

Performance comparison of different digital Modulation schemes:-

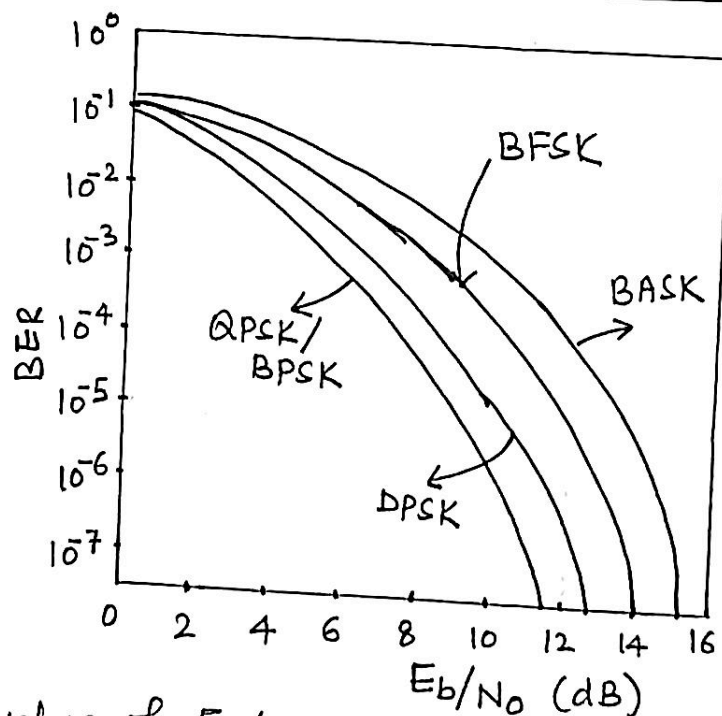
Modulation scheme	Euclidean Distance (d_{12})	Probability of error, P_e		Transmission BW, B_T	
		$Q(x)$	$\text{erfc}(x)$	Min ^m	Max ^m
BASK	$\sqrt{E_b}$	$Q\left(\sqrt{\frac{E_b}{2N_0}}\right)$	$\frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{4N_0}}\right)$	R_b	$2R_b$
BFSK	$\sqrt{2E_b}$	$Q\left(\sqrt{\frac{E_b}{N_0}}\right)$	$\frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{2N_0}}\right)$	$3R_b$	$4R_b$ ($2R_b + f_1 - f_2$)
BPSK	$2\sqrt{E_b}$	$Q\left(\sqrt{\frac{2E_b}{N_0}}\right)$	$\frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{N_0}}\right)$	R_b	$2R_b$
QPSK		$Q\left(\sqrt{\frac{2E_b}{N_0}}\right)$	$\frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{N_0}}\right)$	R_b	R_b
DPSK		$\frac{1}{2} \exp\left[-\frac{E_b}{N_0}\right]$		R_b	$2R_b$

$$Q(x) = \frac{1}{2} \text{erfc}\left(\frac{x}{\sqrt{2}}\right)$$

$$\text{erfc}(x) = \frac{2}{\sqrt{\pi}} \int_x^{\infty} e^{-z^2} dz$$

In, general

$$P_e = Q\left(\sqrt{\frac{d_{12}^2}{2N_0}}\right)$$



Note: (1) For the same value of E_b/N_0 in all modulation schemes, BER is least in BPSK and QPSK.

$\text{erfc}(\uparrow) \rightarrow \text{value} \downarrow \Rightarrow \text{BER} \downarrow$
 $\hookrightarrow \text{Argument} \uparrow$

(2) Euclidean distance $\uparrow \Rightarrow \text{BER} \downarrow$

\Rightarrow Among BASK, BFSK, and BPSK \Rightarrow BPSK is most reliable.

- Channel bandwidth and transmitted power are two primary communication resources and have to be used as efficiently as possible.

⇒ Power utilization efficiency (energy efficiency): Measured by the required E_b/N_0 to achieve a certain P_e .

⇒ Spectrum utilization efficiency (bandwidth efficiency): Measured by the achievable data rate per unit bandwidth

$$\boxed{\eta = \frac{R_b}{B_T}}$$

- It is always desired to maximize bandwidth efficiency at to minimal required E_b/N_0 .

