

EEE-2103: Electronic Devices and Circuits

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Series Voltage Negative Feedback

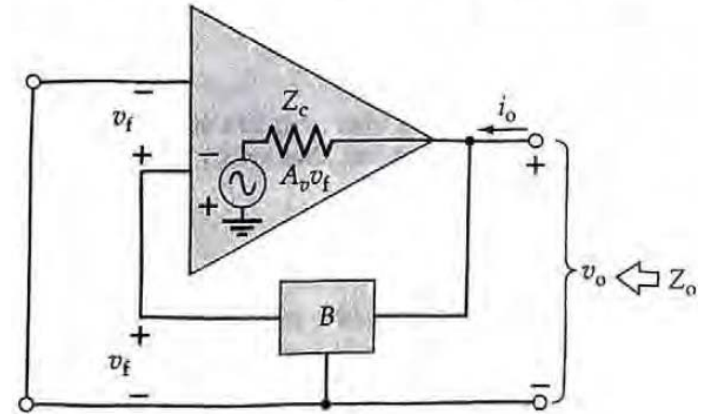
Output impedance:

Output voltage v_o produces feedback voltage v_f
 v_f generates voltage $A_v v_f$ in series with Z_c
 Z_c = output impedance without feedback

$$v_o = i_o Z_c - A_v v_f$$

$$i_o Z_c = v_o + A_v v_f = v_o + A_v B v_o = v_o (1 + A_v B)$$

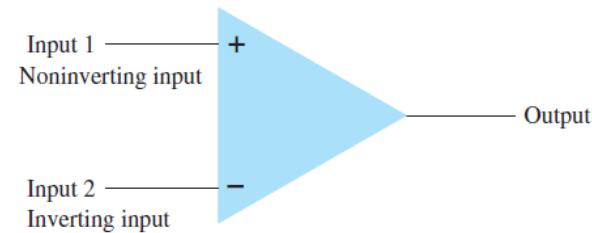
$$Z_o = v_o / i_o = \frac{Z_c}{1 + A_v B}$$



Operational Amplifiers (Op-Amps)

Op-amp → very high gain differential amplifier,
high input impedance,
low output impedance.

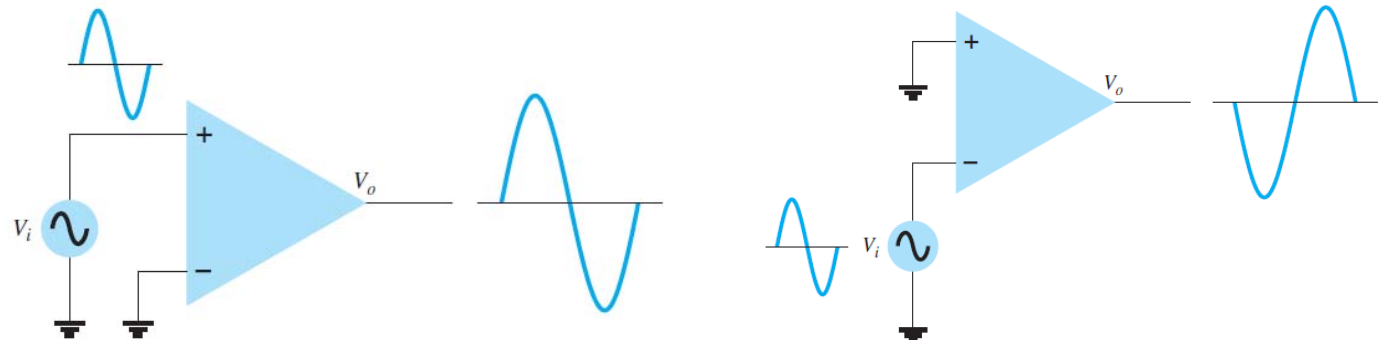
Uses → provide voltage amplitude changes,
oscillators,
filter circuits,
instrumentation circuits.



