would that summability happens for  $|\alpha| > 1$ .

Summability of the impulse response implies that  $y[n]$  is bounded.  $Y(\Omega)$  will exist when both  $|h[n]|$  and  $|x[n]|$  are individually summable, as this ensures that  $|y[n]|$  is summable.

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$$x[n] \longleftrightarrow X(\omega)$$

①  $\dots, 0, x[-2], 0, \dots, x[0], 0, x[2], \dots$



$$y[n] = x[n] \cdot p[n]$$

$$Y(\omega) = X(\omega) \otimes P(\omega)$$

$$p[n] = \frac{1}{2} (1 + (-1)^n)$$

$$P(\omega) = \frac{1}{2} \left[ 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k) + 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - \pi - 2\pi k) \right]$$

$$P(\omega) = \pi \sum_{k=-\infty}^{\infty} \delta(\omega - \pi k)$$

$$Y(\omega) = X(\omega) \otimes P(\omega)$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) P(\omega - \omega) d\omega$$

$$= \frac{1}{2\pi} \pi \int_{-\pi}^{\pi} X(\omega - \omega) \sum_{k=-\infty}^{\infty} \delta(\omega - \pi k) d\omega$$

$$= \frac{1}{2} \sum_{k=-\infty}^{\infty} X(\omega - \pi k)$$

↓  
take  $\omega \in [0, 2\pi)$

$$Y(\omega) = \frac{1}{2} [X(\omega) + X(\omega - \pi)] \quad \text{Ans.}$$

(b)  $0, x[-1], 0, x[1], \dots$

same as part (a) with

$$p[n] = \frac{1}{2} [1 - (-1)^n]$$

$$P(\omega) = \pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k) - \pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k - \pi)$$

$$Y(\omega) = X(\omega) \otimes P(\omega)$$

$$Y(\omega) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega - \omega') \left[ \sum_{k=-\infty}^{\infty} \delta(\omega - \omega' - 2\pi k) - \sum_{k=-\infty}^{\infty} \delta(\omega - \omega' - \pi - 2\pi k) \right] d\omega'$$

$$Y(\omega) = \frac{1}{2} [X(\omega) + X(\omega - \pi)] \quad \underline{\text{Ans}}$$

(c) odd members inverted

$$Y(\omega) = Y_a(\omega) - Y_b(\omega)$$

using part (a) & (b)

$$Y(\omega) = X(\omega - \pi) \quad \underline{\text{Ans.}}$$

(d) even members inverted

$$Y(\omega) = Y_b(\omega) - Y_a(\omega)$$

from part (a) & (b)

$$\underline{\text{Ans}} \quad Y(\omega) = -X(\omega - \pi)$$