To see the ultimate power-bandwidth tradeoff, we use Shannon's channel capacity theorem (1948).

- channel capacity is the theoretical upper bound for the maximum rate at which information could be transmitted without error.
- For a bandlimited AWGN channel, the max^m rate achievable is given by -

$$R_b \leq C = B_T \log_2(1 + \text{SNR}) = B_T \log_2\left(1 + \frac{P_s}{N_0 B_T}\right)$$

$$\frac{E_b}{N_0} = \frac{P_s T_b}{N_0} = \frac{P_s}{R_b N_0} = \frac{P_s B_T}{R_b N_0 B_T} = \text{SNR} \frac{B_T}{R_b}$$

$$\boxed{\frac{E_b}{N_0} = \frac{B_T}{R_b} (2^{R_b/B_T} - 1)}$$

$$\text{As } \frac{R_b}{B_T} \rightarrow 0$$

$$\frac{E_b}{N_0} = \ln 2 = 0.693 = -1.59 \text{ dB}$$

⇒ This value is called the Shannon Limit. Received E_b/N_0 must be $> -1.6 \text{ dB}$ to ensure reliable communications.

- ⑥ Binary data are transmitted at a rate of 10^6 bits per second over a radio link. Assuming channel noise is AWGN with zero mean and PSD at the receiver input is 10^{-10} W/Hz, find the average carrier power required to maintain an average probability of error $P_e \leq 10^{-4}$ for coherent BFSK. Determine the minimum channel bandwidth required.

Ans: $R_b = 10^6$ bps, $\frac{N_0}{2} = 10^{-10} \Rightarrow N_0 = 2 \times 10^{-10}$ W/Hz

$$P_e \leq 10^{-4}$$

$$P_e = \frac{1}{2} \operatorname{erfc} \left(\sqrt{\frac{E_b}{2N_0}} \right)$$

$$10^{-4} = \frac{1}{2} \operatorname{erfc} \left(\sqrt{\frac{E_b}{2 \times 2 \times 10^{-10}}} \right)$$

$$E_b = 3.025 \times 10^{-9} \text{ Joules}$$

$$P_s = \frac{E_b}{T_b} = E_b R_b = 3.025 \text{ mW}$$

$$P_s \geq 3.025 \text{ mW}$$

← carrier power

$$(B_T)_{\min} = 3R_b = 3 \times 10^6 = 3 \text{ MHz.}$$

- ⑦ An FSK system transmits binary data at a rate of 10^6 bps. Assuming channel noise AWGN with zero mean and PSD 2×10^{-20} watts/Hz, determine the average probability of error. Assume coherent detection, and amplitude of received sinusoidal signal for both symbol 1 and 0 to be $1.2 \mu\text{V}$.

Ans: $R_b = 10^6 \text{ bps}$, $\frac{N_0}{2} = 2 \times 10^{-20} \Rightarrow N_0 = 4 \times 10^{-20}$

$A_c = 1.2 \mu\text{V}$

$$P_e = \frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{2N_0}}\right)$$

$$= \frac{1}{2} \text{erfc}\left(\sqrt{\frac{7.2 \times 10^{-19}}{2 \times 4 \times 10^{-20}}}\right)$$

$P_e = \frac{1}{2} \text{erfc}(3)$

$P_e = 1 \times 10^{-5}$

$$E_b = P_s T_b$$

$$= \frac{A_c^2}{2} \times \frac{1}{R_b}$$

$$= 7.2 \times 10^{-19} \text{ Joule}$$

$\text{erf}(x) + \text{erfc}(x) = 1$

error function complementary error fun.

- ⑧ A binary data is transmitted over an AWGN channel using BPSK at a rate of 1 Mbps. It is desired to have average probability of error $P_e \leq 10^{-4}$. Noise power spectral density is 10^{-12} w/Hz. Find the average carrier power to maintain this P_e . Also, $\text{erfc}(3.5) = 0.0002$ given.