Signal is applied to + → same polarity (or phase) output

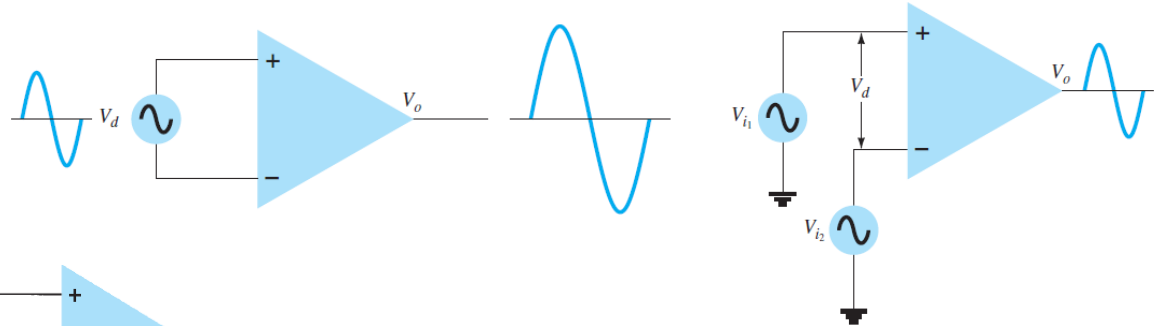
Signal is applied to - → opposite polarity (or phase) output

Single-Ended Input →

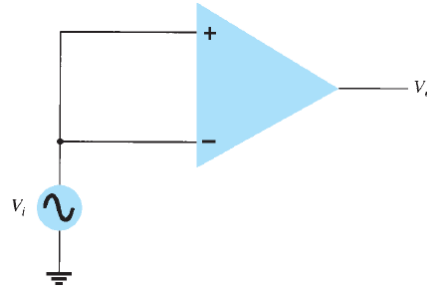


Operational Amplifiers (Op-Amps)

Double-Ended (Differential) Input →



Common-Mode Operation →



Common-Mode Rejection →

difference signal at inputs are highly amplified,
common signals at inputs are slightly amplified = rejected.
noise is common to both inputs.
differential connection provides → attenuation of unwanted input
amplification of difference input.

Voltage Follower Circuits

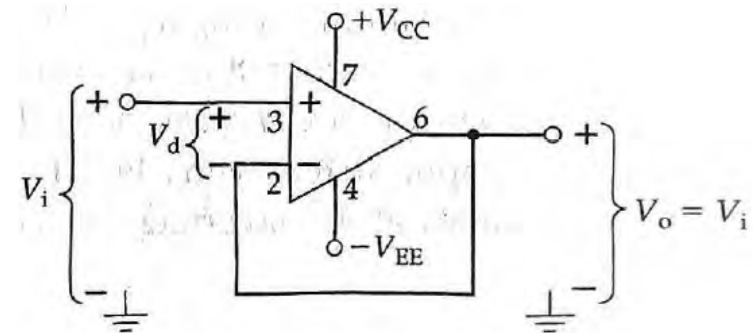
Suppose, $V_o > V_i \rightarrow$

$$V_{B2} = V_2 > V_{B1} = V_i$$

I_{C2} would be increased

V_{RC} will be increased

V_{B3} will be reduced $\rightarrow V_o = V_i$

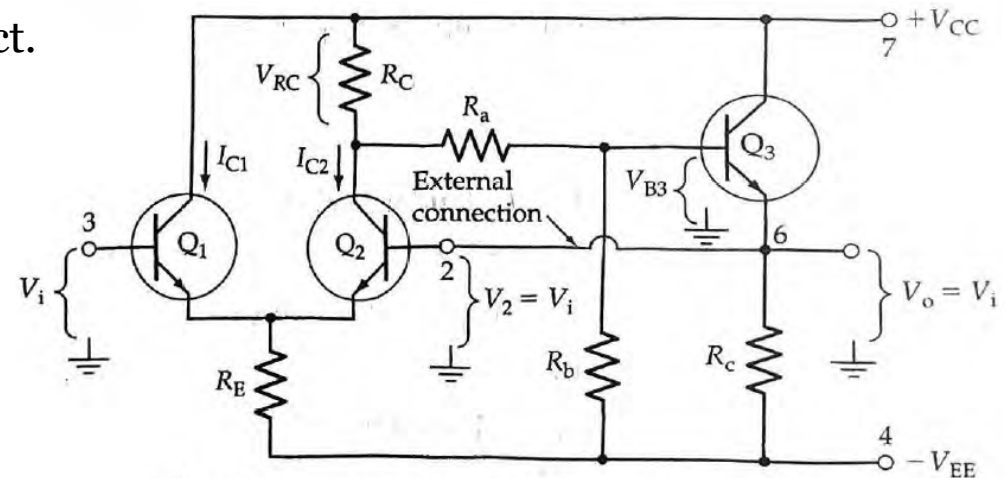


Suppose, $V_o < V_i \rightarrow$

$V_o = V_i$ due to similar -ve feedback effect.

When V_i at 3 is increased or decreased \rightarrow

V_o follows V_i perfectly.



Non-Inverting Amplifiers

Portion of V_o is fed back to inverting input.

V_o changes as necessary $\rightarrow V_{R_3} = V_i = I_2 R_3$

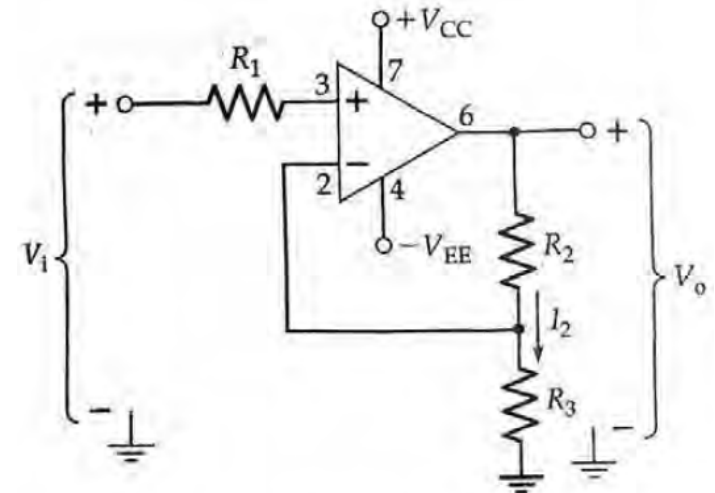
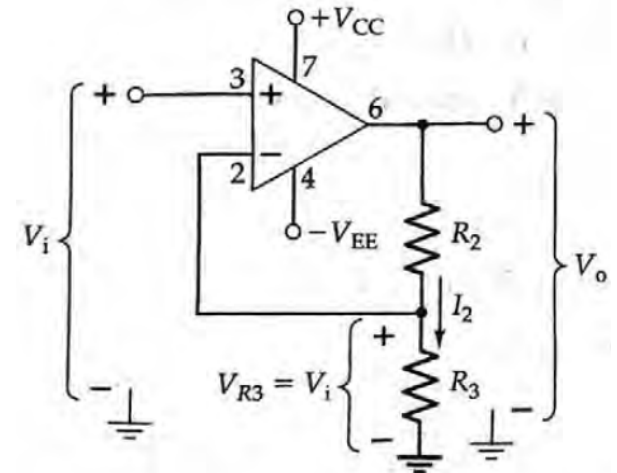
$I_2 \gg$ op-amp input bias current

$$I_2 = \frac{V_o}{R_2 + R_3}$$

$$\text{Circuit voltage gain, } A_{CL} = \frac{V_o}{V_i} = \frac{I_2(R_2 + R_3)}{I_2 R_3} = \frac{R_2 + R_3}{R_3}$$

V_{R1} is produced by input bias current.

$$R_1 = R_2 || R_3$$



Non-Inverting Amplifiers

Problem-45:

Design a direct-coupled non-inverting amplifier using 741 op-amp. The output voltage is to be 2 V when the input is 50 mV. Also, calculate typical input and output impedances for the amplifier. Assume $I_{B(max)} = 500 \text{ nA}$, $A_v = 200000$, $r_i = 2 \text{ M}\Omega$, $r_o = 75 \Omega$.

$$I_{2(min)} = 100I_{B(max)} = 100 \times (500 \times 10^{-9}) = 50 \mu\text{A}$$

$$R_3 = V_i / I_2 = (50 \times 10^{-3}) / (50 \times 10^{-6}) = 1 \text{ k}\Omega$$

$$R_2 + R_3 = V_o / I_2 = 2 / (50 \times 10^{-6}) = 40 \text{ k}\Omega$$

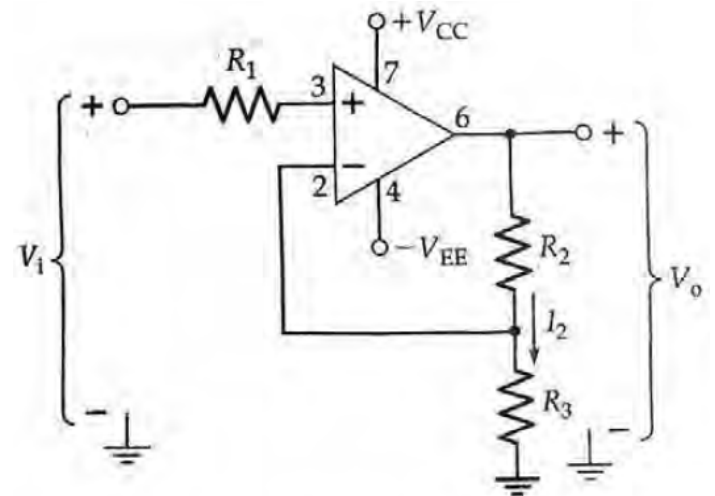
$$R_2 = (R_2 + R_3) - R_3 = 40 - 1 = 39 \text{ k}\Omega$$