Ans: $P_e \leq 10^{-4}$, $\frac{N_0}{2} = 10^{-12} \Rightarrow 2 \times 10^{-12} = N_0$

$R_b = 1 \text{ Mbps}$

$$P_e = \frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_b}{N_0}}\right)$$

$$10^{-4} = \frac{1}{2} \text{erfc}\left(\sqrt{\frac{P_s \times 10^{-6}}{2 \times 10^{-12}}}\right)$$

$$\text{erfc}^{-1}(0.0002) = \sqrt{5 \times 10^5 P_s}$$

$P_s = E_b / T_b$

$$E_b = P_s T_b = \frac{P_s}{R_b} = P_s \times 10^{-6}$$

$$\frac{E_b}{N_0} = \frac{P_s \times 10^{-6}}{2 \times 10^{-12}}$$

$\Rightarrow P_s = 24.5 \mu\text{Watts}$

- ⑨ In a binary communication channel, the symbol 0 is transmitted with $p(0) = 0.6$, and the symbol 1 is transmitted with $p(1) = 0.4$. Determine, the error probability of the channel if conditional probability of detecting symbol 0 is 10^{-4} and that of symbol 1 is 10^{-6} .

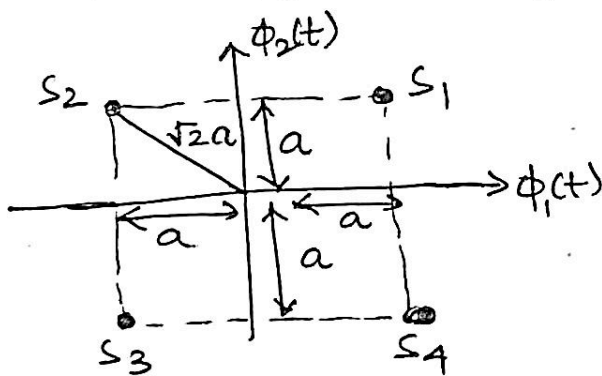
Ans:-

$$P_e = P(e/0) P(0) + P(e/1) P(1)$$

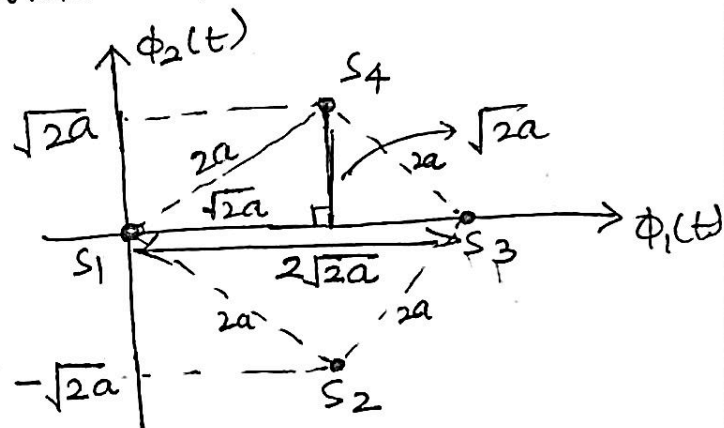
$$= 10^{-4} \times 0.6 + 10^{-6} \times 0.4$$

$$= 0.604 \times 10^{-4}$$

- ⑩ Two signal constellations are shown below. Compare the two modulation schemes in terms of their BER as well as energy efficiency performances.



Constellation A



Constellation B

Solution:- The min^m distance between signal points which are adjacent in the constellation decides the error probability. Here, $d_{min} = 2a$ for both the signal constellations. So both modulation schemes will perform identically in terms of BER.

const. A $\Rightarrow d_{min} = 2a$
 $A/VNB = 2$

const B. $d_{min} = 2a$
 $A/VNB = 2$

$\boxed{BER_A = BER_B}$

For energy efficiency.

const. A. \Rightarrow the average energy

$$E_{avg} = \frac{E_{s1} + E_{s2} + E_{s3} + E_{s4}}{4} = \frac{1}{4} [(\sqrt{2}a)^2 \times 4]$$