$$R_1 = R_2 || R_3 = 39 || 1 \approx 1 \text{ k}\Omega$$

$$B = \frac{R_3}{R_2 + R_3} = \frac{1}{39 + 1} = \frac{1}{40}$$

$$Z_i = (1 + A_v B) r_i = \left[1 + 200000 \frac{1}{40} \right] \times 2 \times 10^6 \approx 10000 \text{ M}\Omega$$

$$Z_o = \frac{r_o}{1 + A_v B} = \frac{75}{1 + (200000/40)} \approx 0.015 \Omega$$



Inverting Amplifiers

Non-inverting input terminal is grounded via R_3
 Inverting input terminal \rightarrow voltage \approx ground.
 virtual ground.

Very small input voltage difference is amplified by open-loop gain.
 Portion of V_o is fed back via R_2 and R_1 to correct changes.

Circuit input current, $I_1 = V_{R1}/R_1 = V_i/R_1$ [$V_{R1} = V_i$]

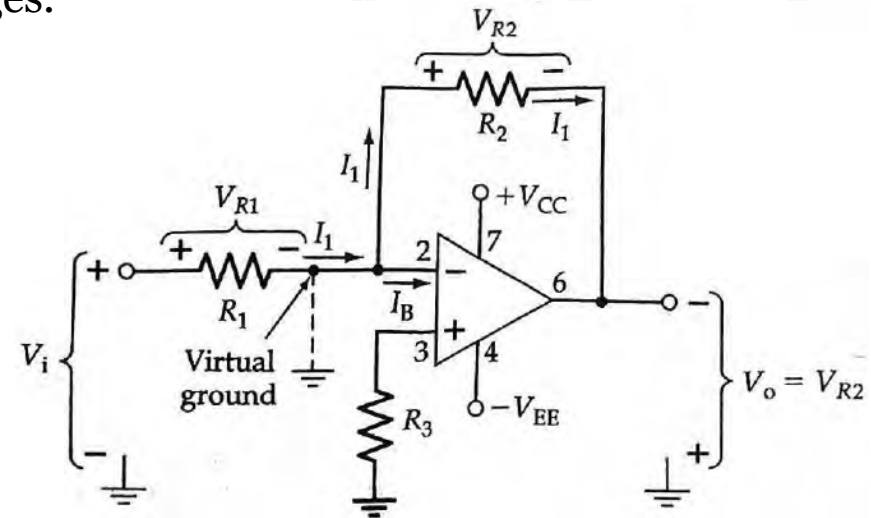
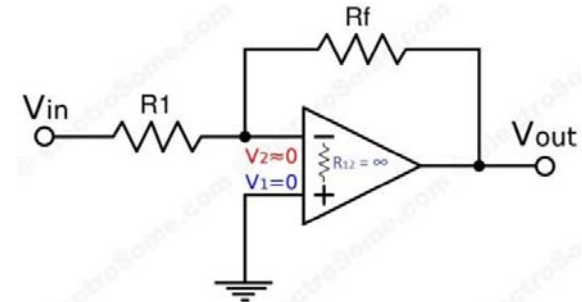
$I_1 \gg$ input bias current I_B

$$V_{R2} = I_1 R_2$$

$$V_o = -I_1 R_2$$

Circuit voltage gain, $A_{CL} = V_o/V_i = -I_1 R_2/I_1 R_1 = -R_2/R_1$

Input impedance, $Z_i = R_1$



Inverting Amplifiers

Problem-47:

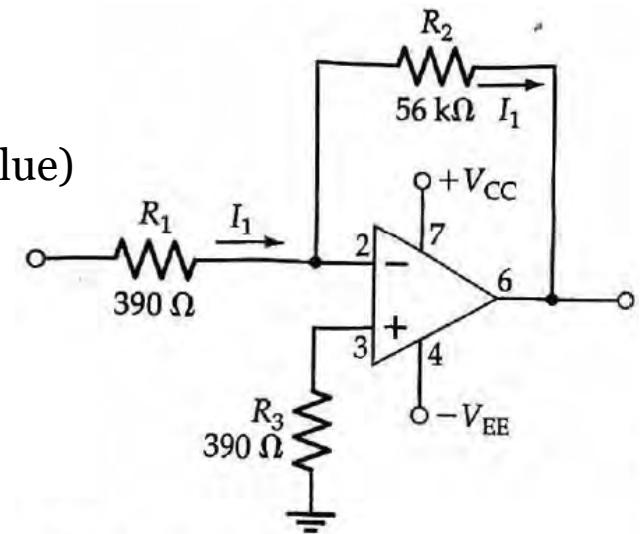
Design a direct-coupled inverting amplifier using a 741 op-amp. The input voltage amplitude is 20 mV and the voltage gain is to be 144. Assume $I_{B(max)} = 500$ nA.