$$E_{s1} = E_{s2} = E_{s3} = E_{s4} = (\sqrt{2}a)^2 = 2a^2$$

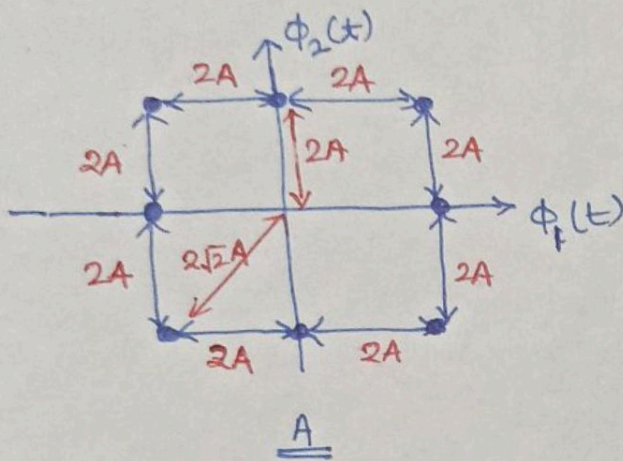
Note: The signal vector amplitude on the signal space is a measure of $\sqrt{E_b}$, where E_b is its energy.

const. B.

$$E_{avg} = \frac{0 + (2a)^2 + (2\sqrt{2}a)^2 + (2a)^2}{4} = 4a^2$$

Conclusion:- Constellation A is more energy efficient. It can give the same BER performance as constellation B, by transmitting half the energy.

- 12) Two sets of signal constellation for some modulation schemes are shown in the figure below. Minimum distance between adjacent points is $2A$. Assume that the signal points are equally probable, which constellation is more power efficient? Explain the nature of modulation shown by these two constellations.



Ans:-

Constellation A

$$E_{avg} = \frac{1}{8} [4(2A)^2 + 4(2\sqrt{2}A)^2] = 6A^2$$

$$d_{min} = 2A$$

$$AVNB = 2$$

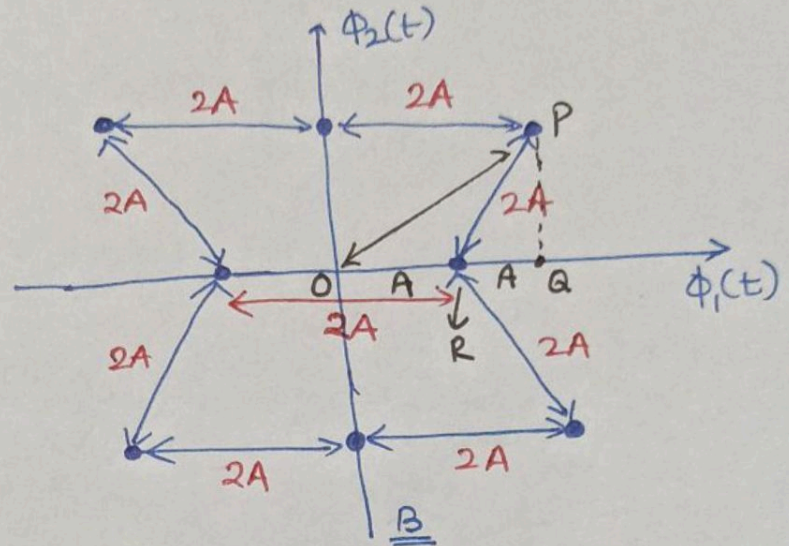
$$\boxed{BER_A = BER_B}$$

$$E_b = P_s \times T_b$$

$$P_s = \frac{E_b}{T_b} \Rightarrow P_{avg} = \frac{E_{avg}}{T_b}$$

$$\boxed{P_{avg} = \frac{6A^2}{T_b}}$$

\Rightarrow Thus the second constellation is more power efficient.



Constellation B

$$\begin{aligned} \Delta PQR \\ PQ &= \sqrt{(2A)^2 - (A)^2} \\ &= \sqrt{4A^2 - A^2} = \sqrt{3}A \end{aligned}$$

ΔOPA

$$OQ^2 + PQ^2 = OP^2$$

$$OP = \sqrt{(\sqrt{3}A)^2 + (2A)^2} = \sqrt{7}A$$

$$\begin{aligned} E_{avg} &= \frac{1}{8} [4(\sqrt{7}A)^2 + 2(A)^2 + 2(\sqrt{3}A)^2] \\ &= 4.5A^2 \end{aligned}$$

$$AVNB = 2$$

$$d_{min} = 2A$$

$$\boxed{P_{avg} = \frac{4.5A^2}{T_b}}$$