$$I_{1(min)} = 100I_{B(max)} = 100 \times (500 \times 10^{-9}) = 50 \mu\text{A}$$

$$R_1 = V_i / I_1 = (20 \times 10^{-3}) / (50 \times 10^{-6}) = 400 \Omega \text{ (use } 390 \Omega \text{ standard value)}$$

$$R_2 = A_{CL} R_1 = 144 \times 390 = 56.2 \text{ k}\Omega \text{ (use } 56 \text{ k}\Omega \text{ standard value)}$$

$$R_3 = R_1 || R_2 = 390 || (56 \times 10^3) \approx 390 \Omega$$



Summing Amplifiers

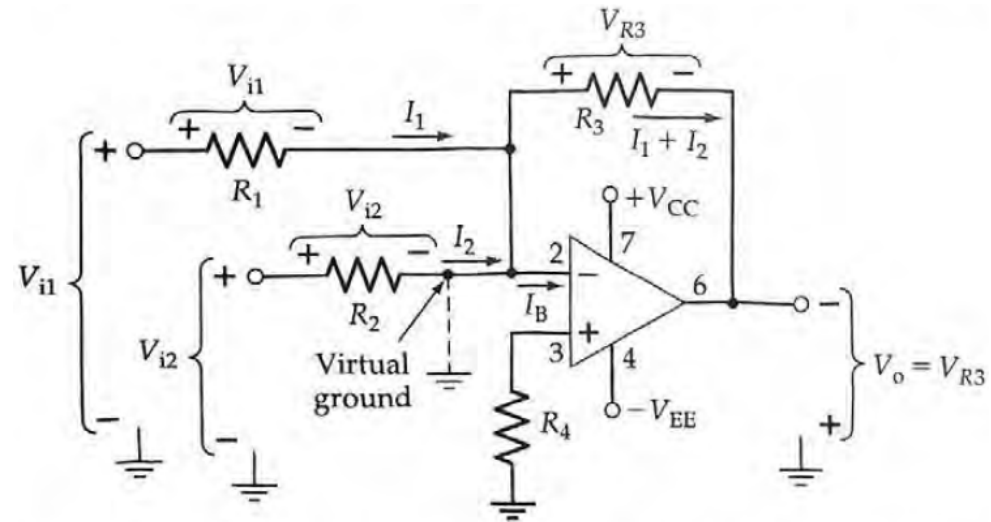
2 inputs applied to R_1 and R_2

2 input voltages, V_{i1} and V_{i2}

2 input currents \rightarrow $I_1 = V_{i1}/R_1$
 $I_2 = V_{i2}/R_2$

Output voltage,

$$\begin{aligned} V_o &= -(I_1 + I_2)R_3 \\ &= -\left[\frac{V_{i1}}{R_1} + \frac{V_{i2}}{R_2}\right]R_3 \\ &= -\frac{R_3}{R_1}[V_{i1} + V_{i2}] \quad [R_1 = R_2] \\ &= -[V_{i1} + V_{i2}] \quad [R_1 = R_2 = R_3] \end{aligned}$$



Summing Amplifiers

Problem-48:

Design a three-input summing amplifier using an op-amp and to have a voltage gain of 3. Calculate the resistor currents and the output voltage when all three inputs are 1 V.

Select $R_4 = 1 \text{ M}\Omega$

Let, $R_1 = R_2 = R_3 = R_4/A_{CL} = 1 \times 10^6/3$
 $= 333 \text{ k}\Omega$ (use 330 k Ω standard value)

$$I_1 = I_2 = I_3 = V_i/R_1 = 1/(330 \times 10^3) = 3.03 \text{ }\mu\text{A}$$

$$I_4 = I_1 + I_2 + I_3 = 3.03 + 3.03 + 3.03 = 9.09 \text{ }\mu\text{A}$$

$$V_o = -I_4 R_4 = -9.09 \times 10^{-6} \times 1 \times 10^6 = -9.09 \text{ V}$$

